PIC Microcontroller and Embedded Systems
Using Assembly and C for PIC18

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PIC MICROCONTROLLER
AND EMBEDDED SYSTEMS

Using Assembly and C for PIC18

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Pearson Education International
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Regard man as a mine rich in gems of inestimable value. Education can, alone, cause it to reveal its treasures, and enable mankind to benefit therefrom.

Baha'u'llah
# BRIEF CONTENTS

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INTRODUCTION

Products using microprocessors generally fall into two categories. The first category uses high-performance microprocessors such as the Pentium in applications where system performance is critical. We have an entire book dedicated to this topic, *The 80x86 IBM PC and Compatible Computers, Volumes I and II*, from Prentice Hall. In the second category of applications, performance is secondary; issues of cost, space, power, and rapid development are more critical than raw processing power. The microprocessor for this category is often called a microcontroller.

This book is for the second category of applications. The PIC18 is a widely used microcontroller. There are many reasons for this, including the existence of massive support in both software and hardware by Microchip Technology. This book is intended for use in college-level courses teaching microcontrollers and embedded systems. It not only establishes a foundation of Assembly language programming, but also provides a comprehensive treatment of PIC18 interfacing for engineering students. From this background, the design and interfacing of microcontroller-based embedded systems can be explored. This book can also be used by practicing technicians, hardware engineers, computer scientists, and hobbyists. It is an ideal source for those building stand-alone projects, or projects in which data is collected and fed into a PC for distribution on a network.

Prerequisites

Readers should have had an introductory digital course. Knowledge of Assembly language would be helpful but is not necessary. Although the book is written for those with no background in Assembly language programming, students with prior Assembly language experience will be able to gain a mastery of PIC18 architecture very rapidly and start on their projects right away. For the PIC18 C programming sections of the book, a basic knowledge of C programming is required. We use the PIC18 C compiler from Microchip Technology throughout the book. The PIC18 C compiler is compatible with MPLAB and is available for free from the Microchip website (www.microchip.com). We encourage you to use the MPLAB to simulate and run the programs in this book.

Overview

A systematic, step-by-step approach is used to cover various aspects of PIC18 C and Assembly language programming and interfacing. Many examples and sample programs are given to clarify the concepts and provide students with an opportunity to learn by doing. Review questions are provided at the end of each section to reinforce the main points of the section.

Chapter 0 covers number systems (binary, decimal, and hex), and provides an introduction to basic logic gates and computer terminology. This is designed especially for students, such as mechanical engineering students who have not taken a digital logic course or those who need to refresh their memory on these topics.

Chapter 1 discusses the history of the PIC18 and features of other PIC family members such as the PIC16. It also provides a list of various members of
the PIC18 family.

Chapter 2 discusses the internal architecture of the PIC18 and explains the use of a PIC18 assembler to create ready-to-run programs. It also explores the stack and the flag register.

In Chapter 3 the topics of loop, jump, and call instructions are discussed, with many programming examples.

Chapter 4 is dedicated to the discussion of I/O ports. This allows students who are working on a project to start experimenting with PIC18 I/O interfacing and start the project as soon as possible.

Chapter 5 is dedicated to arithmetic, logic instructions, and programs.

Chapter 6 covers the PIC18 addressing modes and explains how to access the data stored in the code space of the PIC18, as well as how to do bank switching.

The C programming of the PIC18 is covered in Chapter 7. We use the PIC18 C compiler from Microchip Technology for this and other C programs of the PIC18 family throughout the book. The PIC18 C compiler is compatible with MPLAB and is available for free from the www.microchip.com website.

In Chapter 8 we discuss the hardware connection of the PIC18 chip.

Chapter 9 describes the PIC18 timers and how to use them as event counters.

Chapter 10 is dedicated to serial data communication of the PIC18 and its interfacing to the RS232. It also shows PIC18 communication with COM ports of the x86 IBM PC and compatible computers.

Chapter 11 provides a detailed discussion of PIC18 interrupts with many examples on how to write interrupt handler programs.

Chapter 12 shows PIC18 interfacing with real-world devices such as LCDs and keyboards.

Chapter 13 shows PIC18 interfacing with real-world devices such as DAC chips, ADC chips, and sensors.

In Chapter 14 we cover how to use PIC18 Flash and EEPROM memories for data storage.

Chapter 15 covers the CCP and ECCP modules inside the PIC18 and shows how they are used.

Chapter 16 shows how to connect and program the DS1306 real-time clock chip using the SPI bus protocol.

Finally, Chapter 17 shows basic interfacing to relays, optoisolators, and motors.

The appendices have been designed to provide all reference material required for the topics covered in the book. Appendix A describes each PIC18 instruction in detail, with examples. Appendix B describes basics of wire wrapping. Appendix C covers IC technology and logic families, as well as PIC18 I/O port interfacing and fan-out. Make sure you study this before connecting the PIC18 to an external device. In Appendix D, the use of flowcharts and pseudocode is explored. Appendix E is for students familiar with x86 and 8051 architectures who need to make a rapid transition to PIC18 architecture. Appendix F provides the table of ASCII characters. Appendix G lists resources for assembler shareware, and electronics parts. Appendix H contains data sheets for the PIC18 chip.
Lab Manual

The lab manual covers some very basic labs and can be found at the www.MicroDigitalEd.com website. The more advanced and rigorous lab assignments are left up to the instructors depending on the course objectives, class level, and whether the course is graduate or undergraduate. The support materials for this and other books by the authors can be found on this website, too.

Solutions Manual/PowerPoint® Slides

The end-of-chapter problems cover some very basic concepts. The more challenging and rigorous homework assignments are left up to the instructors depending on the course objectives, class level, and whether the course is graduate or undergraduate. The solutions manual was produced with the help of Mr. Rastì and Prof. Faramarz Mortezaei. The solutions manual and PowerPoint® slides for the drawings are available online for instructors only.

Online Instructor Resources

To access supplementary materials online, instructors need to request an instructor access code. Go to www.prenhall.com, click the Instructor Resource Center link, and then click Register Today for an instructor access code. Within 48 hours after registering you will receive a confirming e-mail including an instructor access code. Once you have received your code, go to the site and log on for full instructions on downloading the materials you wish to use.

Acknowledgments

This book is the result of the dedication and encouragement of many individuals. Our sincere and heartfelt appreciation goes to all of them.

First, we would like to thank Mr. Javad Rasti of Esfahan University. His detailed and thorough reading of the chapters resulted in finding and fixing some of the errors before the book was published. Many of the drawings and tables in this book were recreated from PIC18 data sheets by Pedram Mazidi. Numerous professors, professional engineers, and students found errors or made suggestions in improving this book. We would like to thank all of them sincerely for their enthusiasm and support. They are Javad Rasti (Esfahan University), Vahid Mokhtari (BIHE), Mohammadi Abdar (Azad University), Clyde Knight, Sam Waterman, and Faramarz Mortezaei (all from DeVry University), Frank Fortman, David Goodman, and Maryam Mohseni. Their encouragement meant a great deal to us in writing this book.
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We enjoyed writing this book, and hope you enjoy reading it and using it for your courses and projects. Please let us know if you have any suggestions or find any errors.

**Assemblers/Compiler**

The MPLAB and PIC18 C compilers can be downloaded from the following website:

http://www.microchip.com
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This book is dedicated

to the memory of Mr. N. Akhtar-Khavari and Mr. Z. Mahrami

for their dedication to the cause of world peace.

– Muhammad Ali Mazidi

To Tony and Jim for their friendship and faith in me over the years.

– Rolin D. McKinlay

I dedicate my part to my brother John, who reached out to me even though we
lived in different homes. The experience that was given provided me the
inspiration to look for something more in life.

– Danny Causey
CHAPTER 0

INTRODUCTION TO COMPUTING

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Convert any number from base 2, base 10, or base 16 to any of the other two bases
- Add and subtract hex numbers
- Add binary numbers
- Represent any binary number in 2’s complement
- Represent an alphanumeric string in ASCII code
- Describe the logical operations AND, OR, NOT, XOR, NAND, and NOR
- Use logic gates to diagram simple circuits
- Explain the difference between a bit, a nibble, a byte, and a word
- Give precise mathematical definitions of the terms kilobyte, megabyte, gigabyte, and terabyte
- Explain the difference between RAM and ROM and describe their use
- Describe the purpose of the major components of a computer system
- List the three types of buses found in computers and describe the purpose of each type of bus
- Describe the role of the CPU in computer systems
- List the major components of the CPU and describe the purpose of each
To understand the software and hardware of a microcontroller-based system, one must first master some very basic concepts underlying computer design. In this chapter (which in the tradition of digital computers is called Chapter 0), the fundamentals of numbering and coding systems are presented. After an introduction to logic gates, an overview of the workings inside the computer is given. Finally, in the last section we give a brief history of CPU architecture. Although some readers may have an adequate background in many of the topics of this chapter, it is recommended that the material be scanned, however briefly.

SECTION 0.1: NUMBERING AND CODING SYSTEMS

Whereas human beings use base 10 (decimal) arithmetic, computers use the base 2 (binary) system. In this section we explain how to convert from the decimal system to the binary system, and vice versa. The convenient representation of binary numbers, called hexadecimal, also is covered. Finally, the binary format of the alphanumeric code, called ASCII, is explored.

Decimal and binary number systems

Although there has been speculation that the origin of the base 10 system is the fact that human beings have 10 fingers, there is absolutely no speculation about the reason behind the use of the binary system in computers. The binary system is used in computers because 1 and 0 represent the two voltage levels of on and off. Whereas in base 10 there are 10 distinct symbols, 0, 1, 2, .... 9, in base 2 there are only two, 0 and 1, with which to generate numbers. Base 10 contains digits 0 through 9; binary contains digits 0 and 1 only. These two binary digits, 0 and 1, are commonly referred to as bits.

Converting from decimal to binary

One method of converting from decimal to binary is to divide the decimal number by 2 repeatedly, keeping track of the remainders. This process continues until the quotient becomes zero. The remainders are then written in reverse order to obtain the binary number. This is demonstrated in Example 0-1.

Example 0-1

Convert $25_{10}$ to binary.

<table>
<thead>
<tr>
<th>Quotient</th>
<th>Remainder</th>
</tr>
</thead>
<tbody>
<tr>
<td>25/2 =</td>
<td>12</td>
</tr>
<tr>
<td>12/2 =</td>
<td>6</td>
</tr>
<tr>
<td>6/2 =</td>
<td>3</td>
</tr>
<tr>
<td>3/2 =</td>
<td>1</td>
</tr>
<tr>
<td>1/2 =</td>
<td>0</td>
</tr>
</tbody>
</table>

Therefore, $25_{10} = 11001_{2}$. 
Converting from binary to decimal

To convert from binary to decimal, it is important to understand the concept of weight associated with each digit position. First, as an analogy, recall the weight of numbers in the base 10 system, as shown in the diagram. By the same token, each digit position of a number in base 2 has a weight associated with it:

<table>
<thead>
<tr>
<th>Binary</th>
<th>Decimal</th>
</tr>
</thead>
<tbody>
<tr>
<td>10101</td>
<td>16</td>
</tr>
<tr>
<td>1 × 2^4</td>
<td>16</td>
</tr>
<tr>
<td>1 × 2^3</td>
<td>8</td>
</tr>
<tr>
<td>0 × 2^2</td>
<td>0</td>
</tr>
<tr>
<td>0 × 2^1</td>
<td>0</td>
</tr>
<tr>
<td>1 × 2^0</td>
<td>1</td>
</tr>
</tbody>
</table>

Knowing the weight of each bit in a binary number makes it simple to add them together to get its decimal equivalent, as shown in Example 0-2.

**Example 0-2**

Convert 11001₂ to decimal.

**Solution:**

<table>
<thead>
<tr>
<th>Weight</th>
<th>Digits</th>
<th>Sum</th>
</tr>
</thead>
<tbody>
<tr>
<td>16</td>
<td>1</td>
<td>16</td>
</tr>
<tr>
<td>8</td>
<td>1</td>
<td>8</td>
</tr>
<tr>
<td>4</td>
<td>0</td>
<td>4</td>
</tr>
<tr>
<td>2</td>
<td>0</td>
<td>2</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>

Therefore, 11001₂ = 39₁₀.

Knowing the weight associated with each binary bit position allows one to convert a decimal number to binary directly instead of going through the process of repeated division. This is shown in Example 0-3.

**Example 0-3**

Use the concept of weight to convert 39₁₀ to binary.

**Solution:**

<table>
<thead>
<tr>
<th>Weight</th>
<th>32</th>
<th>16</th>
<th>8</th>
<th>4</th>
<th>2</th>
<th>1</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td></td>
</tr>
</tbody>
</table>

Therefore, 39₁₀ = 10011₀₂.
Hexadecimal system

Base 16, or the *hexadecimal* system as it is called in computer literature, is used as a convenient representation of binary numbers. For example, it is much easier for a human being to represent a string of 0s and 1s such as 100010010110 as its hexadecimal equivalent of 896H. The binary system has 2 digits, 0 and 1. The base 10 system has 10 digits, 0 through 9. The hexadecimal (base 16) system has 16 digits. In base 16, the first 10 digits, 0 to 9, are the same as in decimal, and for the remaining six digits, the letters A, B, C, D, E, and F are used. Table 0-1 shows the equivalent binary, decimal, and hexadecimal representations for 0 to 15.

### Converting between binary and hex

To represent a binary number as its equivalent hexadecimal number, start from the right and group 4 bits at a time, replacing each 4-bit binary number with its hex equivalent shown in Table 0-1. To convert from hex to binary, each hex digit is replaced with its 4-bit binary equivalent. See Examples 0-4 and 0-5.

<table>
<thead>
<tr>
<th>Example 0-4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Represent binary 100111110101 in hex.</td>
</tr>
<tr>
<td><strong>Solution:</strong></td>
</tr>
<tr>
<td>First the number is grouped into sets of 4 bits: 1001 1111 0101. Then each group of 4 bits is replaced with its hex equivalent:</td>
</tr>
<tr>
<td>1001 1111 0101</td>
</tr>
<tr>
<td>9 F 5</td>
</tr>
<tr>
<td>Therefore, 100111110101₂ = 9F5 hexadecimal.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Example 0-5</th>
</tr>
</thead>
<tbody>
<tr>
<td>Convert hex 29B to binary.</td>
</tr>
<tr>
<td><strong>Solution:</strong></td>
</tr>
</tbody>
</table>
| \[
\begin{array}{c|c|c|c}
\hline
2 & 9 & B \\
\hline
0010 & 1001 & 1011 \\
\hline
\end{array}
\] |
| Dropping the leading zeros gives 1010011011. |

### Converting from decimal to hex

Converting from decimal to hex could be approached in two ways:

1. Convert to binary first and then convert to hex. Example 0-6 shows this method of converting decimal to hex.

2. Convert directly from decimal to hex by repeated division, keeping track of the remainders. Experimenting with this method is left to the reader.
Example 0-6

(a) Convert $45_{10}$ to hex.

\[ 45_{10} = 0010 \quad 1101_2 = 2D \text{ hex} \]

(b) Convert $629_{10}$ to hex.

\[ 629_{10} = (512 + 64 + 32 + 16 + 4 + 1) = 0010 \quad 0111 \quad 0101_2 = 275 \text{ hex} \]

(c) Convert $1714_{10}$ to hex.

\[ 1714_{10} = (1024 + 512 + 128 + 32 + 16 + 2) = 0110 \quad 1011 \quad 0010_2 = 6B2 \text{ hex} \]

**Converting from hex to decimal**

Conversion from hex to decimal can also be approached in two ways:

1. Convert from hex to binary and then to decimal. Example 0-7 demonstrates this method of converting from hex to decimal.
2. Convert directly from hex to decimal by summing the weight of all digits.

Example 0-7

Convert the following hexadecimal numbers to decimal.

(a) $6B2_{16} = 0110 \quad 1011 \quad 0010_2$

\[ \begin{array}{cccccccc}
1024 & 512 & 256 & 128 & 64 & 32 & 16 & 8 \\
1 & 1 & 0 & 1 & 0 & 1 & 0 & 0
\end{array} \]

\[ 1024 + 512 + 128 + 32 + 16 + 2 = 1714_{10} \]

(b) $9F2D_{16} = 1001 \quad 1111 \quad 0010 \quad 1101_2$

\[ \begin{array}{cccccccccc}
32768 & 16384 & 8192 & 4096 & 2048 & 1024 & 512 & 256 & 128 & 64 \\
1 & 0 & 0 & 1 & 1 & 1 & 1 & 1 & 0 & 0
\end{array} \]

\[ 32768 + 4096 + 2048 + 1024 + 512 + 256 + 32 + 8 + 4 + 1 = 40749_{10} \]
Table 0-2: Counting in Bases

<table>
<thead>
<tr>
<th>Decimal</th>
<th>Binary</th>
<th>Hex</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>00000</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>00001</td>
<td>1</td>
</tr>
<tr>
<td>2</td>
<td>00010</td>
<td>2</td>
</tr>
<tr>
<td>3</td>
<td>00011</td>
<td>3</td>
</tr>
<tr>
<td>4</td>
<td>00100</td>
<td>4</td>
</tr>
<tr>
<td>5</td>
<td>00101</td>
<td>5</td>
</tr>
<tr>
<td>6</td>
<td>00110</td>
<td>6</td>
</tr>
<tr>
<td>7</td>
<td>00111</td>
<td>7</td>
</tr>
<tr>
<td>8</td>
<td>01000</td>
<td>8</td>
</tr>
<tr>
<td>9</td>
<td>01001</td>
<td>9</td>
</tr>
<tr>
<td>10</td>
<td>01010</td>
<td>A</td>
</tr>
<tr>
<td>11</td>
<td>01011</td>
<td>B</td>
</tr>
<tr>
<td>12</td>
<td>01100</td>
<td>C</td>
</tr>
<tr>
<td>13</td>
<td>01101</td>
<td>D</td>
</tr>
<tr>
<td>14</td>
<td>01110</td>
<td>E</td>
</tr>
<tr>
<td>15</td>
<td>01111</td>
<td>F</td>
</tr>
<tr>
<td>16</td>
<td>10000</td>
<td>10</td>
</tr>
<tr>
<td>17</td>
<td>10001</td>
<td>11</td>
</tr>
<tr>
<td>18</td>
<td>10010</td>
<td>12</td>
</tr>
<tr>
<td>19</td>
<td>10011</td>
<td>13</td>
</tr>
<tr>
<td>20</td>
<td>10100</td>
<td>14</td>
</tr>
<tr>
<td>21</td>
<td>10101</td>
<td>15</td>
</tr>
<tr>
<td>22</td>
<td>10110</td>
<td>16</td>
</tr>
<tr>
<td>23</td>
<td>10111</td>
<td>17</td>
</tr>
<tr>
<td>24</td>
<td>11000</td>
<td>18</td>
</tr>
<tr>
<td>25</td>
<td>11001</td>
<td>19</td>
</tr>
<tr>
<td>26</td>
<td>11010</td>
<td>1A</td>
</tr>
<tr>
<td>27</td>
<td>11011</td>
<td>1B</td>
</tr>
<tr>
<td>28</td>
<td>11100</td>
<td>1C</td>
</tr>
<tr>
<td>29</td>
<td>11101</td>
<td>1D</td>
</tr>
<tr>
<td>30</td>
<td>11110</td>
<td>1E</td>
</tr>
<tr>
<td>31</td>
<td>11111</td>
<td>1F</td>
</tr>
</tbody>
</table>

Counting in bases 10, 2, and 16

To show the relationship between all three bases, in Table 0-2 we show the sequence of numbers from 0 to 31 in decimal, along with the equivalent binary and hex numbers. Notice in each base that when one more is added to the highest digit, that digit becomes zero and a 1 is carried to the next-highest digit position. For example, in decimal, 9 + 1 = 0 with a carry to the next-highest position. In binary, 1 + 1 = 0 with a carry; similarly, in hex, F + 1 = 0 with a carry.

### Addition of binary and hex numbers

The addition of binary numbers is a very straightforward process. Table 0-3 shows the addition of two bits. The discussion of subtraction of binary numbers is bypassed since all computers use the addition process to implement subtraction. Although computers have adder circuitry, there is no separate circuitry for subtractors. Instead, adders are used in conjunction with 2's complement circuitry to perform subtraction. In other words, to implement “x − y”, the computer takes the 2’s complement of y and adds it to x. The concept of 2’s complement is reviewed next. Example 0-8 shows the addition of binary numbers.

Table 0-3: Binary Addition

<table>
<thead>
<tr>
<th>A + B</th>
<th>Carry</th>
<th>Sum</th>
</tr>
</thead>
<tbody>
<tr>
<td>0 + 0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0 + 1</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1 + 0</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1 + 1</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>

Example 0-8

Add the following binary numbers. Check against their decimal equivalents.

Solution:

<table>
<thead>
<tr>
<th>Binary</th>
<th>Decimal</th>
</tr>
</thead>
<tbody>
<tr>
<td>1101</td>
<td>13</td>
</tr>
<tr>
<td>+ 1001</td>
<td>9</td>
</tr>
<tr>
<td>10110</td>
<td>22</td>
</tr>
</tbody>
</table>
2's complement

To get the 2's complement of a binary number, invert all the bits and then add 1 to the result. Inverting the bits is simply a matter of changing all 0s to 1s and 1s to 0s. This is called the 1's complement. See Example 0-9.

<table>
<thead>
<tr>
<th>Example 0-9</th>
</tr>
</thead>
<tbody>
<tr>
<td>Take the 2's complement of 10011101.</td>
</tr>
<tr>
<td>Solution:</td>
</tr>
<tr>
<td>10011101</td>
</tr>
<tr>
<td>01100010</td>
</tr>
<tr>
<td>+ 1</td>
</tr>
<tr>
<td>01100011</td>
</tr>
</tbody>
</table>

Addition and subtraction of hex numbers

In studying issues related to software and hardware of computers, it is often necessary to add or subtract hex numbers. Mastery of these techniques is essential. Hex addition and subtraction are discussed separately below.

Addition of hex numbers

This section describes the process of adding hex numbers. Starting with the least significant digits, the digits are added together. If the result is less than 16, write that digit as the sum for that position. If it is greater than 16, subtract 16 from it to get the digit and carry 1 to the next digit. The best way to explain this is by example, as shown in Example 0-10.

<table>
<thead>
<tr>
<th>Example 0-10</th>
</tr>
</thead>
<tbody>
<tr>
<td>Perform hex addition: 23D9 + 94BE.</td>
</tr>
<tr>
<td>Solution:</td>
</tr>
<tr>
<td>23D9</td>
</tr>
<tr>
<td>+ 94BE</td>
</tr>
<tr>
<td>B897</td>
</tr>
<tr>
<td>MSD: 2 + 9 = B</td>
</tr>
</tbody>
</table>

Subtraction of hex numbers

In subtracting two hex numbers, if the second digit is greater than the first, borrow 16 from the preceding digit. See Example 0-11.

ASCII code

The discussion so far has revolved around the representation of number systems. Because all information in the computer must be represented by 0s and 1s, binary patterns must be assigned to letters and other characters. In the 1960s a standard representation called ASCII (American Standard Code for Information Interchange) was established. The ASCII (pronounced “ask-E”) code assigns bina-
ry patterns for numbers 0 to 9, all the letters of the English alphabet, both uppercase (capital) and lowercase, and many control codes and punctuation marks. The great advantage of this system is that it is used by most computers, so that information can be shared among computers. The ASCII system uses a total of 7 bits to represent each code. For example, 100 0001 is assigned to the uppercase letter “A” and 110 0001 is for the lowercase “a”. Often, a zero is placed in the most-significant bit position to make it an 8-bit code. Figure 0-1 shows selected ASCII codes. A complete list of ASCII codes is given in Appendix F.

Notice that the pattern of ASCII codes was designed to allow for easy manipulation of ASCII data. For example, digits 0 through 9 are represented by ASCII codes 30 through 39. This enables a program to easily convert ASCII to decimal by masking off the “3” in the upper nibble. Also notice that there is a relationship between the uppercase and lowercase letters. The uppercase letters are represented by ASCII codes 41 through 5A while lowercase letters are represented by codes 61 through 7A. Looking at the binary code, the only bit that is different between the uppercase “A” and lowercase “a” is bit 5. Therefore, conversion between uppercase and lowercase is as simple as changing bit 5 of the ASCII code.

### Example 0-11

Perform hex subtraction: 59F − 2B8.

**Solution:**

<table>
<thead>
<tr>
<th>Hex</th>
<th>Symbol</th>
<th>Hex</th>
<th>Symbol</th>
</tr>
</thead>
<tbody>
<tr>
<td>41</td>
<td>A</td>
<td>61</td>
<td>a</td>
</tr>
<tr>
<td>42</td>
<td>B</td>
<td>62</td>
<td>b</td>
</tr>
<tr>
<td>43</td>
<td>C</td>
<td>63</td>
<td>c</td>
</tr>
<tr>
<td>44</td>
<td>D</td>
<td>64</td>
<td>d</td>
</tr>
<tr>
<td></td>
<td>...</td>
<td></td>
<td>...</td>
</tr>
<tr>
<td>59</td>
<td>Y</td>
<td>79</td>
<td>y</td>
</tr>
<tr>
<td>5A</td>
<td>Z</td>
<td>7A</td>
<td>z</td>
</tr>
</tbody>
</table>

**Figure 0-1. Selected ASCII Codes**

<table>
<thead>
<tr>
<th>Hex</th>
<th>Symbol</th>
<th>Hex</th>
<th>Symbol</th>
</tr>
</thead>
<tbody>
<tr>
<td>41</td>
<td>A</td>
<td>61</td>
<td>a</td>
</tr>
<tr>
<td>42</td>
<td>B</td>
<td>62</td>
<td>b</td>
</tr>
<tr>
<td>43</td>
<td>C</td>
<td>63</td>
<td>c</td>
</tr>
<tr>
<td>44</td>
<td>D</td>
<td>64</td>
<td>d</td>
</tr>
<tr>
<td></td>
<td>...</td>
<td></td>
<td>...</td>
</tr>
<tr>
<td>59</td>
<td>Y</td>
<td>79</td>
<td>y</td>
</tr>
<tr>
<td>5A</td>
<td>Z</td>
<td>7A</td>
<td>z</td>
</tr>
</tbody>
</table>

---

### Review Questions

1. Why do computers use the binary number system instead of the decimal system?
2. Convert 34₁₀ to binary and hex.
3. Convert 110101₂ to hex and decimal.
5. Convert 101100₁₀ to its 2’s complement representation.
6. Add 36BH + F6H.
7. Subtract 36BH − F6H.
8. Write “80x86 CPUs” in its ASCII code (in hex form).
SECTION 0.2: DIGITAL PRIMER

This section gives an overview of digital logic and design. First, we cover binary logic operations, then we show gates that perform these functions. Next, logic gates are put together to form simple digital circuits. Finally, we cover some logic devices commonly found in microcontroller interfacing.

Binary logic

As mentioned earlier, computers use the binary number system because the two voltage levels can be represented as the two digits 0 and 1. Signals in digital electronics have two distinct voltage levels. For example, a system may define 0 V as logic 0 and +5 V as logic 1. Figure 0-2 shows this system with the built-in tolerances for variations in the voltage. A valid digital signal in this example should be within either of the two shaded areas.

Logic gates

Binary logic gates are simple circuits that take one or more input signals and send out one output signal. Several of these gates are defined below.

AND gate

The AND gate takes two or more inputs and performs a logic AND on them. See the truth table and diagram of the AND gate. Notice that if both inputs to the AND gate are 1, the output will be 1. Any other combination of inputs will give a 0 output. The example shows two inputs, *x* and *y*. Multiple outputs are also possible for logic gates. In the case of AND, if all inputs are 1, the output is 1. If any input is 0, the output is 0.

<table>
<thead>
<tr>
<th>Inputs</th>
<th>Output</th>
</tr>
</thead>
<tbody>
<tr>
<td>X Y</td>
<td>X AND Y</td>
</tr>
<tr>
<td>0 0</td>
<td>0</td>
</tr>
<tr>
<td>0 1</td>
<td>0</td>
</tr>
<tr>
<td>1 0</td>
<td>0</td>
</tr>
<tr>
<td>1 1</td>
<td>1</td>
</tr>
</tbody>
</table>

OR gate

The OR logic function will output a 1 if one or more inputs is 1. If all inputs are 0, then and only then will the output be 0.

<table>
<thead>
<tr>
<th>Inputs</th>
<th>Output</th>
</tr>
</thead>
<tbody>
<tr>
<td>X Y</td>
<td>X OR Y</td>
</tr>
<tr>
<td>0 0</td>
<td>0</td>
</tr>
<tr>
<td>0 1</td>
<td>1</td>
</tr>
<tr>
<td>1 0</td>
<td>1</td>
</tr>
<tr>
<td>1 1</td>
<td>1</td>
</tr>
</tbody>
</table>

Tri-state buffer

A buffer gate does not change the logic level of the input. It is used to isolate or amplify the signal.

Figure 0-2. Binary Signals
Inverter
The inverter, also called NOT, outputs the value opposite to that input to the gate. That is, a 1 input will give a 0 output, while a 0 input will give a 1 output.

XOR gate
The XOR gate performs an exclusive-OR operation on the inputs. Exclusive-OR produces a 1 output if one (but only one) input is 1. If both operands are 0, the output is 0. Likewise, if both operands are 1, the output is also 0. Notice from the XOR truth table, that whenever the two inputs are the same, the output is 0. This function can be used to compare two bits to see if they are the same.

NAND and NOR gates
The NAND gate functions like an AND gate with an inverter on the output. It produces a 0 output when all inputs are 1; otherwise, it produces a 1 output. The NOR gate functions like an OR gate with an inverter on the output. It produces a 1 if all inputs are 0; otherwise, it produces a 0. NAND and NOR gates are used extensively in digital design because they are easy and inexpensive to fabricate. Any circuit that can be designed with AND, OR, XOR, and INVERTER gates can be implemented using only NAND and NOR gates. A simple example of this is given below. Notice in NAND, that if any input is 0, the output is 1. Notice in NOR, that if any input is 1, the output is 0.

Logic design using gates
Next we will show a simple logic design to add two binary digits. If we add two binary digits there are four possible outcomes:

<table>
<thead>
<tr>
<th>Carry</th>
<th>Sum</th>
</tr>
</thead>
<tbody>
<tr>
<td>0 + 0 =</td>
<td>0</td>
</tr>
<tr>
<td>0 + 1 =</td>
<td>0</td>
</tr>
<tr>
<td>1 + 0 =</td>
<td>0</td>
</tr>
<tr>
<td>1 + 1 =</td>
<td>1</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Logical Inverter</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input</td>
</tr>
<tr>
<td>X</td>
</tr>
<tr>
<td>0</td>
</tr>
<tr>
<td>1</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Logical XOR Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>Inputs</td>
</tr>
<tr>
<td>X Y</td>
</tr>
<tr>
<td>0 0</td>
</tr>
<tr>
<td>0 1</td>
</tr>
<tr>
<td>1 0</td>
</tr>
<tr>
<td>1 1</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Logical NAND Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>Inputs</td>
</tr>
<tr>
<td>X Y</td>
</tr>
<tr>
<td>0 0</td>
</tr>
<tr>
<td>0 1</td>
</tr>
<tr>
<td>1 0</td>
</tr>
<tr>
<td>1 1</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Logical NOR Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>Inputs</td>
</tr>
<tr>
<td>X Y</td>
</tr>
<tr>
<td>0 0</td>
</tr>
<tr>
<td>0 1</td>
</tr>
<tr>
<td>1 0</td>
</tr>
<tr>
<td>1 1</td>
</tr>
</tbody>
</table>
Notice that when we add $1 + 1$ we get $0$ with a carry to the next higher place. We will need to determine the sum and the carry for this design. Notice that the sum column above matches the output for the XOR function, and that the carry column matches the output for the AND function. Figure 0-3(a) shows a simple adder implemented with XOR and AND gates. Figure 0-3(b) shows the same logic circuit implemented with AND and OR gates and inverters.

![Figure 0-3. Two Implementations of a Half-Adder](image)

Figure 0-4 shows a block diagram of a half-adder. Two half-adders can be combined to form an adder that can add three input digits. This is called a full-adder. Figure 0-5 shows the logic diagram of a full-adder, along with a block diagram that masks the details of the circuit. Figure 0-6 shows a 3-bit adder using three full-adders.

![Figure 0-4. Block Diagram of a Half-Adder](image)

![Figure 0-5. Full-Adder Built From a Half-Adder](image)
Decoders

Another example of the application of logic gates is the decoder. Decoders are widely used for address decoding in computer design. Figure 0-7 shows decoders for 9 (1001 binary) and 5 (0101) using inverters and AND gates.

Flip-flops

A widely used component in digital systems is the flip-flop. Frequently, flip-flops are used to store data. Figure 0-8 shows the logic diagram, block diagram, and truth table for a flip-flop.

The D flip-flop is widely used to latch data. Notice from the truth table that a D-FF grabs the data at the input as the clock is activated. A D-FF holds the data as long as the power is on.
Review Questions

1. The logical operation _____ gives a 1 output when all inputs are 1.
2. The logical operation _____ gives a 1 output when one or more of its inputs is 1.
3. The logical operation _____ is often used to compare two inputs to determine whether they have the same value.
4. A _____ gate does not change the logic level of the input.
5. Name a common use for flip-flops.
6. An address ______ is used to identify a predetermined binary address.

SECTION 0.3: INSIDE THE COMPUTER

In this section we provide an introduction to the organization and internal working of computers. The model used is generic, but the concepts discussed are applicable to all computers, including the IBM PC, PS/2, and compatibles. Before embarking on this subject, it will be helpful to review definitions of some of the most widely used terminology in computer literature, such as K, mega, giga, byte, ROM, RAM, and so on.

Some important terminology

One of the most important features of a computer is how much memory it has. Next we review terms used to describe amounts of memory in IBM PCs and compatibles. Recall from the discussion above that a bit is a binary digit that can have the value 0 or 1. A byte is defined as 8 bits. A nibble is half a byte, or 4 bits. A word is two bytes, or 16 bits. The display is intended to show the relative size of these units. Of course, they could all be composed of any combination of zeros and ones.

A kilobyte is $2^{10}$ bytes, which is 1024 bytes. The abbreviation K is often used to represent kilobytes. For example, some floppy disks hold 356K of data. A megabyte, or meg as some call it, is $2^{20}$ bytes. That is a little over 1 million bytes; it is exactly 1,048,576 bytes. Moving rapidly up the scale in size, a gigabyte is $2^{30}$ bytes (over 1 billion), and a terabyte is $2^{40}$ bytes (over 1 trillion). As an example of how some of these terms are used, suppose that a given computer has 16 megabytes of memory. That would be $16 \times 2^{20}$, or $2^4 \times 2^{20}$, which is $2^{24}$. Therefore 16 megabytes is $2^{24}$ bytes.

Two types of memory commonly used in microcomputers are RAM, which stands for "random access memory" (sometimes called read/write memory), and ROM, which stands for "read-only memory." RAM is used by the computer for temporary storage of programs that it is running. That data is lost when the computer is turned off. For this reason, RAM is sometimes called volatile memory. ROM contains programs and information essential to operation of the computer. The information in ROM is permanent, cannot be changed by the user, and is not lost when the power is turned off. Therefore, it is called nonvolatile memory.
Internal organization of computers

The internal working of every computer can be broken down into three parts: CPU (central processing unit), memory, and I/O (input/output) devices (see Figure 0-9). The function of the CPU is to execute (process) information stored in memory. The function of I/O devices such as the keyboard and video monitor is to provide a means of communicating with the CPU. The CPU is connected to memory and I/O through strips of wire called a bus. The bus inside a computer carries information from place to place just as a street bus carries people from place to place. In every computer there are three types of buses: address bus, data bus, and control bus.

For a device (memory or I/O) to be recognized by the CPU, it must be assigned an address. The address assigned to a given device must be unique; no two devices are allowed to have the same address. The CPU puts the address (in binary, of course) on the address bus, and the decoding circuitry finds the device. Then the CPU uses the data bus either to get data from that device or to send data to it. The control buses are used to provide read or write signals to the device to indicate if the CPU is asking for information or sending information. Of the three buses, the address bus and data bus determine the capability of a given CPU.

![Figure 0-9. Inside the Computer](image)

More about the data bus

Because data buses are used to carry information in and out of a CPU, the more data buses available, the better the CPU. If one thinks of data buses as highway lanes, it is clear that more lanes provide a better pathway between the CPU and its external devices (such as printers, RAM, ROM, etc.; see Figure 0-10). By the same token, that increase in the number of lanes increases the cost of construction. More data buses mean a more expensive CPU and computer. The average size of data buses in CPUs varies between 8 and 64. Early personal computers such as Apple 2 used an 8-bit data bus, while supercomputers such as Cray use a 64-bit data bus. Data buses are bidirectional, because the CPU must use them either to receive or to send data. The processing power of a computer is related to the size of its buses, because an 8-bit bus can send out 1 byte a time, but a 16-bit bus can send out 2 bytes at a time, which is twice as fast.
More about the address bus

Because the address bus is used to identify the devices and memory connected to the CPU, the more address buses available, the larger the number of devices that can be addressed. In other words, the number of address buses for a CPU determines the number of locations with which it can communicate. The number of locations is always equal to \(2^x\), where \(x\) is the number of address lines, regardless of the size of the data bus. For example, a CPU with 16 address lines can provide a total of 65,536 \(\left(2^{16}\right)\) or 64K of addressable memory. Each location can have a maximum of 1 byte of data. This is because all general-purpose microprocessor CPUs are what is called byte addressable. As another example, the IBM PC AT uses a CPU with 24 address lines and 16 data lines. Thus, the total accessible memory is 16 megabytes \(\left(2^{24}\right) = 16\) megabytes). In this example there would be \(2^{24}\) locations, and because each location is one byte, there would be 16 megabytes of memory. The address bus is a unidirectional bus, which means that the CPU uses the address bus only to send out addresses. To summarize: The total number of memory locations addressable by a given CPU is always equal to \(2^x\) where \(x\) is the number of address bits, regardless of the size of the data bus.

![Figure 0-10. Internal Organization of a Computer](image)

**CPU and its relation to RAM and ROM**

For the CPU to process information, the data must be stored in RAM or ROM. The function of ROM in computers is to provide information that is fixed and permanent. This is information such as tables for character patterns to be displayed on the video monitor, or programs that are essential to the working of the computer, such as programs for testing and finding the total amount of RAM installed on the system, or for displaying information on the video monitor. In contrast, RAM stores temporary information that can change with time, such as various versions of the operating system and application packages such as word processing or tax calculation packages. These programs are loaded from the hard drive into RAM to be processed by the CPU. The CPU cannot get the information
from the disk directly because the disk is too slow. In other words, the CPU first seeks the information to be processed from RAM (or ROM). Only if the data is not there does the CPU seek it from a mass storage device such as a disk, and then it transfers the information to RAM. For this reason, RAM and ROM are sometimes referred to as primary memory and disks are called secondary memory. Figure 0-11 shows a block diagram of the internal organization of the PC.

![Figure 0-11. Internal Block Diagram of a CPU](image)

### Inside CPUs

A program stored in memory provides instructions to the CPU to perform an action. The action can simply be adding data such as payroll data or controlling a machine such as a robot. The function of the CPU is to fetch these instructions from memory and execute them. To perform the actions of fetch and execute, all CPUs are equipped with resources such as the following:

1. Foremost among the resources at the disposal of the CPU are a number of registers. The CPU uses registers to store information temporarily. The information could be two values to be processed, or the address of the value needed to be fetched from memory. Registers inside the CPU can be 8-bit, 16-bit, 32-bit, or even 64-bit registers, depending on the CPU. In general, the more and bigger the registers, the better the CPU. The disadvantage of more and bigger registers is the increased cost of such a CPU.
2. The CPU also has what is called the ALU (arithmetic/logic unit). The ALU section of the CPU is responsible for performing arithmetic functions such as add, subtract, multiply, and divide, and logic functions such as AND, OR, and NOT.
3. Every CPU has what is called a *program counter*. The function of the program counter is to point to the address of the next instruction to be executed. As each instruction is executed, the program counter is incremented to point to the address of the next instruction to be executed. The contents of the program counter are placed on the address bus to find and fetch the desired instruction. In the IBM PC, the program counter is a register called IP, or the instruction pointer.

4. The function of the *instruction decoder* is to interpret the instruction fetched into the CPU. One can think of the instruction decoder as a kind of dictionary, storing the meaning of each instruction and what steps the CPU should take upon receiving a given instruction. Just as a dictionary requires more pages the more words it defines, a CPU capable of understanding more instructions requires more transistors to design.

**Internal working of computers**

To demonstrate some of the concepts discussed above, a step-by-step analysis of the process a CPU would go through to add three numbers is given next. Assume that an imaginary CPU has registers called A, B, C, and D. It has an 8-bit data bus and a 16-bit address bus. Therefore, the CPU can access memory from addresses 0000 to FFFFH (for a total of 10000H locations). The action to be performed by the CPU is to put hexadecimal value 21 into register A, and then add to register A values 42H and 12H. Assume that the code for the CPU to move a value to register A is 1011 0000 (B0H) and the code for adding a value to register A is 0000 0100 (04H). The necessary steps and code to perform them are as follows.

<table>
<thead>
<tr>
<th>Action</th>
<th>Code</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>Move value 21H into register A</td>
<td>B0H</td>
<td>21H</td>
</tr>
<tr>
<td>Add value 42H to register A</td>
<td>04H</td>
<td>42H</td>
</tr>
<tr>
<td>Add value 12H to register A</td>
<td>04H</td>
<td>12H</td>
</tr>
</tbody>
</table>

If the program to perform the actions listed above is stored in memory locations starting at 1400H, the following would represent the contents for each memory address location:

<table>
<thead>
<tr>
<th>Memory address</th>
<th>Contents of memory address</th>
</tr>
</thead>
<tbody>
<tr>
<td>1400</td>
<td>(B0) code for moving a value to register A</td>
</tr>
<tr>
<td>1401</td>
<td>(21) value to be moved</td>
</tr>
<tr>
<td>1402</td>
<td>(04) code for adding a value to register A</td>
</tr>
<tr>
<td>1403</td>
<td>(42) value to be added</td>
</tr>
<tr>
<td>1404</td>
<td>(04) code for adding a value to register A</td>
</tr>
<tr>
<td>1405</td>
<td>(12) value to be added</td>
</tr>
<tr>
<td>1406</td>
<td>(F4) code for halt</td>
</tr>
</tbody>
</table>

The actions performed by the CPU to run the program above would be as follows:

1. The CPU’s program counter can have a value between 0000 and FFFFH. The program counter must be set to the value 1400H, indicating the address of the
first instruction code to be executed. After the program counter has been loaded with the address of the first instruction, the CPU is ready to execute.

2. The CPU puts 1400H on the address bus and sends it out. The memory circuitry finds the location while the CPU activates the READ signal, indicating to memory that it wants the byte at location 1400H. This causes the contents of memory location 1400H, which is B0, to be put on the data bus and brought into the CPU.

3. The CPU decodes the instruction B0 with the help of its instruction decoder dictionary. When it finds the definition for that instruction it knows it must bring into register A of the CPU the byte in the next memory location. Therefore, it commands its controller circuitry to do exactly that. When it brings in value 21H from memory location 1401, it makes sure that the doors of all registers are closed except register A. Therefore, when value 21H comes into the CPU it will go directly into register A. After completing one instruction, the program counter points to the address of the next instruction to be executed, which in this case is 1402H. Address 1402 is sent out on the address bus to fetch the next instruction.

4. From memory location 1402H the CPU fetches code 04H. After decoding, the CPU knows that it must add the byte sitting at the next address (1403) to the contents of register A. After the CPU brings the value (in this case, 42H) into register A, it provides the contents of register A along with this value to the ALU to perform the addition. It then takes the result of the addition from the ALU’s output and puts it in register A. Meanwhile the program counter becomes 1404, the address of the next instruction.

5. Address 1404H is put on the address bus and the code is fetched into the CPU, decoded, and executed. This code again is adding a value to register A. The program counter is updated to 1406H.

6. Finally, the contents of address 1406 are fetched in and executed. This HALT instruction tells the CPU to stop incrementing the program counter and asking for the next instruction. Without the HALT, the CPU would continue updating the program counter and fetching instructions.

Now suppose that address 1403H contained value 04 instead of 42H. How would the CPU distinguish between data 04 to be added and code 04? Remember that code 04 for this CPU means “move the next value into register A”. Therefore, the CPU will not try to decode the next value. It simply moves the contents of the following memory location into register A, regardless of its value.

**Review Questions**

1. How many bytes is 24 kilobytes?
2. What does “RAM” stand for? How is it used in computer systems?
3. What does “ROM” stand for? How is it used in computer systems?
4. Why is RAM called volatile memory?
5. List the three major components of a computer system.
7. List the three types of buses found in computer systems and state briefly the purpose of each type of bus.
8. State which of the following is unidirectional and which is bidirectional:
   (a) data bus  (b) address bus
9. If an address bus for a given computer has 16 lines, what is the maximum
    amount of memory it can access?
10. What does “ALU” stand for? What is its purpose?
11. How are registers used in computer systems?
12. What is the purpose of the program counter?
13. What is the purpose of the instruction decoder?

SUMMARY

The binary number system represents all numbers with a combination of
the two binary digits, 0 and 1. The use of binary systems is necessary in digital
computers because only two states can be represented: on or off. Any binary num-
ber can be coded directly into its hexadecimal equivalent for the convenience of
humans. Converting from binary/hex to decimal, and vice versa, is a straightforward process that becomes easy with practice. The ASCII code is a binary code
used to represent alphanumeric data internally in the computer. It is frequently
used in peripheral devices for input and/or output.

The logic gates AND, OR, and inverter are the basic building blocks of
simple circuits. NAND, NOR, and XOR gates are also used to implement circuit
design. Diagrams of half-adders and full-adders were given as examples of the use
of logic gates for circuit design. Decoders are used to detect certain addresses.
Flip-flops are used to latch in data until other circuits are ready for it.

The major components of any computer system are the CPU, memory, and
I/O devices. “Memory” refers to temporary or permanent storage of data. In most
systems, memory can be accessed as bytes or words. The terms kilobyte, megabyte, gigabyte, and terabyte are used to refer to large numbers of bytes. There
are two main types of memory in computer systems: RAM and ROM. RAM (ran-
dom access memory) is used for temporary storage of programs and data. ROM
(read-only memory) is used for permanent storage of programs and data that the
computer system must have in order to function. All components of the computer
system are under the control of the CPU. Peripheral devices such as I/O (input/output) devices allow the CPU to communicate with humans or other computer sys-
tems. There are three types of buses in computers: address, control, and data.
Control buses are used by the CPU to direct other devices. The address bus is used
by the CPU to locate a device or a memory location. Data buses are used to send
information back and forth between the CPU and other devices.

Finally, this chapter gave an overview of digital logic.
PROBLEMS

SECTION 0.1: NUMBERING AND CODING SYSTEMS

1. Convert the following decimal numbers to binary:
   (a) 12  (b) 123  (c) 63  (d) 128  (e) 1000
2. Convert the following binary numbers to decimal:
   (a) 100100  (b) 1000001  (c) 11101  (d) 1010  (e) 00100010
3. Convert the values in Problem 2 to hexadecimal.
4. Convert the following hex numbers to binary and decimal:
   (a) 2B9H  (b) F44H  (c) 912H  (d) 2BH  (e) FFFFH
5. Convert the values in Problem 1 to hex.
6. Show the ASCII codes for numbers 0, 1, 2, 3, ..., 9 in both hex and binary.

SECTION 0.2: DIGITAL PRIMER

11. Draw a 3-input OR gate using a 2-input OR gate.
12. Show the truth table for a 3-input OR gate.
13. Draw a 3-input AND gate using a 2-input AND gate.
14. Show the truth table for a 3-input AND gate.
15. Design a 3-input XOR gate with a 2-input XOR gate. Show the truth table for a 3-input XOR.
16. List the truth table for a 3-input NAND.
17. List the truth table for a 3-input NOR.
18. Show the decoder for binary 1100.
19. Show the decoder for binary 11011.
20. List the truth table for a D-FF.

SECTION 0.3: INSIDE THE COMPUTER

21. Answer the following:
   (a) How many nibbles are 16 bits?
   (b) How many bytes are 32 bits?
   (c) If a word is defined as 16 bits, how many words is a 64-bit data item?
   (d) What is the exact value (in decimal) of 1 meg?
(e) How many K is 1 meg?
(f) What is the exact value (in decimal) of 1 gigabyte?
(g) How many K is 1 gigabyte?
(h) How many meg is 1 gigabyte?
(i) If a given computer has a total of 8 megabytes of memory, how many bytes (in decimal) is this? How many kilobytes is this?

22. A given mass storage device such as a hard disk can store 2 gigabytes of information. Assuming that each page of text has 25 rows and each row has 80 columns of ASCII characters (each character = 1 byte), approximately how many pages of information can this disk store?

23. In a given byte-addressable computer, memory locations 10000H to 9FFFFH are available for user programs. The first location is 10000H and the last location is 9FFFFH. Calculate the following:
   (a) The total number of bytes available (in decimal)
   (b) The total number of kilobytes (in decimal)

24. A given computer has a 32-bit data bus. What is the largest number that can be carried into the CPU at a time?

25. Below are listed several computers with their data bus widths. For each computer, list the maximum value that can be brought into the CPU at a time (in both hex and decimal).
   (a) Apple 2 with an 8-bit data bus
   (b) IBM PS/2 with a 16-bit data bus
   (c) IBM PS/2 model 80 with a 32-bit data bus
   (d) Cray supercomputer with a 64-bit data bus

26. Find the total amount of memory, in the units requested, for each of the following CPUs, given the size of the address buses:
   (a) 16-bit address bus (in K)
   (b) 24-bit address bus (in megs)
   (c) 32-bit address bus (in megabytes and gigabytes)
   (d) 48-bit address bus (in megabytes, gigabytes, and terabytes)

27. Regarding the data bus and address bus, which is unidirectional and which is bidirectional?

28. Which register of the CPU holds the address of the instruction to be fetched?

29. Which section of the CPU is responsible for performing addition?

30. List the three bus types present in every CPU.

CHAPTER 0: INTRODUCTION TO COMPUTING
ANSWERS TO REVIEW QUESTIONS

SECTION 0.1: NUMBERING AND CODING SYSTEMS

1. Computers use the binary system because each bit can have one of two voltage levels: on and off.
2. $34_{10} = 100010_2 = 22_{16}$
3. $110101_2 = 35_{16} = 53_{10}$
4. 1110001
5. 010100
6. 461
7. 275
8. 38 30 78 36 20 43 50 55 73

SECTION 0.2: DIGITAL PRIMER

1. AND
2. OR
3. XOR
4. Buffer
5. Storing data
6. Decoder

SECTION 0.3: INSIDE THE COMPUTER

1. 24,576
2. Random access memory; it is used for temporary storage of programs that the CPU is running, such as the operating system, word processing programs, etc.
3. Read-only memory; it is used for permanent programs such as those that control the keyboard, etc.
4. The contents of RAM are lost when the computer is powered off.
5. The CPU, memory, and I/O devices
6. Central processing unit; it can be considered the “brain” of the computer; it executes the programs and controls all other devices in the computer.
7. The address bus carries the location (address) needed by the CPU; the data bus carries information in and out of the CPU; the control bus is used by the CPU to send signals controlling I/O devices.
8. (a) bidirectional (b) unidirectional
9. 64K, or 65,536 bytes
10. Arithmetic/logic unit; it performs all arithmetic and logic operations.
11. They are used for temporary storage of information.
12. It holds the address of the next instruction to be executed.
13. It tells the CPU what steps to perform for each instruction.
CHAPTER 1

THE PIC MICROCONTROLLERS: HISTORY AND FEATURES

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Compare and contrast microprocessors and microcontrollers
- Describe the advantages of microcontrollers for some applications
- Explain the concept of embedded systems
- Discuss criteria for considering a microcontroller
- Explain the variations of speed, packaging, memory, and cost per unit and how these affect choosing a microcontroller
- Compare and contrast the various members of the PIC family
- Compare the PIC with microcontrollers offered by other manufacturers
This chapter begins with a discussion of the role and importance of microcontrollers in everyday life. In Section 1.1 we also discuss criteria to consider in choosing a microcontroller, as well as the use of microcontrollers in the embedded market. Section 1.2 covers various members of the PIC18 family and their features. In addition, we provide a brief discussion of alternatives to the PIC chip such as the 8051, AVR, and 68HC11 microcontrollers.

**SECTION 1.1: MICROCONTROLLERS AND EMBEDDED PROCESSORS**

In this section we discuss the need for microcontrollers and contrast them with general-purpose microprocessors such as the Pentium and other x86 microprocessors. We also look at the role of microcontrollers in the embedded market. In addition, we provide some criteria on how to choose a microcontroller.

**Microcontroller versus general-purpose microprocessor**

What is the difference between a microprocessor and microcontroller? By microprocessor is meant the general-purpose microprocessors such as Intel’s x86 family (8086, 80286, 80386, 80486, and the Pentium) or Motorola’s PowerPC family. These microprocessors contain no RAM, no ROM, and no I/O ports on the chip itself. For this reason, they are commonly referred to as *general-purpose microprocessors*. See Figure 1-1.

![Figure 1-1. Microprocessor System Contrasted With Microcontroller System](image)

A system designer using a general-purpose microprocessor such as the Pentium or the PowerPC must add RAM, ROM, I/O ports, and timers externally to make them functional. Although the addition of external RAM, ROM, and I/O ports makes these systems bulkier and much more expensive, they have the advantage of versatility, enabling the designer to decide on the amount of RAM, ROM, and I/O ports needed to fit the task at hand. This is not the case with microcontrollers. A microcontroller has a CPU (a microprocessor) in addition to a fixed amount of RAM, ROM, I/O ports, and a timer all on a single chip. In other words, the processor, RAM, ROM, I/O ports, and timer are all embedded together on one chip; therefore, the designer cannot add any external memory, I/O, or timer to it. The fixed amount of on-chip ROM, RAM, and number of I/O ports in microcontrollers makes them ideal for many applications in which cost and space are critical. In many applications, for example a TV remote control, there is no need for
the computing power of a 486 or even an 8086 microprocessor. In many applications, the space used, the power consumed, and the price per unit are much more critical considerations than the computing power. These applications most often require some I/O operations to read signals and turn on and off certain bits. For this reason some call these processors IBP, “itty-bitty processors.” (See “Good Things in Small Packages Are Generating Big Product Opportunities” by Rick Grehan, BYTE magazine, September 1994 (http://www.byte.com) for an excellent discussion of microcontrollers.)

It is interesting to note that some microcontroller manufacturers have gone as far as integrating an ADC (analog-to-digital converter) and other peripherals into the microcontroller.

Microcontrollers for embedded systems

In the literature discussing microprocessors, we often see the term embedded system. Microprocessors and microcontrollers are widely used in embedded system products. An embedded product is controlled by its own internal microprocessor (or microcontroller) as opposed to an external controller. Typically, in an embedded system, the microcontroller’s ROM is burned with a purpose for specific functions needed for the system. A printer is an example of an embedded system because the processor inside it performs one task only; namely, getting the data and printing it. Contrast this with a Pentium-based PC (or any x86 IBM-compatible PC), which can be used for any number of applications such as word processor, print-server, bank teller terminal, video game player, network server, or Internet terminal. A PC can also load and run software for a variety of applications. Of course, the reason a PC can perform myriad tasks is that it has RAM memory and an operating system that loads the application software into RAM and lets the CPU run it. In an embedded system, typically only one application software is burned into ROM. An x86 PC contains or is connected to various embedded products such as the keyboard, printer, modem, disk controller, sound card, CD-ROM driver, mouse, and so on. Each one of these peripherals has a microcontroller inside it that performs only one task. For example, inside every mouse a microcontroller performs the task of finding the mouse’s position and sending it to the PC. Table 1-1 lists some embedded products.

x86 PC embedded applications

Although microcontrollers are the preferred choice for many embedded systems, sometimes a microcontroller is inadequate for the task. For this reason, in recent years many manufacturers of general-purpose microprocessors such as Intel, Freescale Semiconductor (formerly Motorola), and AMD (Advanced Micro

<table>
<thead>
<tr>
<th>Home</th>
<th>Appliances</th>
<th>Intercom</th>
<th>Telephones</th>
<th>Security systems</th>
<th>Garage door openers</th>
<th>Answering machines</th>
<th>Fax machines</th>
<th>Home computers</th>
<th>TVs</th>
<th>Cable TV tuner</th>
<th>VCR</th>
<th>Camcorder</th>
<th>Remote controls</th>
<th>Video games</th>
<th>Cellular phones</th>
<th>Musical instruments</th>
<th>Sewing machines</th>
<th>Lighting control</th>
<th>Paging</th>
<th>Camera</th>
<th>Pinball machines</th>
<th>Toys</th>
<th>Exercise equipment</th>
</tr>
</thead>
<tbody>
<tr>
<td>Office</td>
<td>Telephones</td>
<td>Computers</td>
<td>Security systems</td>
<td>Fax machine</td>
<td>Microwave</td>
<td>Copier</td>
<td>Laser printer</td>
<td>Color printer</td>
<td>Paging</td>
<td>Trip computer</td>
<td>Engine control</td>
<td>Air bag</td>
<td>ABS</td>
<td>Instrumentation</td>
<td>Security system</td>
<td>Transmission control</td>
<td>Entertainment</td>
<td>Climate control</td>
<td>Cellular phone</td>
<td>Keyless entry</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 1-1: Some Embedded Products Using Microcontrollers
Devices, Inc.) have targeted their microprocessor for the high end of the embedded market. Intel and AMD push their x86 processors for both the embedded and desktop PC markets. In the early 1990s, Apple computer began using the PowerPC microprocessors (604, 603, 620, etc.) in place of the 680x0 for the Macintosh. The PowerPC microprocessor is a joint venture between IBM and Motorola, and is targeted for the high end of the embedded market as well as the PC market. It must be noted that when a company targets a general-purpose microprocessor for the embedded market it optimizes the processor used for embedded systems. For this reason these processors are often called high-end embedded processors. Another chip widely used in the high end of the embedded system design is the ARM microprocessor. Very often the terms embedded processor and microcontroller are used interchangeably.

One of the most critical needs of an embedded system is to decrease power consumption and space. This can be achieved by integrating more functions into the CPU chip. All the embedded processors based on the x86 and PowerPC 6xx have low power consumption in addition to some forms of I/O, COM port, and ROM, all on a single chip. In high-performance embedded processors, the trend is to integrate more and more functions on the CPU chip and let the designer decide which features to use. This trend is invading PC system design as well. Normally, in designing the PC motherboard we need a CPU plus a chipset containing I/O, a cache controller, a flash ROM containing BIOS, and finally a secondary cache memory. New designs are emerging in industry. For example, many companies have a chip that contains the entire CPU and all the supporting logic and memory, except for DRAM. In other words, we have the entire computer on a single chip.

Currently, because of Linux, MS-DOS, and Windows standardization, many embedded systems use x86 PCs. In many cases, using x86 PCs for the high-end embedded applications not only saves money but also shortens development time because a vast library of software already exists for the Linux, DOS, and Windows platforms. The fact that Windows and Linux are widely used and well-understood platforms means that developing a Windows-based or Linux-based embedded product reduces the cost and shortens the development time considerably.

Choosing a microcontroller

There are five major 8-bit microcontrollers. They are: Freescale Semiconductor’s (formerly Motorola) 68HC08/68HC11, Intel’s 8051, Atmel’s AVR, Zilog’s Z8, and PIC from Microchip Technology. Each of the above microcontrollers has a unique instruction set and register set; therefore, they are not compatible with each other. Programs written for one will not run on the others. There are also 16-bit and 32-bit microcontrollers made by various chip makers. With all these different microcontrollers, what criteria do designers consider in choosing one? Three criteria in choosing microcontrollers are as follows: (1) meeting the computing needs of the task at hand efficiently and cost effectively; (2) availability of software and hardware development tools such as compilers, assemblers, debuggers, and emulators; and (3) wide availability and reliable sources of the microcontroller. Next, we elaborate on each of the above criteria.
Criteria for choosing a microcontroller

1. The first and foremost criterion in choosing a microcontroller is that it must meet the task at hand efficiently and cost effectively. In analyzing the needs of a microcontroller-based project, we must first see whether an 8-bit, 16-bit, or 32-bit microcontroller can best handle the computing needs of the task most effectively. Among other considerations in this category are:
   (a) Speed. What is the highest speed that the microcontroller supports?
   (b) Packaging. Does it come in a 40-pin DIP (dual inline package) or a QFP (quad flat package), or some other packaging format? This is important in terms of space, assembling, and prototyping the end product.
   (c) Power consumption. This is especially critical for battery-powered products.
   (d) The amount of RAM and ROM on the chip.
   (e) The number of I/O pins and the timer on the chip.
   (f) Ease of upgrade to higher-performance or lower-power-consumption versions.
   (g) Cost per unit. This is important in terms of the final cost of the product in which a microcontroller is used. For example, some microcontrollers cost 50 cents per unit when purchased 100,000 units at a time.

2. The second criterion in choosing a microcontroller is how easy it is to develop products around it. Key considerations include the availability of an assembler, debugger, a code-efficient C language compiler, emulator, technical support, and both in-house and outside expertise. In many cases, third-party vendor (i.e., a supplier other than the chip manufacturer) support for the chip is as good as, if not better than, support from the chip manufacturer.

3. The third criterion in choosing a microcontroller is its ready availability in needed quantities both now and in the future. For some designers this is even more important than the first two criteria. Currently, of the leading 8-bit microcontrollers, the 8051 family has the largest number of diversified (multiple source) suppliers. (Supplier means a producer besides the originator of the microcontroller.) In the case of the 8051, which was originated by Intel, several companies also currently produce (or have produced in the past) the 8051.

   Note that Freescale Semiconductor (Motorola), Atmel, Zilog, and Microchip Technology have all dedicated massive resources to ensure wide and timely availability of their products because their products are stable, mature, and single sourced. In recent years, companies have begun to sell Field-Programmable Gate Array (FPGA) and Application-Specific Integrated Circuit (ASIC) libraries for the different microcontrollers.

Mechatronics and microcontrollers

The microcontroller is playing a major role in an emerging field called mechatronics. Here is an excellent summary of what the field of mechatronics is all about, taken from the web site of Newcastle University (http://mechatronics2004.newcastle.edu.au/mech2004), which holds a major conference every year on this subject:

"Many technical processes and products in the area of mechanical and
electrical engineering show an increasing integration of mechanics with electronics and information processing. This integration is between the components (hardware) and the information-driven functions (software), resulting in integrated systems called mechatronic systems.

The development of mechatronic systems involves finding an optimal balance between the basic mechanical structure, sensor and actuator implementation, automatic digital information processing and overall control, and this synergy results in innovative solutions. The practice of mechatronics requires multi-disciplinary expertise across a range of disciplines, such as: mechanical engineering, electronics, information technology, and decision making theories.”

**Review Questions**

1. True or false. Microcontrollers are normally less expensive than microprocessors.
2. When comparing a system board based on a microcontroller and a general-purpose microprocessor, which one is cheaper?
3. A microcontroller normally has which of the following devices on-chip?
   - (a) RAM
   - (b) ROM
   - (c) I/O
   - (d) all of the above
4. A general-purpose microprocessor normally needs which of the following devices to be attached to it?
   - (a) RAM
   - (b) ROM
   - (c) I/O
   - (d) all of the above
5. An embedded system is also called a dedicated system. Why?
6. What does the term *embedded system* mean?
7. Why does having multiple sources of a given product matter?

**SECTION 1.2: OVERVIEW OF THE PIC18 FAMILY**

In this section, we first look at the PIC family of microcontrollers and then examine the PIC18 family in more detail.

**A brief history of the PIC microcontroller**

In 1989, Microchip Technology Corporation introduced an 8-bit microcontroller called the PIC, which stands for Peripheral Interface Controller. This microcontroller had small amounts of data RAM, a few hundred bytes of on-chip ROM for the program, one timer, and a few pins for I/O ports, all on a single chip with only 8 pins. (See Figure 1-2.) It is amazing that a company that began with such a humble product became one of the leading suppliers of 8-bit microcontrollers in less than a decade. At the time of this writing, Microchip is the number-one supplier of 8-bit microcontrollers in the world. Since the introduction of the PIC16xxx, they have introduced an array of 8-bit microcontrollers too numerous to list here. They include the PIC families of 10xxx, 12xxx, 14xxx, 16xxx, 17xxx, and 18xxx. They are all 8-bit processors, meaning that the CPU can work on only 8 bits of data at a time. Data larger than 8 bits has to be broken into 8-bit pieces to be processed by the CPU. One of the problems with the PIC family is that they are not all 100% upwardly compatible in terms of software when going from one family to another family. For example, while the 12xxx/16xxx have 12-bit and 14-bit wide instructions, the PIC18xxx instruction is 16 bits wide with many new instruc-
tions. To run programs written for the PIC12xxx on a PIC18, we must recompile the program and possibly change some register locations before loading it into the PIC18. At the time of this writing, the PIC18xxx family has the highest performance of all the families of 8-bit PIC microcontrollers. The fact that PIC18xxx is available in 18- to 80-pin packages makes it an ideal choice for new designs because it allows an easy migration to more powerful versions of the chip without losing software compatibility. At this time, no 8-pin version of the PIC18xxx exists, and that is the main reason to choose other family members of the 10xxx–16xxx if your design calls for a small package. Because this book is about the PIC18 family, we describe some of the main features of this family and refer the reader to the Microchip web site for other families of PIC10xxx–16xxx. For those who have mastered the PIC18 family, understanding the other families is very easy and straightforward. The following is a brief description of the PIC18 series.

**PIC18 features**

The PIC18 has a RISC architecture that comes with some standard features such as on-chip program (code) ROM, data RAM, data EEPROM, timers, ADC, and USART and I/O ports. See Figure 1-2. Although the size of the program ROM, data RAM, data EEPROM, and I/O ports varies among the family members, they all have peripherals such as timers, ADC, and USART. See Figures 1-3 and 1-4. Due to the importance of these peripherals, we have dedicated an entire chapter to each one of them. The details of the RAM/ROM memory and I/O features of the PIC18 are given in the next few chapters.

Figure 1-2. Simplified View of a PIC Microcontroller
Note 1: Many of the general purpose I/O pins are multiplexed with one or more peripheral module functions. The multiplexing combinations are device dependent.

Figure 1-3. PIC18 Block Diagram
Note 1: The high order bits of the Direct Address for the RAM are from the STATUS register.
2: Not all devices have this feature, please refer to device data sheet.
3: Many of the general purpose I/O pins are multiplexed with one or more peripheral module functions.
The multiplexing combinations are device dependent.

Figure 1-4. PIC16 Block Diagram
**PIC microcontroller program ROM**

In microcontrollers, the ROM is used to store programs and for that reason it is called program or code ROM. Although the PIC18 has 2M (megabytes) of program (code) ROM space, not all family members come with that much ROM installed. The program ROM size can vary from 4K to 128K at the time of this writing, depending on the family member. The PIC18 program ROM is available in different memory types, such as flash, OTP, and masked, all of which have different part numbers. A discussion of the various types of ROM is given in Chapter 14, if you need to refresh your memory on these important memory technologies. Note that although different flavors of the PIC18 exist in terms of speed and amount of on-chip RAM/ROM, they are all compatible with each other as far as the instructions are concerned. This means that if you write your program for one, it will run on any of them regardless of the chip number. Next, we discuss briefly the program ROM type for the PIC18 family.

**PIC microcontroller with UV-EPROM**

Some of the PIC microcontrollers use UV-EPROM, for on-chip program ROM. To use these kinds of chips for development requires access to a PROM burner, as well as a UV-EPROM eraser to erase the contents of ROM. The window on the UV-EPROM chip allows the UV light to erase the ROM. The problem with the UV-EPROM is that it takes around 20 minutes to erase the chip before it can be programmed again. This has led Microchip to introduce a flash version of the PIC family. At this time flash is replacing the UV-EPROM altogether. Table 1-2 shows some members of the PIC18 family.

**PIC18Fxxx with flash**

Many PIC18 chips have on-chip program ROM in the form of flash memory. The flash version uses the letter F in the part number to indicate that the on-chip ROM is flash. PIC18F458 is an example of PIC18 with flash ROM. The flash version is ideal for fast development because flash memory can be erased in seconds compared to the 20 minutes or more needed for the UV-EPROM version. For this reason, the PIC18F has been used in place of the UV-EPROM to eliminate the waiting time needed to erase the chip, thereby speeding up the development time. To use the PIC18F to develop a microcontroller-based system requires a ROM burner that supports flash memory; however, a ROM eraser is not needed, because flash is an EEPROM (electrically erasable PROM). Notice that in flash memory, you must erase the entire contents of ROM in order to program it again. This erasing of flash is done by the ROM programmer itself, and so a separate eraser is not needed. We can also program the PIC18F via the PICKit 2 from MicroChip using the USB port of an IBM PC.

**OTP version of the PIC**

OTP (one-time-programmable) versions of the PIC are also available from Microchip. PIC16C432 chip uses OTP for program ROM. Contrast the PIC16C432 and PIC18F252. The letter C indicates the OTP ROM, while the letter F is for the flash. The flash version is typically used for product development. When a product is designed and absolutely finalized, the OTP version of the PIC
is used for mass production because it is cheaper than flash in terms of price per unit. The problem with the OTP is that you cannot reprogram it if you want to modify your program.

**Masked version of PIC**

Microchip Corporation provides a service in which you can send in your program and they will burn the program into the PIC chip during the fabrication process of the chip. This chip is commonly referred to as masked PIC, which is one of the stages of IC fabrication. Masked PIC is the cheapest of all types, if the unit numbers are high enough. This is because there is a minimum order for the masked version of the PIC microcontrollers.

**PIC microcontroller data RAM and EEPROM**

While ROM is used to store program (code), the RAM space is for data storage. The PIC18 has a maximum of 4096 bytes (4K) of data RAM space. Not all of the family members come with that much RAM. The data RAM size for the PIC18 varies from 256 bytes to 4096 bytes. As we will see in the next chapter, the data RAM space has two components: General-Purpose RAM (GPR) and Special Function Registers (SFRs). Because the SFRs are fixed and every microcontroller must have them, it is the GPR’s size that varies from chip to chip. For this reason, the Microchip web site gives only the GPR size. The RAM GPR space is used for read/write scratch pad and data manipulation and is divided into banks of 256 bytes each, as we will see in Chapter 6. The GPR size given for the PIC18 is always a multiple of 256 bytes. In some of the PIC18 family members, we also have a small amount of EEPROM to store critical data that does not need to be changed very often. While every PIC18 must have some data RAM for scratch pad, the EEPROM is optional, so not all versions of the PIC18 come with EEPROM. EEPROM is used mainly for storage of critical data, as we will see in Chapter 14.

<table>
<thead>
<tr>
<th>Part Num</th>
<th>Code ROM</th>
<th>Data RAM</th>
<th>Data EEPROM</th>
<th>I/O pins</th>
<th>ADC</th>
<th>Timers</th>
<th>Pin numbers &amp; Package</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F1220</td>
<td>4K (flash)</td>
<td>256</td>
<td>256</td>
<td>16</td>
<td>10-bit</td>
<td>4</td>
<td>18 DIP</td>
</tr>
<tr>
<td>PIC18F2420</td>
<td>16K (flash)</td>
<td>768</td>
<td>0</td>
<td>25</td>
<td>10-bit</td>
<td>4</td>
<td>28 DIP</td>
</tr>
<tr>
<td>PIC18F2220</td>
<td>4K (flash)</td>
<td>512</td>
<td>256</td>
<td>25</td>
<td>10-bit</td>
<td>4</td>
<td>28 DIP</td>
</tr>
<tr>
<td>PIC18F452</td>
<td>32K (flash)</td>
<td>1536</td>
<td>256</td>
<td>34</td>
<td>10-bit</td>
<td>4</td>
<td>40 DIP</td>
</tr>
<tr>
<td>PIC18F4520</td>
<td>32K (flash)</td>
<td>1536</td>
<td>256</td>
<td>36</td>
<td>10-bit</td>
<td>4</td>
<td>40 DIP</td>
</tr>
<tr>
<td>PIC18F458</td>
<td>32K (flash)</td>
<td>1536</td>
<td>256</td>
<td>34</td>
<td>10-bit</td>
<td>4</td>
<td>40 DIP</td>
</tr>
<tr>
<td>PIC18F4580</td>
<td>32K (flash)</td>
<td>1536</td>
<td>256</td>
<td>36</td>
<td>10-bit</td>
<td>4</td>
<td>40 DIP</td>
</tr>
<tr>
<td>PIC18F8722</td>
<td>128K (flash)</td>
<td>3936</td>
<td>1024</td>
<td>70</td>
<td>10-bit</td>
<td>5</td>
<td>80 TQFP</td>
</tr>
</tbody>
</table>

Notes:
1. All ROM, RAM, and EEPROM memories are in bytes.
2. Data RAM (General-Purpose RAM) is the amount of RAM available for data manipulation (scratch pad) in addition to the Special Function Registers (SFRs) space.
3. All the above chips have USART for serial data transfer.
**PIC microcontroller I/O pins**

The PIC18 can have from 16 to 72 pins dedicated for I/O. The number of I/O pins depends on the number of pins in the package itself. The number of pins for the PIC18 package goes from 18 to 80 at this time. In the case of the 18-pin PIC18F1220, we have 16 pins for I/O, while in the case of the 80-pin PIC18F8722, we can use up to 72 pins for I/O. We will study I/O pins and programming in Chapter 4.

**PIC microcontroller peripherals**

All the members of the PIC18 family come with ADC (analog-to-digital converter), timers, and USART (Universal Synchronous Asynchronous Receiver Transmitter) as standard peripherals. As we will see in Chapter 13, the ADC is 10-bit and the number of ADC channels in each PIC chip varies from 5 to 16, depending on the number of pins in the package. The PIC18 can have up to 4 timers besides the watchdog timer. We will examine timers in Chapter 9. The USART peripheral allows us to connect the PIC18-based system to serial ports such as the COM port of the IBM PC, as we will see in Chapter 10. Many of the PIC18 family members come with the I²C and CAN bus as well.

**PIC trainer**

In Chapter 8, we discuss the design of the PIC18F458 trainer extensively. This trainer is programmed using the PICkit 2. The MDEPIC trainer is also compatible with other 40-pin devices from Microchip.

**Other Microcontrollers**

There are many other popular 8-bit microcontrollers besides the PIC chip. Among them are the 8051, 68HC11, AVR, and Z8. Besides Intel, a number of other companies make the 8051 family, as seen in Table 1-4. The AVR is made by Atmel Corp. Freescale (Motorola) makes the 68HC11 and many of its variations. Zilog produces the Z8 microcontroller. To contrast the PIC18 family with the 8051/52 chip, examine Table 1-3. For a comprehensive treatment of the 8051 microcontroller, see “The 8051 Microcontroller and Embedded Systems” by Mazidi, et. al.

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**Table 1-3: Comparison of 8051 and PIC18 Family (40-pin package)**

<table>
<thead>
<tr>
<th>Feature</th>
<th>8051/52</th>
<th>PIC18xxx</th>
</tr>
</thead>
<tbody>
<tr>
<td>Program ROM (maximum space)</td>
<td>64K</td>
<td>2M</td>
</tr>
<tr>
<td>Data RAM (maximum space)</td>
<td>256 bytes</td>
<td>4K</td>
</tr>
<tr>
<td>Timers</td>
<td>3</td>
<td>4</td>
</tr>
<tr>
<td>I/O pins</td>
<td>32</td>
<td>33</td>
</tr>
<tr>
<td>Serial port</td>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>
Table 1-4: Some of the Companies that Produce Widely Used 8-bit Microcontrollers

<table>
<thead>
<tr>
<th>Company</th>
<th>Web Site</th>
<th>Architecture</th>
</tr>
</thead>
<tbody>
<tr>
<td>Microchip</td>
<td><a href="http://www.microchip.com">http://www.microchip.com</a></td>
<td>PIC16xxx/18xxx</td>
</tr>
<tr>
<td>Intel</td>
<td><a href="http://www.intel.com/design/mcs51">http://www.intel.com/design/mcs51</a></td>
<td>8051</td>
</tr>
<tr>
<td>Atmel</td>
<td><a href="http://www.atmel.com">http://www.atmel.com</a></td>
<td>AVR and 8051</td>
</tr>
<tr>
<td>Philips/Signetics</td>
<td><a href="http://www.semiconductors.philips.com">http://www.semiconductors.philips.com</a></td>
<td>8051</td>
</tr>
<tr>
<td>Zilog</td>
<td><a href="http://www.zilog.com">http://www.zilog.com</a></td>
<td>Z8 and Z80</td>
</tr>
<tr>
<td>Dallas Semi/Maxim</td>
<td><a href="http://www.maxim-ic.com">http://www.maxim-ic.com</a></td>
<td>8051</td>
</tr>
<tr>
<td>Freescale Semi</td>
<td><a href="http://www.freescale.com">http://www.freescale.com</a></td>
<td>68HC11/68HC08</td>
</tr>
</tbody>
</table>

See http://www.microcontroller.com for a complete list.

See the following web sites for PIC microcontrollers and PIC Trainer:

http://www.microchip.com

http://www.MicroDigitalEd.com

Review Questions

1. Name three features of the PIC18xxx.
2. What is the main difference between the PIC18Fxxx and PIC18Cxxx microcontrollers?
3. Give the size of RAM in each of the following:
   (a) PIC18F2420  (b) PIC18F4520
4. Give the size of the on-chip program ROM in each of the following:
   (a) PIC18F2420  (b) PIC18F4520
5. The PIC18 is a(n) _______-bit microprocessor.

SUMMARY

This chapter discussed the role and importance of microcontrollers in everyday life. Microprocessors and microcontrollers were contrasted and compared. We discussed the use of microcontrollers in the embedded market. We also discussed criteria to consider in choosing a microcontroller such as speed, memory, I/O, packaging, and cost per unit. The second section of this chapter described various families of the PIC, such as the PIC18 and PIC16, and their features. In addition, we discussed various members of the PIC18 family such as the PIC18F252 and PIC18F458.
PROBLEMS

SECTION 1.1: MICROCONTROLLERS AND EMBEDDED PROCESSORS

1. True or False. A general-purpose microprocessor has on-chip ROM.
2. True or False. Generally, a microcontroller has on-chip ROM.
3. True or False. A microcontroller has on-chip I/O ports.
4. True or False. A microcontroller has a fixed amount of RAM on the chip.
5. What components are usually put together with the microcontroller onto a single chip?
6. Intel's Pentium chips used in Windows PCs need external and chips to store data and code.
7. List three embedded products attached to a PC.
8. Why would someone want to use an x86 as an embedded processor?
9. Give the name and the manufacturer of some of the most widely used 8-bit microcontrollers.
10. In Question 9, which one has the most manufacture sources?
11. In a battery-based embedded product, what is the most important factor in choosing a microcontroller?
12. In an embedded controller with on-chip ROM, why does the size of the ROM matter?
13. In choosing a microcontroller, how important is it to have multiple sources for that chip?
14. What does the term "third-party support" mean?
15. Suppose that a microcontroller architecture has both 8-bit and 16-bit versions. Which of the following statements is true?
   (a) The 8-bit software will run on the 16-bit system.
   (b) The 16-bit software will run on the 8-bit system.

SECTION 1.2: OVERVIEW OF THE PIC18 FAMILY

16. The PIC18F458 has ______ bytes of on-chip program ROM.
17. The PIC18F2420 has ______ bytes of on-chip data RAM.
18. The PIC18F452 has ______ on-chip timer(s).
19. The PIC18F458 has ______ bytes of on-chip data RAM.
20. Check the Microchip web site to see if we have a ROMless version of the PIC18. Give the part number if there is one.
21. The PIC18F458 has ______ pins for I/O.
22. The PIC18Fxxx has circuitry to support ______ serial ports.
23. The PIC18F458 on-chip program ROM is of type ______.
24. The PIC16C432 on-chip program ROM is of type ______.
25. The PIC18F452 on-chip program ROM is of type ______.
26. The PIC18F8772 on-chip program ROM is of type ______.
27. Give the amount of program ROM and data RAM for the following chips:
   (a) PIC18F2420  (b) PIC18F458  (c) PIC18F8772
28. Of the PIC18 family, which memory type is the most cost effective if you are using a million of them in an embedded product?
29. What are the main differences between the PIC18F2420 and PIC18F2220?
30. The PIC18F458/4580 has ________ bytes of data EEPROM.

ANSWERS TO REVIEW QUESTIONS

SECTION 1.1: MICROCONTROLLERS AND EMBEDDED PROCESSORS

1. True
2. A microcontroller-based system
3. (d)
4. (d)
5. It is dedicated because it is dedicated to doing one type of job.
6. Embedded system means that the application and processor are combined into a single system.
7. Having multiple sources for a given part means you are not hostage to one supplier. More importantly, competition among suppliers brings about lower cost for that product.

SECTION 1.2: OVERVIEW OF THE PIC18 FAMILY

1. 4K of RAM space, 2M of on-chip ROM space, and a large number of I/O pins.
2. C is OTP while F is the flash ROM.
3. The PIC18F2420 has 768 bytes of RAM and the PIC18F4520 has 1536 bytes.
4. (a) 16K
   (b) 32K
5. 8
CHAPTER 2

PIC ARCHITECTURE & ASSEMBLY LANGUAGE PROGRAMMING

OBJECTIVES

Upon completion of this chapter, you will be able to:

» Examine the data RAM file register of the PIC microcontroller
» Manipulate data using the WREG and MOVE instructions
» Perform simple operations such as ADD and MOVE using the file register and access bank in the PIC microcontroller
» Explain the purpose of the status register
» Discuss data RAM memory space allocation in the PIC microcontroller
» List SFRs (special function registers) of the PIC microcontroller
» Code simple PIC Assembly language instructions
» Describe PIC data types and directives
» Assemble and run a PIC program using MPLAB
» Describe the sequence of events that occur upon PIC power-up
» Examine programs in PIC ROM code
» Explain the PIC ROM memory map
» Detail the execution of PIC Assembly language instructions
» Understand the RISC and Harvard architectures of the PIC microcontroller
» Examine the PIC’s registers and data RAM using the MPLAB simulator
CPUs use many registers to store data temporarily. To program in Assembly language, we must understand the registers and architecture of a given CPU and the role they play in processing data. In Section 2.1 we look at the WREG register of the PIC. We demonstrate the use of one of the most widely used registers of the PIC with simple instructions such as MOVE and ADD. Allocation of RAM memory inside the PIC and the access bank of the PIC18 are discussed in Section 2.2. Programming the access bank is examined in Section 2.3. In Section 2.4 we discuss the status register’s flag bits and how they are affected by arithmetic instructions. In Section 2.5 we look at some widely used Assembly language directives, pseudocode, and data types related to the PIC. In Section 2.6 we examine Assembly language and machine language programming and define terms such as mnemonics, opcode, operand, and so on. The process of assembling and creating a ready-to-run program for the PIC is discussed in Section 2.7. Step-by-step execution of a PIC program and the role of the program counter are examined in Section 2.8. The merits of RISC architecture are examined in Section 2.9. Assembling and running of the PIC programs with MPLAB are discussed in Section 2.10. In that section we also examine the registers and memory of the PIC using the MPLAB simulator.

SECTION 2.1: THE WREG REGISTER IN THE PIC

PIC microcontrollers have many registers for arithmetic and logic operations. Among them is the WREG register. Because there are a large number of registers inside the PIC, we will concentrate on the widely used register WREG in this section. General-purpose registers are covered in the next section, as well as special function registers. In this section we examine the WREG register of the PIC and show its use with the simple instructions MOVE and ADD.

WREG register

In the CPU, registers are used to store information temporarily. That information could be a byte of data to be processed, or an address pointing to the data to be fetched. The vast majority of PIC registers are 8-bit registers. In the PIC there is only one data type: 8-bit. The 8 bits of a register are shown in the diagram below. These range from the MSB (most-significant bit) D7 to the LSB (least-significant bit) D0. With an 8-bit data type, any data larger than 8 bits must be broken into 8-bit chunks before it is processed.

```
+---+---+---+---+---+---+---+---+
| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
```

The 8-bit WREG register is the most widely used register in the PIC microcontroller. WREG stands for working register, as there is only one. The WREG register is the same as the accumulator in other microprocessors. The WREG register is used for all arithmetic and logic instructions. To understand the use of the WREG register, we will show it in the context of two simple instructions: MOVE and ADD.
MOVLW instruction

Simply stated, the MOVLW instruction moves 8-bit data into the WREG register. It has the following format:

\[ \text{MOVLW K } \text{;move literal value K into WREG} \]

K is an 8-bit value that can range from 0–255 in decimal, or 00–FF in hex. The L stands for literal, which means, literally, a number must be used. In other words, if we see the word literal in any instruction, we are dealing with an actual value that must be provided right there with the instruction. This is similar to the immediate value we see in other microprocessors. Notice that in MOVLW, the letter L (literal) comes first and then the letter W (WREG), which means “move a literal value to WREG,” the destination. The following instruction loads the WREG register with a literal value of 25H (i.e., 25 in hex).

\[ \text{MOVLW 25H } \text{;move value 25H into WREG (WREG = 25H)} \]

The following instruction loads the WREG register with value 87H (87 in hex).

\[ \text{MOVLW 87H } \text{;load 87H into WREG (WREG = 87H)} \]

The following instruction loads the WREG register with value 15H (15 in hex and 21 in decimal).

\[ \text{MOVLW 15H } \text{;load 15H into WREG (WREG = 15H)} \]

ADDLW instruction

The ADDLW instruction has the following format:

\[ \text{ADDLW K } \text{;ADD literal value K to WREG} \]

The ADD instruction tells the CPU to add the literal value K to register WREG and put the result back in the WREG register. Notice that in ADDLW, first comes the letter L (literal) and then the letter W (WREG), which means “add a literal value to WREG,” the destination. To add two numbers such as 25H and 34H, one can do the following:

\[ \text{MOVLW 25H } \text{;load 25H into WREG} \]
\[ \text{ADDLW 34H } \text{;add value 34 to W (W = W + 34H)} \]

Executing the above lines results in \( \text{WREG} = 59H \) (25H + 34H = 59H)

Figure 2-1 shows the literal value and WREG being fed to the PIC ALU.
The following program will add values 12H, 16H, 31H, and 43H:

MOVLW 12H ;load value 12H into WREG (WREG = 12H)
ADDLW 16H ;add 16 to WREG (WREG = 28H)
ADDLW 11H ;add 11 to WREG (WREG = 39H)
ADDLW 43H ;add 43 to WREG (WREG = 7CH)

When programming the WREG register of the PIC microcontroller with a literal value, the following points should be noted:

1. Values can be loaded directly into the WREG register. There is no need for a preceding pound sign or dollar sign to indicate that a value is an immediate value as is the case with some other microcontrollers.
2. If values 0 to F are moved into an 8-bit register such as WREG, the rest of the bits are assumed to be all zeros. For example, in "MOVLW 5H" the result will be WREG = 05H; that is, WREG = 00000101 in binary.
3. Moving a value larger than 255 (FF in hex) into the WREG register will truncate the upper byte and cause a warning in the .err file.

MOVLW 7F2H ;ILLEGAL 7F2H > 8 bits (FFH), becomes F2H
MOVLW 456H ;ILLEGAL 456H > FFH, becomes 56H
MOVLW 60A5H ;ILLEGAL but becomes A5H
Review Questions

1. Write instructions to move value 34H into the WREG register.
2. Write instructions to add the values 16H and CDH. Place the result in the WREG register.
3. True or false. No value can be moved directly into the WREG register.
4. What is the largest hex value that can be moved into an 8-bit register? What is the decimal equivalent of that hex value?
5. The vast majority of registers in the PIC are _____-bit.

SECTION 2.2: THE PIC FILE REGISTER

The PIC microcontroller has many other registers in addition to the WREG register. They are called data memory space to distinguish them from program (code) memory space. The data memory space in PIC is a read/write (static RAM) memory. In the PIC microcontroller literature, the data memory is also called the file register. In this section, we examine the various locations of file register data RAM in the PIC family and discuss their usage with simple instructions such as ADD and MOVE.

File register (data RAM) space allocation in PIC

The file register is read/write memory used by the CPU for data storage, scratch pad, and registers for internal use and functions. As with WREG, we can perform arithmetic and logic operations on many locations of the file register data RAM. The PIC microcontrollers’ file register size ranges from 32 bytes to several thousand bytes depending on the chip. Even within the same family, the size of the file register data RAM varies from chip to chip. Notice that the file register data RAM has a byte-size width, just like WREG. The file register data RAM in PIC is divided into two sections: (a) Special Function Registers (SFR), and (b) General-Purpose Registers (GPR). The general-purpose register section is also referred to as General-Purpose RAM (GP RAM). We examine each section separately.

SFRs (Special Function Registers)

The Special Function Registers (SFRs) are dedicated to specific functions such as ALU status, timers, serial communication, I/O ports, ADC, and so on. The function of each SFR is fixed by the CPU designer at the time of design because it is used for control of the microcontroller or peripheral. The PIC SFRs are 8-bit registers. The number of locations in the file register set aside for SFR depends on the pin numbers and peripheral functions supported by that chip. That number can vary from chip to chip even among members of the same family. Some have as few as 7 (8-pin PIC12C508 with no on-chip analog-to-digital converter) and some have over a hundred (40-pin PIC18F458 with on-chip analog-to digital converter). For example, the more timers we have in a PIC chip, the more SFR registers we will have. We will study and use many SFRs in future chapters.
The general-purpose registers are a group of RAM locations in the file register that are used for data storage and scratch pad. Each location is 8 bits wide and can be used to store any data we want as long as it is 8-bit. Again, the number of RAM locations in the file register that are set aside for general-purpose registers can vary from chip to chip, even among members of the same family. In the PIC controllers, the space that is not allocated to the SFRs typically is used for general-purpose registers. That means in a PIC chip with a thousand-byte file register, no more than 100 bytes are used for SFRs and the rest are used for general-purpose registers. A larger GPR size means more difficulties in managing these registers if you use Assembly language programming. In today’s high-performance microcontroller, however, with over a thousand bytes of GPR, the job of managing them is handled by the C compilers. Indeed, the C compilers are the very reason we need a large GPR since it makes it easier for C compilers to store parameters and perform their jobs much faster. See Table 2-1 for a comparison of file registers among various PIC chips. Also see Figure 2-2.

### Table 2-1: File Register Size for PIC Chips

<table>
<thead>
<tr>
<th>PIC</th>
<th>File Register (Bytes)</th>
<th>SFR (Bytes)</th>
<th>Available space for GPR (Bytes)</th>
</tr>
</thead>
<tbody>
<tr>
<td>12F508</td>
<td>32</td>
<td>7</td>
<td>25</td>
</tr>
<tr>
<td>16F84</td>
<td>80</td>
<td>12</td>
<td>68</td>
</tr>
<tr>
<td>18F1220</td>
<td>512</td>
<td>256</td>
<td>256</td>
</tr>
<tr>
<td>18F452</td>
<td>1792</td>
<td>256</td>
<td>1536</td>
</tr>
<tr>
<td>18F2220</td>
<td>768</td>
<td>256</td>
<td>512</td>
</tr>
<tr>
<td>18F458</td>
<td>1792</td>
<td>256</td>
<td>1536</td>
</tr>
<tr>
<td>18F8722</td>
<td>4096</td>
<td>158</td>
<td>3938</td>
</tr>
</tbody>
</table>

Extracted from http://www.microchip.com

### GP RAM vs. EEPROM in PIC chips

Note that there are two RAM columns in the chip information section of the Microchip web site. One refers to the general-purpose registers’ (GP RAM) size, and the other is the EEPROM size. GP RAM (which constitutes most of the file register) must not be confused with the EEPROM data memory. The GPRs are used by the CPU for internal data storage, whereas the EEPROMs are considered as an add-on memory that one can also add externally to the chip. In other words, while many PIC chips have zero bytes of EEPROM data memory, it is impossible for a microcontroller to have zero size for the file register. The EEPROM memory of PIC chips is covered in Chapter 14.

The Microchip website provides the data RAM size, which is the same as GPR size.
Figure 2-2. File Registers of PIC12, PIC16, and PIC18

<table>
<thead>
<tr>
<th>PIC12F508</th>
<th>PIC16F84</th>
<th>PIC18F8722</th>
</tr>
</thead>
<tbody>
<tr>
<td>00h</td>
<td>00h</td>
<td>0000h</td>
</tr>
<tr>
<td>07h</td>
<td>0Bh</td>
<td>F7Fh</td>
</tr>
<tr>
<td>08h</td>
<td>0Ch</td>
<td>F80h</td>
</tr>
<tr>
<td>1Fh</td>
<td>4Fh</td>
<td>FFFh</td>
</tr>
</tbody>
</table>

8-bit

0000h
0001h
0002h

FFDh
FFEh
FFFh

a) Maximum space of file register (data RAM) in PIC18F (4096 bytes)

8-bit

0000h

GP RAM

4096-128 = 3968 bytes

Not all members have this amount

F7Fh
F80h

SFR Region (128 bytes)

b) File register allocation between GP RAM and SFR

8-bit

0000h

Various PIC18 members have different amount

All PIC18F chips have this section

FFFh

See Chapter 6 for more on how to access the entire 4096 of the file register

8-bit

Bank 0

GPRAM

128 bytes

Access Bank

Seg 0

Seg 1

Bank 1

Bank 14

Bank 15

SFR

128 bytes

c) Data memory map

d) Access Bank

8-bit

0000h

128 bytes

Access Bank

GPRAM

SFR

07Fh

0F0h

F80h

FFFh

d) Access Bank

Figure 2-3. File Register for PIC18 Family
File register and access bank in the PIC18

The file register of the PIC18 family can have a maximum of 4096 (4K) bytes. With 4096 bytes, the file register has addresses of 000–FFFH. The file register in the PIC18 is divided into 256-byte banks. Therefore, we can have up to a maximum of 16 banks (16 × 256 = 4096). Although not all members of the PIC18 family have that many banks, every PIC18 family member has at least one bank for the file register. This bank is called the access bank and is the default bank when we power up the PIC18 chip. To simplify the discussion of how to use the file register in the PIC family, we focus on this single bank that is found in every member of the PIC18 family. You can examine the file registers in other PIC families such as PIC12 and PIC16 at the Microchip website. In this book we concentrate on the PIC18 series with their large file register, although the insight gained in the process can be applied to the PIC16 and PIC12 series.

Examine the access bank for the PIC18 in Figure 2-3. The 256-byte access bank is divided into two equal sections of 128 bytes. These 128-byte sections are given to the general-purpose registers and special function registers. The 128 bytes from locations 00H to 7FH are set aside for general-purpose registers and are used for read/write storage, or what is normally called a scratch pad. These 128 locations of RAM are widely used for storing data and parameters by PIC18 programmers and C compilers. Each location of this 128-byte RAM of general-purpose registers can be accessed directly by its address. We will use these locations in future chapters to store data brought into the CPU via I/O and serial ports. We will also use them to define counters for time delay in Chapter 3. The other 128 bytes of the access bank is used for SFRs. It has addresses of F80H to FFFH, as shown in Figure 2-4. One might wonder why the memory space of the SFRs and GPRs in the access bank is not contiguous. The reason is to allow the RAM space between 080H and F7FH to be used for the general-purpose registers by various members of the PIC18 if they implement a larger data RAM size for the file register. A file register of more than 256 bytes will necessitate bank switching. Bank switching is a method used to access all the banks of the file register for PIC18 family members that have more than the minimum access bank. PIC18 members with a file register of more than 256 bytes will be discussed in more detail in Chapter 6 when we discuss bank switching.

Notice that the I/O port SFRs, PORTA, PORTB, PORTC, PORTD, and associated registers are among the most widely used SFRs in PIC. See Chapter 4 for additional information on the special function registers.
Review Questions

1. True or false. Data space in PIC is SRAM memory, whereas program (code) space is of the ROM type.
2. The general-purpose RAM and SFRs together are called ________.
3. True or false. The larger the file register, the more difficult it is to manage.
4. True or false. The more file register space that is assigned to the SFRs, the less is available for the GP RAM.
5. The SFR registers in PIC are ______-bit.
6. The file register space in PIC18 is divided into _____-byte banks.
7. The file register space in PIC18 can be a maximum of _____ bytes.
SECTION 2.3: USING INSTRUCTIONS WITH THE DEFAULT ACCESS BANK

The instructions we have used so far are the literal (constant) value of K and the WREG register. They also used the WREG register as their destination. We saw simple examples of using MOVFW and ADDLW earlier in Section 2.1. The PIC allows direct access to other locations in the file register for ALU and other operations. In this section we show the instructions using various locations of the file register. This is one of the most important sections in the book for mastering the topic of PIC Assembly language programming.

MOVWF instruction

As we discussed in the last section, the access bank of the file register is the default bank upon powering up the PIC18. The term file register must be emphasized because the instructions have the letter F in their mnemonics. In instructions such as MOVWF, the F stands for a location in the file register, while W means WREG. The MOVWF instruction tells the CPU to move (in reality, copy) the source register of WREG to a destination in the file register (F). After this instruction is executed, the location in the file register will have the same value as register WREG. The location in the file register can be one of the SFRs or a location in the general purpose registers region. For example, the “MOVWF PORTA” instruction will move the contents of WREG into the SFR register called PORTA. The following program first loads the WREG register with value 55H, then moves this value around to various SFRs of ports B, C, and D:

```
MOVWF PORTB ;copy WREG to Port B (Port B = 55H)
MOVWF PORTC ;copy WREG to Port C (Port C = 55H)
MOVWF PORTD ;copy WREG to Port D (Port D = 55H)
```

PORTB, PORTC, and PORTD are part of the special function registers in the file register, as was shown in Figure 2-4. They can be connected to the I/O pins of the PIC microcontroller as we will see in Chapter 4. We can also move (copy) the contents of WREG into any location in the general-purpose registers (RAM) region of the file registers. The following program will put 99H into locations 0–4 of the GPR region in the file register:

```
MOVWF 0H  ;move (copy) WREG contents to location 0h
MOVWF 1H  ;move (copy) WREG contents to location 1h
MOVWF 2H
MOVWF 3H
MOVWF 4H
```

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>99</td>
</tr>
<tr>
<td>001</td>
<td>99</td>
</tr>
<tr>
<td>002</td>
<td>99</td>
</tr>
<tr>
<td>003</td>
<td>99</td>
</tr>
<tr>
<td>004</td>
<td>99</td>
</tr>
</tbody>
</table>

The chart indicates the contents of addresses 0–4 after execution of the code.
Example 2-1

State the contents of file register RAM locations after the following program:

```assembly
MOVLW 99H ;load WREG with value 99H
MOVF 12H
MOVLW 85H ;load WREG with value 85H
MOVF 13H
MOVLW 3FH ;load WREG with value 3FH
MOVF 14H
MOVLW 63H ;load WREG with value 63H
MOVF 15H
MOVLW 12H ;load WREG with value 12H
MOVF 16H
```

Solution:
After the execution of MOVWF 12H fileReg RAM location 12H has value 99H;
After the execution of MOVWF 13H fileReg RAM location 13H has value 85H;
After the execution of MOVWF 14H fileReg RAM location 14H has value 3FH;
After the execution of MOVWF 15H fileReg RAM location 15H has value 63H;
And so on, as shown in the chart.

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>012</td>
<td>99</td>
</tr>
<tr>
<td>013</td>
<td>85</td>
</tr>
<tr>
<td>014</td>
<td>3F</td>
</tr>
<tr>
<td>015</td>
<td>63</td>
</tr>
<tr>
<td>016</td>
<td>12</td>
</tr>
</tbody>
</table>

Notice that you cannot move literal (immediate) values directly into the general-purpose RAM locations in the PIC18. They must be moved there via WREG.

More instructions involving the WREG and the access bank

There is a group of logic and arithmetic instructions that involve both the WREG and a location in the file register. The ADDWF instruction is one of them. The ADDWF instruction adds together the contents of WREG and a file register location. The file register location can be one of the SFRs or a general-purpose register. The destination for the result can be the WREG or the file register. The following format indicates the destination:

```
ADDWF fileReg, D
```

where fileReg is the file register location and D indicates the destination bit. The D bit can be 0 or 1. If D = 0, it means that the destination is WREG. If D = 1, then the result will be placed in the file register.
The following will first put value 22H into GP RAM locations 5, 6, and 7, then add them together and put the result in WREG:

```
MOVLW 22H   ;WREG = 22H
MOVWF 5H    ;move (copy) WREG contents to location 5H
MOVWF 6H    ;move (copy) WREG contents to location 6H
MOVWF 7H    ;move (copy) WREG contents to location 7H
ADDWF 5H, 0 ;add W and loc 5, result in WREG (W = 44H)
ADDWF 6H, 0 ;add W and loc 6, result in WREG (W = 66H)
ADDWF 7H, 0 ;add W and loc 7, result in WREG (W = 88H)
```

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>005</td>
<td>22</td>
</tr>
<tr>
<td>006</td>
<td>22</td>
</tr>
<tr>
<td>007</td>
<td>22</td>
</tr>
</tbody>
</table>

GPR after the execution up to “MOVWF 7H”
WREG = 22H

Now look at the same program where the result is put into file register location 7:

```
MOVLW 22H   ;WREG = 22H
MOVWF 5H    ;move (copy) WREG contents to location 5H
MOVWF 6H    ;move (copy) WREG contents to location 6H
MOVWF 7H    ;move (copy) WREG contents to location 7H
ADDWF 5, 0  ;add W and loc 5, result in WREG (W = 44H)
ADDWF 6, 0  ;add W and loc 6, result in WREG (W = 66H)
ADDWF 7, 1  ;add W and loc 7, result in location 7H
            ;now location 7 has 88H and WREG = 66H
```

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>005</td>
<td>22</td>
</tr>
<tr>
<td>006</td>
<td>22</td>
</tr>
<tr>
<td>007</td>
<td>22</td>
</tr>
</tbody>
</table>

GP RAM after the execution up to “MOVWF 7H”
WREG = 22H

```
MOVLW 22H   ;WREG = 22H
MOVWF 5H    ;move (copy) WREG contents to location 5H
MOVWF 6H    ;move (copy) WREG contents to location 6H
MOVWF 7H    ;move (copy) WREG contents to location 7H
ADDWF 5, 0  ;add W and loc 5, result in WREG (W = 44H)
ADDWF 6, 0  ;add W and loc 6, result in WREG (W = 66H)
ADDWF 7, 1  ;add W and loc 7, result in location 7H
            ;now location 7 has 88H and WREG = 66H
```

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>005</td>
<td>22</td>
</tr>
<tr>
<td>006</td>
<td>22</td>
</tr>
<tr>
<td>007</td>
<td>88</td>
</tr>
</tbody>
</table>

GP RAM after the execution up to “ADDWF 7H, 0”
WREG = 88H

GPR after the execution up to “ADDWF 7H, 0”
WREG = 88H
To make things less confusing as far as the D bit is concerned, the PIC assembler allows us to use the letters W or F instead of 0 or 1 to indicate the destination. Look at the following two formats:

```
ADDWF fileReg, w ;add WREG and fileReg. WREG = the result
ADDWF fileReg, f ;add WREG and fileReg
    ;fileReg = the result
```

This format is much easier and will help us to avoid confusion about the destination. Look at the rewrite of the last program with the new format for the ADDWF instruction:

```
MOVWF 5H ;move (copy) WREG contents to location 5H
MOVWF 6H ;move (copy) WREG contents to location 6H
MOVWF 7H ;move (copy) WREG contents to location 7H
ADDWF 5H,W ;add W and loc 5, result in WREG (W = 44H)
ADDWF 6H,W ;add W and loc 6, result in WREG (W = 66H)
ADDWF 7H,F ;add W and loc 7, result in location 7
    ;now location 7 has 88H and WREG = 66H
```

The above concept is important and must be understood since there are a large number of instructions with this format. Compare Examples 2-2 and 2-3.

---

**Example 2-2**

State the contents of file register RAM locations 12H and WREG after the following program:

```
MOVWF 0 ;move 0 WREG to clear it (WREG = 0)
MOVWF 12H ;move WREG to location 12 to clear it
MOVWF 22H ;load WREG with value 22H
ADDWF 12H, F ;add WREG to loc 12H, loc 12 = sum
ADDWF 12H, F ;add WREG to loc 12H, loc 12 = sum
ADDWF 12H, F ;add WREG to loc 12H, loc 12 = sum
ADDWF 12H, F ;add WREG to loc 12H, loc 12 = sum
```

**Solution:**

The program clears both the WREG and RAM location 12H in the file register. Then it loads WREG with value 22H. From then on, it adds the WREG register and location 12 together and saves the result in location 12H. It does that four times. At the end, location 12H of GP RAM has the value of 88H (4 x 22H = 88H) and WREG = 22H.

After each “ADDWF 12, F” instruction

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
<th>Address</th>
<th>Data</th>
<th>Address</th>
<th>Data</th>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>011</td>
<td>011</td>
<td>012</td>
<td>012</td>
<td>012</td>
<td>012</td>
<td>012</td>
<td>012</td>
</tr>
<tr>
<td>012</td>
<td>22</td>
<td>012</td>
<td>44</td>
<td>012</td>
<td>66</td>
<td>012</td>
<td>88</td>
</tr>
<tr>
<td>013</td>
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<td>013</td>
<td>013</td>
<td>013</td>
<td>013</td>
<td>013</td>
<td>013</td>
</tr>
</tbody>
</table>

WREG = 22H  WREG = 22H  WREG = 22H  WREG = 22H

---

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Example 2-3

Rewrite the last example to place the sum in WREG as you add the file register locations and the WREG register.

```plaintext
MOVLW 0       ;move 0 WREG to clear it (WREG = 0)
MOVWF 12H      ;move WREG to location 12 to clear it
MOVLW 22H      ;load WREG with value 22H
ADDWF 12H, W   ;add WREG and loc 12H, WREG = sum
ADDWF 12H, W   ;add WREG and loc 12H, WREG = sum
ADDWF 12H, W   ;add WREG and loc 12H, WREG = sum
```

Solution:
The program adds WREG and location 12H together and saves the result in WREG each time. At the end, location 12H has a value of 22H and WREG = 88H (4 × 22H = 88H).

After each “ADDWF 12, W” instruction:

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>011</td>
<td></td>
</tr>
<tr>
<td>012</td>
<td>22</td>
</tr>
<tr>
<td>013</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Address</th>
<th>Data</th>
</tr>
</thead>
<tbody>
<tr>
<td>011</td>
<td></td>
</tr>
<tr>
<td>012</td>
<td>22</td>
</tr>
<tr>
<td>013</td>
<td></td>
</tr>
</tbody>
</table>

WREG = 22H     WREG = 44H     WREG = 66H     WREG = 88H

Figure 2-5. WREG, fileReg, and ALU in PIC18
Now examine the instructions in Tables 2-2 and 2-3. The instructions in Table 2-2 operate on both WREG and a file register location and then give you the option of placing the result in WREG or a file register location. The instructions in Table 2-3, however, operate on the file register only and then give you the option of placing the result in WREG or a file register location.

Table 2-2: ALU Instructions Using Both WREG and fileReg

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDWF</td>
<td>ADD WREG and fileReg</td>
</tr>
<tr>
<td>ADDWFC</td>
<td>ADD WREG and fileReg with Carry</td>
</tr>
<tr>
<td>ANDWF</td>
<td>AND WREG with fileReg</td>
</tr>
<tr>
<td>IORWF</td>
<td>OR WREG with fileReg</td>
</tr>
<tr>
<td>SUBFWB</td>
<td>Subtract fileReg from WREG with borrow</td>
</tr>
<tr>
<td>SUBWF</td>
<td>Subtract WREG from fileReg</td>
</tr>
<tr>
<td>SUBWFB</td>
<td>Subtract WREG from fileReg with borrow</td>
</tr>
<tr>
<td>XORWF</td>
<td>Exclusive-OR WREG with fileReg</td>
</tr>
</tbody>
</table>

**Note:** The d bit selects the destination for the operation. If d = w; the result is stored in WREG (d = 0). If d = F; the result is stored in the fileReg (d = 1). The default is F. That means “ADDWF myfile” is the same as “ADDWF myfile,F.”

See Chapter 5 for examples of the instructions in Table 2-2.

Table 2-3: File Register Instructions Using fileReg or WREG as Destination

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>COMF</td>
<td>Complement fileReg</td>
</tr>
<tr>
<td>DECF</td>
<td>Decrement fileReg</td>
</tr>
<tr>
<td>DECFSZ</td>
<td>Decrement fileReg and skip if zero</td>
</tr>
<tr>
<td>DECFSNZ</td>
<td>Decrement fileReg and skip if not zero</td>
</tr>
<tr>
<td>INCF</td>
<td>Increment fileReg</td>
</tr>
<tr>
<td>INCFSZ</td>
<td>Increment fileReg and skip if zero</td>
</tr>
<tr>
<td>INCSNZ</td>
<td>Increment fileReg and skip if not zero</td>
</tr>
<tr>
<td>MOVF</td>
<td>Move fileReg</td>
</tr>
<tr>
<td>NEGF</td>
<td>Negative fileReg</td>
</tr>
<tr>
<td>RLCF</td>
<td>Rotate left fileReg through carry</td>
</tr>
<tr>
<td>RLNCF</td>
<td>Rotate left fileReg (No carry)</td>
</tr>
<tr>
<td>RRCF</td>
<td>Rotate right fileReg through carry</td>
</tr>
<tr>
<td>RRNCF</td>
<td>Rotate right fileReg (No carry)</td>
</tr>
<tr>
<td>SWAPF</td>
<td>Swap nibbles in fileReg</td>
</tr>
<tr>
<td>BTG</td>
<td>Bit Toggle fileReg</td>
</tr>
</tbody>
</table>

**Note:** The d bit selects the destination for the operation. If d = w; the result is stored in the WREG (d = 0). If d = F; the result is stored in the fileReg (d = 1). The default is F. That means “DECF myfile” is the same as “DECF myfile,F.”

Chapters 3 through 6 will show how to use the instructions in Table 2-3.
COMF instruction

The “COMF fileReg, d” instruction complements (inverts) the contents of fileReg and places the result in WREG or fileReg. This is an example of what is called Read - Modify - Write and we will see more of this in future chapters. In the following program, we put 55H into WREG and then send it to SFR location of Port B. Then the content of Port B is complemented, which becomes AA in hex. The 01010101 (55H) is inverted and becomes 10101010 (AAH).

```assembly
MOVLW 55H    ;WREG = 55h
MOVWF PORTB ;Move WREG to Port B SFR (PB = 55h)
COMF PORTB, F ;complement Port B (PB = AAh)
```

Examine Example 2-4.

**Example 2-4**

Write a simple program to toggle the SFR of PORT B continuously forever.

**Solution:**

```assembly
MOVLW 55H    ;WREG = 55h
MOVWF PORTB ;move WREG to Port B SFR (PB = 55h)
B1  COMF PORTB, F ;complement Port B and place it in Port B
      GOTO B1 ;repeat forever (See Chapter 3 for GOTO)
```

DECF instruction

The “DECF fileReg, d” instruction decrements (subtracts one from) the contents of fileReg and places the result in WREG or fileReg. In the following program, we put the value 3 into fileReg location 0x20. Then the value in location 0x20 is decremented and placed in fileReg.

```assembly
MOVLW 3       ;WREG = 3
MOVWF 20H     ;move WREG to loc 20H (loc 20H = 3)
DECF 0x20, F  ;loc 20H has 2
DECF 0x20, F  ;loc 20H has 1
DECF 0x20, F  ;loc 20H has 0 and WREG = 3
```

Now, contrast the above code with the following:

```assembly
MOVLW 3       ;WREG = 3
MOVWF 20H     ;move WREG to loc 20H (loc 20H = 3)
DECF 0x20, w  ;loc 20H has 3 (WREG = 2)
DECF 0x20, w  ;loc 20H has 3 (WREG = 2)
DECF 0x20, w  ;loc 20H has 3 (WREG = 2)
```

The above concept will be used in loops in the next chapter.
**MOVF instruction**

The MOVF mnemonic is intended to perform MOVFW. It has the following format:

\[
\text{MOVF} \quad \text{fileReg}, \ D
\]

If \( D = 0 \), it copies the content of fileReg to WREG. If \( D = 1 \), the content of fileReg is copied to itself. While typically we use the MOVF instruction to bring data into WREG from I/O pins, we sometimes use it to copy fileReg to itself for the purpose of testing fileReg contents. Examine the difference between the MOVWF and MOVF instructions. We used the MOVWF instruction earlier to move data to SFRs such as Port B. We also saw how it is widely used to load fixed (literal) data into the RAM locations of the file register because there is no way we can load data into the file register directly. In contrast, the MOVF instruction is widely used to bring data from I/O ports such as Port B into the CPU. We also use the MOVF instruction to bring data into WREG from any SFRs or from any location in the GP RAM in order to perform arithmetic and operations on them. Examine Examples 2-5 and 2-6. Note that the only time we use the “MOVF fileReg, F” instruction to copy data from fileReg to itself is when we want to affect the flag bits of the status register. The status register bits are discussed in the next section, and Chapter 3 shows how to use them.

**Example 2-5**

Write a program to get data from the SFRs of Port B and send it to the SFRs of PORT C continuously.

**Solution:**

```
AGAIN MOVF PORTB, W ;bring data from PortB into WREG
MOVWF PORTC ;send it to Port C
GOTO AGAIN ;keep doing it forever
```

In Example 2-5 we use GOTO to repeat an action indefinitely. We will study looping in Chapter 3. The details of I/O ports are discussed in Chapter 4.

**Example 2-6**

Write a program to get data from the SFRs of Port B. Add the value 5 to it and send it to the SFRs of Port C.

**Solution:**

```
MOVF PORTB,W ;bring data from Port B into WREG
ADDLW 05H ;add 5 to WREG
MOVWF PORTC ;copy WREG to Port C
```
MOVFF instruction

The MOVFF instruction copies data from one location in fileReg to another location in fileReg. The fileReg location for source and destination can be any of the 4096 locations in the data RAM space of the PIC18. The MOVFF instruction allows us to move data within the 4K space of the data RAM without going through the WREG register. (See Figure 2-6.) Compare Examples 2-5 and 2-7.

![Diagram of MOVFF instruction](image)

Figure 2-6. Moving Data Directly Among the fileReg Locations

**Example 2-7**

Write a program to get data from the SFRs of Port B and send it to the SFRs of PORT C continuously using MOVFF. Compare this to Example 2-5 and explain the difference.

**Solution:**

```
AGAIN MOVFF PORTB, PORTC ; copy data from Port B to Port C
GOTO AGAIN ; keep doing it forever
```

In Example 2-5 we have:

```
AGAIN MOVF PORTB, W ; bring data from Port B into WREG
MOVWF PORTC ; send it to Port C
GOTO AGAIN ; keep doing it forever
```

Using MOBVFF we simply copy data from one location to another location. But when we use WREG we can perform arithmetic and logic operations on data before it is moved.
Review Questions

1. True or false. The access bank is 256 bytes divided evenly between GPRs and SFRs.
2. Write instructions to add the values 16H and CDH. Place the result in location 0 of the file register.
3. True or false. No value can be moved directly into general-purpose RAM.
4. What is the largest hex value that can be moved into a location in the file register? What is the decimal equivalent of the hex value?
5. “ADDWF PORTB, W” puts the result in _____.

SECTION 2.4: PIC STATUS REGISTER

Like all other microprocessors, the PIC has a flag register to indicate arithmetic conditions such as the carry bit. The flag register in the PIC is called the status register. In this section, we discuss various bits of this register and provide some examples of how it is altered. Chapters 3 and 5 show how the flag bits of the status register are used.

PIC18 status register

The status register is an 8-bit register. It is also referred to as the flag register. Although the status register is 8 bits wide, only 5 bits of it are used by the PIC18. The three unused bits are unimplemented and read as 0. The five flags are called conditional flags, meaning that they indicate some conditions that result after an instruction is executed. These five flags are C (carry), DC (digital carry), Z (zero), OV (overflow), and N (negative). See Figure 2-7 for the bits of the status register. Each of the conditional flags can be used to perform a conditional branch (jump), as we will see in Chapter 3.

<table>
<thead>
<tr>
<th>D7</th>
<th>D6</th>
<th>D5</th>
<th>D4</th>
<th>D3</th>
<th>D2</th>
<th>D1</th>
<th>D0</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>X</td>
<td>X</td>
<td>N</td>
<td>OV</td>
<td>Z</td>
<td>DC</td>
<td>C</td>
</tr>
</tbody>
</table>

C – Carry flag
DC – Digital Carry flag
Z – Zero flag
OV – Overflow flag
N – Negative flag
X – D5, D6, and D7 are not implemented, and reserved for future use.

Figure 2-7. Bits of Status Register

The following is a brief explanation of the flag bits of the status register. The impact of instructions on this register is then discussed.

CHAPTER 2: PIC ARCHITECTURE & ASSEMBLY LANGUAGE
C, the carry flag

This flag is set whenever there is a carry out from the D7 bit. This flag bit is affected after an 8-bit addition or subtraction. Chapter 5 shows how the carry flag is used.

DC, the digital carry flag

If there is a carry from D3 to D4 during an ADD or SUB operation, this bit is set; otherwise, it is cleared. This flag bit is used by instructions that perform BCD (binary coded decimal) arithmetic. In some microprocessors this is called the AC flag (Auxiliary Carry flag). See Chapter 5 for more information.

Z, the zero flag

The zero flag reflects the result of an arithmetic or logic operation. If the result is zero, then Z = 1. Therefore, Z = 0 if the result is not zero. See Chapter 3 to see how we use the Z flag for looping.

OV, the overflow flag

This flag is set whenever the result of a signed number operation is too large, causing the high-order bit to overflow into the sign bit. In general, the carry flag is used to detect errors in unsigned arithmetic operations while the overflow flag is used to detect errors in signed arithmetic operations. The OV and N flag bits are used for the signed number arithmetic operations and are discussed in Chapter 5.

N, the negative flag

Binary representation of signed numbers uses D7 as the sign bit. The negative flag reflects the result of an arithmetic operation. If the D7 bit of the result is zero, then N = 0 and the result is positive. If the D7 bit is one, then N = 1 and the result is negative. The negative and OV flag bits are used for the signed number arithmetic operations and are discussed in Chapter 5.

ADDLW instruction and the status register

Next we examine the impact of the ADDLW instruction on the flag bits C, DC, and Z of the status register. Some examples should clarify their meanings. Although all the flag bits C, Z, DC, OV, and N are affected by the ADDLW instruction, we will focus on flags C, DC, and Z for now. The other flag bits are discussed in Chapter 5, because they relate only to signed number operations. Examine Examples 2-8 through 2-10 to see the impact of the ADD instruction on selected flag bits.

Not all instructions affect the flags

Some instructions affect all the five flag bits C, DC, Z, OV, and N (e.g., ADDWL). But some instructions affect no flag bits at all. The move instructions are in this category (except MOVF). And some instructions affect only the Z or N flag bits, or both. The logic instructions are in this category (e.g., ANDWL).
Example 2-8

Show the status of the C, DC, and Z flags after the addition of 38H and 2FH in the following instructions:

```
MOVLW 38H
ADDLW 2FH ; add 2FH to WREG
```

Solution:

<table>
<thead>
<tr>
<th>38H</th>
<th>0011 1000</th>
</tr>
</thead>
<tbody>
<tr>
<td>+ 2FH</td>
<td>0010 1111</td>
</tr>
<tr>
<td>67H</td>
<td>0110 0111</td>
</tr>
</tbody>
</table>

WREG = 67H

C = 0 because there is no carry beyond the D7 bit.
DC = 1 because there is a carry from the D3 to the D4 bit.
Z = 0 because the WREG has a value other than 0 after the addition.

Example 2-9

Show the status of the C, DC, and Z flags after the addition of 9CH and 64H in the following instructions:

```
MOVLW 9CH
ADDLW 64H ; add 64H to WREG
```

Solution:

<table>
<thead>
<tr>
<th>9CH</th>
<th>1001 1100</th>
</tr>
</thead>
<tbody>
<tr>
<td>+ 64H</td>
<td>0110 0100</td>
</tr>
<tr>
<td>100H</td>
<td>0000 0000</td>
</tr>
</tbody>
</table>

WREG = 00

C = 1 because there is a carry beyond the D7 bit.
DC = 1 because there is a carry from the D3 to the D4 bit.
Z = 1 because the WREG has a value 0 in it after the addition.

Example 2-10

Show the status of the C, DC, and Z flags after the addition of 88H and 93H in the following instructions:

```
MOVLW 88H
ADDLW 93H ; add 93H to WREG
```

Solution:

<table>
<thead>
<tr>
<th>88H</th>
<th>1000 1000</th>
</tr>
</thead>
<tbody>
<tr>
<td>+ 93H</td>
<td>1001 0011</td>
</tr>
<tr>
<td>11BH</td>
<td>0001 1011</td>
</tr>
</tbody>
</table>

WREG = 1BH

C = 1 because there is a carry beyond the D7 bit.
DC = 0 because there is no carry from the D3 to the D4 bit.
Z = 0 because the WREG has a value other than 0 after the addition.
Table 2-4 shows the instructions and the flag bits affected by them. Appendix A provides a complete list of all the instructions and their associated flag bits.

Table 2-4: Instructions That Affect Flag Bits

<table>
<thead>
<tr>
<th>Instruction</th>
<th>C</th>
<th>DC</th>
<th>Z</th>
<th>OV</th>
<th>N</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDLW</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>ADDWF</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>ADDWFC</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>ANDLW</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>ANDWF</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>CLRF</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>COMF</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>DAW</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DECFS</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>INCFS</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>IORLWS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>IORWFs</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>MOVFS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>NEGFs</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>RLCS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>RLNCS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>RCFs</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>RRNCS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>SUBFWS</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>SUBLW</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>SUBWFs</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>SUBWFBS</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>XORLWS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
<tr>
<td>XORWS</td>
<td>X</td>
<td></td>
<td></td>
<td></td>
<td>X</td>
</tr>
</tbody>
</table>

Note: X can be 0 or 1.

See Chapter 5 for how to use these instructions.

Flag bits and decision making

Because status flags are also called conditional flags, there are instructions that will make a conditional jump (branch) based on the status of the flag bits. Table 2-5 provides the list. Chapter 3 will discuss the conditional branch instructions and how they are used.

Table 2-5: PIC18 Branch (Jump) Instructions Using Flag Bits

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Action</th>
</tr>
</thead>
<tbody>
<tr>
<td>BC</td>
<td>Branch if ( C = 1 )</td>
</tr>
<tr>
<td>BNC</td>
<td>Branch if ( C \neq 0 )</td>
</tr>
<tr>
<td>BZ</td>
<td>Branch if ( Z = 1 )</td>
</tr>
<tr>
<td>BNZ</td>
<td>Branch if ( Z \neq 0 )</td>
</tr>
<tr>
<td>BN</td>
<td>Branch if ( N = 1 )</td>
</tr>
<tr>
<td>BNC</td>
<td>Branch if ( N \neq 0 )</td>
</tr>
<tr>
<td>BOV</td>
<td>Branch if ( OV = 1 )</td>
</tr>
<tr>
<td>BNOV</td>
<td>Branch if ( OV \neq 0 )</td>
</tr>
</tbody>
</table>
Review Questions

1. The flag register in the PIC is called the ________.
2. What is the size of the flag register in the PIC?
3. Which bits of the status register are unused?
4. Find the C, Z, and DC flag bits for the following code:
   \[
   \text{MOVLW } 9FH \\
   \text{ADDLW } 61H
   \]
5. Find the C, Z, and DC flag bits for the following code:
   \[
   \text{MOVLW } 82H \\
   \text{ADDLW } 22H
   \]
6. Find the C, Z, and DC flag bits for the following code:
   \[
   \text{MOVLW } 67H \\
   \text{ADDLW } 99H
   \]

SECTION 2.5: PIC DATA FORMAT AND DIRECTIVES

In this section we look at some widely used data formats and directives supported by the PIC assembler.

PIC data type

The PIC microcontroller has only one data type. It is 8 bits, and the size of each register is also 8 bits. It is the job of the programmer to break down data larger than 8 bits (00 to FFH, or 0 to 255 in decimal) to be processed by the CPU. For examples of how to process data larger than 8 bits, see Chapter 5. The data types used by the PIC can be positive or negative. A discussion of signed numbers is given in Chapter 5 also. The bit-addressable data is discussed in Chapters 4 and 6.

Data format representation

There are four ways to represent a byte of data in the PIC assembler. The numbers can be in hex, binary, decimal, or ASCII formats. The following are examples of how each work.

Hex numbers

There are four ways to show hex numbers:

1. We can use h (or H) right after the number like this: MOV LW 99H
2. Put 0x (or 0X) in front of the number like this: MOV LW 0x99
3. Put nothing in front or back of the number like this: MOV LW 99
4. Put h in front of the number, but with single quotes around the number like this: MOV LW '99'

We use all four of these methods in this book, because many application notes out there use one of them and we need to get used to them. Notice that some PIC assemblers might give you a warning (but no error) when you use 99H because the assembler already knows that data is in hex and there is no need to remind it. We do that simply to remind ourselves (and it is a good reminder) when we do coding in Assembly.
Here are a few lines of code that use the hex format:

MOVLW 25 ;WREG = 25H
ADDLW 0x11 ;WREG = 25H + 11H = 36H
ADDLW 12H ;WREG = 36H + 12H = 48H
ADDLW H'2A' ;WREG = 48H + 2AH = 72H
ADDLW 2CH ;WREG = 72H + 2CH = 9EH

The following are invalid:

MOVLW E5H ;invalid, it must be MOVLW 0E5H
ADDLW C6 ;invalid, it must be ADDLW 0C6

Notice in the last two instructions that if the value starts with the hex digits A–F, then it must be preceded with a zero. However, the following is valid:

MOVLW 0F ;valid, WREG = 0FH (or 00001111 in binary)

**Binary numbers**

There is only one way to represent binary numbers in a PIC assembler. It is as follows:

MOVLW B'10011001' ;WREG = 10011001 or 99 in hex

The lowercase b will also work. Note that ' is the single quote key, which is on the same key as the double quote ". This is different from other assemblers such as the 8051 and x86. Here are some examples of how to use it:

MOVLW B'00100101' ;WREG = 25H
ADDLW B'00010001' ;WREG = 25H + 11H = 36H

**Decimal numbers**

There are two ways to represent decimal numbers in a PIC assembler. One way is as follows:

MOVLW D'12' ;WREG = 00001100 or 0C in hex

The lowercase d will work also. This is different from other assemblers such as the 8051 and x86. In those assemblers, to indicate decimal numbers we simply use the decimal (e.g., 12) and nothing before or after it, while in the PIC assembler, 12 is the default for hex numbers. Here are some examples of how to use it:

MOVLW D'37' ;WREG = 25H (37 in decimal is 25 in hex)
ADDLW D'17' ;WREG = 37 + 17 = 54 where 54 in dec is 36H

The other way to represent decimal numbers is to use ".value" as seen in some application notes for PIC microcontrollers. This is shown as follows:

MOVLW .12 ;WREG = 00001100 = 0CH = 12
**ASCII character**

To represent ASCII data in a PIC assembler we use the letter A as follows:

```assembly
MOVLW A'2' ; WREG = 00110010 or 32 in hex (See Appendix F)
```

Lowercase ‘a’ will work as well. Again, this is different from other assemblers such as the 8051 and x86. In those assemblers, single quotes are used for single ASCII characters and double quotes are used for a string. Here are some more examples:

```assembly
MOVLW A'9' ; WREG = 39H, which is hex number for ASCII '9'
ADDLW A'1' ; WREG = 39H + 31H = 70H
            ; (31 hex is for ASCII '1')
MOVLW '9'  ; WREG = 39H another way for ASCII
```

To define ASCII strings (more than one character), we use the DB (define byte) directive. We will see DB usage in Chapter 6.

**Assembler directives**

While instructions tell the CPU what to do, directives (also called pseudo-instructions) give directions to the assembler. For example, the MOVLW and ADDLW instructions are commands to the CPU, but EQU, ORG, and END are directives to the assembler. The following sections present some more widely used directives of the PIC and how they are used.

**EQU (equate)**

This is used to define a constant value or a fixed address. The EQU directive does not set aside storage for a data item, but associates a constant number with a data or an address label so that when the label appears in the program, its constant will be substituted for the label. The following uses EQU for the counter constant, and then the constant is used to load the WREG register:

```assembly
COUNT EQU 0x25
            
            MOV LW COUNT ; WREG = 25H
```

When executing the above instruction "MOVLW COUNT", the register WREG will be loaded with the value 25H. What is the advantage of using EQU? Assume that a constant (a fixed value) is used throughout the program, and the programmer wants to change its value everywhere. By the use of EQU, the programmer can change it once and the assembler will change all of its occurrences throughout the program, rather than search the entire program trying to find every occurrence.

**SET**

This directive is used to define a constant value or a fixed address. In this regard, the SET and EQU directives are identical. The only difference is the value assigned by the SET directive may be reassigned later.
Using EQU for fixed data assignment

To get more practice using EQU to assign fixed data, examine the following:

; in hexadecimal
DATA1 EQU 39 ; hex data is the default
DATA2 EQU 0x39 ; another way for hex
DATA3 EQU 39H ; another way for hex (redundant)
DATA4 EQU H'39' ; another way for hex
DATA5 EQU h'39' ; another way for hex

; in binary
DATA6 EQU b'00110101' ; binary (35 in hex)
DATA7 EQU B'00110101' ; binary (35 in hex)

; in decimal
DATA8 EQU D'28' ; decimal numbers (1C in hex)
DATA9 EQU d'28' ; second way for decimal

; in ASCII
DATA10 EQU A'2' ; ASCII characters
DATA11 EQU a'2' ; another way for ASCII char
DATA12 EQU '2' ; another way for ASCII char

We use DB to allocate code ROM memory locations for fixed data such as ASCII strings. See Chapter 6 for more examples.

Using EQU for SFR address assignment

EQU is also widely used to assign SFR addresses. Examine the following code:

COUNTER EQU 0x00 ; counter value 00
PORTB EQU 0xFF6 ; SFR Port B address
MOVLW COUNTER ; WREG = 00H
MOVWF PORTB ; Port B now has 00 too
INCF PORTB, F ; Port B has 01
INCF PORTB, F ; increment Port B (Port B = 02)
INCF PORTB, F ; increment Port B (Port B = 03)

The above is for the PIC18 family. If you use a different PIC controller such as PIC16F, where Port B is a different address, then change the EQU address for Port B and re-assemble the program and run it.

COUNTER EQU 0x00 ; counter value 00
PORTB EQU 0x07 ; Port B addr in PIC16F
MOVLW COUNTER ; WREG = 00H
MOVWF PORTB ; Port B now has 00 too
INCF PORTB, F ; Port B has 01
INCF PORTB, F ; Port B has 02
INCF PORTB, F ; Port B has 03
and so on
Using EQU for RAM address assignment

Another common usage of EQU is for the address assignment of the general-purpose region of the file register. Examine the following rewrite of an earlier example using EQU:

```
MYREG EQU 0x12 ;assign RAM loc to MYREG
MOVLW 0 ;clear WREG (WREG = 0)
MOVWF MYREG ;clear MYREG (loc 12H has 0)
MOVLW 22H ;WREG = 22H
ADDWF MYREG, F ;MYREG = WREG + MYREG
ADDWF MYREG, F ;MYREG = WREG + MYREG
ADDWF MYREG, F ;MYREG = WREG + MYREG
ADDWF MYREG, F ;MYREG = WREG + MYREG
```

This is especially helpful when the address needs to be changed in order to use a different PIC chip for a given project. It is much easier to refer to a name than a number when accessing RAM address locations.

The following program will move value 9 into RAM locations 0–4, then add them together and place the sum in location 10H:

```
MYVAL EQU 9 ;MYVAL = 9
R0 EQU 0 ;assign RAM addresses to R0
R1 EQU 1 ;to R1
R2 EQU 2
R3 EQU 3
R4 EQU 4
SUM EQU 10H

MOVLW MYVAL ;WREG = 9
MOVWF R0 ;RAM loc 0 has 9
MOVWF R1 ;RAM loc 1 has 9
MOVWF R2 ;RAM loc 2 has 9
MOVWF R3 ;RAM loc 3 has 9
MOVWF R4 ;RAM loc 4 has 9
MOVLW 0 ;WREG = 0
ADDWF R0, W ;WREG = R0 + WREG
ADDWF R1, W ;WREG = R1 + WREG
ADDWF R2, W ;WREG = R2 + WREG
ADDWF R3, W ;WREG = R3 + WREG
ADDWF R4, W ;WREG = R4 + WREG
MOVWF SUM
```

**ORG (origin)**

The ORG directive is used to indicate the beginning of the address. It can be used for both code and data. The number that comes after ORG must be in hex.
END directive

Another important pseudocode is the END directive. This indicates to the assembler the end of the source (asm) file. The END directive is the last line of the PIC program, meaning that anything after the END directive in the source code is ignored by the assembler.

LIST directive

Unlike ORG and END, which are used by all assemblers, the LIST directive is unique to the PIC assembler. It indicates to the assembler the specific PIC chip for which the program should be assembled. It is used as follows:

LIST P=18F458

The above tells the PIC assembler to assemble the program specifically for the PIC18F458 microcontroller. We use LIST to state the target chip.

#include directive

The #include directive tells the PIC assembler to use the libraries associated with the specific chip for which we are compiling the program.

_config directive

The _config directive tells the assembler the configuration bits for the targeted PIC chip. It is important to use the correct _config directive, because incorrect use may make the chip unusable. The configuration bits are read during power-up of the PIC device and are stored at location 300000H. Microchip has defined the _config directive symbols to ease the configuration. These symbols are located in the .INC file for the device that is being used. See Chapter 8.
	radix directive

We can use the radix directive to indicate whether the numbering system is hexadecimal or decimal. The default is hex if we do not use the radix directive. If we use “radix dec”, the default representation will change to decimal and any unformatted number will be interpreted as decimal rather than hex, as seen before.

Rules for labels in Assembly language

By choosing label names that are meaningful, a programmer can make a program much easier to read and maintain. There are several rules that names must follow. First, each label name must be unique. The names used for labels in Assembly language programming consist of alphabetic letters in both upper and lower case, the digits 0 through 9, and the special characters question mark (?), period (.), at (@), underline (_), and dollar sign ($). The first character of the label must be an alphabetic character. In other words, it cannot be a number. Every assembler has some reserved words that must not be used as labels in the program. Foremost among the reserved words are the mnemonics for the instructions. For example, “MOVWL” and “ADDWL” are reserved because they are instruction mnemonics. In addition to the mnemonics there are some other reserved words. Check your assembler for the list of reserved words.
Review Questions

1. Give three ways for hex data representation in the PIC assembler.
2. Show how to represent decimal 99 in formats of (a) hex, (b) decimal, and (c) binary in the PIC assembler.
3. What is the advantage in using the EQU directive to define a constant value?
4. Show the hex number value used by the following directives:
   (a) ASC_DATA EQU A'4' (b) MY_DATA EQU B'00011111'
5. Give the value in WREG for the following:
   MYCOUNT EQU 15
   MOVlw MYCOUNT
6. Give the value in fileReg 0x20 for the following:
   MYCOUNT EQU 0x95
   MYREG EQU 0x20
   MOVlw MYCOUNT
   MOVWF MYREG
7. Give the value in fileReg 0x63 for the following:
   MYDATA EQU D'12'
   MYREG EQU 0x63
   FACTOR EQU 0x10
   MOVlw MYDATA
   ADDlw FACTOR
   MOVWF MYREG

SECTION 2.6: INTRODUCTION TO PIC ASSEMBLY PROGRAMMING

In this section we discuss Assembly language format and define some widely used terminology associated with Assembly language programming.

While the CPU can work only in binary, it can do so at a very high speed. It is quite tedious and slow for humans, however, to deal with 0s and 1s in order to program the computer. A program that consists of 0s and 1s is called machine language. In the early days of the computer, programmers coded programs in machine language. Although the hexadecimal system was used as a more efficient way to represent binary numbers, the process of working in machine code was still cumbersome for humans. Eventually, Assembly languages were developed, which provided mnemonics for the machine code instructions, plus other features that made programming faster and less prone to error. The term mnemonic is frequently used in computer science and engineering literature to refer to codes and abbreviations that are relatively easy to remember. Assembly language programs must be translated into machine code by a program called an assembler. Assembly language is referred to as a low-level language because it deals directly with the internal structure of the CPU. To program in Assembly language, the programmer must know all the registers of the CPU and the size of each, as well as other details.

Today, one can use many different programming languages, such as BASIC, Pascal, C, C++, Java, and numerous others. These languages are called high-level languages because the programmer does not have to be concerned with the internal details of the CPU. Whereas an assembler is used to translate an
Assembly language program into machine code (sometimes also called object code or opcode for operation code), high-level languages are translated into machine code by a program called a compiler. For instance, to write a program in C, one must use a C compiler to translate the program into machine language. Next we look at PIC Assembly language format.

Structure of Assembly language

An Assembly language program consists of, among other things, a series of lines of Assembly language instructions. An Assembly language instruction consists of a mnemonic, optionally followed by one or two operands. The operands are the data items being manipulated, and the mnemonics are the commands to the CPU, telling it what to do with those items.

An Assembly language program (see Program 2-1) is a series of statements, or lines, which are either Assembly language instructions such as ADDLW and MOVWF, or statements called directives. While instructions tell the CPU what to do, directives (also called pseudo-instructions) give directions to the assembler. For example, in Program 2-1 while the MOVWF and ADDLW instructions are commands to the CPU, ORG and END are directives to the assembler. The directive ORG tells the assembler to place the opcode at memory location 0 while END indicates the end of the source code to the assembler. In other words, one directive is for the start of the program and the other for the end of the program.

An Assembly language instruction consists of four fields:

[label] mnemonic [operands] [;comment]

Brackets indicate that a field is optional and not all lines have them. Brackets should not be typed in. Regarding the above format, the following points should be noted:

1. The label field allows the program to refer to a line of code by name. The label field cannot exceed a certain number of characters. Check your assembler for the rule.
2. The Assembly language mnemonic (instruction) and operand(s) fields together perform the real work of the program and accomplish the tasks for which the program was written. In Assembly language statements such as

```
MOVLW 55H
ADDLW 67H
```

ADDLW and MOVLW are the mnemonics that produce opcodes; the "55H" and "67H" are the operands. Instead of a mnemonic and an operand, these two fields could contain assembler pseudo-instructions, or directives. Remember that directives do not generate any machine code (opcode) and are used only by the assembler, as opposed to instructions that are translated into machine
code (opcode) for the CPU to execute. In Program 2-1 the commands ORG (origin) and END are examples of directives. More of these pseudo-instructions are discussed in future chapters.

3. The comment field begins with a semicolon comment indicator “;”. Comments may be at the end of a line or on a line by themselves. The assembler ignores comments, but they are indispensable to programmers. Although comments are optional, it is recommended that they be used to describe the program in a way that makes it easier for someone else to read and understand.

4. Notice the label “HERE” in the label field in Program 2-1. In the GOTO the PIC is told to stay in this loop indefinitely. If your system has a monitor program you do not need this line and it should be deleted from your program. In Section 2.7 we will see how to create a ready-to-run program.

| ;PIC Assembly Language Program To Add Some Data. |
| ;store SUM in fileReg location 10H. |
| SUM EQU 10H ;RAM loc 10H for SUM |
| ORG 0H ;start at address 0 |
| MOVLW 25H ;WREG = 25 |
| ADDLW 0x34 ;add 34H to WREG |
| ADDLW 11H ;add 11H to WREG |
| ADDLW D'18' ;W = W + 12H = 7CH |
| ADDLW 1CH ;W = W + 1CH = 98H |
| ADDLW B'00000110' ;W = W + 6 = 9EH |
| MOVWF SUM ;save the SUM in loc 10H |
| HERE GOTO HERE ;stay here forever |
| END ;end of asm source file |

Program 2-1: Sample of an Assembly Language Program

Review Questions

1. What is the purpose of pseudo-instructions?
2. ___________ are translated by the assembler into machine code, whereas ___________ are not.
3. True or false. Assembly language is a high-level language.
4. Which of the following instructions produces opcode? List all that do.
   (a) MOVLW 25H (b) ADDLW 12 (c) ORG 2000H (d) GOTO HERE
5. Pseudo-instructions are also called ___________.
6. True or false. Assembler directives are not used by the CPU itself. They are simply a guide to the assembler.
7. In Question 4, which one is an assembler directive?
SECTION 2.7: ASSEMBLING AND LINKING A PIC PROGRAM

Now that the basic form of an Assembly language program has been given, the next question is: How it is created, assembled, and made ready to run? The steps to create an executable Assembly language program (Figure 2-8) are outlined as follows:

1. First we use a text editor to type in a program similar to Program 2-1. In the case of the PIC microcontrollers, we use the MPLAB IDE, which has a text editor, assembler, linker, simulator, and much more all in one software package. It is an excellent development software that supports all the PIC chips and is free. Many editors or word processors are also available that can be used to create or edit the program. Some widely used editors are the MS-DOS EDIT, and Notepad in Windows, which comes with all Microsoft operating systems. Notice that the editor must be able to produce an ASCII file. For assemblers, the file names follow the usual DOS conventions, but the source file has the extension “asm”. The “asm” extension for the source file is used by an assembler in the next step.

2. The “asm” source file containing the program code created in step 1 is fed to the PIC assembler. The assembler converts the instructions into machine code. The assembler will produce an object file and an error file. The extension for

---

![Figure 2-8. Steps to Create a Program](image)
the object file is “o”. The extension for the error file, which contains any syntax errors and their line numbers, is “err”. The error file can be viewed with any text editor.

3. Assemblers require a third step called linking. The link program takes one or more object files and produces a hex file, a list file, a map file, an intermediate object file, and a debug file. The hex file has the extension “hex”, the list file extension is “lst”, the map file extension is “map”, the intermediate object file extension is “out”, and the debug file extension is “cod”. After a successful link, the hex file is ready to be burned into the PIC’s program ROM and is downloaded into the PIC Trainers. See Chapter 8 for more details.

The MPLAB IDE, a Windows-based program, combines steps 2 and 3 into one step after the program has been typed.

More about asm, err, and object files

The asm file is also called the source file and must have the “asm” extension. As mentioned earlier, this file is created with a text editor such as MS-DOS EDIT or Windows Notepad. Many assemblers come with a text editor. The assembler converts the asm file’s Assembly language instructions into machine language and provides the o (object) file. The PIC assembler produces the object and error files. The object file as mentioned earlier, has an “o” as it extension. In modular programming, we use the linker to link many object files together to create a ready-to-burn hex file as we will see in Chapter 6. But before we can link a program to create a ready-to-run program, we must make sure that it is error free. The PIC assembler provides us the error file with the extension of “err” and this is the file we examine to see the nature of syntax errors. The linker will not link the program until all the syntax errors are fixed. We can print the error file or use Notepad to examine the nature of the errors. Then we go back to the asm file and correct all the errors before we assemble it again. A sample of an error file is shown on the next page.

```
;PIC Assembly Language Program To Add Some Data.
;store sum in fileReg location 10H.

SUM EQU 10H ;RAM loc 10H for sum

ORG 0H ;start at address 0
MOVLW 25H ;WREG = 25
ADDLW 0x34 ;add 34H to WREG
ADDLW 11H ;add 11H to WREG
ADDLW D'18' ;W = W + 12H = 7CH
ADDLW 1CH ;W = W + 1CH = 98H
ADDLW B'00000110' ;W = W + 6 = 9EH
MOVLW SUM ;save the sum in loc 10H
GOTO HERE ;stay here forever
END ;end of asm source file
```

Program 2-1: Sample of a PIC Assembly Source Code (asm file)
Warning [207] C:\MDEPIC\EXAMPLE 2-1.ASM 6 : Found label after column 1. (R4)
Warning [207] C:\MDEPIC\EXAMPLE 2-1.ASM 13 : Found label after column 1. (movle)
Error [122] C:\MDEPIC\EXAMPLE 2-1.ASM 13 : Illegal opcode (d)
Warning [207] C:\MDEPIC\EXAMPLE 2-1.ASM 17 : Found label after column 1. (DEC)
Error [122] C:\MDEPIC\EXAMPLE 2-1.ASM 17 : Illegal opcode (COUNT)
Warning [203] C:\MDEPIC\EXAMPLE 2-1.ASM 20 : Found opcode in column 1. (movwf)
Warning [207] C:\MDEPIC\EXAMPLE 2-1.ASM 21 : Found label after column 1. (addl)
Error [108] C:\MDEPIC\EXAMPLE 2-1.ASM 21 : Illegal character (0)
Error [116] C:\MDEPIC\EXAMPLE 2-1.ASM 29 : Address label duplicated or different in second pass (AGAIN)

Program 2-1: Sample of a PIC Error (err file)

“lst” and “map” files

The lst (list) and map files are very useful to the programmer. The list shows the binary and source code. The map file shows the memory layout of used and unused memory locations. These files can be accessed by an editor such as Notepad and displayed on the monitor, or sent to the printer to get a hard copy. The programmer uses the list and map files to ensure correct system design.

<table>
<thead>
<tr>
<th>LOC</th>
<th>OBJECT CODE LINE</th>
<th>SOURCE TEXT</th>
<th>VALUE</th>
</tr>
</thead>
<tbody>
<tr>
<td>00001</td>
<td></td>
<td>;PIC Asm Language Program To Add Some Data</td>
<td></td>
</tr>
<tr>
<td>00002</td>
<td>;store SUM in fileReg location 10H</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000003</td>
<td>SUM EQU 10H</td>
<td>;RAM loc 10H for Sum</td>
<td></td>
</tr>
<tr>
<td>00005</td>
<td></td>
<td>;start at address 0</td>
<td></td>
</tr>
<tr>
<td>000000</td>
<td>ORG 0H</td>
<td>;add 34H to WREG</td>
<td></td>
</tr>
<tr>
<td>000000</td>
<td>MOVW 25H</td>
<td>;add 34H to WREG</td>
<td></td>
</tr>
<tr>
<td>000002</td>
<td>ADDLW 0x34</td>
<td>;add 11H to WREG</td>
<td></td>
</tr>
<tr>
<td>000003</td>
<td>ADDLW 11H</td>
<td>;add 11H to WREG</td>
<td></td>
</tr>
<tr>
<td>000006</td>
<td>ADDLW D'18'</td>
<td>;W = W + 12H = 7CH</td>
<td></td>
</tr>
<tr>
<td>000008</td>
<td>ADDLW 1CH</td>
<td>;W = W + 1CH = 98H</td>
<td></td>
</tr>
<tr>
<td>00000A</td>
<td>ADDLW B'00000110'</td>
<td>;W = W + 6 = 9EH</td>
<td></td>
</tr>
<tr>
<td>00000C</td>
<td>MOVWF SUM</td>
<td>;save the SUM in loc 10H</td>
<td></td>
</tr>
<tr>
<td>00000E</td>
<td>HERE GOTO HERE</td>
<td>;stay here forever</td>
<td></td>
</tr>
<tr>
<td>000015</td>
<td>END</td>
<td>;end of asm source file</td>
<td></td>
</tr>
</tbody>
</table>

Program 2-1: List File

Review Questions

1. True or false. The MPLAB, MS-DOS Edit, and Windows Notepad text editors all produce an ASCII file.
2. True or false. The extension for the source file is “asm”.
3. Which of the following files can be produced by the text editor?
   (a) myprog.asm (b) myprog.o (c) myprog.hex (d) myprog.lst (e) myprog.err
4. Which of the following files is produced by an assembler?
   (a) myprog.asm (b) myprog.o (c) myprog.hex (d) myprog.lst (e) myprog.err
5. Which of the following files lists syntax errors?
   (a) myprog.asm (b) myprog.o (c) myprog.hex (d) myprog.lst (e) myprog.err
SECTION 2.8: THE PROGRAM COUNTER AND PROGRAM ROM SPACE IN THE PIC

In this section we discuss the role of the program counter (PC) in executing a program and show how the code is fetched from ROM and executed. We will also discuss the program (code) ROM space for various PIC family members. Finally, we examine the Harvard architecture of the PIC18.

Program counter in the PIC

Another important register in the PIC microcontroller is the PC (program counter). The program counter is used by the CPU to point to the address of the next instruction to be executed. As the CPU fetches the opcode from the program ROM, the program counter is incremented automatically to point to the next instruction. The wider the program counter, more the memory locations a CPU can access. That means that a 14-bit program counter can access a maximum of 16K (2\(^{14}\) = 16K) of code from addresses 0000–3FFFH. The PIC family 16F has 14-bit program counters, while the program counter in PIC12F is 12-bit. In the case of a 16-bit program counter, the code space is 64K (2\(^{16}\) = 64K), which occupies the 0000–FFFFH address range. The 8051 microcontrollers have a 16-bit program counter. The program counter in the PIC18 family is 21-bit. This means that the PIC18 family can access program addresses 000000 to FFFFFFH, a total of 2M of code. However, not all members of the PIC18 family have the entire 2M (2\(^{21}\) = 2M) of on-chip ROM installed. See Table 2-6. The 14-bit program counter in the PIC16C family had imposed the maximum code size of 16K. To overcome this major limitation, PIC designers had to introduce the tedious job of page switching in the later members of the PIC16 family. They learned their lessons and solved the problem for the PIC18 family by expanding the program counter to 21-bit for that family. See Figure 2-9. The 2M code space is plenty of space for many years to come. The data in Table 2-6 is from the Microchip web site.

ROM memory map in the PIC18 family

As we just saw, some family members have only a few K of on-chip ROM (PIC18F2220) and some, such as the PIC18F6680, have 64K of ROM. PIC18F458 has 32K of on-chip ROM. The point to remember is that no member of the PIC family can access more than 2M of opcode because the program counter in the PIC is 21 bits wide (000000 to FFFFFF address range). It must be noted that while the first location of program ROM inside the PIC has the address of 000000, the last location can be different depending on the size of the ROM on the chip. (See Figure 2-10.) Among the PIC18 family members, the PIC18F2220 has 4K of on-chip ROM. This 4K ROM memory has memory addresses of 00000 to 000FFFH. Therefore, the first location of on-chip

<table>
<thead>
<tr>
<th>0</th>
<th>0</th>
<th>0</th>
<th>PCU</th>
<th>PCH</th>
<th>PCL</th>
</tr>
</thead>
<tbody>
<tr>
<td>8-bit</td>
<td>8-bit</td>
<td>8-bit</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>21-bit PC (Program Counter)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Figure 2-9. Program Counter in PIC18
ROM of this PIC has an address of 000000 and the last location has the address of 0FFFH. Look at Example 2-11 to see how this is computed.

### Table 2-6: PIC18 On-chip ROM Size and Address Space

<table>
<thead>
<tr>
<th>On-Chip Code ROM (Bytes)</th>
<th>Code Address Range (Hex)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F2220 4K</td>
<td>00000–00FFF</td>
</tr>
<tr>
<td>PIC18F2410 16K</td>
<td>00000–03FFF</td>
</tr>
<tr>
<td>PIC18F458 32K</td>
<td>00000–07FFF</td>
</tr>
<tr>
<td>PIC18F6680 64K</td>
<td>00000–0FFFF</td>
</tr>
<tr>
<td>PIC18F8722 128K</td>
<td>00000–1FFFF</td>
</tr>
</tbody>
</table>

### Example 2-11

Find the ROM memory address of each of the following PIC chips:
(a) PIC18F2220 with 4 KB
(b) PIC18F2410 with 16 KB
(c) PIC18F458 with 32 KB

**Solution:**

(a) With 4K of on-chip ROM memory space, we have 4096 bytes \( (4 \times 1024 = 4096) \). This maps to address locations of 0000 to 0FFFH. Notice that 0 is always the first location.

(b) With 16K of on-chip ROM memory space, we have 16,384 bytes \( (16 \times 1024 = 16,384) \), which gives 0000–3FFFH.

(c) With 32K we have 32,768 bytes \( (32 \times 1024 = 32,768) \). Converting 32,768 to hex, we get 8000H; therefore, the memory space is 0000 to 7FFFH.

### Figure 2-10. PIC18 On-Chip Program (code) ROM Address Range
Where the PIC wakes up when it is powered up

One question that we must ask about any microcontroller (or microprocessor) is: At what address does the CPU wake up when power is applied? Each microprocessor is different. In the case of the PIC family, that is, all members regardless of the family and variation, the microcontroller wakes up at memory address 0000 when it is powered up. By powering up we mean applying $V_{CC}$ to the RESET pin as discussed in Chapter 8. In other words, when the PIC is powered up, the PC (program counter) has the value of 00000 in it. This means that it expects the first opcode to be stored at ROM address 00000H. For this reason, in the PIC system, the first opcode must be burned into memory location 00000H of program ROM because this is where it looks for the first instruction when it is booted. We achieve this by using the ORG statement in the source program as shown earlier. Next we discuss the step-by-step action of the program counter in fetching and executing a sample program.

Placing code in program ROM

To get a better understanding of the role of the program counter in fetching and executing a program, we examine the action of the program counter as each instruction is fetched and executed. First, we examine once more the list file of the sample program and show how the code is placed in the ROM of PIC chip. As we can see, the opcode and operand for each instruction are listed on the left side of the list file.

After the program is burned into ROM of a PIC family member such as PIC18F452 or PIC18F458, the opcode and operand are placed in ROM memory locations starting at 0000 as shown in the Program 2-1 list file.

The list shows that address 0000 contains 0E, which is the opcode for moving a value into WREG, and address 0001 contains the operand (in this case 25H) to be moved to WREG. Therefore, the instruction “MOVLW 25H” has a machine code of “0E25”, where 0E is the opcode and 25 is the operand. Similarly, the machine code “0F34” is located in memory locations 0002 and 0003 and represents the opcode and the operand for the instruction “ADDLW 34H”. In the same way, machine code “0F11” is located in memory locations 0004 and 0005 and rep-
resents the opcode and the operand for the instruction “ADDLW 11”. The memory location 0006 has the opcode of 0F, which is the opcode for the instruction “MOVLW” and memory location 0007 has the content 12, which is the operand for the decimal 18 in the ADDLW D'18' instruction. The opcode for instruction “ADDLW 1CH” is located at address 0008 and the operand 1CH at address 0009. The memory locations 000A and 000B have the opcode and operand for the ADDLW B'00000110' instruction. The opcode for instruction “MOVWF SUM” is located at address 000C and its address of 10H at address 000D. The opcode for “GOTO HERE” and its target address are located in locations 000E, F, 10, and 11. While all the instructions in this program are 2-byte instructions, the GOTO instruction is a 4-byte instruction. The reasons are explained at the end of this section.

<table>
<thead>
<tr>
<th>LOC</th>
<th>OBJECT CODE LINE SOURCE TEXT</th>
<th>VALUE</th>
</tr>
</thead>
<tbody>
<tr>
<td>00001</td>
<td>;PIC Asm Language Program To Add Some Data</td>
<td></td>
</tr>
<tr>
<td>00002</td>
<td>;store sum in fileReg location 10H</td>
<td></td>
</tr>
<tr>
<td>00003</td>
<td>SUM EQU 10H</td>
<td></td>
</tr>
<tr>
<td>00004</td>
<td>;RAM loc 10H for sum</td>
<td></td>
</tr>
<tr>
<td>00005</td>
<td>000000 ORG 0H</td>
<td></td>
</tr>
<tr>
<td>00002</td>
<td>MOVLW 25H</td>
<td></td>
</tr>
<tr>
<td>00004</td>
<td>ADDLW 0x34</td>
<td></td>
</tr>
<tr>
<td>00005</td>
<td>ADDLW 11H</td>
<td></td>
</tr>
<tr>
<td>00006</td>
<td>ADDLW D'18'</td>
<td></td>
</tr>
<tr>
<td>00008</td>
<td>ADDLW 1CH</td>
<td></td>
</tr>
<tr>
<td>0000A</td>
<td>ADDLW B'00000110'</td>
<td></td>
</tr>
<tr>
<td>0000C</td>
<td>MOVWF SUM</td>
<td></td>
</tr>
<tr>
<td>0000E</td>
<td>HERE GOTO HERE</td>
<td></td>
</tr>
<tr>
<td>00015</td>
<td>END</td>
<td></td>
</tr>
</tbody>
</table>

Program 2-1: List File

<table>
<thead>
<tr>
<th>ROM Address</th>
<th>Machine Language</th>
<th>Assembly Language</th>
</tr>
</thead>
<tbody>
<tr>
<td>00000</td>
<td>0E25</td>
<td>MOVLW 25H</td>
</tr>
<tr>
<td>00002</td>
<td>0F34</td>
<td>ADDLW 34</td>
</tr>
<tr>
<td>00004</td>
<td>0F11</td>
<td>ADDLW 11H</td>
</tr>
<tr>
<td>00006</td>
<td>0F12</td>
<td>ADDLW D'18'</td>
</tr>
<tr>
<td>00008</td>
<td>0F1C</td>
<td>ADDLW 1CH</td>
</tr>
<tr>
<td>0000A</td>
<td>0F06</td>
<td>ADDLW B'00000110'</td>
</tr>
<tr>
<td>0000C</td>
<td>6E10</td>
<td>MOVWF SUM</td>
</tr>
<tr>
<td>0000E</td>
<td>EF07 F000</td>
<td>HERE GOTO HERE</td>
</tr>
</tbody>
</table>
Executing a program byte by byte

Assuming that the above program is burned into the ROM of a PIC18 chip, the following is a step-by-step description of the action of the PIC upon applying power to it:

1. When the PIC is powered up, the PC (program counter) has 00000 and starts to fetch the first opcode from location 000000 of the program ROM. In the case of the above program the first opcode is 0E, which is the code for moving an operand to WREG. Upon executing the opcode, the CPU places the value of 25 in WREG. Now one instruction is finished. Then the program counter is incremented to point to 000002 (PC = 000002), which contains opcode 0F, the opcode for the instruction “ADDLW 34H”.

2. Upon executing the opcode 0F, the value 34H is added to WREG. Then the program counter is incremented to 0004.

3. ROM location 0004 has the opcode for instruction “ADDLW 11H”. This instruction is executed and now PC = 0006.

4. This process goes on until all the instructions up to “MOVWF SUM” are fetched and executed. Notice that all the above instructions are 2-byte instructions; that is, each one takes two memory locations.

5. Now PC = 000E points to the next instruction, which is “GOTO HERE”. This is a 4-byte instruction. It takes ROM addresses of 0E, 0F, 10, and 11. After the execution of this instruction, PC = 000E. This keeps the program in an infinite loop. If your PIC Trainer has a monitor program you do not have to use the GOTO instruction, and the program will go back to the monitor program. The fact that the program counter points at the next instruction to be executed explains why some microprocessors (notably the x86) call the program counter the instruction pointer.

ROM width in the PIC18

A microprocessor’s memory that holds code is byte-addressable and under the control of the program counter, as we have seen so far in this section. That means that each location of the address space holds only one byte. If we have 16 address lines, it will give us 216 locations, which is 64K of memory space with an address map of 0000–FFFFH. CPUs with 8-bit data will fetch one byte at a time. This was the case in the first IBM PC and Apple computers. To bring in more information (code or data) into the CPU we can increase the width of the data bus to 16 bits. That is what IBM did to the PC AT in 1984. To increase performance even further, Intel Corporation increased the data bus width to 32 bits for the 386
and 64 bits for the Pentium. In a sense, the data bus is like traffic lanes on the highway where each lane is 8 bits wide. The more lanes, the more information we can bring into the CPU for processing. For the PIC18, the internal data bus between the code ROM and the CPU is 16 bits, as shown in Figure 2-12. Therefore, the 64K ROM space is shown as 32K \times 16 using a 16-bit word data size. The same rule applies to the entire program address space of PIC18, which is 2M, organized as 1M \times 16. The widening of the data path between the program ROM and the CPU is another way in which the PIC designers increased the processing power of the PIC18 family. Another reason to make the code ROM 16 bits wide is to match it with the instruction width of the PIC18 because the vast majority of the instructions are 2-byte instructions. This way, the CPU brings in an instruction from ROM everytime it makes a trip to the program ROM. That will make instruction fetch a single cycle, as we will see in the next chapter when instruction timing is discussed.

The PIC18 designers have made all instructions either 2-byte or 4-byte; there are no 1-byte or 3-byte instructions, as is the case with the x86 and 8051 chips. This is part of the RISC architectural philosophy, which we will study in the next section. It must be noted that not all the program ROMs in the PIC microcontrollers have 16-bit width. The PIC16 ROM width is 14-bit, while the PIC12 has a 12-bit width. It must also be noted that the data memory SRAM for the file register in the PIC microcontroller is 8-bit, and just like program ROM, it is byte addressable.
Little endian vs. big endian war

Examine the placing of the code in the PIC18 ROM, shown in Figure 2-13. The low byte goes to the low memory location and the high byte goes to the high memory address. This convention is called little endian to contrast it with big endian. The origin of the terms big endian and little endian is from a Gulliver’s Travels story about how an egg should be opened: from the big end or the little end. In the big endian method, the high byte goes to the low address, whereas in the little endian method, the high byte goes to the high address and the low byte to the low address. All Intel microprocessors and many microcomputers, notably the Digital VAX, use the little endian convention. Freescale (Motorola) microprocessors (used in the Macintosh), along with some mainframes, use big endian. The difference might seem as trivial as whether to break an egg from the big end or the little end, but it is a nuisance in converting software from one camp to be run on a computer of the other camp. Some microprocessors, such as the PowerPC from IBM/Freescale (Motorola), let the software designer choose little endian or big endian convention.

Harvard architecture in the PIC

Every microprocessor must have memory space to store program (code) and data. As we have seen so far, the PIC is no exception with its code ROM space and data RAM (file register) space. While code provides instructions to the CPU, the data provides the information to be processed. The CPU uses buses (wire traces) to access the code ROM and data RAM memory spaces. The early computers used the same bus for accessing both the code and data. Such an architecture is commonly referred to as von Neumann (Princeton) architecture. That means for von Neumann computers, the process of accessing the code or data could cause them to get in each other’s way and slow down the processing speed of the CPU, because each had to wait for the other to finish fetching. To speed up the process of program execution, some CPUs use what is called Harvard architecture. In Harvard architecture, we have separate buses for the code and data memory. That means that we need four sets of buses: (1) a set of data buses for
carrying data into and out of the CPU, (2) a set of address buses for accessing the data, (3) a set of data buses for carrying code into the CPU, and (4) an address bus for accessing the code. See Figure 2-14. This is easy to implement inside an IC chip such as a microcontroller where both ROM code and data RAM are internal (on-chip) and distances are on the micron and millimeter scale. But to implement Harvard architecture for systems such as x86 IBM PC-type computers is very expensive because the RAM and ROM that hold code and data are external to the CPU. Separate wire traces for data and code on the motherboard will make the board large and expensive. For example, for a Pentium microprocessor with a 64-bit data bus and a 32-bit address bus we will need about 100 wire traces on the mother board if it is von Neumann architecture (96 for address and data, plus a few others for control signals of read and write and so on). But the number of wire traces will double to 200 if we use Harvard architecture. Harvard architecture will also necessitate a large number of pins coming out of the microprocessor itself. For this reason you do not see Harvard architecture implemented in the world of PCs and workstations. This is also the reason that microcontrollers such as PIC use Harvard architecture internally, but they still use von Neumann architecture if they need external memory for code and data space. The von Neumann architecture was developed at Princeton University, while the Harvard architecture was the work of Harvard University.

Figure 2-14. von Neumann vs. Harvard Architecture
Instruction size of the PIC18

Recall that PIC18 program memory is byte-addressable, and the instructions are either 2-byte or 4-byte. Almost all the instructions in the PIC18 are 2-byte instructions. The exceptions are MOVFF, GOTO, and a few others. Next we explore the instruction size and formation for a few of the instructions we have used in this chapter. This should give you some insights into the instructions of the PIC18.

MOVLW instruction formation

The MOVLW is a 2-byte (16-bit) instruction. Of the 16 bits, the first 8 bits are set aside for the opcode and the other 8 bits are used for the literal value of 00 to FFH. This is shown below.

\[
\begin{array}{c}
0000 \quad 1110 \quad \text{k}k\text{k}k \\
0 \leq k \leq \text{FF}
\end{array}
\]

ADDLW instruction formation

The ADDLW is a 2-byte (16-bit) instruction. Of the 16 bits, the first 8 bits are set aside for the opcode and the other 8 bits are used for the literal value of 00 to FFH. This is shown below.

\[
\begin{array}{c}
0000 \quad 1111 \quad \text{k}k\text{k}k \\
0 \leq k \leq \text{FF}
\end{array}
\]

MOVWF instruction formation

The MOVWF is a 2-byte (16-bit) instruction. Of the 16 bits, the first 8 bits are set aside for the opcode and the other 8 bits are used for the location of the file register in the data RAM. The LSB bit of the opcode is designated by the letter a to signify the access from the access bank or the other bank in the 4096 location. If \(a = 0\), the fileReg is in the access bank. If \(a = 1\), then we have to use bank switching, which is covered in Chapter 6. This is shown below.

\[
\begin{array}{c}
0110 \quad 111a \quad \text{f}f\text{f}f \\
0 \leq f \leq \text{FF}
\end{array}
\]

\(a = 0\) : access bank is used.
\(a = 1\) : access bank is specified by the BSR register.

See Chapter 6.
MOVFF instruction formation

The MOVFF is a 4-byte (32-bit) instruction. Of the 32 bits, the first 16 bits are set aside for the opcode and the address of the source fileReg and the other 16 bits are used for the opcode and the address of the destination. This is shown below.

<table>
<thead>
<tr>
<th>1100</th>
<th>ssss</th>
<th>ssss</th>
<th>ssss</th>
<th>Source (fs)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1111</td>
<td>dddd</td>
<td>dddd</td>
<td>dddd</td>
<td>Destination (fd)</td>
</tr>
</tbody>
</table>

0 ≤ fs ≤ FFF
0 ≤ fd ≤ FFF

Notice that for both the source and destination parts of the instruction, 12 bits are used for the file register address of the PIC18. The 12 bits cover the entire range of the addresses 000–FFFH for the file register, which has 4096 bytes (4K) of data RAM space. That means that MOVFF can move the contents of any location in the file register to any other location directly. This is done without going through the WREG, as we saw in Section 2.3.

GOTO instruction formation

The GOTO is a 4-byte (32-bit) instruction. Of the 32 bits, only 12 bits are set aside for the opcode and the rest (20 bits) are used for the target address of the GOTO. This is shown below.

<table>
<thead>
<tr>
<th>1110</th>
<th>1111</th>
<th>k7kkk</th>
<th>kkkk0</th>
</tr>
</thead>
<tbody>
<tr>
<td>1111</td>
<td>k19kkk</td>
<td>kkkk</td>
<td>kkkk8</td>
</tr>
</tbody>
</table>

0 ≤ k ≤ FFFFF

However, the 20-bit address gives us only 1M of address space and the PIC18 has 2M of ROM space. This is solved by making the least-significant bit (LSB) of the GOTO instruction 0, as shown below.

Setting the LSB of the target address to zero will make sure that the target address is an even address. As we saw in the last section, that is exactly what we want because all the instructions are either 2-byte or 4-byte. This should also avoid landing at the middle of an instruction.
Coming from other microprocessors to the PIC18

If you have a background in programming other microprocessors/microcontrollers, making the transition from these devices to the PIC18 can be easier if we remember some facts about the PIC18. They are:

1. The access bank section of the register of the PIC18 with its address range of 00–7FH can be viewed as large a number of registers, except they do not have names like other processors. We can assign any register names we want, however, as long as we are not using any of the reserved names used by SFRs, WREG, and so on. Here is an example if we are used to the 8051 or some other RISC processor:

\[
\begin{align*}
R0 & \text{ EQU 0} \\
R1 & \text{ EQU 1} \\
R2 & \text{ EQU 2} \\
R3 & \text{ EQU 3} \\
\ldots & \ldots \\
\end{align*}
\]

Or look at the following for the x86:

\[
\begin{align*}
BL & \text{ EQU 0} \\
BH & \text{ EQU 1} \\
CL & \text{ EQU 2} \\
CH & \text{ EQU 4} \\
DL & \text{ EQU 5} \\
DH & \text{ EQU 6} \\
\end{align*}
\]

In both of the above we can use any fileReg address of 00–7FH.

2. WREG is exactly like the accumulator in other microprocessors. It must be involved in all the arithmetic and logic operations.

3. To move data to locations in the file register or SFR, we must first move it to WREG. As was shown, we use the MOVLW instruction first to load the value into the WREG and then use MOVWF to move it to a desired location in the fileReg. In other words, no value can be moved directly to SFR or fileReg.

Review Questions

1. In the PIC18, the program counter is _____ bits wide.
2. True or false. Every member of the PIC18 family, regardless of the program ROM size, wakes up at memory 0000H when it is powered up.
3. At what ROM location do we store the first opcode of a PIC18 program?
4. The instruction “MOVLW 44H” is a _____-byte instruction.
5. What is the ROM address space for the PIC18F458?
6. The instruction “GOTO label” is a _____-byte instruction.
7. True or false. All the instructions in the PIC18 are 2- or 4-byte instructions.
SECTION 2.9: RISC ARCHITECTURE IN THE PIC

There are three ways available to microprocessor designers to increase the processing power of the CPU:

1. Increase the clock frequency of the chip. One drawback of this method is that the higher the frequency, the more power and heat dissipation. Power and heat dissipation is especially a problem for hand-held devices.
2. Use Harvard architecture by increasing the number of buses to bring more information (code and data) into the CPU to be processed. While in the case of x86 and other general purpose microprocessors this architecture is very expensive and unrealistic, in today’s microcontrollers this is not a problem. As we saw in the last section, the PIC18 has Harvard architecture.
3. Change the internal architecture of the CPU and use what is called RISC architecture.

Microchip used all three methods to increase the processing power of the PIC18 microcontrollers. In this section we discuss the merits of RISC architecture and examine how the PIC18 microcontrollers have adapted it.

RISC architecture

In the early 1980s, a controversy broke out in the computer design community, but unlike most controversies, it did not go away. Since the 1960s, in all mainframes and minicomputers, designers put as many instructions as they could think of into the CPU. Some of these instructions performed complex tasks. An example is adding data memory locations and storing the sum into memory. Naturally, microprocessor designers followed the lead of minicomputer and mainframe designers. Because these microprocessors used such a large number of instructions and many of them performed highly complex activities, they came to be known as CISC (complex instruction set computer). According to several studies in the 1970s, many of these complex instructions etched into the brain of the CPU were never used by programmers and compilers. The huge cost of implementing a large number of instructions (some of them complex) into the microprocessor, plus the fact that a good portion of the transistors on the chip are used by the instruction decoder, made some designers think of simplifying and reducing the number of instructions. As this concept developed, the resulting processors came to be known as RISC (reduced instruction set computer).

Features of RISC

The following are some of the features of RISC as implemented by the PIC18 microcontroller.

Feature 1

RISC processors have a fixed instruction size. In a CISC microcontroller such as the 8051, instructions can be 1, 2, or even 3 bytes. For example, look at
the following instructions in the 8051:

CLR C ;Clear Carry flag , a 1-byte instruction
ADD Accumulator, #mybyte ;a 2-byte instruction
LJMP target_address ;a 3-byte instruction

This variable instruction size makes the task of the instruction decoder very difficult because the size of the incoming instruction is never known. In a RISC architecture, the size of all instructions is fixed. Therefore, the CPU can decode the instructions quickly. This is like a bricklayer working with bricks of the same size as opposed to using bricks of variable sizes. Of course, it is much more efficient to use bricks of the same size. In the last section we saw how the PIC18 uses 2-byte instructions with very few 4-byte instructions.

**Feature 2**

One of the major characteristics of RISC architecture is a large number of registers. All RISC architectures have at least 32 registers. Of these 32 registers, only a few are assigned to a dedicated function. One advantage of a large number of registers is that it avoids the need for a large stack to store parameters. Although a stack can be implemented on a RISC processor, it is not as essential as in CISC because so many registers are available. In the PIC microcontrollers the use of a 256-byte bank for the file register satisfies this RISC feature. The stack for the PIC18 is covered in the next chapter.

**Feature 3**

RISC processors have a small instruction set. RISC processors have only the basic instructions such as ADD, SUB, MUL, LOAD, STORE, AND, OR, EXOR, CALL, JUMP, and so on. The limited number of instructions is one of the criticisms leveled at the RISC processor because it makes the job of Assembly language programmers much more tedious and difficult compared to CISC Assembly language programming. This is one reason that RISC is used more commonly in high-level language environments such as the C programming language rather than Assembly language environments. It is interesting to note that some defenders of CISC have called it “complete instruction set computer” instead of “complex instruction set computer” because it has a complete set of every kind of instruction. How many of these instructions are used and how often is another matter. The limited number of instructions in RISC leads to programs that are large. Although these programs can use more memory, this is not a problem because memory is cheap. Before the advent of semiconductor memory in the 1960s, however, CISC designers had to pack as much action as possible into a single instruction to get the maximum bang for their buck. In the PIC16 we have around 35 instructions, while the PIC18 has 75 instructions. We will examine more of the instruction set for the PIC18 in future chapters.

**Feature 4**

At this point, one might ask, with all the difficulties associated with RISC programming, what is the gain? The most important characteristic of the RISC processor is that more than 95% of instructions are executed with only one clock cycle, in contrast to CISC instructions. Even some of the 5% of the RISC instruct-
tions that are executed with two clock cycles can be executed with one clock cycle by juggling instructions around (code scheduling). Code scheduling is most often the job of the compiler. We will examine the instruction cycle time and pipelining of the PIC18 in Chapter 3.

**Feature 5**

RISC processors have separate buses for data and code. In all the x86 processors, like all other CISC computers, there is one set of buses for the address (e.g., A0–A24 in the 80286) and another set of buses for data (e.g., D0–D15 in the 80286) carrying opcodes and operands in and out of the CPU. To access any section of memory, regardless of whether it contains code or data operands, the same address bus and data bus are used. In RISC processors, there are four sets of buses: (1) a set of data buses for carrying data (operands) in and out of the CPU, (2) a set of address buses for accessing the data, (3) a set of buses to carry the opcodes, and (4) a set of address buses to access the opcodes. The use of separate buses for code and data operands is commonly referred to as Harvard architecture. In the last section we examined the Harvard architecture of the PIC18.

**Feature 6**

Because CISC has such a large number of instructions, each with so many different addressing modes, microinstructions (microcode) are used to implement them. The implementation of microinstructions inside the CPU takes more than 40–60% of transistors in many CISC processors. In the case of RISC, however, due to the small set of instructions, they are implemented using the hardware method. Hardwiring of RISC instructions takes no more than 10% of the transistors.

**Feature 7**

RISC uses load/store architecture. In CISC microprocessors, data can be manipulated while it is still in memory. For example, in instructions such as “ADD Reg, Memory”, the microprocessor must bring the contents of the external memory location into the CPU, add it to the contents of the register, then move the result back to the external memory location. The problem is there might be a delay in accessing the data from external memory. Then the whole process would be stalled, preventing other instructions from proceeding in the pipeline. In RISC, designers did away with these kinds of instructions. In RISC, instructions can only load from external memory into registers or store registers into external memory locations. There is no direct way of doing arithmetic and logic operations between a register and the contents of external memory locations. All these instructions must be performed by first bringing both operands into the registers inside the CPU, then performing the arithmetic or logic operation, and then sending the result back to memory. This idea was first implemented by the Cray 1 supercomputer in 1976 and is commonly referred to as load/store architecture. In the last section, we saw that the arithmetic and logic operations are between the fileReg (an internal memory) and WREG but none involve a ROM location and a fileReg location. For example, there is no “ADDW ROM-Loc” instruction in PIC18.
In concluding this discussion of RISC processors, it is interesting to note that RISC technology was explored by the scientists in IBM in the mid-1970s, but it was David Patterson of the University of California at Berkeley who in 1980 brought the merits of RISC concepts to the attention of computer scientists. It must also be noted that in recent years CISC processors such as the Pentium have used some of the RISC features in their design. This was the only way they could enhance the processing power of the x86 processors and stay competitive. Of course, they had to use lots of transistors to do the job, because they had to deal with all the CISC instructions of the 8086/286/386 processors and the legacy software of DOS.

Review Questions

1. What do RISC and CISC stand for?
2. True or false. The CISC architecture executes the vast majority of its instructions in 2, 3, or more clock cycles, while RISC executes them in one clock.
3. RISC processors normally have a _____ (large, small) number of general-purpose registers.
4. True or false. Instructions such as “ADD WREG, ROMmemory” do not exist in RISC microcontrollers such as the PIC18.
5. How many instructions does the PIC18 have? Does it qualify as RISC?
6. True or false. While CISC instructions are of variable sizes, RISC instructions are all the same size.
7. Which of the following operations do not exist for the ADD instruction in RISC?
   (a) register to register  (b) immediate to register  (c) memory to memory
8. True or false. Harvard architecture uses the same address and data buses to fetch both code and data.

SECTION 2.10: VIEWING REGISTER AND MEMORY WITH MPLAB SIMULATOR

The PIC microcontroller has one of the best tools and support systems, many of them free or inexpensive. MPLAB is an assembler, linker, and simulator provided for free by Microchip Corporation and can be downloaded from the www.microchip.com web site. See http://www.MicroDigitalEd.com for tutorials on how to use the MPLAB assembler and simulators.

Many assemblers and C compilers come with a simulator. Simulators allow us to view the contents of registers and memory after executing each instruction (single-stepping). It is strongly recommended to use a simulator to single-step some of the programs in this chapter and future chapters. Single-stepping a program with a simulator gives us a deeper understanding of microcontroller architecture, in addition to the fact that we can use it to find the nature of error in our programs. Figures 2-15 through 2-17 show screen-shots for PIC simulators from MPLAB.
### Special Function Registers

<table>
<thead>
<tr>
<th>Address</th>
<th>SFR Name</th>
<th>Hex</th>
<th>Decimal</th>
<th>Binary</th>
<th>Char</th>
</tr>
</thead>
<tbody>
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<td>00000000</td>
<td></td>
</tr>
<tr>
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<td>PORTB</td>
<td>00</td>
<td>0</td>
<td>00000000</td>
<td></td>
</tr>
<tr>
<td>0F82</td>
<td>PORTC</td>
<td>00</td>
<td>0</td>
<td>00000000</td>
<td></td>
</tr>
<tr>
<td>0F83</td>
<td>PORTD</td>
<td>00</td>
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</tr>
<tr>
<td>0F84</td>
<td>PORTE</td>
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<td>0F89</td>
<td>LATA</td>
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<td>0</td>
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**Figure 2-15. SFR Window in MPLAB Simulator**

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**Figure 2-16. File Register (Data RAM) Window in MPLAB Simulator**

### Program Memory: 1

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<th>Line</th>
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<th>Disassembly</th>
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<td>0000</td>
<td>G0A</td>
<td>MOVVL 0x3</td>
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<tr>
<td>2</td>
<td>0002</td>
<td>E25</td>
<td>MOVWF 0x25, ACCESS</td>
</tr>
<tr>
<td>3</td>
<td>0004</td>
<td>E00</td>
<td>MOVW 0</td>
</tr>
<tr>
<td>4</td>
<td>0006</td>
<td>F03</td>
<td>ADDLW 0x3</td>
</tr>
<tr>
<td>5</td>
<td>0008</td>
<td>E25</td>
<td>DECF 0x25, F, ACCESS</td>
</tr>
<tr>
<td>6</td>
<td>000A</td>
<td>E1FD</td>
<td>BNZ 0x6</td>
</tr>
<tr>
<td>7</td>
<td>000C</td>
<td>E81</td>
<td>MOVWF 0xF81, ACCESS</td>
</tr>
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</table>

**Figure 2-17. Program (Code) ROM Window in MPLAB Simulator**
See the following web site for a tutorial on using MPLAB:

http://www.MicroDigitalEd.com

SUMMARY

This chapter began with an exploration of the major registers of the PIC, including WREG, SFRs, and general-purpose data RAM, and the program counter. The use of these registers was demonstrated in the context of programming examples. The process of creating an Assembly language program was described from writing the source file, to assembling it, linking, and executing the program. The PC (program counter) register always points to the next instruction to be executed. The way the PIC uses program ROM space was explored because PIC Assembly language programmers must be aware of where programs are placed in ROM, and how much memory is available.

An Assembly language program is composed of a series of statements that are either instructions or pseudo-instructions, also called directives. Instructions are translated by the assembler into machine code. Pseudo-instructions are not translated into machine code: They direct the assembler in how to translate instructions into machine code. Some pseudo-instructions, called data directives, are used to define data. Data is allocated in byte-size increments. The data can be in binary, hex, decimal, or ASCII formats.

Flags are useful to programmers because they indicate certain conditions, such as carry or zero, that result from execution of instructions. The concepts of the RISC and Harvard architectures were also explored.

The RISC architecture allows the design of much more powerful microcontrollers. It has a simple instruction set and uses of a large number of registers. Harvard architecture allows us to bring more code and data to the CPU faster. The use of a wider data bus in the PIC18 allows us to fetch an instruction every cycle because the PIC instructions are typically 2 bytes.
PROBLEMS

SECTION 2.1: THE WREG REGISTER IN THE PIC

1. PIC18 is a(n) _____-bit microcontroller.
2. Register WREG is _____ bits wide.
3. The literal value in MOVLW is _____ bits wide.
4. The largest number that can be loaded into WREG is _____ in hex.
5. To load WREG with the value 65H, the pound sign is (not necessary, optional, necessary) in the instruction "MOVLW #65H".
6. What is the result of the following code and where is it kept?
   
   MOVLW 15H
   ADDLW 13H

7. Which of the following is (are) illegal?
   (a) MOVLW 500 (b) MOVLW 50 (c) MOVLW 00
   (d) MOVLW 255H (e) MOVLW 25H (f) MOVLW F5H
   (g) MOVLW mybyte,50H

8. Which of the following is (are) illegal?
   (a) ADDLW 300H (b) ADDLW 50H (c) ADDLW $500
   (d) ADDLW 255H (e) ADDLW 12H (f) ADDLW 0F5H
   (g) ADDWL 25H

9. What is the result of the following code and where is it kept?
   MOVLW 25H
   ADDLW 1FH

10. What is the result of the following code and where is it kept?
    MOVLW 15H
    ADDLW 0EAH

11. The largest number that K can take for the instruction "ADDWL K" is ____ in hex.

12. True or false. We have many WREG registers in the PIC18.

SECTION 2.2: THE PIC FILE REGISTER

13. PIC data RAM consists of _______ (EEPROM, SRAM).

14. True or false. Data RAM in PIC is also called the file register.

15. True or false. The SFRs are part of the file register memory space.

16. True or false. The general-purpose RAM is not part of the file register memory space.

17. True or false. All members of PIC18 family have the same size file register.

18. If we add the SFR and general-purpose RAM sizes together we should get the total space for the ______.

19. Find the file register size for the following PIC chips:
   (a) PIC12508 (b) PIC16F84 (c) PIC18F8772

20. What is the difference between the EEPROM and data RAM space in the PIC18?

21. Can we have a PIC chip with no EEPROM?

22. Can we have a PIC chip with no file register?
23. The access bank has _____ bytes space.
24. Give the address map of the SFR and GP RAM section of the access bank.
25. What is the maximum number of banks that the PIC18 can have?
26. What is the maximum number of bytes that the PIC18 can have for the file register?

SECTION 2.3: USING INSTRUCTIONS WITH THE DEFAULT ACCES BANK

27. What is the address range for the scratch pad section of the access bank?
28. Show a simple code to load values 30H and 97H into locations 5 and 6 respectively.
29. Show a simple code to load value 55H into locations 0–8.
30. Show a simple code to load value 5FH into Port B SFR.
31. True or false. We can not load literal values into the scratch pad area directly.
32. True or false. The “ADDWF fileReg, D” instruction involves a fileReg and WREG.
33. In Question 32, to place the result in WREG, the D bit must be ____.
34. In Question 32, to place the result in fileReg, the D bit must be ____.
35. Show a simple code to (a) load value 11H into locations 0–5, and (b) add the values together and place the result in WREG as they are added.
36. Repeat Problem 35, except place the result in location 5 after the addition is done.
37. Show a simple code to (a) load value 15H into location 7, and (b) add it to WREG five times and place the result in WREG as the values are added. WREG should be zero before the addition starts.
38. Repeat Problem 37, except place the result in location 7 as numbers are being added together.
39. What is the difference between the MOVWF and MOVF instructions?
40. Write a simple code to complement the contents of location 8 and place the result in WREG.
41. True or false. We can use MOVFF to copy data from any location to any location in the file register.
42. Write a simple code to copy data from location 8 to PORTC (a) using WREG and (b) without using WREG.

SECTION 2.4: PIC STATUS REGISTER

43. The status register is a(n) ____ -bit register.
44. Which bits of the status register are used for the C and DC flag bits, respectively?
45. Which bits of the status register are used for the OV and N flag bits, respectively?
46. In the ADDLW instruction, when is C raised?
47. In the ADDLW instruction, when is DC raised?
48. What is the status of the C and Z flags after the following code?
   MOVFLW  FFH
   ADDLW  1
49. Find the C flag value after each of the following codes:
   (a) MOVLW 54H  (b) MOVLW 00  (c) MOVLW FFH
       ADDLW 0C4H  ADDLW FFH  ADDLW 05H
50. Write a simple program in which the value 55H is added 5 times.

SECTION 2.5: PIC DATA FORMAT AND DIRECTIVES

51. State the value (in hex) used for each of the following data:
    MYDAT_1 EQU 55
    MYDAT_2 EQU D'98'
    MYDAT_3 EQU A'G'
    MYDAT_4 EQU 0x50
    MYDAT_5 EQU D'200'
    MYDAT_6 EQU A'A'
    MYDAT_7 EQU AAH
    MYDAT_8 EQU D'255'
    MYDAT_9 EQU B'10010000'
    MYDAT_10 EQU B'01111110'
    MYDAT_11 EQU D'10'
    MYDAT_12 EQU D'15'

52. State the value (in hex) for each of the following data:
    DAT_1 EQU 22
    DAT_2 EQU 56H
    DAT_3 EQU B'10011001'
    DAT_4 EQU D'32'
    DAT_5 EQU 0xF6
    DAT_6 EQU B'11111011'

53. Show a simple code to (a) load value 11H into locations 0–5, and (b) add them together and place the result in WREG as the values are added. Use EQU to assign the names R0–R5 to locations 0–5.

SECTION 2.6: INTRODUCTION TO PIC ASSEMBLY PROGRAMMING

and

SECTION 2.7: ASSEMBLING AND LINKING A PIC PROGRAM

54. Assembly language is a _______ (low, high)-level language while C is a _______ (low, high)-level language.

55. Of C and Assembly language, which is more efficient in terms of code generation (i.e., the amount of ROM space it uses)?

56. Which program produces the o file?

57. True or false. The source file has the extension "asm".

58. Which file provides the listing of error messages?

59. True or false. The source code file can be a non-ASCII file.

60. True or false. Every source file must have ORG and END directives.

61. Do the ORG and END directives produce opcodes?

62. Why are the ORG and END directives also called pseudocode?

63. True or false. The ORG and END directives appear in the ".lst" file.
64. True or false. The linker produces the file with the extension "asm".
65. True or false. The linker produces the file with the extension "hex".
66. The file with the _____ extension is downloaded into PIC ROM.
67. Give three file extensions produced by MPLAB.

SECTION 2.8: THE PROGRAM COUNTER AND PROGRAM ROM SPACE IN THE PIC

68. Every PIC18 family member wakes up at address _____ when it is powered up.
69. A programmer puts the first opcode at address 100H. What happens when the microcontroller is powered up?
70. Find the number of bytes each of the following instructions takes:
   (a) MOVLLW 5H
   (b) MOVLLW 9FH
   (c) ADDLW 50H
   (d) ADDLW 0
   (e) MOVLLW 0x41
   (f) MOVLLW 20
   (g) ADDLW d'2001'
   (h) GOTO
71. Write a program to (a) place each of your 5-digit ID numbers into a RAM locations starting at address 0, (b) add each digit to WREG and store the sum in RAM location 6, and (c) use the program listing and show the ROM memory addresses and their contents.
72. Use the program listing of your choice and show the ROM memory addresses and their contents.
73. Find the address of the last location of on-chip program ROM for each of the following:
   (a) PIC with 48 KB
   (b) PIC with 96 KB
   (c) PIC with 64 KB
   (d) PIC with 16 KB
   (f) PIC with 128 KB
74. Show the lowest and highest values (in hex) that the PIC18 program counter can take.
75. A given PIC has 7FFFH as the address of the last location of its on-chip ROM. What is the size of on-chip ROM for this PIC?
76. Repeat Question 75 for 3FFH.
77. Find the on-chip program ROM size in K for the PIC18 with the following address ranges:
   (a) 0000–1FFF
   (b) 0000–3FFF
   (c) 0000–5FFF
   (d) 0000–BFFF
   (e) 0000–FFFF
   (f) 00000–1FFFF
   (g) 00000–2FFFF
   (h) 00000–3FFFF
78. Find the on-chip program ROM size in K for the PIC18 with the following address ranges:
   (a) 00000–4FFFF
   (b) 00000–3FFFF
   (c) 00000–5FFFF
   (d) 00000–7FFFF
   (e) 00000–BFFFF
   (f) 00000–FFFFF
   (g) 00000–17FFFF
   (h) 00000–1FFFFF
Some of the above might not be in production yet.
79. How wide is the ROM in the PIC18 chip?
80. How wide is the data bus between the CPU and the program ROM in PIC18?
81. Show the even and odd address designation for 4K x 16. See Figure 2-11.
82. In Question 81, what is the ROM size in K?
83. Show the even and odd address designation for 16K x 16. See Figure 2-11.
84. In Question 83, what is the ROM size in K?
85. Explain Harvard architecture and how it makes processing of code and data faster.
86. What is the drawback of using Harvard architecture for memories external to the CPU?
87. In instruction “MOVlw K” explain why the K value cannot be larger than 255 decimal.
88. In instruction “ADDlw K” explain why the K value cannot be larger than 255 decimal.
89. In “MOVWF fileReg”, explain what the size of the instruction is and how it allows one to cover the entire range of the file register in PIC18.
90. In “MOVFF source, dest”, explain what the size of the instruction is and how it allows one to cover the entire range of the file register in PIC18.
91. In the instruction “GOTO target-addr” explain why the lowest bit of the program counter is 0.
92. Explain why the instruction “GOTO target-addr” will not land at an odd address.
93. In Question 92, explain why it should not.
94. Explain how the instruction “GOTO target-addr” is able to cover the entire 2M address space of the PIC18.

SECTION 2.9. RISC ARCHITECTURE IN THE PIC

95. What do RISC and CISC stand for?
96. In ______ (RISC, CISC) architecture we can have 1-, 2-, 3-, or 4-byte instructions.
97. In ______ (RISC, CISC) architecture instructions are fixed in size.
98. In ______ (RISC, CISC) architecture instructions are mostly executed in one or two cycles.
99. In ______ (RISC, CISC) architecture we can have an instruction to ADD a register to external memory.
100. True or false. Most instructions in CISC are executed in one or two cycles.

ANSWERS TO REVIEW QUESTIONS

SECTION 2.1: THE WREG REGISTER IN THE PIC

1. MOVlw 0x34
2. MOVlw 0x16
   ADDlw 0xCD
3. False
4. FF hex and 255 in decimal
5. 8
SECTION 2.2: THE PIC FILE REGISTER

1. True
2. File register
3. True
4. True
5. 8
6. 256
7. 4096

SECTION 2.3: USING INSTRUCTIONS WITH THE DEFAULT ACCESS BANK

1. True
2. MOV LW 0x16
   MOVWF 0
   MOV LW 0xFD
   ADDWF 0, F
3. True
4. FF, 255
5. WREG

SECTION 2.4: PIC STATUS REGISTER

1. Status register
2. 8 bits
3. D5, D6, and D7
4.
   Hex   binary
   9F    1001 1111
   + 61  + 0110 0001
   100   10000 0000  This leads to C = 1, DC = 1, and Z = 1.
5.
   Hex   binary
   82    1000 0010
   + 22  + 0010 0010
   A4    1010 0100  This leads to C = 0, DC = 0, and Z = 0.
6.
   Hex   binary
   67    0110 0111
   + 99  + 1001 1001
   100   10000 0000  This leads to C = 1, DC = 1, and Z = 1.

SECTION 2.5: PIC DATA FORMAT AND DIRECTIVES

1. DATA1 EQU 9FH
   DATA2 EQU 0x9F
   DATA3 EQU H'9F'
2. DATA1 EQU 99H
   DATA2 EQU D'99'
   DATA3 EQU B'10011001'
3. If the value is to be changed later, it can be done once in one place instead of at every occurrence.
4. (a) 34H      (b) 1FH
5. WREG = 15H
6. value of location 0x20 = (0x95)
7. 0CH + 10H = 1CH will be in fileReg location 63H
SECTION 2.6: INTRODUCTION TO PIC ASSEMBLY PROGRAMMING

1. The real work is performed by instructions such as MOV and ADD. Pseudo-instructions, also called assembly directives, instruct the assembler in doing its job.
2. The instruction mnemonics, pseudo-instructions
3. False
4. All except (c)
5. Assembler directives
6. True
7. (c)

SECTION 2.7: ASSEMBLING AND LINKING A PIC PROGRAM

1. True
2. True
3. (a)
4. (b) through (e)
5. (d) and (e)

SECTION 2.8: THE PROGRAM COUNTER AND PROGRAM ROM SPACE IN THE PIC

1. 21
2. True
3. 0000H
4. 2
5. With 32K, we have 32768 (32 x 1024) bytes, and the ROM space is 0000 to 7FFFH.
6. 4
7. True

SECTION 2.9: RISC ARCHITECTURE IN THE PIC

1. CISC stands for complex instruction set computer; RISC is reduced instruction set computer.
2. True
3. Small
4. True
5. 75, yes
6. True
7. (c)
8. False
CHAPTER 3

BRANCH, CALL, AND TIME DELAY LOOP

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Code PIC Assembly language instructions to create loops
- Code PIC Assembly language conditional branch instructions
- Explain conditions that determine each conditional branch instruction
- Code GOTO (long jump) instructions for unconditional jumps
- Calculate target addresses for conditional branch instructions
- Code PIC subroutines
- Describe the stack and its use in subroutines
- Discuss pipelining in the PIC
- Discuss crystal frequency versus instruction cycle time in the PIC
- Code PIC programs to generate a time delay
In the sequence of instructions to be executed, it is often necessary to transfer program control to a different location. There are many instructions in PIC to achieve this. This chapter covers the control transfer instructions available in PIC Assembly language. In Section 3.1, we discuss instructions used for looping, as well as instructions for conditional and unconditional branches (jumps). In the second section, we examine the stack and the CALL instruction. In Section 3.3, instruction pipelining of the PIC18 is examined. Instruction timing and time delay subroutines are also discussed in Section 3.3.

SECTION 3.1: BRANCH INSTRUCTIONS AND LOOPING

In this section we first discuss how to perform a looping action in PIC and then the branch (jump) instructions, both conditional and unconditional.

Looping in PIC

Repeating a sequence of instructions or an operation a certain number of times is called a loop. The loop is one of most widely used programming techniques. In the PIC, there are several ways to repeat an operation many times. One way is to repeat the operation over and over until it is finished, as shown below:

```assembly
MOVLW 0 ;WREG = 0
ADDLW 3 ;add value 3 to WREG
ADDLW 3 ;add value 3 to WREG(W = 6)
ADDLW 3 ;add value 3 to WREG(W = 9)
ADDLW 3 ;add value 3 to WREG(W = 0Ch)
ADDLW 3 ;add value 3 to WREG(W = 0Fh)
```

In the above program, we add 3 to WREG 5 times. That makes $5 \times 3 = 15 = 0$Fh. One problem with the above program is that too much code space would be needed to increase the number of repetitions to 50 or 100. A much better way is to use a loop. There are two ways to do a loop in PIC. Next, we describe each method.

**DECFSZ instruction and looping**

The DECFSZ (decrement fileReg skip zero) instruction is a widely used instruction supported across all PIC families of microcontrollers from PIC12 to PIC18. It has the following format:

```
DECFSZ fileReg , d ;decrement fileReg and skip next instruction if 0
```

In this instruction, the fileReg is decremented, and if its content is zero, it skips the next instruction. By placing the “GOTO target” instruction right below it we can create a loop. The target address of the “GOTO target” instruction is the beginning of the loop, as shown in Examples 3-1 and 3-2. Figure 3-1 shows the flowchart for the DECFSZ instruction. Study the flowchart structure in Appendix D to get familiar with the symbols. The flowchart is a widely used method to represent a sequence of actions pictorially. Its usage for program design is recommended very strongly.
Example 3-1

Write a program to (a) clear WREG, and (b) add 3 to WREG ten times and place the result in SFR of PORTB. Use the DECFSZ instruction to perform looping.

Solution:

;this program adds value 3 to WREG ten times

COUNT EQU 0x25 ;use loc 25H for counter

MOVWF COUNT ;load the counter
MOVlw 0 ;WREG = 0
AGAIN ADDlw 3 ;add 03 to WREG (WREG = sum)
DECFSZ COUNT,F ;decrement counter, skip if count = 0
GOTO AGAIN ;repeat until count becomes 0
MOVWF PORTB ;send sum to PORTB SFR

Notice that the DECFSZ instruction will decrement the counter (fileReg loc 0x25), which has 10 in it. It becomes 9. Because it is not zero, it will execute the “GOTO AGAIN” instruction. The “GOTO AGAIN” goes back to the start of the loop. Next, it decrements, our counter becomes 8, and, because it is not zero, it executes the GOTO. It goes on like that until the counter becomes zero. Upon the counter becoming zero, it skips the GOTO, which gets it out of the loop, and executes the “MOVWF PORTB” instruction. Notice that we use “DECFSZ COUNT, F” and not “DECFSZ COUNT, W” because we want the count value to change for the next iteration. We will never get out of the loop if we use “DECFSZ COUNT, W” because COUNT = 9 and the decrement value is placed in WREG.
Using instruction BNZ for looping

The BNZ (branch if not zero) instruction is supported by the PIC18 family and not earlier families such as PIC16 or PIC12. It uses the zero flag in the status register. The BNZ instruction is used as follows:

BACK ............ ; start of the loop
............ ; body of the loop
............ ; body of the loop
DECF ; decrement fileReg, Z = 1 if fileReg = 0
BNZ BACK ; branch to BACK if Z = 0

In the last two instructions, the fileReg is decremented; if it is not zero, it branches (jumps) back to the target address referred to by the label. Prior to the start of the loop, the fileReg is loaded with the counter value for the number of repetitions. Notice that the BNZ instruction refers to the Z flag of the status register affected by the previous instruction, DECF. This is shown in Example 3-2.

In the program in Example 3-2, fileReg location 0x25 is used as a counter. The counter is first set to 10. In each iteration, the DEC instruction decrements the fileReg and sets the flag bits accordingly. If fileReg is not zero (Z ≠ 0), it jumps to the target address associated with the label "AGAIN". This looping action continues until fileReg COUNT becomes zero. After fileReg COUNT becomes zero (Z = 0), it falls through the loop and executes the instruction immediately below it, in this case "MOVWF PORTB". See Figure 3-2.

Notice, in the "DECF COUNT, F" instruction, that fileReg 25H is used as a register to hold the count as it decrements instead of WREG. If WREG is used as the destination of the DECF instruction, then you have an infinite loop because COUNT remains at its original value of 10.

Example 3-2

Write a program to (a) clear WREG, then (b) add 3 to WREG ten times.

Use the zero flag and BNZ.

Solution:

; this program adds value 3 to the WREG ten times

COUNT EQU 0x25 ; use loc 25H for counter
MOVLW d'10' ; WREG = 10 (decimal) for counter
MOVWF COUNT ; load the counter
MOVLW 0 ; WREG = 0
AGAIN ADDLW 3 ; add 03 to WREG (WREG = sum)
DECF COUNT, F ; decrement counter
BNZ AGAIN ; repeat until COUNT = 0
MOVWF PORTB ; send sum to PORTB SFR

100
Example 3-3

What is the maximum number of times that the loop in Example 3-2 can be repeated?

Solution:

Because location COUNT in fileReg is an 8-bit register, it can hold a maximum of FFH (255 decimal); therefore, the loop can be repeated a maximum of 255 times. See Example 3-4 to bypass this limitation.
Loop inside a loop

As shown in Example 3-2, the maximum count is 255. What happens if we want to repeat an action more times than 255? To do that, we use a loop inside a loop, which is called a nested loop. In a nested loop, we use two registers to hold the count. See Example 3-4.

Example 3-4

Write a program to (a) load the PORTB SFR register with the value 55H, and (b) complement Port B 700 times.

Solution:

Because 700 is larger than 255 (the maximum capacity of any register), we use two registers to hold the count. The following code shows how to use fileReg locations 25H and 26H as a register for counters.

```
R1 EQU 0x25
R2 EQU 0x26
COUNT_1 EQU d'10'
COUNT_2 EQU d'70'
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB ;PORTB = 55H
MOVLW COUNT_1 ;WREG = 10, outer loop count value
MOVWF R1 ;load 10 into loc 25H (outer loop count)
LOP_1 MOVLW COUNT_2 ;WREG = 70, inner loop count value
MOVWF R2 ;load 70 into loc 26H
LOP_2 COMPF PORTB, F ;complement Port B SFR
DEC R2, F ;dec fileReg loc 26 (inner loop)
BNZ LOP_2, F ;repeat it 70 times
DEC R1, F ;dec fileReg loc 25 (outer loop)
BNZ LOP_1 ;repeat it 10 times
```

In this program, fileReg location 0x26 is used to keep the inner loop count. In the instruction “BNZ LOP_2”, whenever location 26H becomes 0 it falls through and “DEC R1, F” is executed. This instruction forces the CPU to load the inner count with 70 if it is not zero, and the inner loop starts again. This process will continue until location 25 becomes zero and the outer loop is finished.

<table>
<thead>
<tr>
<th>MEMORY LOCATION</th>
<th>VALUE</th>
</tr>
</thead>
<tbody>
<tr>
<td>25</td>
<td>10 R1</td>
</tr>
<tr>
<td>26</td>
<td>70 R2</td>
</tr>
</tbody>
</table>
Figure 3-3. Flowchart for Example 3-4

INSTRUCTIONS

LOAD WREG

MOVWF PORTB

LOAD PORTB

LOAD WREG

MOVWF R1

LOAD COUNTER 1

MOVLW COUNT_1

LOAD COUNTER 2

MOVWF R1

LOAD WREG

MOVF PORTB, F

TOGGLE PORTB

LOP_1

DECF R2, F

DECREMENT COUNTER 2

LOP_2

BNZ LOP_2

IS COUNTER 2 ZERO?

BNZ LOP_1

YES

NO

YES

NO

DEC R1, F

END

BNZ LOP_1

CHAPTER 3: BRANCH, CALL, AND TIME DELAY LOOP
Looping 100,000 times

Because two registers give us a maximum value of 65025 (255 \times 255 = 65025), we can use three registers to get up to more than 16 million (2^{24}) iterations. The following code repeats an action 100,000 times:

```
R1 EQU 0x1
R2 EQU 0x2
R2 EQU 0x3
COUNT_1 EQU D'100'
COUNT_2 EQU D'100'
COUNT_3 EQU D'10'

MOVLW 0x55
MOVWF PORTB
MOVWF COUNT_3
MOVWF R3
LOP_3 MOVLW COUNT_2
MOVWF R2
LOP_2 MOVLW COUNT_1
MOVWF R1
LOP_1 COMPF PORTB, F
    DECF R1, F
    BNZ LOP_1
    DECF R2, F
    BNZ LOP_2
    DECF R3, F
    BNZ LOP_3
```

Other conditional jumps

Conditional branches for the PIC are summarized in Table 3-1. More details of each instruction are provided in Appendix A. In Table 3-1, notice that some of the instructions, such as BZ (Branch if Z = 1) and BC (Branch if C = 1), jump only if a certain condition is met. Next, we examine some conditional branch instructions with examples.

**BZ (Branch if Z = 1)**

In this instruction, the Z flag is checked. If it is high, it jumps to the target address. For example, look at the following code.

```
OVER MOVF PORTB,W ;read Port B and put it in WREG
    JZ OVER ;jump if WREG is zero
```

In this program, if PORTB is zero, it jumps to the label OVER. It stays there until PORTB has a value other than zero. Notice that the BZ instruction can be used to see whether any fileReg or WREG is zero. More importantly, you don’t have to perform an arithmetic instruction such as decrement to use the BZ instruction. See Example 3-5.

---

**Table 3-1: PIC Conditional Branch (Jump) Instructions**

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Action</th>
</tr>
</thead>
<tbody>
<tr>
<td>BC</td>
<td>Branch if C = 1</td>
</tr>
<tr>
<td>BNC</td>
<td>Branch if C ≠ 0</td>
</tr>
<tr>
<td>BZ</td>
<td>Branch if Z = 1</td>
</tr>
<tr>
<td>BNZ</td>
<td>Branch if Z ≠ 0</td>
</tr>
<tr>
<td>BN</td>
<td>Branch if N = 1</td>
</tr>
<tr>
<td>BNN</td>
<td>Branch if N ≠ 0</td>
</tr>
<tr>
<td>BOV</td>
<td>Branch if OV = 1</td>
</tr>
<tr>
<td>BNOV</td>
<td>Branch if OV ≠ 0</td>
</tr>
</tbody>
</table>
Example 3-5

Write a program to determine if fileReg location 0x30 contains the value 0. If so, put 55H in it.

Solution:

```assembly
MYLOC EQU 0x30
MOVF MYLOC,F ;copy MYLOC to itself
BNZ NEXT ;branch if MYLOC is not zero
MOVLW 0x55
MOVWF MYLOC ;put 0x55 if MYLOC has zero value
NEXT ...
```

**BNC (branch if no carry, branch if CY = 0)**

In this instruction, the carry flag bit in the Status register is used to make the decision whether to jump. In executing “BNC label”, the processor looks at the carry flag to see if it is raised (C = 1). If it is not, the CPU starts to fetch and execute instructions from the address of the label. If C = 1, it will not branch but will execute the next instruction below BNC. Study Example 3-6 to see how BNC is used to add numbers together when the sum is higher than FFH. Note that there

Example 3-6

Find the sum of the values 79H, F5H, and E2H. Put the sum in fileReg locations 5 (low byte) and 6 (high byte).

Solution:

```assembly
L_Byte EQU 0x5 ;assign RAM loc 5 to L_byte of sum
H_Byte EQU 0x6 ;assign RAM loc 6 to H_byte of sum

ORG 0h
MOVLW 0x0 ;clear WREG(WREG = 0)
MOVF H_Byte ;H_Byte = 0
ADDLW 0x79 ;WREG = 0 + 79H = 79H, C = 0
BNC N_1 ;if C = 0, add next number
INCF H_Byte,F ;C = 1, increment (now H_Byte = 0)
N_1 ADDLW 0xF5 ;WREG = 79 + F5 = 6E and C = 1
BNC N_2 ;branch if CY = 0
INCF H_Byte,F ;C = 1, increment (now H_Byte = 1)
N_2 ADDLW 0xE2 ;WREG = 6E + E2 = 50 and C = 1
BNC OVER ;branch if C = 0
INCF H_Byte,F ;C = 1, increment (now H_Byte = 2)
OVER MOVWF L_Byte ;now L_Byte = 50H, and H_Byte = 02
END
```

**MEMORY LOCATION**

<p>| | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>4</td>
<td></td>
<td></td>
</tr>
<tr>
<td>5</td>
<td></td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

WREG = 79H  WREG = 6EH  WREG = 50H

L_Byte  H_Byte
is also a “BC label” instruction. In the BC instruction, if C = 1 it jumps to the target address. We will give more examples of these instructions in the context of some applications in Chapter 5.

The other conditional branch instructions in Table 3-1 are discussed in Chapter 5 when arithmetic operations with signed numbers are discussed.

**All conditional branches are short jumps**

It must be noted that all conditional jumps are short jumps, meaning that the address of the target must be within 256 bytes of the contents of the program counter (PC). This concept is discussed next.

---

**Example 3-7**

Using the following list file of Example 3-6, verify the jump forward address calculation.

<table>
<thead>
<tr>
<th>Line</th>
<th>PC</th>
<th>Opcode</th>
<th>Mnemonic</th>
<th>Operand</th>
</tr>
</thead>
<tbody>
<tr>
<td>LOC</td>
<td>OBJECT LINE CODE</td>
<td>SOURCE TEXT</td>
<td></td>
<td></td>
</tr>
<tr>
<td>VALUE</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000005</td>
<td>00001</td>
<td>L_byte EQU 0x5</td>
<td>assign RAM Loc 5 to L_byte of sum</td>
<td></td>
</tr>
<tr>
<td>00000006</td>
<td>00002</td>
<td>H_byte EQU 0x6</td>
<td>assign RAM Loc 6 to H_byte of sum</td>
<td></td>
</tr>
<tr>
<td>00000000</td>
<td>00004</td>
<td>ORG 0h</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000000</td>
<td>00005</td>
<td>MOVWF H_byte</td>
<td>H_byte = 0</td>
<td></td>
</tr>
<tr>
<td>00000004</td>
<td>00006</td>
<td>ADDLW 0x79</td>
<td>WREG = 0 + 79H = 79H, C = 0</td>
<td></td>
</tr>
<tr>
<td>00000006</td>
<td>00008</td>
<td>BNC N_1</td>
<td>if C = 0, add next number</td>
<td></td>
</tr>
<tr>
<td>00000008</td>
<td>00009</td>
<td>INCF H_byte,F</td>
<td>C = 1, increment (now H_byte = 0)</td>
<td></td>
</tr>
<tr>
<td>0000000A</td>
<td>00010</td>
<td>ADDLW 0xF5</td>
<td>WREG = 79 + F5 = 6E and C = 1</td>
<td></td>
</tr>
<tr>
<td>0000000C</td>
<td>00011</td>
<td>BNC N_2</td>
<td>branch if CY = 0</td>
<td></td>
</tr>
<tr>
<td>0000000E</td>
<td>00012</td>
<td>INCF H_byte,F</td>
<td>C = 1, increment (now H_byte = 1)</td>
<td></td>
</tr>
<tr>
<td>00000010</td>
<td>00013</td>
<td>ADDLW 0xE2</td>
<td>WREG = 6E + E2 = 50 and C = 1</td>
<td></td>
</tr>
<tr>
<td>00000012</td>
<td>00014</td>
<td>BNC OVER</td>
<td>branch if C = 0</td>
<td></td>
</tr>
<tr>
<td>00000014</td>
<td>00015</td>
<td>INCF H_byte,F</td>
<td>C = 1, increment (now H_byte = 2)</td>
<td></td>
</tr>
<tr>
<td>00000016</td>
<td>00016</td>
<td>OVER MOVWF H_byte</td>
<td>now H_byte = 50H, and H_byte = 02</td>
<td></td>
</tr>
<tr>
<td>000017</td>
<td>END</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Solution:**

First notice that the BNC instruction jumps forward. The target address for a forward jump is calculated by adding the PC of the following instruction to the second byte of the branch instruction times 2. Recall that each instruction takes 2 bytes. In line 6 the instruction “BNC N_1” has an opcode of E3 and an operand of 01 at the addresses of 000006 and 000007. The 01 × 02 = 02 is the relative address, relative to the address of the next instruction INCF, which is 000008. By adding 000002 to 000008, the target address of the label N_1, which is 00000A, is generated. In the same way for line 000011, the “BNC N_2” instruction, and line 000014, the “BNC OVER” instruction jumps forward because the relative value is positive.
Calculating the short branch address

All conditional branches such as BNC, BZ, and BNZ are short branches due to the fact that they are all 2-byte instructions. In these instructions the first byte is the opcode and the second byte is the relative address. The target address is relative to the value of the program counter. If the second byte is positive, the jump is forward. If the second byte is negative, then the jump is backwards. The second byte can be a value from \(-127\) to \(+128\). To calculate the target address, we add the second byte of the instruction times 2 to the PC of the next instruction [target address = \((\text{2nd byte of instruction} \times 2) + \text{PC}\)]. See Example 3-7. We do the same thing for the backward branch, although the second byte is negative. That is, we multiply the negative number by two and add it to the PC value of the next instruction. See Example 3-8.

Example 3-8

Verify the calculation of backward jumps for the listing of Example 3-2, shown below.

Solution:

<table>
<thead>
<tr>
<th>LOC</th>
<th>VALUE</th>
<th>OBJECT</th>
<th>LINE</th>
<th>SOURCE</th>
<th>TEXT</th>
</tr>
</thead>
<tbody>
<tr>
<td>00000025</td>
<td>00001</td>
<td>COUNT</td>
<td>EQU 0x25</td>
<td>; use loc 25H for counter</td>
<td></td>
</tr>
<tr>
<td>000000</td>
<td>00002</td>
<td>ORG 0h</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000000</td>
<td>0E0A</td>
<td>MOVLM d'10' ; WREG = 10 (decimal) for counter</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000002</td>
<td>6E25</td>
<td>MOVWF COUNT ; load the counter</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000004</td>
<td>0E00</td>
<td>MOVLM 0 ; WREG = 0</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000006</td>
<td>0F03</td>
<td>AGAIN</td>
<td>ADDLW 3 ; add 03 to WREG (WREG = sum)</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000008</td>
<td>0E25</td>
<td>DECF COUNT, F ; decrement counter</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000A</td>
<td>E1FD</td>
<td>BNZ AGAIN ; repeat until COUNT = 0</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000CC</td>
<td>6E81</td>
<td>MOVWF PORTB ; send sum to PORTB SFR</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>00010</td>
<td>END</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

In the program list, “BNZ AGAIN” has opcode E1 and relative address FDH. The FDH gives us \(-3\), which means the displacement is \(-3 \times 2 = -6\). When the relative address of \(-6\) is added to 00000CH, the address of the instruction below the byte, we have \(-6 + 0CH = 06H\) (the carry is dropped). Notice that 000006 is the address of the label AGAIN. FDH is a negative number and that means it will branch backward. For further discussion of the addition of negative numbers, see Chapter 5.

Although we can use BNZ along with DECF to perform a loop, it is better to use an instruction such as DCFSNZ, because it combines the decrement and jump into a single instruction.
Unconditional branch instruction

The unconditional branch is a jump in which control is transferred unconditionally to the target location. In the PIC18 there are two unconditional branches: GOTO (go to) and BRA (branch). Deciding which one to use depends on the target address. Each instruction is explained next.

**GOTO (GOTO is a long jump)**

GOTO is an unconditional jump that can go to any memory location in the 2M address space of the PIC18. It is a 4-byte (32-bit) instruction in which 12 bits are used for the opcode, and the other 20 bits represent the 20-bit address of the target location. The 20-bit target address allows a jump to 1M of memory locations from 00000 to FFFFFH, instead of 2M. This problem is solved by making the lowest bit of the program counter A0 = 0, and the 20-bit target address of the GOTO becomes address bits A21–A1. In this way, the GOTO can cover the entire 2M address space of 00000–1FFFFH and also makes sure that the target address lands on an even address location. Because all the PIC18 instructions are 2-byte or 4-byte instructions, the GOTO will not land at the middle of an instruction. See Figure 3-4.

![Figure 3-4. GOTO Instruction](image)

Remember that although the program counter in the PIC18 is 21-bit (thereby giving a ROM address space of 2M), not all PIC18 family members have that much on-chip program ROM. Some of the PIC18 family members have only 4K–32K of on-chip ROM for program space; consequently, every byte is precious. For this reason there is also a BRA (branch) instruction, which is a 2-byte instruction as opposed to the 4-byte GOTO instruction. This can save some bytes of memory in many applications where ROM memory space is in short supply. BRA is discussed next.

**BRA (branch)**

In this 2-byte (16-bit) instruction, the first 5 bits are the opcode and the rest (lower 11 bits) is the relative address of the target location. The relative address range of 000–FFFH is divided into forward and backward jumps; that is, within...
-1024 to +1023 bytes of memory relative to the address of the current PC (program counter). If the jump is forward, then the target address is positive. If the jump is backward, then the target address is negative. In this regard, BRA is like the conditional branch instructions except that 11 bits are used for the offset address instead of 8. This is shown in detail in Figure 3-5.

![Figure 3-5. BRA (Branch Unconditionally) Instruction Address Range](image)

Notice that this is a 2-byte instruction, and is preferred over the GOTO because it takes less ROM space. Chapter 5 examines signed numbers.

**GOTO to itself using $ sign**

In cases where there is no monitor program, we use the GOTO (jump) to itself in order to keep the microcontroller busy. A simple way of doing that is to use the $ sign. That means in place of this:

```
HERE GOTO HERE
```

we can use the following:

```
GOTO $
```

This will also work for the BRA instruction, as shown below:

```
OVER BRA OVER
```

which is the same as:

```
BRA $ ;$ means same line
```

**Review Questions**

1. The mnemonic BNZ stands for ________.
2. True or false. "BNZ BACK" makes its decision based on the last instruction affecting the Z flag.
3. "BNZ HERE" is a ____-byte instruction.
4. In "JZ NEXT", which register’s content is checked to see if it is zero?
5. GOTO is a(n) ____-byte instruction.
SECTION 3.2: CALL INSTRUCTIONS AND STACK

Another control transfer instruction is the CALL instruction, which is used to call a subroutine. Subroutines are often used to perform tasks that need to be performed frequently. This makes a program more structured in addition to saving memory space. In the PIC18 there are two instructions for call: CALL (long call) and RCALL (relative call). Deciding which one to use depends on the target address. Each instruction is explained next.

CALL

In this 4-byte (32-bit) instruction, the 12 bits are used for the opcode and the other 20 bits, A21–A1, are used for the address of the target subroutine. Just as in the GOTO instruction, the lowest bit of the program counter is 0 automatically to ensure it lands on an even address. Therefore, CALL can be used to call subroutines located anywhere within the 2M address space of 00000–1FFFFH for the PIC18, as shown in Figure 3-6.

![Figure 3-6. CALL Instruction](image)

To make sure that the PIC knows where to come back to after execution of the called subroutine, the microcontroller automatically saves on the stack the address of the instruction immediately below the CALL. When a subroutine is called, control is transferred to that subroutine, and the processor saves the PC (program counter) of the next on the stack and begins to fetch instructions from the new location. After finishing execution of the subroutine, the instruction RETURN transfers control back to the caller. Every subroutine needs RETURN as the last instruction.

Stack and stack pointer in the PIC18

The stack is read/write memory (RAM) used by the CPU to store some very critical information temporarily. This information usually is an address, but it could be data as well. The CPU needs this storage area because there are only a limited number of registers. The stack in the PIC18 is 21-bit because the program counter is 21-bit. This means that it is used for the CALL instruction to make sure that the PIC knows where to come back to after execution of the called subroutine.
A 21-bit stack can take values of 00000 to 1FFFFFH, just like the program counter. If the stack is RAM, there must be a register inside the CPU to point to it. The register used to access the stack is called the SP (stack pointer) register. The PIC18 has a 5-bit stack pointer, which can take values of 00 to 1FH. That gives us a total of 32 locations where each location is 21 bits wide. This is shown in Figure 3-7. When the PIC18 is powered up, the SP register contains value 0. This means that stack location 1 is the first location used for the stack because the SP points to the last-used location. That means that location 0 of the stack is not available and we have only 31 stack locations in the PIC18.

![Figure 3-7. PIC Stack 31 × 21](image)

**How stacks are accessed in the PIC18**

The storing of CPU information such as the program counter on the stack is called a PUSH, and loading the contents of the stack back into a CPU register is called a POP. In other words, a register is pushed onto the stack to save it and popped off the stack to retrieve it. The following describes each process.

**Pushing onto the stack**

In the PIC, the stack pointer (SP) is pointing to the last used location of the stack. The last-used location of the stack is referred to as the top of the stack (TOS). As data is pushed onto the stack, the stack pointer is incremented. Notice that this is different from many other microprocessors, notably x86 processors, in which the SP is decremented when data is pushed onto the stack. Examining Example 3-9, we see that as each CALL is executed, the contents of the program counter are saved on the stack and SP is incremented. Notice that for every program counter saved on the stack, SP is incremented only once.

**Popping from the stack**

Popping the contents of the stack back into a given register such as the program counter is the opposite process of pushing. When the RETURN instruction
at the end of the subroutine is executed, the top location of the stack is copied back to the program counter and the stack pointer is decremented once. That means the stack is LIFO (Last-In-First-Out) memory.

**CALL instruction and the role of the stack**

In the PIC, the CPU uses the stack to save the address of the instruction just below the CALL instruction. This is how the CPU knows where to resume when it returns from the called subroutine. To understand the importance of the stack in microcontrollers, examine the contents of the stack and stack pointer for Example 3-9. This is shown in Example 3-10.

The following points should be noted for the program in Example 3-9:

1. Notice the DELAY subroutine. Upon executing the first "CALL DELAY", the address of the instruction right below it, "MOVLW 0xAA", is pushed onto the stack, and the PIC starts to execute instructions at address 000300H.
2. In the DELAY subroutine, first the counter MYREG is set to 255 (MYREG = FFH); therefore, the loop is repeated 256 times. When MYREG becomes 0, control falls to the RETURN instruction, which pops the address from the top of the stack into the program counter and resumes executing the instructions after the CALL.

The amount of time delay in Example 3-9 depends on the frequency of the PIC. How to calculate the exact time will be explained in the last section of this chapter.

**Example 3-9**

Toggle all the bits of the SFR register of Port B by sending to it the values 55H and AAH continuously. Put a time delay in between each issuing of data to Port B.

**Solution:**

```assembly
MYREG EQU 0x08 ;use location 08 as counter
ORG 0
BACK
MOVLW 0x55 ;load WREG with 55H
MOVF PORTB ;send 55H to port B
CALL DELAY ;time delay
MOVLW 0xAA ;load WREG with AA (in hex)
MOVF PORTB ;send AAH to port B
CALL DELAY
GOTO BACK ;keep doing this indefinitely

;this is the delay subroutine
ORG 300H ;put time delay at address 300H
DELAY
MOVLW 0xFF ;WREG = 255, the counter
MOVF MYREG
AGAIN
NOP ;no operation wastes clock cycles
NOP
DECFF MYREG, F
BNZF AGAIN ;repeat until MYREG becomes 0
RETURN ;return to caller
END ;end of asm file
```

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Example 3-10

Analyze the stack for the CALL instructions in the following program.

Solution:

When the first CALL is executed, the address of the instruction "MOVLW 0xAA" is saved (pushed) on the stack. The last instruction of the called subroutine must be a RETURN instruction, which directs the CPU to pop the contents of the top location of the stack into the PC and resume executing at address 000007. The diagrams show the stack frame after the CALL and RETURN instructions.

LOC  OBJECT CODE  LINE  SOURCE  TEXT
VALUE
00001  #DEFINE PORTB 0xF81
00000008  00002  MYREG EQU 0x08 ;use location 08 as counter
00003
00004
00005
00000  00006  ORG 0
000000  00006  BACK MOVLW 0x55 ;load WREG with 55H
000002  00007  MOVWF PORTB ;send 55H to port B
000004  00008  CALL DELAY ;time delay
000008  00009  MOVLW 0xAA ;load WREG with AA (in hex)
00000A  00010  MOVWF PORTB ;send AAH to port B
00000C  00011  CALL DELAY
000010  00012  GOTO BACK ;keep doing this indefinitely
00013
00014 ;—— this is the delay subroutine
00015
000300  00016  ORG 300H ;put delay at address 300H
000300  00017  DELAY MOVLW 0xFF ;WREG = 255, the counter
000302  00018  MOVWF MYREG
000304  00019  AGAIN NOP ;no op wastes clock cycles
000306  00020  NOP
000308  00021  DECF MYREG, F
00030A  00022  BNZ AGAIN ;repeat until MYREG becomes 0
00030C  00023  RETURN ;return to caller
00030D  00024  END ;end of asm file

BEFORE  THE  FIRST  CALL
AFTER  THE  FIRST  CALL
AFTER  RETURN
AFTER  THE  SECOND  CALL
AFTER  RETURN

<p>| | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>4</td>
<td>4</td>
<td>4</td>
<td>4</td>
</tr>
<tr>
<td>3</td>
<td>3</td>
<td>3</td>
<td>3</td>
</tr>
<tr>
<td>2</td>
<td>2</td>
<td>2</td>
<td>2</td>
</tr>
<tr>
<td>1</td>
<td>0008</td>
<td>1</td>
<td>000010</td>
</tr>
</tbody>
</table>

SP = 0
SP = 1
SP = 0
SP = 1
SP = 0

CHAPTER 3: BRANCH, CALL, AND TIME DELAY LOOP 113
The upper limit of the stack

As mentioned earlier, in the PIC18 there are only 31 RAM locations for the stack and they are mapped to addresses 01 to 1FH. That limits the number of calls inside of calls for a given program to 31 because stack location 00 is not available. In PIC, the stack is used for calls and interrupts. We must remember that upon calling a subroutine, the stack keeps track of where the CPU should return after completing the subroutine. For this reason, we must be very careful not to manipulate the stack contents. See Chapter 6 for more on this.

**Calling many subroutines from the main program**

In Assembly language programming, it is common to have one main program and many subroutines that are called from the main program. (See Figure 3-8.) This allows you to make each subroutine into a separate module. Each module can be tested separately and then brought together with the main program. More importantly, in a large program the modules can be assigned to different programmers in order to shorten development time. See Chapter 6 for discussion of modules.

<table>
<thead>
<tr>
<th>;MAIN program calling subroutines</th>
</tr>
</thead>
<tbody>
<tr>
<td>ORG 0</td>
</tr>
<tr>
<td>MAIN CALL SUBR_1</td>
</tr>
<tr>
<td>CALL SUBR_2</td>
</tr>
<tr>
<td>CALL SUBR_3</td>
</tr>
<tr>
<td>HERE BRA HERE ;stay here</td>
</tr>
<tr>
<td>;-------------------------------</td>
</tr>
<tr>
<td>SUBR_1 . . .</td>
</tr>
<tr>
<td>. . . RETURN</td>
</tr>
<tr>
<td>;-------------------------------</td>
</tr>
<tr>
<td>SUBR_2 . . .</td>
</tr>
<tr>
<td>. . . RETURN</td>
</tr>
<tr>
<td>;-------------------------------</td>
</tr>
<tr>
<td>SUBR_3 . . .</td>
</tr>
<tr>
<td>. . . RETURN</td>
</tr>
<tr>
<td>;-------------------------------</td>
</tr>
</tbody>
</table>

**Figure 3-8. PIC Assembly Main Program That Calls Subroutines**

It needs to be emphasized that in using CALL, the target address of the subroutine can be anywhere within the 2M memory space of the PIC18. (See Example 3-11.) This is not the case for the other call instruction, RCALL, which is explained next.

**RCALL (relative call)**

RCALL is a 2-byte instruction in contrast to CALL, which is 4 bytes.
Example 3-11

Write a program to count up from 00 to FFH and send the count to SFR of Port B. Use one CALL subroutine for sending the data to Port B and another one for time delay. Put a time delay in between each issuing of data to Port B.

Solution:

<table>
<thead>
<tr>
<th>LOC</th>
<th>OBJECT CODE</th>
<th>LINE</th>
<th>SOURCE TEXT</th>
</tr>
</thead>
<tbody>
<tr>
<td>00001</td>
<td>list F=PIC18F458</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00002</td>
<td>#include P18F458.INC</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00003</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000007</td>
<td>COUNT EQU 0x07 ;use location 07 for count-up</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000008</td>
<td>MYREG EQU 0x08 ;use location 08 for delay</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00006</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000000</td>
<td>ORG 0</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000000</td>
<td>MOVWF COUNT</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00000004</td>
<td>BACK CALL DISPLAY</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000009</td>
<td>MOVWF COUNT ;count = 0</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000010</td>
<td>GOTO BACK</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000011</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000012</td>
<td>;------ increment and put it in PORTB</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000013</td>
<td>DISPLAY INCF COUNT,F ;increment count</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000014</td>
<td>MOVFF COUNT,PORTB ;send it to PORTB</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000015</td>
<td>CALL DISPLAY</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000016</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000017</td>
<td>RET ;return to caller</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000018</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000019</td>
<td>;-------- this is the delay subroutine</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000020</td>
<td>ORG 300H ;put time delay at address 300H</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000300</td>
<td>DELAY MOVWL 0xFF ;WREG = 255, the counter</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000300</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000302</td>
<td>MOVWF MYREG</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000304</td>
<td>AGAIN NOP ;no operation wastes clock cycles</td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000306</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>0000308</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000030A</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000030C</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000030E</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000028</td>
<td>RETURN ;return to caller</td>
<td></td>
<td></td>
</tr>
<tr>
<td>000029</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>000030</td>
<td>END ;end of asm file</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Because RCALL is a 2-byte instruction, the target address of the subroutine must be within 2K because only 11 bits of the 2 bytes are used for the address. There is no difference between RCALL and CALL in terms of saving the program counter on the stack or the function of the RETURN instruction. The only difference is that the target address for CALL can be anywhere within the 2M address space of the PIC18 while the target address of RCALL must be within a 2K range. In many variations of the PIC18 marketed by Microchip Corporation, on-chip ROM is as low as 4K. In such cases, the use of RCALL instead of CALL can save a number of bytes of program ROM space.

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Of course, in addition to using compact instructions, we can program efficiently by having a detailed knowledge of all the instructions supported by a given microprocessor, and using them wisely. Look at Example 3-12.

**Example 3-12**

Rewrite the main part of Example 3-9 as efficiently as you can.

**Solution:**

```assembly
MYREG EQU 0x08
ORG 0
MOVLW 0x55 ;load WREG with 55H
BACK MOVWF PORTB ;issue value in PORTB SFR
RCALL DELAY ;time delay
COMPF PORTB,F ;complement Port B SFR
BRA BACK ;keep doing this indefinitely
;——this is the delay subroutine
DELAY MOVLW 0xFF ;WREG = 255, the counter
MOVF MYREG
AGAIN NOP ;no operation wastes clock cycles
NOP
DECF MYREG,F
BNZ AGAIN ;repeat until MYREG becomes 0
RETURN ;return to caller (MYREG = 0)
END ;end of asm file
```

**Example 3-13**

A developer is using the PIC18 microcontroller chip for a product. This chip has only 4K of on-chip flash ROM. Which of the instructions, CALL or RCALL, is more useful in programming this chip?

**Solution:**

The RCALL instruction is more useful because it is a 2-byte instruction. It saves two bytes each time the call instruction is used. However, we must use CALL if the target address is beyond the 2K boundary.

**Review Questions**

1. How wide is the size of the stack in the PIC18?
2. True or false. In the PIC18, control can be transferred anywhere within the 2M of code space by using the CALL instruction.
3. The CALL instruction is a(n) ___-byte instruction.
4. True or false. In the PIC18, control can be transferred anywhere within the 2M of code space by using the RCALL instruction.
5. With each CALL instruction, the stack pointer register, SP, is __________ (incremented, decremented).
6. With each RETURN instruction, the SP is __________ (incremented, decremented).
7. On power-up, the PIC uses location ______ as the first location of the stack.
8. How deep is the size of the stack in the PIC18?
9. The RCALL instruction is a(n) ___-byte instruction.
10. ______ (RCALL, CALL) takes more ROM space.
SECTION 3.3: PIC18 TIME DELAY AND INSTRUCTION PIPELINE

In the last section we used the DELAY subroutine. In this section we discuss how to generate various time delays and calculate exact delays for the PIC18. We will also discuss instruction pipelining and its impact on execution time.

Delay calculation for the PIC18

In creating a time delay using Assembly language instructions, one must be mindful of two factors that can affect the accuracy of the delay:

1. The crystal frequency: The frequency of the crystal oscillator connected to the OSC1 and OSC2 input pins is one factor in the time delay calculation. The duration of the clock period for the instruction cycle is a function of this crystal frequency.

2. The PIC design: Since the 1970s, both the field of IC technology and the architectural design of microprocessors have seen great advancements. Due to the limitations of IC technology and limited CPU design experience for many years, the instruction cycle duration was longer. Advances in both IC technology and CPU design in the 1980s and 1990s have made the single instruction cycle a common feature of many microcontrollers. Indeed, one way to increase performance without losing code compatibility with the older generation of a given family is to reduce the number of instruction cycles it takes to execute an instruction. One might wonder how microprocessors such as PIC are able to execute an instruction in one cycle. There are three ways to do that: (a) Use Harvard architecture to get the maximum amount of code and data into the CPU, (b) use RISC architecture features such as fixed-size instructions, and finally (c) use pipelining to overlap fetching and execution of instructions. We have examined the Harvard and RISC architectures in Chapter 2. Next, we discuss pipelining.

Pipelining

In early microprocessors such as the 8085, the CPU could either fetch or execute at a given time. In other words, the CPU had to fetch an instruction from memory, then execute it, and then fetch the next instruction, execute it, and so on. The idea of pipelining in its simplest form is to allow the CPU to fetch and execute at the same time, as shown in Figure 3-9.

<table>
<thead>
<tr>
<th>Non-pipeline</th>
<th>fetch 1</th>
<th>exec 1</th>
<th>fetch 2</th>
<th>exec 2</th>
<th>fetch 3</th>
<th>exec 3</th>
</tr>
</thead>
<tbody>
<tr>
<td>Pipeline</td>
<td>fetch 1</td>
<td>exec 1</td>
<td>fetch 2</td>
<td>exec 2</td>
<td>fetch 3</td>
<td>exec 3</td>
</tr>
<tr>
<td></td>
<td>fetch 3</td>
<td>exec 3</td>
<td>fetch 4</td>
<td>exec 4</td>
<td>fetch 5</td>
<td>exec 5</td>
</tr>
</tbody>
</table>

Figure 3-9. Pipeline vs. Non-pipeline
Instruction cycle time for the PIC

It takes a certain amount of time for the CPU to execute an instruction. In the PIC, this time is referred to as instruction cycles (referred to as machine cycles in some other CPUs). Because all the instructions in the PIC18 are either 2-byte or 4-byte, most instructions take no more than one or two instruction cycles to execute. (Notice, however, that some instructions such as BTFSS could take up to three instruction cycles.) Appendix A provides a list of PIC18 instructions and their cycles. In the PIC family, the length of the instruction cycle depends on the frequency of the oscillator connected to the PIC system. The crystal oscillator, along with on-chip circuitry, provide the clock source for the PIC CPU (see Chapter 8). In the PIC18, one instruction cycle consists of four oscillator periods. Therefore, to calculate the instruction cycle for the PIC, we take 1/4 of the crystal frequency, then take its inverse, as shown in Example 3-14.

<table>
<thead>
<tr>
<th>Example 3-14</th>
</tr>
</thead>
<tbody>
<tr>
<td>The following shows the crystal frequency for three different PIC-based systems. Find the period of the instruction cycle in each case.</td>
</tr>
<tr>
<td>(a) 4 MHz</td>
</tr>
</tbody>
</table>

Solution:
(a) 4/4 = 1 MHz; instruction cycle is 1/1 MHz = 1 μs (microsecond)
(b) 16 MHz/4 = 4 MHz; instruction cycle = 1/4 MHz = 0.25 μs = 250 ns (nanosecond)
(c) 20 MHz/4 = 5 MHz; instruction cycle = 1/5 MHz = 0.2 μs = 200 ns

Branch penalty

The overlapping of fetch and execution of the instruction is widely used in today’s microcontrollers such as PIC. For the concept of pipelining to work, we need a buffer or queue in which an instruction is prefetched and ready to be executed. In some circumstances, the CPU must flush out the queue. For example, when a branch instruction is executed, the CPU starts to fetch codes from the new memory location and the code in the queue that was fetched previously is discarded. In this case, the execution unit must wait until the fetch unit fetches the new instruction. This is called a branch penalty. The penalty is an extra instruction cycle to fetch the instruction from the target location instead of executing the instruction right below the branch. Remember that the instruction below the branch has already been fetched and is next in line to be executed when the CPU branches to a different address. This means that while the vast majority of PIC instructions take only one instruction cycle, some instructions take two or three instruction cycles. These are GOTO, BRA, CALL, and all the conditional branch instructions such as BNZ, BC, and so on. The conditional branch instruction can take only one instruction cycle if it does not jump. For example, the BNZ will jump if Z = 0 and that takes two instruction cycles. If Z = 1, then it falls through and it takes only one instruction cycle. See Examples 3-15 and 3-16.
Example 3-15

For a PIC18 system of 4 MHz, find how long it takes to execute each of the following instructions:

(a) MOVLLW  (b) DECF  (c) MOVWF
(d) ADDLW   (e) NOP   (f) GOTO
(g) CALL    (h) BNZ

Solution:

The machine cycle for a system of 4 MHz is 1 μs, as shown in Example 3-14. Appendix A shows instruction cycles for each of the above instructions. Therefore, we have:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Instruction cycles</th>
<th>Time to execute</th>
</tr>
</thead>
<tbody>
<tr>
<td>(a) MOVLLW</td>
<td>1</td>
<td>1 x 1 μs = 1 μs</td>
</tr>
<tr>
<td>(b) DECF</td>
<td>1</td>
<td>1 x 1 μs = 1 μs</td>
</tr>
<tr>
<td>(c) MOVWF</td>
<td>1</td>
<td>1 x 1 μs = 1 μs</td>
</tr>
<tr>
<td>(d) ADDLW</td>
<td>1</td>
<td>1 x 1 μs = 1 μs</td>
</tr>
<tr>
<td>(e) NOP</td>
<td>1</td>
<td>1 x 1 μs = 1 μs</td>
</tr>
<tr>
<td>(f) GOTO</td>
<td>2</td>
<td>2 x 1 μs = 2 μs</td>
</tr>
<tr>
<td>(g) CALL</td>
<td>2</td>
<td>2 x 1 μs = 2 μs</td>
</tr>
<tr>
<td>(h) BNZ</td>
<td>2/1</td>
<td>(2 μs taken, 1 μs if it falls through)</td>
</tr>
</tbody>
</table>

Example 3-16

Find the size of the delay of the code snippet below if the crystal frequency is 4 MHz:

Solution:

From Appendix A, we have the following machine cycles for each instruction of the DELAY subroutine:

```
MYREG EQU 0x08 ; use location 08 as counter

DELAY MOVLLW 0xFF 1
       MOVWF MYREG 1

AGAIN NOP 1
         NOP 1
         DECF MYREG,F 1
         BNZ AGAIN 2

RETURN 1
```

Therefore, we have a time delay of \([(255 \times 5) + 1 + 1 + 1] \times 1 \, \mu s = 1278 \, \mu s.
Notice that BNZ takes two instruction cycles if it jumps back, and takes only one when falling through the loop. That means the above number should be 1277 μs.
Delay calculation for PIC18

As seen in the last section, a delay subroutine consists of two parts: (1) setting a counter, and (2) a loop. Most of the time delay is performed by the body of the loop, as shown in Examples 3-17 and 3-18.

Example 3-18

Find the size of the delay in the following program if the crystal frequency is 4 MHz:

```assembly
MYREG EQU 0x08 ;use location 08 as counter
ORG 0
BACK MOVLLW 0x55 ;load WREG with 55H
MOVWF PORTB ;send 55H to port B
CALL DELAY ;time delay
MOVLLW 0xAA ;load WREG with AA (in hex)
MOVWF PORTB ;send AAH to port B
CALL DELAY
GOTO BACK ;keep doing this indefinitely

;— this is the delay subroutine
ORG 300H ;put time delay at address 300H
DELAY MOVLLW 0xFA ;WREG = 250, the counter
MOVWF MYREG
AGAIN NOP ;no operation wastes clock cycles
NOP
NOP
DECF MYREG, F ;repeat until MYREG becomes 0
BNZ AGAIN ;return to caller
RETURN
END ;end of asm file
```

Solution:

From Appendix A, we have the following machine cycles for each instruction of the DELAY subroutine:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Cycle</th>
</tr>
</thead>
<tbody>
<tr>
<td>DELAY</td>
<td>MOVLLW 0xFA</td>
</tr>
<tr>
<td></td>
<td>MOVWF MYREG</td>
</tr>
<tr>
<td>AGAIN</td>
<td>NOP</td>
</tr>
<tr>
<td></td>
<td>NOP</td>
</tr>
<tr>
<td></td>
<td>NOP</td>
</tr>
<tr>
<td></td>
<td>DECF MYREG, F</td>
</tr>
<tr>
<td></td>
<td>BNZ AGAIN</td>
</tr>
<tr>
<td></td>
<td>RETURN</td>
</tr>
</tbody>
</table>

Therefore, we have a time delay of \([(250 \times 6) + 1 + 1 + 1] \times 1 \mu s = 1503 \mu s\).

Very often we calculate the time delay based on the instructions inside the loop and ignore the clock cycles associated with the instructions outside the loop.
In Example 3-16, the largest value the MYREG register can take is 255; therefore, one way to increase the delay is to use NOP instructions in the loop. NOP, which stands for "no operation," simply wastes time, but takes 2 bytes of program ROM space and that is too heavy a price to pay for just one instruction cycle. A better way is to use a nested loop.

**Loop inside a loop delay**

Another way to get a large delay is to use a loop inside a loop, which is also called a *nested loop*. See Example 3-18. Compare that with Example 3-19 to see the disadvantage of using many NOPs.

### Example 3-18

For a instruction cycle of 1 μs, find the time delay in the following subroutine:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Cycle</th>
</tr>
</thead>
<tbody>
<tr>
<td>R2 EQU 0x7</td>
<td></td>
</tr>
<tr>
<td>R3 EQU 0x8</td>
<td></td>
</tr>
<tr>
<td>DELAY</td>
<td></td>
</tr>
<tr>
<td>MOVLW D'200'</td>
<td>1</td>
</tr>
<tr>
<td>MOVWF R2</td>
<td>1</td>
</tr>
<tr>
<td>AGAIN</td>
<td></td>
</tr>
<tr>
<td>MOVLW D'250'</td>
<td>1</td>
</tr>
<tr>
<td>MOVWF R3</td>
<td>1</td>
</tr>
<tr>
<td>HERE</td>
<td></td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>DECF R3, F</td>
<td>1</td>
</tr>
<tr>
<td>BNZ HERE</td>
<td>2</td>
</tr>
<tr>
<td>DECF R2, F</td>
<td>1</td>
</tr>
<tr>
<td>BNZ AGAIN</td>
<td>2</td>
</tr>
<tr>
<td>RETURN</td>
<td>1</td>
</tr>
</tbody>
</table>

**Solution:**

For the HERE loop, we have \((5 \times 250)\) 1 μs = 1250 μs. The AGAIN loop repeats the HERE loop 200 times; therefore, we have \(200 \times 1250 \mu s = 250000 \mu s\), if we do not include the overhead. However, the following instructions of the outer loop add to the delay:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Cycle</th>
</tr>
</thead>
<tbody>
<tr>
<td>AGAIN</td>
<td></td>
</tr>
<tr>
<td>MOVLW D'250'</td>
<td>1</td>
</tr>
<tr>
<td>MOVWF R3</td>
<td>1</td>
</tr>
<tr>
<td>..........</td>
<td></td>
</tr>
<tr>
<td>DECF R2, F</td>
<td>1</td>
</tr>
<tr>
<td>BNZ AGAIN</td>
<td>2</td>
</tr>
</tbody>
</table>

The above instructions at the beginning and end of the AGAIN loop add \(5 \times 200 \times 1 \mu s = 1000 \mu s\) to the time delay. We should also subtract 200 μs for the times BNZ HERE falls through. As a result we have \(250000 + 1000 - 200 = 250800 \mu s = 250.8\) milliseconds for the total time delay associated with the above DELAY subroutine. Notice that in the case of a nested loop, as in all other time delay loops, the time is approximate because we have ignored the first few instructions and the last instruction, RETURN, in the subroutine. NOP is a 2-byte instruction. There are 11 instructions in the above DELAY program, and all the instructions are 2-byte instructions. That means that the loop delay takes 22 bytes of ROM code space.
Example 3-19

Find the time delay for the following subroutine, assuming a crystal frequency of 4 MHz. Discuss the disadvantage of this over Example 3-18.

```
MYREG EQU 0x8

DELAY MOVLW D'200'
       MOVWF MYREG
       NOP
       NOP
       NOP
       NOP
       NOP
       NOP
       NOP
       NOP
       NOP
       NOP
       DECF MYREG, F
       BNZ AGAIN

RETURN
```

**Machine Cycle**

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Machine Cycle</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOVLW D'200'</td>
<td>1</td>
</tr>
<tr>
<td>MOVWF MYREG</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>NOP</td>
<td>1</td>
</tr>
<tr>
<td>DECF MYREG, F</td>
<td>1</td>
</tr>
<tr>
<td>BNZ AGAIN</td>
<td>2</td>
</tr>
<tr>
<td>RETURN</td>
<td>1</td>
</tr>
</tbody>
</table>

**Solution:**

The time delay inside the AGAIN loop is \(200(13 + 2) \times 1 \mu s = 3000 \mu s\). NOP is a 2-byte instruction, even though it does not do anything except to waste cycle time. There are 17 instructions in the above DELAY program, and all the instructions are 2-byte instructions. This means the loop delay takes 34 bytes of ROM code space, and gives us only a 3000 \(\mu s\) delay. That is the reason we use a nested loop instead of NOP instructions to create a time delay. Chapter 9 shows how to use PIC timers to create delays much more efficiently.

From these discussions we conclude that the use of instructions in generating time delay is not the most reliable method. To get more accurate time delay we use timers, as described in Chapter 9. We can use MPLAB’s simulator to verify delay time and number of cycles used. Meanwhile, to get an accurate time delay for a given PIC microcontroller, we must use an oscilloscope to measure the exact time delay.
Example 3-20

Write a program to toggle all the bits of SFR PORTB every 1 s. Assume that the crystal frequency is 10 MHz and the system is using a PIC18F458.

Solution:

; tested using MPLAB with PIC18F458 operating at 10 MHz
R2 EQU 0x2
R3 EQU 0x3
R4 EQU 0x4

MOVLW 0x55 ; load WREG with 55H
MOVWF PORTB ; send 55H to PORTB B
BACK CALL DELAY_500MS ; time delay
COMF PORTB ; complement PORTB
GOTO BACK ; keep doing this indefinitely

; this is the delay subroutine
DELAY_500MSEC

MOVLW D'20'  
MOVWF R4
BACK MOVLW D'100'
MOVWF R3
AGAIN MOVLW D'250'
MOVWF R2
HERE NOP
NOP
DBCF R2, F
BNZ HERE
DBCF R4, F
BNZ BACK
RETURN

Delay \(20 \times 100 \times 250 \times 5 \times 400 \text{ ns} = 1,000,000,000 \text{ ns} = 1,000,000 \mu\text{s} = 1 \text{ s.}\)

In this calculation, we have not included the overhead associated with the two outer loops. Use the MPLAB simulator to verify the delay.

**PIC multistage execution pipeline**

We can use a superpipeline to speed up execution of instructions. In superpipelining, the process of executing instructions is split into many small steps that are all executed in parallel. In this way, the execution of many instructions is overlapped. One limitation of superpipelining is that the speed of execution is limited to the slowest stage of the pipeline. Compare this to making pizza. You can split the process of making pizza into many stages, such as flattening the dough, putting on the toppings, and baking, but the process is limited to the slowest stage, baking, no matter how fast the rest of the stages are performed. What happens if we use two or three ovens for baking pizzas to speed up the process? This may
work for making pizza but not for executing programs, because in the execution of instructions we must make sure that the sequence of instructions is kept intact and that there is no out-of-step execution. In the PIC18, the execution unit takes 4 clock periods of the oscillator, as shown in Figure 3-10.

![Figure 3-10. Pipeline Activity After the Instruction Has Been Fetch](image)

Figure 3-10 explains why we divide the oscillator by 4 to get the instruction cycle. In Q1, we decode the instruction that is already fetched and sitting in the queue. In Q2, the operand is fetched from the file register. In Q3, the operation is performed: The adding of the two numbers is done. In Q4, the result is written into the destination register. In reality, one can construct the PIC18 superpipeline for four instructions, and is shown in Figure 3-11.

![Figure 3-11. Pipeline Activity for Both Fetch and Execute](image)

Notice, in many computer architecture books the process stage is referred to as execute and write to is called write back.

**Review Questions**

1. True or false. In the PIC18, the instruction cycle lasts 4 clock periods of the crystal frequency.
2. The minimum number of instruction cycles needed to execute a PIC18 instruction is _____.
3. For Question 2, what is the maximum number of cycles needed, and for which instructions?
4. Find the instruction cycle for a crystal frequency of 12 MHz.
5. Assuming a crystal frequency of 4 MHz, find the time delay associated with the loop section of the following DELAY subroutine:

```
DELAY
  MOVLW D'100'
  MOVWF MYREG
  HERE
  NOP
  NOP
  NOP
  NOP
  NOP
  DECF MYREG, F
  BNZ HERE
  RETURN
```

6. True or false. In the PIC18, the instruction cycle lasts 6 clock periods of the crystal frequency.

7. Find the machine cycle for a PIC18 if the crystal frequency is 8 MHz.

8. True or false. In the PIC, the instruction fetching and execution are done at the same time.

9. True or false. BRA and CALL will always take 2 instruction cycles.

10. True or false. The BNZ instruction will always take 2 instruction cycles.

---

**SUMMARY**

The flow of a program proceeds sequentially, from instruction to instruction, unless a control transfer instruction is executed. The various types of control transfer instructions in Assembly language include conditional and unconditional branches, and call instructions.

Looping in PIC Assembly language is performed using an instruction to decrement a counter and to jump to the top of the loop if the counter is not zero. This is accomplished with the BNZ instruction. Other branch instructions jump conditionally, based on the value of the carry flag, the Z flag, or other bits of the status register. Unconditional branches can be long or short, depending on the location of the target address. Special attention must be given to the effect of CALL and RCALL instructions on the stack.

---

**PROBLEMS**

**SECTION 3.1: BRANCH INSTRUCTIONS AND LOOPING**

1. In the PIC, looping action with the instruction “BNZ target” is limited to ____ iterations.

2. If a conditional branch is not taken, what is the next instruction to be executed?

3. In calculating the target address for a branch, a displacement is added to the contents of register ________.

4. The mnemonic BRA stands for ________ and it is a(n) __-byte instruction.

5. The GOTO instruction is a(n) ____-byte instruction.

6. What is the advantage of using BRA over GOTO?
7. True or false. The target of a BNZ can be anywhere in the 2M address space.
8. True or false. All PIC branch instructions can branch to anywhere in the 2M
   address space.
9. Which of the following instructions are 2-byte instructions.
   (a) BZ   (b) BNC   (c) GOTO   (d) BRA
10. Dissect the BRA instruction, indicating how many bits are used for the operand
    and the opcode, and indicate how far it can branch.
11. True or false. All conditional branches are 2-byte instructions.
12. Show code for a nested loop to perform an action 1,000 times.
13. Show code for a nested loop to perform an action 100,000 times.
14. Find the number of times the following loop is performed:
    
15. The target address of a BNZ is backward if the second byte of opcode is
    _______ (negative, positive).
16. The target address of a BNZ is forward if the second byte of opcode is
    _______ (negative, positive).

SECTION 3.2: CALL INSTRUCTIONS AND STACK

17. CALL is a(n) ___-byte instruction.
18. RCALL is a(n) ___-byte instruction.
19. True or false. The RCALL target address can be anywhere in the 2M address
    space.
20. True or false. The CALL target address can be anywhere in the 2M address
    space.
21. When CALL is executed, how many locations of the stack are used?
22. When RCALL is executed, how many locations of the stack are used?
23. Upon reset, the first available location of the stack is ________.
24. Describe the action associated with the RETURN instruction.
25. Give the size of the stack in PIC18.
26. In PIC18, which address is pushed into the stack and the stack pointer when a
    call instruction is executed.

SECTION 3.3: PIC18 TIME DELAY AND INSTRUCTION PIPELINE

27. Find the oscillator frequency if the instruction cycle = 1.25 μs.
28. Find the instruction cycle if the crystal frequency is 20 MHz.
29. Find the instruction cycle if the crystal frequency is 10 MHz.
30. Find the instruction cycle if the crystal frequency is 16 MHz.
31. True or false. The CALL and RCALL instructions take the same amount of time to execute even though one is a 4-byte instruction and the other is a 2-byte instruction.

32. Find the time delay for the delay subroutine shown below if the system has a PIC18 with a frequency of 4 MHz:

```
MOVLW    D'200'
MOVWF    REGA
BACK
MOVLW    D'100'
MOVWF    REGB
HERE
NOP
DECF    REGB, F
BNZ    HERE
DECF    REGA, F
BNZ    BACK
```

33. Find the time delay for the delay subroutine shown below if the system has a PIC18 with a frequency of 16 MHz:

```
MOVLW    D'200'
MOVWF    REGA
BACK
MOVLW    D'100'
MOVWF    REGB
HERE
NOP
NOP
DECF    REGB, F
BNZ    HERE
DECF    REGA, F
BNZ    BACK
```

34. Find the time delay for the delay subroutine shown below if the system has a PIC18 with a frequency of 4 MHz:

```
MOVLW    D'200'
MOVWF    REGA
BACK
MOVLW    D'250'
MOVWF    REGB
HERE
NOP
DECF    REGB
BNZ    HERE
DECF    REGA
BNZ    BACK
```

35. Find the time delay for the delay subroutine shown below if the system has a PIC18 with a frequency of 10 MHz:

```
MOVLW    D'200'
MOVWF    REGA
BACK
MOVLW    D'100'
MOVWF    REGB
NOP
NOP
NOP
HERE
DECF    REGB, F
BNZ    HERE
DECF    REGA, F
BNZ    BACK
```
ANSWERS TO REVIEW QUESTIONS

SECTION 3.1: BRANCH INSTRUCTIONS AND LOOPING

1. Branch if not zero
2. True
3. 2
4. Z flag of status register
5. 4

SECTION 3.2: CALL INSTRUCTIONS AND STACK

1. 21-bit
2. True
3. 4
4. False
5. Incremented
6. Decremented
7. 1
8. 31 locations (21 × 31)
9. 2
10. CALL

SECTION 3.3: PIC18 TIME DELAY AND INSTRUCTION PIPELINE

1. True
2. 1
3. 2 and CALL. Also, the DECFSZ instruction can take up to 3 cycles
4. 12 MHz / 4 = 3 MHz, and MC = 1/3 MHz = 0.333 μs
5. \( [100 (1 + 1 + 1 + 1 + 1 + 1 + 2)] \times 1 \mu s = 800 \mu s = 0.8 \text{ milliseconds} \)
7. 8 MHz / 4 = 2 MHz; machine cycle is 1 / 2 MHz = 500 ns
8. True
9. True
10. False. Only if it branches to the target address.
CHAPTER 4

PIC I/O PORT PROGRAMMING

OBJECTIVES

Upon completion of this chapter, you will be able to:

- List all the ports of the PIC18
- Describe the dual role of PIC18 pins
- Code Assembly language to use the ports for input or output
- Explain the dual role of Ports A, B, C, and D
- Code PIC instructions for I/O handling
- Code I/O bit-manipulation programs for the PIC
- Explain the bit-addressability of PIC ports
This chapter describes I/O port programming of the PIC18 with many examples. In Section 4-1, we describe I/O access using byte-size data and in Section 4-2, bit manipulation of the I/O ports is discussed in detail.

**SECTION 4.1: I/O PORT PROGRAMMING IN PIC18**

In the PIC18 family, there are many ports for I/O operations, depending on which family member you choose. Examine Figure 4-1 for the PIC18F458 40-pin chip. A total of 33 pins are set aside for the five ports PORTA, PORTB, PORTC, PORTD, and PORTE. The rest of the pins are designated as \( V_{dd} \) (\( V_{cc} \)), \( V_{ss} \) (GND), OSC1, OSC2, MCLR (reset), and another set of \( V_{dd} \) and \( V_{ss} \). They are discussed in Chapter 8.

![Figure 4-1. PICF458 Pin Diagram](image)

**I/O port pins and their functions**

The number of ports in the PIC18 family varies depending on the number of pins on the chip. The 18-pin PIC18 has ports A and B only, while the 64-pin version has ports A through F, and the 80-pin PIC18 has ports A through L, as shown in Table 4-1. The 40-pin PIC18F458 has five ports. They are PORTA, PORTB, PORTC, PORTD, and PORTE. To use any of these ports as an input or output port, it must be programmed, as we will explain throughout this section. In addition to
Table 4-1: Number of Ports in PIC18 Family Members

<table>
<thead>
<tr>
<th>Pins</th>
<th>18-pin</th>
<th>28-pin</th>
<th>40-pin</th>
<th>64-pin</th>
<th>80-pin</th>
</tr>
</thead>
<tbody>
<tr>
<td>Chip</td>
<td>PIC18F1220</td>
<td>PIC18F2220</td>
<td>PIC18F458</td>
<td>PIC18F6525</td>
<td>PIC18F8525</td>
</tr>
<tr>
<td>Port A</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port B</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port C</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port D</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port E</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port F</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port G</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port H</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port I</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
<tr>
<td>Port J</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
<td>X</td>
</tr>
</tbody>
</table>

Note: X indicates that the port is available.

being used for simple I/O, each port has some other functions such as ADC, timers, interrupts, and serial communication pins. Figure 4-1 also shows alternate functions for the PIC18F458 pins. We will study all these alternate functions in future chapters. In this chapter we focus on the simple I/O function of the PIC18 family. Not all ports have 8 pins. For example, in the PIC18F458, Port A has 7 pins; Ports B, C, and D each have 8 pins; and Port E has only 3 pins. Each port has three SFRs associated with it, as shown in Table 4-2. They are designated as PORTx, TRISx, and LATx. For example, for Port B we have PORTB, TRISB, and LATB. Note that TRIS stands for TRISTate and LAT stands for LATch. Next, we describe how to access the SFRs associated with the ports.

**TRIS register role in outputting data**

Each of the Ports A–E in the PIC18F458 can be used for input or output. The TRISx SFR is used solely for the purpose of making a given port an input or output port. For example, to make a port an output, we write 0s to the TRISx register. In other words, to output data to any of the pins of the Port B, we must first put 0s into the TRISB register to make it an output port, and then send the data to the Port B SFR itself.

Table 4-2: Ports' SFR Addresses for PIC18F458

<table>
<thead>
<tr>
<th>Port</th>
<th>Address</th>
</tr>
</thead>
<tbody>
<tr>
<td>PORTA</td>
<td>F80H</td>
</tr>
<tr>
<td>PORTB</td>
<td>F81H</td>
</tr>
<tr>
<td>PORTC</td>
<td>F82H</td>
</tr>
<tr>
<td>PORTD</td>
<td>F83H</td>
</tr>
<tr>
<td>PORTE</td>
<td>F84H</td>
</tr>
<tr>
<td>LATA</td>
<td>F89H</td>
</tr>
<tr>
<td>LATB</td>
<td>F8AH</td>
</tr>
<tr>
<td>LATC</td>
<td>F8BH</td>
</tr>
<tr>
<td>LATD</td>
<td>F8CH</td>
</tr>
<tr>
<td>LATE</td>
<td>F8DH</td>
</tr>
<tr>
<td>TRISA</td>
<td>F92H</td>
</tr>
<tr>
<td>TRISB</td>
<td>F93H</td>
</tr>
<tr>
<td>TRISC</td>
<td>F94H</td>
</tr>
<tr>
<td>TRISD</td>
<td>F95H</td>
</tr>
<tr>
<td>TRISE</td>
<td>F96H</td>
</tr>
</tbody>
</table>
The following code will toggle all 8 bits of Port B forever with some time delay in between “on” and “off” states:

```
MOVLW 0x0 ;WREG = 00
MOVWF TRISB ;make Port B an output port 0000 0000
L1
MOVLW 0x55 ;WREG = 55h
MOVWF PORTB ;put 55h on port B pins
CALL DELAY
MOVLW 0xAA ;WREG = AAh
MOVWF PORTB ;put AAh on port B pins
CALL DELAY
GOTO L1
```

It must be noted that unless we activate the TRIS bit (set it to zero), the data will not go from the port register to the pins of the PIC. This means that if we remove the first two lines of the above code, the 55H and AAH values will not get to the pins. They will be sitting in the SFR of Port B inside the CPU.

To see the role of the TRISx register in allowing the data to go from Portx to the pins, examine Figure 4-3 and Figure 4-4. If you are unfamiliar with the internals of logic gates, see Appendix C for an overview. Notice that the CMOS “on” and “off” states in Figure 4-2 are taken from Appendix C.

Note that upon reset, all ports have value FFH in their TRIS registers. This means all ports are configured as input as we will see next.

---

Figure 4-2. CMOS States for P and N Transistors
Figure 4-3. Outputting (Writing) 0 to a Pin in the PIC18

Figure 4-4. Outputting (Writing) 1 to a Pin in the PIC18
TRIS register role in inputting data

To make a port an input port, we must first put 1s into the TRISx register for that port, and then bring in (read) the data present at the pins. Notice that 0 stands for out and 1 for in. This is easy to remember because 0 and 0 look alike the same way that I looks like 1. The following code will get the data present at the pins of port C and send it to port B indefinitely, after adding the value 5 to it:

```
MOVLW B'00000000' ;WREG = 00000000 (binary)
MOVWF TRISB ;Port B an output port (0 for O)
MOVLW B'11111111' ;WREG = 11111111 (binary)
MOVWF TRISC ;Port C an input port (1 for I)
L2 MOVF PORTC,W ;move data from Port C to WREG
ADDLW 5 ;add some value to it
MOVWF PORTB ;send it to Port B
GOTO L2 ;continue forever
```

Another, more efficient, version of the program is as follows:

```
CLRF TRISB ;clear TRISB (Port B an output port)
SETF TRISC ;set TRISC (Port C an input port)
L2 MOVF PORTC,W ;get data from port C
ADDLW 5 ;add some value
MOVWF PORTB ;send it to port B
BRA L2
```

Again, it must be noted that unless we activate the TRIS bits (by putting 1s there), the data will not be brought into the WREG register from the pins of Port C. To see the role of the TRISx register in allowing the data to come into the CPU from the pins, examine Figures 4-5 and 4-6.

Figure 4-5. Inputting (Reading) 0 from a Pin in the PIC18
Port A

Port A occupies a total of 7 pins (RA0–RA6), but for the PIC18F458, pin A6 is used for the OSC2 pin. A6 is not available if we use a crystal oscillator to provide frequency to the PIC18 chip, as we will see in Chapter 8.

To use the pins of Port A as both input and output ports, each bit must be connected externally to the pin by enabling the bits of the TRISA register. For example, the following code will continuously send out to Port A the alternating values of 55H and AAH:

```
;toggle all bits of PORTA

MOVLW B'00000000'; WREG = 00000000 (binary)
MOVWF TRISA ; make Port A an output port (0 for Out)
L1 MOVLW 0x55 ; WREG = 55H
MOVWF PORTA ; put 55h on Port A pins
CALL DELAY
MOVLW 0xAA ; WREG = AAh
MOVWF PORTA ; put AAh on Port A pins
CALL DELAY
GOTO L1
```

It must be noted that 55H (01010101) when complemented becomes AAH (10101010). Although by sending 55H and AAH to Port A continuously, we toggle all 8 bits of the Port A register, only 6 pins (RA0–RA5) will show the toggling data.

Figure 4-6. Inputting (Reading) 1 from a Pin in the PIC18

CHAPTER 4: PIC I/O PORT PROGRAMMING 135
Port A as input

In order to make all the bits of Port A an input, TRISA must be programmed by writing 1 to all the bits. In the following code, Port A is configured first as an input port by writing all 1s to register TRISA, and then data is received from Port A and saved in some RAM location of the fileReg:

MYREG EQU 0X20 ;save it here

MOVLW B'11111111' ;WREG = 11111111 (binary)
MOVWF TRISA ;make Port A an input port (1 for In)
MOVF PORTA,W ;move from fileReg of Port A to WREG
MOVWF MYREG ;save it in fileReg of MYREG

Port B

Port B occupies a total of 8 pins (RB0–RB7). To use the pins of Port B as both input and output ports, each bit must be connected externally to the pin by enabling the bits of register TRISB.

For example, the following code will continuously send out the alternating values of 55H and AAH to Port B:

;toggle all bits of PORTB

MOVLW B'00000000' ;WREG = 00
MOVWF TRISB ;make Port B an output port
L1
MOVLW 0x55 ;WREG = 55h
MOVWF PORTB ;put 55h on port B pins
CALL DELAY
MOVLW 0xAA ;WREG = AAh
MOVWF PORTB ;put AAh on port B pins
CALL DELAY
GOTO L1

Port B as input

In order to make all the bits of Port B an input, TRISB must be programmed by writing 1 to all the bits. In the following code, Port B is configured first as an input port by writing all 1s to register TRISB, and then data is received from Port B and saved in some RAM location of the fileReg:

MYREG EQU 0X25 ;save it here

MOVLW B'11111111' ;WREG = 11111111 (binary)
MOVWF TRISB ;make Port B an input port (1 for In)
MOVF PORTB,W ;move from fileReg of Port B to WREG
MOVWF MYREG ;save it in fileReg
Dual role of Ports A and B

The PIC18 multiplexes an analog-to-digital converter through Port A to save I/O pins. The alternate functions of the pins for Port A are shown in Table 4-3. We will show how to use Port A's ADC in Chapter 13. Because many projects use an ADC, we do not use Port A for simple I/O functions. The PIC18 multiplexes some other functions through Port B to save pins. The alternate functions of the pins for Port B are shown in Table 4-4. We will show how to use the alternate functions of Port B in future chapters.

<table>
<thead>
<tr>
<th>Bit</th>
<th>Function</th>
<th>Bit</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0</td>
<td>AN0/CVREF</td>
<td>RB0</td>
<td>INT0</td>
</tr>
<tr>
<td>RA1</td>
<td>AN1</td>
<td>RB1</td>
<td>INT1</td>
</tr>
<tr>
<td>RA2</td>
<td>AN2/VREF-</td>
<td>RB2</td>
<td>INT2/CANTX</td>
</tr>
<tr>
<td>RA3</td>
<td>AN3/VREF+</td>
<td>RB3</td>
<td>CANRX</td>
</tr>
<tr>
<td>RA4</td>
<td>TOCKI</td>
<td>RB4</td>
<td></td>
</tr>
<tr>
<td>RA5</td>
<td>AN4/SS/LVDIN</td>
<td>RB5</td>
<td>PGM</td>
</tr>
<tr>
<td>RA6</td>
<td>OSC2/CLKO</td>
<td>RB6</td>
<td>PGC</td>
</tr>
<tr>
<td></td>
<td></td>
<td>RB7</td>
<td>PGD</td>
</tr>
</tbody>
</table>

Table 4-3: Port A Alternate Functions

Table 4-4: Port B Alternate Functions

Port C

Port C occupies a total of 8 pins (RC0–RC7). To use the pins of Port C as both input and output ports, each bit must be connected externally to the pin by enabling the bits of register TRISC. For example, the following code will continuously send out the alternating values of 55H and AAH to Port C:

```
;toggle all bits of PORTB

MOVLW B'00000000' ;WREG = 00
MOVF TRISC ;make Port C an output port
L1
MOVLW 0x55 ;WREG = 55h
MOVWF PORTC ;put 55h on Port C pins
CALL DELAY
MOVLW 0xAA ;WREG = AAh
MOVWF PORTC ;put AAh on Port C pins
CALL DELAY
GOTO L1
```
Port C as input

In order to make all the bits of Port C an input, TRISC must be programmed by writing 1 to all the bits. In the following code, Port C is configured first as an input port by writing all 1s to register TRISC, and then data is received from Port C and saved in some RAM location of the fileReg:

```
MYREG EQU 0x20 ; save it here
MOVLW B'11111111' ; WREG = 11111111 (binary)
MOVWF TRISC ; make Port C an input port (1 for In)
MOVF PORTC,W ; move from fileReg of Port C to WREG
MOVWF MYREG ; save it in fileReg
```

Port D

Port D occupies a total of 8 pins (RD0–RD7). To use the pins of Port D as both input and output ports, each bit must be connected externally to the pin by enabling the bits of register TRISD. For example, the following code will continuously send out to Port D the alternating values of 55H and AAH:

```
; toggle all bits of PORTD
CLR D TRISD ; make Port D an output port
L1
MOVLW 0x55 ; WREG = 55h
MOVWF PORTD ; put 55h on Port D pins
CALL DELAY
MOVLW 0xAA ; WREG = AAh
MOVWF PORTD ; put AAh on Port D pins
CALL DELAY
BRA L1 ; we can use GOTO
```

Port D as input

In order to make all the bits of Port D an input, TRISD must be programmed by writing 1 to all the bits. In the following code, Port D is configured first as an input port by writing all 1s to register TRISD, and then data is received from Port D and saved in some RAM location of the fileReg:

```
MYREG EQU 0x20 ; save it here
SETF TRISD ; TRISD = 11111111 (binary) = PORTD = Input
MOVF PORTD,W ; move from Port D to WREG
MOVWF MYREG ; save it in fileReg
```
**Dual role of Ports C and D**

The alternate functions of the pins for Port C are shown in Table 4-5. We will show how to use Port C’s alternate functions in future chapters. The alternate functions of the pins for Port D are shown in Table 4-6. We will show how to use Port D’s alternate functions in future chapters.

<table>
<thead>
<tr>
<th>Table 4-5: Port C Alternate Functions</th>
<th>Table 4-6: Port D Alternate Functions</th>
</tr>
</thead>
<tbody>
<tr>
<td>Bit</td>
<td>Function</td>
</tr>
<tr>
<td>RC0</td>
<td>T1OSO/T1CKI</td>
</tr>
<tr>
<td>RC1</td>
<td>T1OSI</td>
</tr>
<tr>
<td>RC2</td>
<td>CCP1</td>
</tr>
<tr>
<td>RC3</td>
<td>SCK/SCL</td>
</tr>
<tr>
<td>RC4</td>
<td>SDI/SDA</td>
</tr>
<tr>
<td>RC5</td>
<td>SDO</td>
</tr>
<tr>
<td>RC6</td>
<td>TX/CK</td>
</tr>
<tr>
<td>RC7</td>
<td>RX/DT</td>
</tr>
</tbody>
</table>

**Port E**

Port E occupies a total of 3 pins (RE0–RE2) in the PIC18F458/4580. Port E is used for 3 additional analog inputs or simple I/O: AN5, AN6, and AN7. Just like other ports, Port E has alternate functions. We will show how to use them in future chapters.

**Different ways of accessing the entire 8 bits**

In the following code, as in many previous I/O examples, the entire 8 bits of Port B are accessed:

```assembly
;toggle all bits of PORTB
MOVLW 0x0 ;WREG = 00
MOVWF TRISB ;make Port B an output port
L1 MOVLW 0x55 ;WREG = 55h
MOVWF PORTB ;put 55h on Port B pins
CALL DELAY
MOVLW 0xA ;WREG = AAh
MOVWF PORTB ;put AAh on Port B pins
CALL DELAY
GOTO L1
```
The previous code toggles all the bits of Port B continuously. Another variation of the above code follows:

```
CLRF TRISB ; make Port B an output port  
L1  MOVLW 0x55 ; WREG = 55h  
    MOVWF PORTB ; put 55h on Port B pins  
    CALL DELAY  
    MOVLW 0xAA ; WREG = AAh  
    MOVWF PORTB ; put AAh on Port B pins  
    CALL DELAY  
    GOTO LI
```

The following is another way of doing the same thing:

```
CLRF TRISB ; make Port B an output port  
MOVLW 0x55 ; WREG = 55h  
MOVWF PORTB ; put 55h on Port B pins  
L2  COMF PORTB, F ; toggle bits of Port B  
    CALL DELAY  
    BRA L2
```

The above code uses a technique called *read-modify-write*.

**Read followed by write I/O operation**

Due to the timing issue, we must be careful not to have two I/O operations one right after the other. Examine the following rewrite of an earlier code fragment in which data was read from PORTC and sent to PORTB:

```
CLRF TRISB ; clear TRISB to make PORTB an output port  
SETF TRISC ; set TRISC all Is (Port C as Input)  
L4  MOVF PORTC,W ; get data from Port C into WREG  
    NOP ; NEED some NOP to ensure data is in WREG  
    MOVWF PORTB ; before it is sent to Port B  
    BRA L4 ; keep doing it
```

We need a NOP (or some other instruction) to make sure that the data is written into WREG before it is read for outputting to Port B. This is called *data dependency* in CPU design. This type of data dependency is commonly referred to as RAW (Read-After-Write). The NOP will introduce a bubble into the pipeline to remove data dependency due to RAW. See Figure 4-7. One way to avoid this problem is to use the MOVFF instruction, which is a 4-byte instruction. This is coded as follows:

```
CLRF TRISB ; make Port B an output port  
SETF TRISC ; TRISC = FFh (Port C Input)  
L5  MOVFF PORTC, PORTB ; get from Port C and send to PORTB  
    BRA L5 ; keep doing it
```
The RAW (Read – After – Write) for two consecutive instructions.

INSTRUCTION

**Fetch 1**

MOVF PORTC,W ;Read PORTC into WREG

**Fetch 2**

MOVWF PORTB ;Write WREG to PORTB

The RAW (Read – After – Write) for two consecutive instructions.

INSTRUCTION

**Fetch 1**

MOVF PORTC,W

**Fetch 2**

NOP ;Bubble in Pipeline

**Fetch 3**

MOVWF PORTB

N = No Operation
D = Decode the instruction
R = Read the operand
P = Process
W = Write the result to destination register

**Figure 4-7. Pipeline for Read Followed by Write I/O**

**Ports status upon reset**

Upon reset, all ports have value FFH on their TRIS register, as shown in Table 4-7. This makes them input ports upon reset.

**Table 4-7: Reset Values of TRIS Registers for PIC18**

<table>
<thead>
<tr>
<th>Register</th>
<th>Reset Value (Binary)</th>
</tr>
</thead>
<tbody>
<tr>
<td>TRISA</td>
<td>11111111</td>
</tr>
<tr>
<td>TRISB</td>
<td>11111111</td>
</tr>
<tr>
<td>TRISC</td>
<td>11111111</td>
</tr>
<tr>
<td>TRISD</td>
<td>11111111</td>
</tr>
</tbody>
</table>

*Note: All ports are input ports upon reset.*

**Review Questions**

1. There are total of _____ ports in the PIC18F458.
2. True or false. All of the PIC18F458 ports have 8 pins.
3. List all PIC18F458 ports that have 8 pins.
4. True or false. Upon power-up, the I/O pins are configured as output ports.
5. Code a simple program to send 99H to Port B and Port C.
6. To make Port B an output port, we must place _____ in register ______.
7. To make Port B an input port, we must place _____ in register ______.
Example 4-1

Write a test program for the PIC18 chip to toggle all the bits of PORTB, PORTC, and PORTD every 1/4 of a second. Assume a crystal frequency of 4 MHz.

Solution:

; tested with MPLAB for the PIC18F458 and XTAL = 4 MHz

list P=PIC18F458
#include P18F458.INC

R1 equ 0x07
R2 equ 0x08

ORG 0
    CLRF TRISB ; make Port B an output port
    CLRF TRISC ; make Port C an output port
    CLRF TRISD ; make Port D an output port
    MOVWF PORTB ; WREG = 55h
    MOVWF PORTC ; put 55h on Port B pins
    MOVWF PORTD ; put 55h on Port C pins
    MOVWF PORTD ; put 55h on Port D pins
L3    COMF PORTB,F ; toggle bits of Port B
    COMF PORTC,F ; toggle bits of Port C
    COMF PORTD,F ; toggle bits of Port D
    CALL QDELAY ; quarter of a second delay
    BRA L3

;---------- 1/4 SECOND DELAY

QDELAY
    MOVLW D'200'
    MOVF R1
D1    MOVLW D'250'
    MOVF R2
D2    NOP
    NOP
    DECF R2, F
    BNZ D2
    DECF R1, F
    BNZ D1
    RETURN
END

Calculations:

4 MHz / 4 = 1 MHz
1 / 1 MHz = 1 µs
Delay = 250 × 200 × 5 MC × 1 µs = 250,000 µs (if we include the overhead, we will have 250,800. See Example 3-17 in the previous chapter.)

Use the MPLAB simulator to verify the delay size.
SECTION 4.2: I/O BIT MANIPULATION PROGRAMMING

In this section we further examine the PIC18 I/O instructions. We pay special attention to I/O bit manipulation because it is a powerful and widely used feature of the PIC family.

I/O ports and bit-addressability

Sometimes we need to access only 1 or 2 bits of the port instead of the entire 8 bits. A powerful feature of PIC I/O ports is their capability to access individual bits of the port without altering the rest of the bits in that port. For all PIC ports, we can access either all 8 bits or any single bit without altering the rest. Table 4-8 lists the single-bit instructions for the PIC18. Although the instructions in Table 4-9 can be used for any registers in the data RAM file register, I/O port operations use them most often. We will see the use of these instructions throughout future chapters.

Table 4-8: Single-Bit (Bit-Oriented) Instructions for PIC18

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>BSF fileReg,bit</td>
<td>Bit Set fileReg (set the bit: bit = 1)</td>
</tr>
<tr>
<td>BCF fileReg,bit</td>
<td>Bit Clear fileReg (clear the bit: bit = 0)</td>
</tr>
<tr>
<td>BTG fileReg,bit</td>
<td>Bit Toggle fileReg (complement the bit)</td>
</tr>
<tr>
<td>BTFSC fileReg,bit</td>
<td>Bit test fileReg, skip if clear (skip next instruction if bit = 0)</td>
</tr>
<tr>
<td>BTFSS fileReg,bit</td>
<td>Bit test fileReg, skip if set (skip next instruction if bit = 1)</td>
</tr>
</tbody>
</table>

Table 4-9: Single-Bit Addressability of Ports for PIC18F458/4580

<table>
<thead>
<tr>
<th>PORT</th>
<th>PORTB</th>
<th>PORTC</th>
<th>PORTD</th>
<th>PORTE</th>
<th>Port Bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0</td>
<td>RB0</td>
<td>RC0</td>
<td>RD0</td>
<td>RE0</td>
<td>D0</td>
</tr>
<tr>
<td>RA1</td>
<td>RB1</td>
<td>RC1</td>
<td>RD1</td>
<td>RE1</td>
<td>D1</td>
</tr>
<tr>
<td>RA2</td>
<td>RB2</td>
<td>RC2</td>
<td>RD2</td>
<td>RE2</td>
<td>D2</td>
</tr>
<tr>
<td>RA3</td>
<td>RB3</td>
<td>RC3</td>
<td>RD3</td>
<td></td>
<td>D3</td>
</tr>
<tr>
<td>RA4</td>
<td>RB4</td>
<td>RC4</td>
<td>RD4</td>
<td></td>
<td>D4</td>
</tr>
<tr>
<td>RA5</td>
<td>RB5</td>
<td>RC5</td>
<td>RD5</td>
<td></td>
<td>D5</td>
</tr>
<tr>
<td>RA6</td>
<td>RB6</td>
<td>RC6</td>
<td>RD6</td>
<td></td>
<td>D6</td>
</tr>
<tr>
<td>RA7</td>
<td>RB7</td>
<td>RC7</td>
<td>RD7</td>
<td></td>
<td>D7</td>
</tr>
</tbody>
</table>

Next we describe all these instructions and examine their usage.

BSF (bit set fileReg)

To set HIGH a single bit of a given fileReg, we use the syntax “BSF fileReg, bit_num” where fileReg can be any location in the file register and bit_num is the desired bit number from 0 to 7. Although the bit-oriented instructions can be used for manipulation of bits D0–D7 of any file register, they are mostly used for I/O ports in embedded systems. For example, “BSF PORTB, 5” sets HIGH bit 5 of Port B.
Example 4-2

An LED is connected to each pin of Port D. Write a program to turn on each LED from pin D0 to pin D7. Call a delay module before turning on the next LED.

Solution:

```assembly
CLRF TRISD   ;make PORTD an output port
BSF PORTD, 0 ;bit set turns on RD0  
CALL DELAY  ;delay before next one
BSF PORTD, 1 ;turn on RD1        
CALL DELAY  ;delay before next one
BSF PORTD, 2
CALL DELAY
BSF PORTD, 3
CALL DELAY
BSF PORTD, 4
CALL DELAY
BSF PORTD, 5
CALL DELAY
BSF PORTD, 6
CALL DELAY
BSF PORTD, 7
CALL DELAY

BCF (bit clear fileReg)

To clear a single bit of a given fileReg, we use the syntax “BCF fileReg, bit_number”. Remember that for I/O ports, we must activate the appropriate bit in the TRISx register if we want the pin to reflect the changes. For example, the following code toggles pin RB2 continuously:

```assembly
BCF TRISB, 2   ;bit = 0, make RB2 an output pin
AGAIN BSF PORTB, 2;bit set (RB2 = high)
CALL DELAY
BCF PORTB, 2  ;bit clear(RB2 = low)
CALL DELAY
BRA AGAIN
```
Write the following programs:
(a) Create a square wave of 50% duty cycle on bit 0 of Port C.
(b) Create a square wave of 66% duty cycle on bit 3 of Port C.

Solution:

(a) The 50% duty cycle means that the “on” and “off” states (or the high and low portions of the pulse) have the same length. Therefore, we toggle RC0 with a time delay between each state.

```
BCF TRISC,0 ;clear TRIS bit for RC0 = out
HERE BSF PORTC,0 ;set to HIGH RC0 (RC0 = 1)
CALL DELAY ;call the delay subroutine
BCF PORTC,0 ;RC0 = 0
CALL DELAY ;keep doing it
```

Another way to write the above program is:

```
BCF TRISC,0 ;make RC0 = out
HERE BTG PORTC,0 ;complement bit 0 of PORTC
CALL DELAY ;call the delay subroutine
BRA HERE ;keep doing it
```

(b) A 66% duty cycle means that the “on” state is twice the “off” state.

```
BCF TRISC,3 ;clear TRISC3 bit for output
BACK BSF PORTC,3 ;RC3 = 1
CALL DELAY ;call the delay subroutine
CALL DELAY ;twice for 66%
BCF PORTC,3 ;RC3 = 0
CALL DELAY ;call delay once for 33%
BRA BACK ;keep doing it
```
BTG (bit toggle fileReg)

To toggle a single bit of a given fileReg, we use the syntax “BTG fileReg, bit_number”.

```
BCF TRISB, 2          ;make RB2 an output pin
BACK BTG PORTB, 2    ;toggle pin RB2 only
CALL DELAY
BRA BACK
```

Notice that RB2 is the third bit of Port B (the first bit is RB0, the second bit is RB1, etc.). This is shown in Table 4-9. See Example 4-2 for an example of bit manipulation of I/O bits.

Notice in Example 4-2 that unused portions of Port C are undisturbed. This single-bit addressability of I/O ports is one of most powerful features of the PIC microcontroller and is one of the reasons that many designers choose the PIC over other microcontrollers. We will see the use of the bit-addressability of I/O ports in future chapters.

Checking an input pin

To make decisions based on the status of a given bit in the file register, we use the instructions BTFSC (bit test fileReg skip if clear) and BTFSS (bit test fileReg skip if set). These single-bit instructions are widely used for I/O operations. They allow you to monitor a single pin and make a decision depending on whether it is 0 or 1. Again it must be noted that the instructions BTFSC and BTFSS can be used for any bits of the file register, including the I/O ports A, B, C, D, and so on.

BTFSS (bit test fileReg, skip if set)

To monitor the status of a single bit for HIGH, we use the BTFSS instruction. This instruction tests the bit and skips the next instruction if it is HIGH. Example 4-4 shows how it is used.

```
| 1010 | bbba | ffff | ffff |
```

\[0 \leq f \leq FF\]
\[0 \leq b \leq 7\]

BTFSC (bit test fileReg, skip if clear)

To monitor the status of a single bit for LOW, we use the BTFSC instruction. This instruction tests the bit and skips the instruction right below it if the bit is LOW. Example 4-5 shows how it is used.

```
| 1011 | bbba | ffff | ffff |
```

\[0 \leq f \leq FF\]
\[0 \leq b \leq 7\]
Example 4-4

Write a program to perform the following:
(a) Keep monitoring the RB2 bit until it becomes HIGH;
(b) When RB2 becomes HIGH, write value 45H to Port C, and also send a HIGH-to-LOW pulse to RD3.

Solution:

BSF TRISB, 2 ; make RB2 an input
CLRF TRISC ; make PORTC an output port
BCF PORTD, 3 ; make RD3 an output
MOVLW 0x45 ; WREG = 45h
AGAIN BTFSS PORTB, 2 ; bit test RB2 for HIGH
BRA AGAIN ; keep checking if LOW
MOVWF PORTC ; issue WREG to Port C
BSF PORTD, 3 ; bit set fileReg RD3 (H-to-L)
BCF PORTD, 3 ; bit clear fileReg RD3 (L)

In this program, instruction “BTFSS PORTB, 2” stays in the loop as long as RB2 is LOW. When RB2 becomes HIGH, it skips the branch instruction to get out of the loop, and writes the value 45H to Port C. It also sends a HIGH-to-LOW pulse to RD3.

Example 4-5

Assume that bit RB3 is an input and represents the condition of a door alarm. If it goes LOW, it means that the door is open. Monitor the bit continuously. Whenever it goes LOW, send a HIGH-to-LOW pulse to port RC5 to turn on a buzzer.

Solution:

BSF TRISB, 3 ; make RB3 an input
BCF TRISC, 5 ; make RC5 an output
HERE BTFSC PORTB, 3 ; keep monitoring RB3 for HIGH
BRA HERE ; stay in the loop
BSF PORTC, 5 ; make RC5 HIGH
BCF PORTC, 5 ; make RC5 LOW for H-to-L
BRA HERE
Monitoring a single bit

We can also use the bit test instructions to monitor the status of a single bit and make a decision to perform an action. See Examples 4-6 and 4-7.

**Example 4-6**

A switch is connected to pin RB2. Write a program to check the status of SW and perform the following:
(a) If SW = 0, send the letter ‘N’ to PORTD.
(b) If SW = 1, send the letter ‘Y’ to PORTD.

**Solution:**

```assembly
BSF TRISB, 2 ; make RB2 an input
CLRF TRISD ; make PORTD an output port
AGAIN BTFSS PORTB, 2 ; bit test RB2 for HIGH
BRA OVER ; it must be LOW
MOVLW A'Y' ; WREG = 'Y' ASCII letter Y
MOVWF PORTD ; issue WREG to PORTD
GOTO AGAIN ; we can use BRA too
OVER MOVLW A'N' ; WREG = 'N' ASCII letter N
MOVWF PORTD ; issue WREG to PORTD
GOTO AGAIN ; we can use BRA too
```

---

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Example 4-7

A switch is connected to pin RB2. Write a program to check the status of SW and perform the following:
(a) If \( SW = 0 \), send letter ‘N’ to PORTD.
(b) If \( SW = 1 \), send letter ‘Y’ to PORTD.

Use the BTFSC instruction to check the SW status. This is another version of Example 4-6 using the BTFSC instruction instead of BTFSS.

Solution:

```
BSF TRISB,2 ;make RB2 an input
CLRF TRISD ;make PORTD an output port
AGAIN BTFSC PORTB,2 ;bit test RB2 for LOW
    BRA OVER ;it must be HIGH
    MOVLW A'N' ;WREG = 'N' ASCII letter N
    MOVWF PORTD ;issue WREG to PORTD
    BRA AGAIN ;we can use GOTO
OVER MOVLW A'Y' ;WREG = 'Y' ASCII letter Y
    MOVWF PORTD ;issue WREG to PORTD
    BRA AGAIN ;we can use GOTO
```

```
Diagram:
```

CHAPTER 4: PIC I/O PORT PROGRAMMING 149
Reading a single bit

We can also use the bit test instructions to read the status of a single bit and send it to another bit or save it. This is shown in Examples 4-8 and 4-9.

Example 4-8

A switch is connected to pin RB0 and an LED to pin RB7. Write a program to get the status of SW and send it to the LED.

Solution:

```assembly
BSF TRISB,0 ; make RB0 an input
BCF TRISB,7 ; make RB7 an output
AGAIN BTFSS PORTB,0 ; bit test RB0 for HIGH
GOTO OVER ; it must be LOW (BRA is OK too)
BSF PORTB,7
GOTO AGAIN ; we can use BRA too
OVER BCF PORTB,7
GOTO AGAIN ; we can use BRA too
```

Example 4-9

A switch is connected to pin RB0. Write a program to get the status of SW and save it in D0 of fileReg location 0x20.

Solution:

```assembly
MYBITREG EQU 0x20 ; set aside loc 0x20 reg

BSF TRISB,0 ; make RB0 an input
AGAIN BTFSS PORTB,0 ; bit test RB0 for HIGH
GOTO OVER ; it must be LOW (BRA is OK too)
BSF MYBITREG,0 ; set bit 0 of fileReg
GOTO AGAIN ; we can use BRA too
OVER BCF MYBITREG,0 ; clear bit 0 of fileReg
GOTO AGAIN ; we can use BRA too
```
Reading input pins vs. LATx port

In reading a port, some instructions read the status of the port pins while others read the status of an internal port latch called LATx. Therefore, when reading ports there are two possibilities:

1. Read the status of the input pin.
2. Read the internal latch of the LAT register.

We must make a distinction between these two categories of instructions because confusion between them is a major source of errors in PIC programming, especially where external hardware is concerned. We will discuss these instructions shortly. However, readers must study and understand the material on this topic and on the internal working of ports that is given in Appendix C.2. Examine the structure of the ports in Figure 4-8 once again. In addition to the PORTx and TRISx register, the LATx register is the third important register associated with the PIC18 ports.

![Figure 4-8. LATx Register Role in Reading a Port or Latch](image)

Reading LATx for ports

Some instructions read the contents of an internal port latch instead of reading the status of an external pin. Table 4-10 provides a list of these instructions. For example, consider the “COMP PORTB” instruction. The sequence of actions taken when such an instruction is executed is as follows:

1. The instruction reads the internal latch of the LATB and brings that data into the CPU.
2. This data is complemented.
3. The result is rewritten back to the LATB latch.
4. The data on the pins are changed only if the TRISB bits are cleared to 0s.

It is very rare that we use an instruction to read the latch register, such as “COMF LATB, F”, although it is a valid instruction.

From the above discussion, we conclude that the instructions that read the port latch normally read a latch value, perform an operation, then rewrite it back to the port latch. This is called read-modify-write. To use the read-modify-write, the port must be configured as output.

### Table 4-10: Some of the Read-Modify-Write Instructions

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDWF</td>
<td>fileReg,d Add WREG to f</td>
</tr>
<tr>
<td>BSF</td>
<td>fileReg,bit Bit Set fileReg (set the bit: bit = 1)</td>
</tr>
<tr>
<td>BCF</td>
<td>fileReg,bit Bit Clear fileReg (clear the bit: bit = 0)</td>
</tr>
<tr>
<td>COMF</td>
<td>fileReg,d Complement f</td>
</tr>
<tr>
<td>INCF</td>
<td>fileReg,d Increment f</td>
</tr>
<tr>
<td>SUBWF</td>
<td>fileReg,d Subtract WREG from f</td>
</tr>
<tr>
<td>XORWF</td>
<td>fileReg,d Exclusive-OR WREG with f</td>
</tr>
</tbody>
</table>

### Review Questions

1. True or false. The instruction “BSF PORTB, 1” makes pin RB1 HIGH while leaving other pins of PORTB unchanged, if bit 1 of the TRISB bits is configured for output.
2. Show one way to toggle the pin RB7 continuously using PIC instructions.
3. Using the instruction “BTFSS PORTC, 5” assumes that bit RC5 is an ________ (input, output) pin.
4. Write instructions to get the status of RB2 and put it on RB0.
5. Write instructions to toggle both bits of RD7 and RD0 continuously.

---

**CAUTION**

We strongly recommend that you study Section C.2 (Appendix C) before connecting any external hardware to your PIC system. Failure to use the right instruction or the right connection to port pins can damage the ports of your PIC chip.
SUMMARY

This chapter focused on the I/O ports of the PIC. The five ports of the PIC18F458, PORTA, PORTB, PORTC, PORTD, and PORTE, were explored. These ports can be used for input or output. All the ports have alternate functions. The three registers associated with each port are PORTx, TRISx, and LATx. Their role in I/O manipulation was examined. Then, I/O instructions of the PIC were explained, and numerous examples were given. We also showed the bit-addressability of PIC ports.

PROBLEMS

SECTION 4.1: I/O PORT PROGRAMMING IN PIC18

1. The PIC18F458 has a DIP package of ___pins.
2. In PIC18F458, how many pins are assigned to VCC and GND?
3. In the PIC18F458, how many pins are designated as I/O port pins?
4. How many pins are designated as PORTA in the 40-pin DIP package and what are their numbers?
5. How many pins are designated as PORTB in the 40-pin DIP package and what are their numbers?
6. How many pins are designated as PORTC in the 40-pin DIP package and what are their numbers?
7. How many pins are designated as PORTD in the 40-pin DIP package and what are their numbers?
8. Upon reset, all the bits of ports are configured as ____ (input, output).
9. For the PIC18, which register must be programmed in order to be used as simple I/O?
10. Explain the role of TRISx and PORTx in I/O operations.
11. Write a program to get 8-bit data from PORTC and send it to ports PORTB and PORTD.
12. Write a program to get 8-bit data from PORTD and send it to ports PORTB and PORTC.
13. Which pins are for RxD and TxD?
14. Give RAM data location in the file register assigned to Ports A–C and their TRIS registers for the PIC18F458.
15. Write a program to toggle all the bits of PORTB and PORTC continuously (a) using AAH and 55H (b) using the COMF instruction.

SECTION 4.2: I/O BIT MANIPULATION PROGRAMMING

16. Which ports of the PIC18 are bit-addressable?
17. What is the advantage of bit-addressability for PIC ports?
18. When RB2 is accessed as a single-bit port, it is designated as _____.
19. Is the instruction “COMF PORTB” a valid instruction?
20. Write a program to toggle RB2 and RB5 continuously without disturbing the rest of the bits.

21. Write a program to toggle RD3, RD7, and RC5 continuously without disturbing the rest of the bits.

22. Write a program to monitor bit RC3. When it is HIGH, send 55H to PORTD.

23. Write a program to monitor the RB7 bit. When it is LOW, send 55H and AAH to PORTC continuously.

24. Write a program to monitor the RE0 bit. When it is HIGH, send 99H to PORTB. If it is LOW, send 66H to PORTC.

25. Write a program to monitor the RB5 bit. When it is HIGH, make a LOW-to-HIGH-to-LOW pulse on RB3.

26. Write a program to get the status of RC3 and put it on RC4.

27. The RB4 refers to which bit of PORTB?

28. Create a flowchart and write a program to get the statuses of RD7 and RD6 and put them on RC0 and RC7, respectively.

ANSWERS TO REVIEW QUESTIONS

SECTION 4.1: I/O PORT PROGRAMMING IN PIC18

1. 5
2. False
3. PORTB, PORTC, and PORTD
4. False
5. MOVLW 0x99
   MOVWF PORTB
   MOVWF PORTC
6. 00, TRISB
7. FFH, TRISB

SECTION 4.2: I/O BIT MANIPULATION PROGRAMMING

1. True
2. BCF TRISB, 7
   H1 BTG PORTB, 7
   BRA H1
3. Input
4. BSF TRISB, 2
   BCF TRISB, 0
   AGAIN
   BTFSS PORTB, 2
   BRA OVER
   BSF PORTB, 0
   BRA AGAIN
   OVER
   BCF PORTB, 0
   BRA AGAIN
5. BCF TRISD, 0
   BCF TRISD, 7
   H2 BTG PORTD, 0
   BTG PORTD, 7
   BRA H2
CHAPTER 5

ARITHMETIC, LOGIC INSTRUCTIONS, AND PROGRAMS

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Define the range of numbers possible in PIC unsigned data
- Code addition and subtraction instructions for unsigned data
- Perform addition of BCD data
- Code PIC unsigned data multiplication instructions
- Code PIC programs for division
- Code PIC Assembly language logic instructions AND, OR, and EX-OR
- Use PIC logic instructions for bit manipulation
- Use compare and skip instructions for program control
- Code PIC rotate instructions and data serialization
- Explain the BCD (binary coded decimal) system of data representation
- Contrast and compare packed and unpacked BCD data
- Code PIC programs for ASCII and BCD data conversion
This chapter describes all PIC arithmetic and logic instructions. Program examples are given to illustrate the application of these instructions. In Section 5.1 we discuss instructions and programs related to addition, subtraction, multiplication, and division of unsigned numbers. Signed numbers are discussed in Section 5.2. In Section 5.3, we discuss the logic instructions AND, OR, and XOR, as well as the COMPARE instruction. The ROTATE instruction and data serialization are discussed in Section 5.4. In Section 5.5 we provide some real-world applications such as BCD and ASCII conversion.

SECTION 5.1: ARITHMETIC INSTRUCTIONS

Unsigned numbers are defined as data in which all the bits are used to represent data, and no bits are set aside for the positive or negative sign. This means that the operand can be between 00 and FFH (0 to 255 decimal) for 8-bit data.

Addition of unsigned numbers

In order to add numbers together in the PIC, the WREG register must be involved. One form of the ADD instruction is

\[ \text{ADDLW } K \; ; \text{WREG} = \text{WREG} + K \]

The sum is stored in the WREG register. The instruction could change any of the C, DC, Z, N, or OV bits of the status register, depending on the operands involved. The effect of the ADDLW instruction on N and OV is discussed in Section 5.3 because these bits are relevant mainly in signed number operations. Look at Example 5-1.

<table>
<thead>
<tr>
<th>Example 5-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>Show how the flag register is affected by the following instructions.</td>
</tr>
<tr>
<td>MOVLW 0xF5 ;WREG = F5 hex</td>
</tr>
<tr>
<td>ADDLW 0xFB ;WREG = F5 + 0B = 00 and C = 1</td>
</tr>
</tbody>
</table>

**Solution:**

\[
\begin{array}{c}
F5H \quad \text{1111 0101} \\
+ \quad \text{0BH} \quad + \quad \text{0000 1011} \\
100H \quad \text{0000 0000}
\end{array}
\]

After the addition, register WREG contains 00 and the flags are as follows:

- C = 1 because there is a carry out from D7.
- Z = 1 because the result in WREG is zero.
- DC = 1 because there is a carry from D3 to D4.

**ADDWF and addition of individual bytes**

Instruction "ADDWF fileReg, d" allows the addition of WREG and individual bytes residing in RAM locations of the file register. Notice that WREG must be involved because memory-to-memory arithmetic operations are never
allowed in PIC Assembly language. To calculate the sum of any number of operands, the carry flag should be checked after the addition of each operand. Example 5-2 uses location 7 of the file register to accumulate carries as the operands are added to WREG. In Chapter 6, the loop version of this program will be shown for any number of bytes.

Example 5-2

Assume that file register RAM locations 40–43H have the following hex values. Write a program to find the sum of the values. At the end of the program, location 6 of the file register should contain the low byte and location 7 the high byte of the sum.

40 = (7D)
41 = (EB)
42 = (C5)
43 = (5B)

Solution:

L_Byte EQU 0x6 ; assign RAM location 6 to L_byte of sum
H_Byte EQU 0x7 ; assign RAM location 7 to H_byte of sum

MOVLW 0 ; clear WREG (WREG = 0)
MOVF H_Byte ; H_Byte = 0
ADDWF 0x40,W ; WREG = 0 + 7DH = 7DH , C = 0
BNC N_1 ; branch if C = 0
INCF H_Byte,F ; increment (now H_Byte = 0)
N_1 ADDWF 0x41,W ; WREG = 7D + EB = 68H and C = 1
BNC N_2 ;
INCF H_Byte,F ; C = 1, increment (now H_Byte = 1)
N_2 ADDWF 0x42,W ; WREG = 68 + C5 = 2D and C = 1
BNC N_3 ;
INCF H_Byte ; C = 1, increment (now H_Byte = 2)
N_3 ADDWF 0x43,W ; WREG = 2D + 5B = 88H and C = 0
BNC N_4 ;
INCF H_Byte,F ; (H_Byte = 2)
N_4 MOVWF L_Byte ; now L_Byte = 88h

At the end the fileReg location 6 = (8B), and location 7 = (02) because 7D + EB + C5 + 5B + 30 = 28BH. We can use the register indirect addressing mode to do this program much more efficiently. Chapter 6 shows how to do that.

**ADDWFC and addition of 16-bit numbers**

When adding two 16-bit data operands, we need to be concerned with the propagation of a carry from the lower byte to the higher byte. This is called *multi-byte addition* to distinguish it from the addition of individual bytes. The instruction ADDWFC (ADDW and fileReg with carry) is used on such occasions.
For example, look at the addition of 3CE7H + 3B8DH, as shown next.

\[
\begin{array}{c}
\text{1} \\
\text{3C E7} \\
+ \text{3B 8D} \\
\hline
\text{78 74}
\end{array}
\]

When the first byte is added, there is a carry (E7 + 8D = 74, CY = 1). The carry is propagated to the higher byte, which results in 3C + 3B + 1 = 78 (all in hex). Example 5-3 shows the above steps in a PIC program.

Example 5-3

Write a program to add two 16-bit numbers. The numbers are 3CE7H and 3B8DH. Assume that fileReg location 6 = (8D) and location 7 = (3B). Place the sum in fileReg locations 6 and 7; location 6 should have the lower byte.

Solution:

;location 6 = (8D)
;location 7 = (3B)

MOVLW 0xE7 ;load the low byte now (WREG = E7H)
ADDWF 0x6,F ;F = W + F = E7 + 8D = 74 and CY = 1
MOVLW 0x3C ;load the high byte (WREG = 3CH)
ADDWFC 0x7,F ;F = W + F + carry, adding the upper byte
;with Carry from lower byte
;F = 3C + 3B + 1 = 78H (all in hex)

Notice the use of ADDWF for the lower byte and ADDWFC for the higher byte.

BCD (binary coded decimal) number system

BCD stands for binary coded decimal. BCD is needed because in everyday life we use the digits 0 to 9 for numbers, not binary or hex numbers. Binary representation of 0 to 9 is called BCD (see Figure 5-1). In computer literature, one encounters two terms for BCD numbers: (1) unpacked BCD, and (2) packed BCD. We describe each one next.

Unpacked BCD

In unpacked BCD, the lower 4 bits of the number represent the BCD number, and the rest of the bits are 0. Example: “0000 1001” and “0000 0101” are unpacked BCD for 9 and 5, respectively. Unpacked BCD requires 1 byte of memory, or an 8-bit register, to contain it.

Packed BCD

In packed BCD, a single byte has two BCD numbers in it: one in the lower 4 bits, and one in the upper 4 bits. For example, “0101 1001” is packed BCD for 59H. Only 1 byte of memory is needed to store the packed BCD operands. Thus
one reason to use packed BCD is that it is twice as efficient in storing data.

There is a problem with adding BCD numbers, which must be corrected. The problem is that after adding packed BCD numbers, the result is no longer BCD. Look at the following.

```
MOV LW 0x17
ADD LW 0x28
```

Adding these two numbers gives 0011 1111B (3FH), which is not BCD! A BCD number can only have digits from 0000 to 1001 (or 0 to 9). In other words, adding two BCD numbers must give a BCD result. The result above should have been 17 + 28 = 45 (0100 0101). To correct this problem, the programmer must add 6 (0110) to the low digit: 3F + 06 = 45H. The same problem could have happened in the upper digit (for example, in 52H + 87H = D9H). Again, 6 must be added to the upper digit (D9H + 60H = 139H) to ensure that the result is BCD (52 + 87 = 139). This problem is so pervasive that most microprocessors such as the PIC18 have an instruction to deal with it. In the PIC18 instruction “DAW” is designed to correct the BCD addition problem. This is discussed next.

**DAW instruction**

The DAW (decimal adjust WREG) instruction in the PIC18 is provided to correct the aforementioned problem associated with BCD addition. The mnemonic “DAW” works only with an operand in the WREG register. The DAW instruction will add 6 to the lower nibble or higher nibble if needed; otherwise, it will leave the result alone. The following example will clarify these points.

```
MOV LW 0x47 ;WREG = 47H first BCD operand
ADD LW 0x25 ;hex(binary) addition (WREG = 6CH)
DAW ;adjust for BCD addition (WREG = 72H)
```

After the program is executed, register WREG will contain 72H (47 + 25 = 72). Note that the “DAW” instruction works only on WREG.

**Summary of DAW action**

After any instruction,
1. If the lower nibble (4 bits) is greater than 9, or if DC = 1, add 0110 to the lower 4 bits.
2. If the upper nibble is greater than 9, or if C = 1, add 0110 to the upper 4 bits.

In reality there is no use for the DC (auxiliary carry) flag bit other than for BCD addition and correction.

```
MOV LW 0x00 ;WREG = 0
ADD LW 0x09 ;WREG = 0x09
ADD LW 0x08 ;WREG = 0x11, DC = 1
DAW ;WREG = 0x17 (9 + 8 = 17)
```

As another example, examine the case of adding 55H and 77H. This will result in CCH, which is incorrect as far as BCD is concerned.
\textbf{Hex} \hspace{1cm} \textbf{BCD} \\
57 \hspace{1cm} 0101 0111 \\
+ 77 \hspace{1cm} + 0111 0111 \\
CE \hspace{1cm} 1100 1110 \\
+ 66 \hspace{1cm} + 0110 0110 \\
134 \hspace{1cm} 1 0011 0100 \hspace{1cm} \text{Note } C = 1

Note that unlike other processors, the PIC does not require the use of arithmetic instructions prior to execution of the “DAW” instruction. Look at the following case where no arithmetic instruction is used.

\begin{verbatim}
MOVLW 0x0C ;WREG = 00001100
DAW ;WREG = 00001100 + 00000110 = 00010010 = 0x12
\end{verbatim}

Examine Example 5-4.

\begin{table}[h]
\centering
\begin{tabular}{|c|c|}
\hline
Example 5-4 & \\
\hline
\end{tabular}
\caption{Assume that 5 BCD data items are stored in RAM locations starting at 40H, as shown below. Write a program to find the sum of all the numbers. The result must be in BCD.}
\end{table}

\textbf{Solution:}

\begin{verbatim}
L_Byte EQU 0x6 ;assign RAM loc 6 to L_Byte of sum
H_Byte EQU 0x7 ;assign RAM loc 7 to H_Byte of sum

MOVLW 0 ;clear WREG (WREG = 0)
MOVWF H_Byte ;H_Byte = 0
ADDWF 0x40,W ;WREG = 0 + 71H = 71H, C = 0
DAW ;WREG = 71H
BNC N_1 ;branch if C = 0
INCF H_Byte,F ;
N_1 ADDWF 0x41,W ;WREG = 71 + 88 = F9H
DAW ;WREG = 59H AND C = 1
BNC N_2 ;
INCF H_Byte,F ;C = 1, increment (now H_Byte = 1)
N_2 ADDWF 0x42,W ;WREG = 59 + 69 = C2 and Carry = 0
DAW ;WREG = 28 and C = 1
BNC N_3 ;
INCF H_Byte ;C = 1, increment (now H_Byte = 2)
N_3 ADDWF 0x43,W ;WREG = 28 + 97 = BFH and C = 0
DAW ;WREG = 25 and C = 1
BNC N_4 ;
INCF H_Byte,F ;(now H_Byte = 3)
N_4 MOVWF L_Byte ;Now L_Byte = 25H
\end{verbatim}

After this code executes, fileReg location 6 = (03), and WREG = 25 because 71 + 88 + 69 + 97 = 325H. We can use the register indirect addressing mode and looping to do this program much more efficiently. Chapter 6 shows how to do that.
Subtraction of unsigned numbers

In many microprocessors, there are two different instructions for subtraction: SUB and SUBB (subtract with borrow). In the PIC18 we have four instructions for subtraction: SUBLW, SUBWF, SUBWFB, and SUBFWB. The last two are subtract with borrow. Notice that we use the C (carry) flag for the borrow. We now will examine each of these commands.

**SUBLW K (WREG = K – WREG)**

In subtraction, the PIC microcontrollers (indeed, all modern CPUs) use the 2’s complement method. Although every CPU contains adder circuitry, it would be too cumbersome (and take too many transistors) to design separate subtracter circuitry. For this reason, the PIC uses adder circuitry to perform the subtraction command. Assuming that the PIC is executing a simple subtract instruction and that C = 0 prior to the execution of the instruction, one can summarize the steps of the hardware of the CPU in executing the SUBLW instruction for unsigned numbers as follows:

1. Take the 2’s complement of the subtrahend (WREG operand).
2. Add it to the minuend (K operand).

These two steps are performed for every SUB instruction by the internal hardware of the CPU, regardless of the source of the operands, provided that the addressing mode is supported. It is after these two steps that the result is obtained and the flags are set. Example 5-5 illustrates the two steps.

**Example 5-5**

Show the steps involved in the following.

```
MOV LW 0x23 ; load 23H into WREG (WREG = 23H)
SUBLW 0x3F ; WREG = 3F – WREG
```

**Solution:**

```
K = 3F 0011 1111
- WREG = 23 0010 0011 + 1101 1101 (2’s complement)
1C
1 0001 1100
C = 1, D7 = N = 0 (result is positive)
```

The flags would be set as follows: C = 1, N = 0 (notice that D7 is the negative flag). The programmer must look at the N (or C) flag to determine if the result is positive or negative.

After the execution of SUB, if N = 0 (or C = 1), the result is positive; if N = 1 (or C = 0), the result is negative and the destination has the 2’s complement of the result. Normally, the result is left in 2’s complement, but the NEGF (negate, which is 2’s complement) instruction can be used to change it. Another SUB instruction in PIC is SUBWF (Destination = fileReg – WREG). This is shown in Example 5-6 along with the NEGF instruction.
Example 5-6

Write a program to subtract 4C - 6E.

Solution:

```
MYREG EQU 0x20
MOVLW 0x4C ; load WREG (WREG = 4CH)
MOVWF MYREG ; MYREG = 4CH
MOVLW 0x6E ; WREG = 6EH
SUBWF MYREG,W ; WREG = MYREG - WREG. 4C - 6E = DE, N = 1
BNN NEXT ; if N = 0 (C = 1), jump to NEXT target
NEGF WREG ; take 2's complement of WREG
NEXT MOVWF MYREG ; save the result in MYREG
```

The following are the steps after the SUBWF instruction:

<table>
<thead>
<tr>
<th>Dec</th>
<th>0100</th>
<th>1100</th>
</tr>
</thead>
<tbody>
<tr>
<td>4C</td>
<td>1100</td>
<td>0100</td>
</tr>
<tr>
<td>-6E</td>
<td>1110</td>
<td>0110 2's comp</td>
</tr>
<tr>
<td>-22</td>
<td></td>
<td>1101 1110</td>
</tr>
</tbody>
</table>

After SUBWF, we have N = 1 (or C = 0), and the result is negative, in 2’s complement. Then it falls through and NEGF will be executed. The NEGF instruction will take the 2's complement, and we have MYREG = 22H.

**SUBWFB (dest = fileReg - W - Borrow) subtract with borrow**

This instruction is used for multibyte numbers and will take care of the borrow of the lower byte. If C = 0 prior to executing the SUBWFB instruction, it also subtracts 1 from the result. See Example 5-7.

**SUBFWB (dest = WREG - fileReg - Borrow)**

This instruction is also used for multibyte numbers and will take care of the borrow of the lower byte. Notice the difference between SUBWFB and SUBFWB. See Appendix A for the description of these two instructions.

Example 5-7

Write a program to subtract two 16-bit numbers. The numbers are 2762H - 1296H. Assume fileReg location 6 = (62) and location 7 = (27). Place the difference in fileReg locations 6 and 7; loc 6 should have the lower byte.

Solution:

```
loc 6 = (62)
loc 7 = (27)

MOVLW 0x96 ; load the low byte (WREG = 96H)
SUBWF 0x6,F ; F = F - W = 62 - 96 = CCH, C = borrow = 0, N = 1
MOVLW 0x12 ; load the high byte (WREG = 12H)
SUBWFB 0x7,F ; F = F - W - B, sub byte with the borrow
              ; F = 27 - 12 - 1 = 14H
```

After the SUBWFB, loc 6 has = 62H - 96H = CCH and the carry flag is set to 0, indicating there is a borrow (notice, N = 1). Because C = 0, when SUBWFB is executed the fileReg location 7 has = 27H - 12H - 1 = 14H. Therefore, we have 2762H - 1296H = 14CCH.
The C flag in subtraction for PIC

Notice that the PIC18 is different from other CPU’s such as the x86 and the 8051 when it comes to the carry flag in subtract operations. In those CPUs, the carry is inverted by the CPU itself and we examine the C flag to see if the result is positive or negative. In the PIC18, if C = 0, the result is negative. That is the reason in subtract with borrow we have F = F – W – 5. Use the MPLAB simulator to gain additional insight into this important issue.

Multiplication of unsigned numbers

The PIC supports byte-by-byte multiplication only. The bytes are assumed to be unsigned data. The syntax is as follows:

\[
\text{MULLW } K \quad ; W \times K \text{ and 16-bit is result is in PRODH:PRODL}
\]

In byte-by-byte multiplication, one of the operands must be in the WREG register, and the second operand must be a literal K value. After multiplication, the result is in the special function registers PRODH and PRODL; the lower byte is in PRODL, and the upper byte is in PRODH. See Table 5-1. The following example multiplies 25H by 65H.

\[
\begin{align*}
\text{MOVLW } 0x25 & \quad ; \text{load } 25H \text{ to WREG (WREG = 25H)} \\
\text{MULLW } 0x65 & \quad ; 25H \times 65H = E99 \text{ where} \\
 & \quad ; \text{PRODH = 0EH and PRODL = 99H}
\end{align*}
\]

Table 5-1: Unsigned Multiplication Summary (MULLW K)

<table>
<thead>
<tr>
<th>Byte x Byte</th>
<th>Byte 1</th>
<th>Byte 2</th>
<th>Result</th>
</tr>
</thead>
<tbody>
<tr>
<td>WREG K</td>
<td>PRODH = high byte, PRODL = low byte</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Note: Multiplication of operands larger than 8-bit takes some manipulation.

Division of unsigned numbers

There is no single instruction for the division of byte/byte numbers in the PIC18. We can write a program to perform division by repeated subtraction. In dividing a byte by a byte, the numerator is placed in a fileReg and the denominator is subtracted from it repeatedly. The quotient is the number of times we subtracted and the remainder is in fileReg upon completion. See the following example.

\[
\begin{align*}
\text{NUM} & \quad \text{EQU } 0x19 \quad ; \text{set aside fileReg} \\
\text{MYQ} & \quad \text{EQU } 0x20 \\
\text{MYNMB} & \quad \text{EQU } \text{D'95'} \\
\text{MYDEN} & \quad \text{EQU } \text{D'10'} \\
\text{CLR} & \quad \text{MYQ} \quad ; \text{quotient } = 0 \\
\text{MOVLW} & \quad \text{MYNMB} \quad ; \text{WREG } = 95 \\
\text{MOVWF} & \quad \text{NUM} \quad ; \text{numerator } = 95 \\
\text{MOVLW} & \quad \text{MYDEN} \quad ; \text{WREG } = \text{denominator } = 10 \\
\text{B1} & \quad \text{INCF} \text{ MYQ},F \quad ; \text{increment quotient for every 10 subtr} \\
\text{SUBWF} & \quad \text{NUM,F} \quad ; \text{subtract } 10 \ (F = F - W) \\
\text{BC} & \quad \text{B1} \quad ; \text{keep doing it until } C = 0 \\
\text{DEC} & \quad \text{MYQ,F} \quad ; \text{once too many} \\
\text{ADDWF} & \quad \text{NUM,F} \quad ; \text{add 10 back to get remainder}
\end{align*}
\]
An application for division

Sometimes an ADC (analog-to-digital converter) is connected to a port and the ADC represents some quantity such as temperature or pressure. The 8-bit ADC provides data in hex in the range of 00–FFH. This hex data must be converted to decimal. We do that by dividing it by 10 repeatedly, saving the remainders, as shown in Example 5-8.

Example 5-8

Assume that file register location 0x15 has value FD (hex). Write a program to convert it to decimal. Save the digits in locations 0x22, 0x23, and 0x24, where the least-significant digit is in 0x22

Solution:

```
#include <P18F458.INC>
;PIC Assembly Language Program for division (by repeated subtraction)
;(Byte/Byte)

NUME EQU 0x15 ;RAM location for NUME
QU EQU 0x20 ;RAM location for quotient
RMND_L EQU 0x22
RMND_M EQU 0x23
RMND_H EQU 0x24
MYNUM EQU 0xFD ;FDH = 253 in decimal
MYDEN EQU D'10';253/10
ORG 0H ;start at address 0
MOVLW MYNUM ;WREG = 253, the numerator
MOVWF NUME ;load numerator
MOVLW MYDEN ;WREG = 10, the denominator
CLRF QU,F ;clear quotient
D_1
INCF QU,F ;increment quotient for every sub
SUBWF NUME ;sub WREG from NUME value
BC D_1 ;if positive go back (C = 1 for positive)
ADDWF NUME ;once too many, this is our first digit
DECF QU,F ;once too many for quotient
MOVFF NUME,RMND_L ;save the first digit
MOVFF QU,NUME ;repeat the process one more time
CLRF QU ;clear QU
D_2
INCF QU,F
SUBWF NUME ;sub WREG from NUME value
BC D_2 ;(C = 1 for positive)
ADDWF NUME ;once too many
DECF QU,F
MOVFF NUME,RMND_M ;2nd digit
MOVFF QU,RMND_H ;3rd digit
HERE
GOTO HERE ;stay here forever
END ;end of asm source file
```

To convert a single decimal digit to ASCII format, we OR it with 30H, as shown in Sections 6.4 and 6.5.
Example 5-9

Analyze the program in Example 5-8 for a numerator of 253.

Solution:

To convert a binary (hex) value to decimal, we divide it by 10 repeatedly until the quotient is less than 10. After each division the remainder is saved. In the case of an 8-bit binary, such as FDH, we have 253 decimal, as shown below.

<table>
<thead>
<tr>
<th>Quotient</th>
<th>Remainder</th>
</tr>
</thead>
<tbody>
<tr>
<td>253/10</td>
<td>25</td>
</tr>
<tr>
<td>25/10</td>
<td>5</td>
</tr>
<tr>
<td></td>
<td>2</td>
</tr>
</tbody>
</table>

Therefore, we have FDH = 253. In order to display this data, it must be converted to ASCII, which is described in a later section in this chapter.

Review Questions

1. In multiplication of two bytes in the PIC18, we can place one byte in register ______ and for the other one we can use value K.
2. In unsigned byte-by-byte multiplication, the product will be placed in register(s) ________.
3. Is "MULLW F" a valid PIC18 instruction? Explain your answer.
4. In PIC18, the largest two numbers that can be multiplied are ______ and ________.
5. True or false. The DAW instruction works on WREG only.
6. Is "DAW fileReg, d" a valid PIC18 instruction? Explain your answer.
7. The instruction "ADDLW K" places the sum in ________.
8. Why is the following ADD instruction illegal? "ADDLW fileReg"
9. Rewrite the instruction above to add WREG to the fileReg.
10. The instruction "ADDFWC fileReg, W" places the sum in ________.
11. Find the value of the DC and C flags in each of the following.
   (a) MOVLW 0x4F          (b) MOVLW 0x9C
      ADDLW 0xB1          ADDLW 0x63
12. Show how the CPU would subtract 05H from 43H.
13. If C = 1, WREG = 95H, and fileReg = 4FH prior to the execution of "SUBFWB fileReg, F", what will be the contents of WREG and fileReg after the subtraction?
SECTION 5.2: SIGNED NUMBER CONCEPTS AND ARITHMETIC OPERATIONS

All data items used so far have been unsigned numbers, meaning that the entire 8-bit operand was used for the magnitude. Many applications require signed data. In this section the concept of signed numbers is discussed along with related instructions. If your applications do not involve signed numbers, you can bypass this section.

Concept of signed numbers in computers

In everyday life, numbers are used that could be positive or negative. For example, a temperature of 5 degrees below zero can be represented as \(-5\), and 20 degrees above zero as \(+20\). Computers must be able to accommodate such numbers. To do that, computer scientists have devised the following arrangement for the representation of signed positive and negative numbers: The most significant bit (MSB) is set aside for the sign (+ or -), while the rest of the bits are used for the magnitude. The sign is represented by 0 for positive (+) numbers and 1 for negative (-) numbers. Signed byte representation is discussed below.

Signed 8-bit operands

In signed byte operands, D7 (MSB) is the sign and D0 to D6 are set aside for the magnitude of the number. If D7 = 0, the operand is positive, and if D7 = 1, it is negative. The N flag in the status register is the D7 bit.

Positive numbers

The range of positive numbers that can be represented by the format shown in Figure 5-2 is 0 to +127. If a positive number is larger than +127, a 16-bit operand must be used. Because the PIC18 does not support 16-bit data, we will not discuss it.

Negative numbers

For negative numbers, D7 is 1; however, the magnitude is represented in its 2's complement. Although the assembler does the conversion, it is still important to understand how the conversion works. To convert to negative number representation (2's complement), follow these steps:

1. Write the magnitude of the number in 8-bit binary (no sign).
2. Invert each bit.
3. Add 1 to it.

Examples 5-10, 5-11, and 5-12 demonstrate these three steps.
Example 5-10

Show how the PIC would represent -5.

Solution:

Observe the following steps.

1. 0000 0101 5 in 8-bit binary
2. 1111 1010 invert each bit
3. 1111 1011 add 1 (which becomes FB in hex)

Therefore, \(-5 = FBH\), the signed number representation in 2’s complement for \(-5\). The D7 = N = 1 indicates that the number is negative.

---

Example 5-11

Show how the PIC would represent -34H.

Solution:

Observe the following steps.

1. 0011 0100 34H given in binary
2. 1100 1011 invert each bit
3. 1100 1100 add 1 (which is CC 34H in hex)

Therefore, \(-34 = CCH\), the signed number representation in 2’s complement for 34H. The D7 = N = 1 indicates that the number is negative.

---

Example 5-12

Show how the PIC would represent -128.

Solution:

Observe the following steps.

1. 1000 0000 128 in 8-bit binary
2. 0111 1111 invert each bit
3. 1000 0000 add 1 (which becomes 80 in hex)

Therefore, \(-128 = 80H\), the signed number representation in 2’s complement for \(-128\). The D7 = N = 1 indicates that the number is negative. Notice that 128 (binary 10000000) in unsigned representation is the same as signed \(-128\) (binary 10000000).
From the examples above, it is clear that the range of byte-sized negative numbers is -1 to -128. The following lists byte-sized signed number ranges:

<table>
<thead>
<tr>
<th>Decimal</th>
<th>Binary</th>
<th>Hex</th>
</tr>
</thead>
<tbody>
<tr>
<td>-128</td>
<td>1000 0000</td>
<td>80</td>
</tr>
<tr>
<td>-127</td>
<td>1000 0001</td>
<td>81</td>
</tr>
<tr>
<td>-126</td>
<td>1000 0010</td>
<td>82</td>
</tr>
<tr>
<td>...</td>
<td>...........</td>
<td>...</td>
</tr>
<tr>
<td>-2</td>
<td>1111 1110</td>
<td>FE</td>
</tr>
<tr>
<td>-1</td>
<td>1111 1111</td>
<td>FF</td>
</tr>
<tr>
<td>0</td>
<td>0000 0000</td>
<td>00</td>
</tr>
<tr>
<td>+1</td>
<td>0000 0001</td>
<td>01</td>
</tr>
<tr>
<td>+2</td>
<td>0000 0010</td>
<td>02</td>
</tr>
<tr>
<td>...</td>
<td>...........</td>
<td>...</td>
</tr>
<tr>
<td>+127</td>
<td>0111 1111</td>
<td>7F</td>
</tr>
</tbody>
</table>

The above explains the mystery behind the relative address of −128 to +127 in the BNZ and other conditional branch instructions discussed in Chapter 3.

**Overflow problem in signed number operations**

When using signed numbers, a serious problem arises that must be dealt with. This is the overflow problem. The PIC indicates the existence of an error by raising the OV (overflow) flag, but it is up to the programmer to take care of the erroneous result. The CPU understands only 0s and 1s and ignores the human convention of positive and negative numbers. What is an overflow? If the result of an operation on signed numbers is too large for the register, an overflow has occurred and the programmer must be notified. Look at Example 5-13.

**Example 5-13**

Examine the following code and analyze the result, including the N and OV flags.

```
MOVLW +D'96'
ADDLW +D'70'
```

Solution:

```
+96 0110 0000  
+70 0100 0110  
+166 1010 0110  N = 1 (negative) and OV = 1. Sum = -90
```

According to the CPU, the result is negative (N = 1), which is wrong. The CPU sets OV = 1 to indicate the overflow error. Remember that the N flag is the D7 bit. If N = 0, the sum is positive, but if N = 1, the sum is negative.

In Example 5-13, +96 is added to +70 and the result, according to the CPU, was -90. Why? The reason is that the result was larger than what WREG could contain. Like all other 8-bit registers, WREG could only contain up to +127. The designers of the CPU created the overflow flag specifically for the purpose of informing the programmer that the result of the signed number operation is erroneous. The N flag is D7 of the result. If N = 0, the sum is positive (+) and if N = 1, then the sum is negative.
When is the OV flag set?

In 8-bit signed number operations, OV is set to 1 if either of the following two conditions occurs:

1. There is a carry from D6 to D7 but no carry out of D7 ($C = 0$).
2. There is a carry from D7 out ($C = 1$) but no carry from D6 to D7.

In other words, the overflow flag is set to 1 if there is a carry from D6 to D7 or from D7 out, but not both. This means that if there is a carry both from D6 to D7 and from D7 out, $OV = 0$. In Example 5-13, because there is only a carry from D6 to D7 and no carry from D7 out, $OV = 1$. Study Examples 5-14, 5-15, and 5-16 to understand the overflow flag in signed arithmetic.

### Example 5-14

Observe the following, noting the role of the OV and N flags:

| MOVLW -D'128' | ;WREG = 1000 0000 (WREG = 80H) |
| ADDLW -D'2'   | ;W = (-128) + (-2) |
|               | ;W = 1000000 + 11111110 = 0111 1110, |
|               | ;N = 0, W = 7EH = +126, invalid |

Solution:

| -128 | 1000 0000 |
| +  -2| 1111 1110 |
| -130 | 0111 1110 |

$N = 0$ (positive) and $OV = 1$.

According to the CPU, the result is +126, which is wrong, and $OV = 1$ indicates that.

### Example 5-15

Observe the following, noting the OV and N flags:

| MOVLW -D'2' | ;WREG = 1111 1110 (WREG = FEH) |
| ADDLW -D'5' | ;WREG = (-2) + (-5) = -7 or F9H |
|             | ;correct, since $OV = 0$ |

Solution:

| -2  | 1111 1110 |
| + -5| 1111 1011 |
| -7  | 1111 1001 |

and $OV = 0$ and $N = 1$. Sum is negative.

According to the CPU, the result is -7, which is correct, and the OV flag indicates that. ($OV = 0$).
Example 5-16

Examine the following, noting the role of the OV and N flags:

\[
\text{MOVLW } +D'7' \quad ; \text{WREG} = 0000 \ 0111
\]
\[
\text{ADD LW} +D'18' \quad ; \text{W} = (+7) + (+18)
\]
\[
\quad ; \text{W} = 00000111 + 00010010 = 0001 \ 1001
\]
\[
\quad ; \text{W} = (+7) + (+18) = +25, \text{ N} = 0, \text{ positive and}
\]
\[
\quad ; \text{correct, OV} = 0
\]

Solution:

\[
\begin{align*}
+ \ 7 & \ 0000 \ 0111 \\
+ \ +18 & \ 0001 \ 0010 \\
\text{+25} & \ 0001 \ 1001 \\
\text{N} & = 0 \text{ (positive 25) and OV} = 0
\end{align*}
\]

According to the CPU, this is +25, which is correct and OV = 0 indicates that.

From the above examples, we conclude that in any signed number addition, OV indicates whether the result is valid or not. If OV = 1, the result is erroneous; if OV = 0, the result is valid. We can state emphatically that in unsigned number addition, we must monitor the status of C (carry flag), and in signed number addition, the OV (overflow) flag must be monitored by the programmer. In the PIC, instructions such as BNC and BC allow the program to branch right after the addition of unsigned numbers, as we saw in Section 5.1. There are also the BOV and the BNOV instructions for the OV flag that allow us to correct the signed number error. We also have two branch instructions for the N flag (negative), BN and BNN.

**Instructions to create 2's complement**

The PIC18 does have a special instruction to make the 2's complement of a number. It is called NEG fileReg (negate fileReg) and is covered in the next section.

**Review Questions**

1. In an 8-bit operand, bit _____ is used for the sign bit.
2. Convert -16H to its 2's complement representation.
3. The range of byte-sized signed operands is - _____ to + _____.
4. Show +9 and -9 in binary.
5. Explain the difference between a carry and an overflow.
SECTION 5.3: LOGIC AND COMPARE INSTRUCTIONS

Apart from I/O and arithmetic instructions, logic instructions are some of most widely used instructions. In this section we cover Boolean logic instructions such as AND, OR, Exclusive-OR (XOR), and complement. We will also study the compare instruction.

**AND**

\[
\text{ANDLW } K \quad ; \text{WREG} = \text{WREG} \text{ AND } K
\]

This instruction will perform a logical AND on the two operands and place the result in WREG. There is also the "ANDWF fileReg, d" instruction where the destination can be WREG or fileReg. The fileReg operand can be any register in the data RAM file register. See Appendix A. The AND instruction will affect the Z and N flags. N is D7 of the result, and Z = 1 if the result is zero. The AND instruction is often used to mask (set to 0) certain bits of an operand. See Example 5-17.

### Logical AND Function

<table>
<thead>
<tr>
<th>Inputs</th>
<th>Output</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>Y</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>

**Example 5-17**

Show the results of the following.

\[
\text{MOVLW } 0x35 \quad ; \text{WRBG = 35H} \\
\text{ANDLW } 0x0F \quad ; \text{W = W AND 0FH (now W = 05)}
\]

**Solution:**

\[
\begin{align*}
35H & \quad 0 \quad 0 \quad 1 \quad 1 \quad 0 \quad 1 \quad 0 \\
0FH & \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 1 \quad 1 \\
05H & \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 0 \quad 1 \\
\end{align*}
\]

;35H AND 0FH = 05H, Z = 0, N = 0

**OR**

\[
\text{IORLW } K \quad ; \text{WREG = WREG Inclusive-OR } K
\]

This instruction will perform a logical OR on the two operands and place the result in WREG. There is also an "IORWF fileReg, d" instruction where the destination can be WREG or fileReg. The fileReg operand can be any register in the data RAM file register. See Appendix A. The OR instruction will affect the Z and N flags. N is D7 of the result and Z = 1 if the result is zero. The OR instruction can be used to set certain bits of an operand to 1. See Example 5-18.

### Logical OR Function

<table>
<thead>
<tr>
<th>Inputs</th>
<th>Output</th>
</tr>
</thead>
<tbody>
<tr>
<td>X</td>
<td>Y</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>

**Example 5-18**

Show the results of the following.

\[
\text{MOVLW } 0x35 \quad ; \text{WRBG = 35H} \\
\text{IORLW } 0x0F \quad ; \text{W = W OR 0FH (now W = 05)}
\]

**Solution:**

\[
\begin{align*}
35H & \quad 0 \quad 0 \quad 1 \quad 1 \quad 0 \quad 1 \quad 0 \\
0FH & \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 1 \quad 1 \\
05H & \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 0 \quad 1 \\
\end{align*}
\]

;35H OR 0FH = 05H, Z = 0, N = 0
Example 5-18

(a) Show the results of the following:

\[
\begin{align*}
&\text{MOVLW 0x04} \quad ; \text{WREG = 04} \\
&\text{IORLW 0x30} \quad ; \text{now WREG = 34H}
\end{align*}
\]

(b) Assume that Port B bit RB2 is used to control an outdoor light, and bit RB5 to control a light inside a building. Show how to turn "on" the outdoor light and turn "off" the inside one.

**Solution:**

(a) 04H 0000 0100
30H 0011 0000
34H 0011 0100 04 OR 30 = 34H, Z = 0 and N = 0

(b)

\[
\begin{align*}
&\text{BCF TRISB, 2} \quad ; \text{make RB2 an output} \\
&\text{BCF TRISB, 5} \quad ; \text{make RB5 an output} \\
&\text{MOVLW B'00000100'} \quad ; \text{D2 = 1} \\
&\text{IORWF PORTB, F} \quad ; \text{make RB2 = 1 only} \\
&\text{MOVLW B'11011111'} \quad ; \text{D5 = 0} \\
&\text{ANDWF PORTB, F} \quad ; \text{mask RB5 = 0 only}
\end{align*}
\]

Of course, the above method is unnecessary in PIC, since we can manipulate individual bits using bit-oriented operations. This is shown in Section 6.4.

**EX-OR**

\[
\text{XORLW K} \quad ; \text{WREG = WREG XOR K}
\]

This instruction will perform a logical EX-OR on the two operands and place the result in WREG. There is also an "XORWF fileReg, d" instruction where the destination can be WREG or fileReg. The fileReg operand can be any register in the data RAM file register. See Appendix A. The EX-OR instruction will affect the Z and N flags. N is D7 of the result and Z = 1 if result is zero. See Examples 5-19 and 5-20.

EX-OR can also be used to see if two registers have the same value. "XORWF fileReg, W" will EX-OR the WREG register and a fileReg location, and put the result in WREG. If both registers have the same value, 00 is placed in WREG. Then we can use the BZ instruction to make a decision based on the result. See Examples 5-20 and 5-21.
Another widely used application of EX-OR is to toggle the bits of an operand.

```
MOVLW 0xFF           ;WREG = FFH
XORWF PORTC,F       ;EX-OR PORTC with 1111 1111 will
                      ;change all the bits of Port C to
                      ;opposite
```

Example 5-19

Show the results of the following:
```
MOVLW 0x54
XORLW 0x78
```

Solution:
```
54H 0 1 0 1 0 1 0 0
78H 0 1 1 1 1 0 0 0
2CH 0 0 1 0 1 1 0 0  54H XOR 78H = 2CH, Z = 0, N = 0
```

Example 5-20

The EX-OR instruction can be used to test the contents of a register by EX-ORing it with a known value. In the following code, we show how EX-ORing value 45H with itself will raise the Z flag:

```
OVER MOVF PORTB,W    ;get a byte from PORTB into WREG
XORLW 0x45
BNZ OVER             ;branch if not zero
```

Solution:
```
45H 01000101
45H 01000101
00 00000000
```

EX-ORing a number with itself sets it to zero with $Z = 1$. We can use the BNZ instruction to make the decision. EX-ORing with any other number will result in a non-zero value.

Example 5-21

Read and test PORTB to see whether it has value 45H. If it does, send 99H to PORTC; otherwise, it stays cleared.

Solution:
```
CLRF TRISC           ;Port C = output
CLRF PORTC           ;Port C = 00
SBTF TRISB           ;Port B = input
MOVLW 0x45
XORWF PORTB,W       ;EX-OR with 0x45, Z = 1 if yes
BNZ EXIT            ;branch if PORTB has value other than 0
MOVLW 0x99
MOVWF PORTC         ;Port C = 99h
EXIT:...            
```
COMF (complement fileReg)

This instruction complements the contents of a file register. The complement action changes the 0s to 1s and the 1s to 0s. This is also called 1's complement.

```assembly
CLRF TRISB ; Port B = Output
MOVLW 0x55
MOVWF PORTB
COMF PORTB,F ; now PORTB = AAH
```

NEGФ (negate fileReg)

This instruction takes the 2’s complement of a file register. See Example 5-22.

**Example 5-22**

Find the 2’s complement of the value 85H. Note that 85H is –123.

**Solution:**

<table>
<thead>
<tr>
<th>MYREG</th>
<th>EQU 0x10</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOVLW</td>
<td>0x85</td>
</tr>
<tr>
<td>MOVWF</td>
<td>MYREG</td>
</tr>
<tr>
<td>NEGF</td>
<td>MYREG</td>
</tr>
<tr>
<td>2's comp</td>
<td>0111 1011 = 7BH</td>
</tr>
</tbody>
</table>

Compare instructions

The PIC18 has three instructions for the compare operation, as shown in Table 5-2. These instructions compare a value in the file register with the contents of the WREG register, and make decisions based on whether fileReg is greater than, equal to, or less than WREG. The compare instruction is really a subtraction, except that the values of the operands do not change. In PIC18, flags are not changed either after the compare instruction. It must be emphasized again that in compare instructions, the operands are not affected, regardless of the result of the comparison. We describe each of the instructions in Table 5-2 with an example.

**Table 5-2: PIC18 Compare Instructions**

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>CPFSGT</td>
<td>Compare FileReg with WREG, skip if greater than FileReg &gt; WREG</td>
</tr>
<tr>
<td>CPFSEQ</td>
<td>Compare FileReg with WREG, skip if equal</td>
</tr>
<tr>
<td>CPFSLT</td>
<td>Compare fileReg with WREG, skip if less than FileReg &lt; WREG</td>
</tr>
</tbody>
</table>

*Note: These instructions have no effect on the flag bits of the status register. Also the values in fileReg and WREG remain unchanged.*

**CPFSGT instruction**

The CPFSGT compares a fileReg with WREG and skips the next instruction if fileReg is greater than WREG (F > W). See Figure 5-3 and Example 5-23.
Figure 5-3. Flowchart for CPFSGT

Example 5-23

Write a program to find the greater of the two values 27 and 54, and place it in file register location 0x20.

Solution:

```
VAL_1 EQU D'27'
VAL_2 EQU D'54'
GREG EQU 0x20

MOVLW VAL_1  ;WREG = 27
MOVWF GREG  ;GREG = 27
MOVLW VAL_2  ;WREG = 54
CPFSEQ GREG  ;skip if GREG > WREG
MOVWF GREG  ;place the greater in GREG
```

**CPFSEQ instruction**

The CPFSEQ compares a fileReg with WREG and skips the next instruction if they are equal (F = W). See Example 5-24 and Figure 5-4.

Example 5-24

Write a program to monitor PORTD continuously for the value 63H. It should stop monitoring only if PORTD = 63H.

Solution:

```
SETF TRISD  ;PORTD = input
MOVLW 0x63  ;WREG = 63H
BACK CPFSEQ PORTD  ;skip BRA instruction if PORTD = 63H
BRA BACK
.....
```

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Figure 5-4. Flowchart for CPFSEQ

CPFSLT instruction

The CPFSLT compares a fileReg with WREG and skips the next instruction if fileReg is less than WREG (F < W). See Example 5-25 and Figure 5-5.

Example 5-25

Write a program to find the smaller of the two values 27 and 54, and place it in file register location 0x20.

Solution:

```
VAL_1 EQU D'27'
VAL_2 EQU D'54'
LREG EQU 0x20 ;location for smaller of two

MOVLW VAL_1 ;WREG = 27
MOVWF LREG ;LREG = 27
MOVLW VAL_2 ;WREG = 54
CPFSLT LREG ;skip if LREG < WREG
MOVWF LREG ;place the smaller value in LREG
```
**Example 5-26**

Assume that Port D is an input port connected to a temperature sensor. Write a program to read the temperature and test it for the value 75. According to the test results, place the temperature value into the registers indicated by the following.

- If $T = 75$ then $WREG = 75$
- If $T > 75$ then $GREG = T$
- If $T < 75$ then $LREG = T$

**Solution:**

```assembly
LREG EQU 0x20
GREG EQU 0x21

SETF TRISD ; PORTD = input
MOVLW D'75'; WREG = 75 decimal
CPFSGT PORTD ; skip BRA instruction if PORTD > 75
BRA LEQ
MOVFF PORTD, GREG
BRA OVER
LEQ CPFSLT PORTD ; skip if PORTD < 75
BRA OVER
MOVFF PORTD, LREG
OVER ...... ; it must be equal, WREG = 75
```

Figure 5-5. Flowchart for CPFSLT
Write code to determine if data on PORTB contains the value 99H. If so, write letter ‘Y’ to PORTC; otherwise, make PORTC = ‘N’.

Solution:

```
CLRF TRISC ; PORTC = output
MOVLW A'N' ; WREG = 'N' (ASCII)
MOVWF PORTC ; PORTC = 'N'
SETF TRISB ; PORTB = input
MOVLW 0x99 ; WREG = 99H
CPFSEQ PORTB ; skip BRA instruction if PORTB = WREG
BRA OVER
MOVW 'Y'
MOVWF PORTC ; PORTC = 'Y'
OVER ........
```

Review Questions

1. Find the content of register WREG after the following code in each case:
   (a) MOVLW 0x37  (b) MOVLW 0x37  (c) MOVLW 0x37
      ANDLW 0xCA  IORLW 0xCA  XORLW 0xCA

2. To mask certain bits of the WREG, we must AND it with _____.
3. To set certain bits of the WREG to 1, we must OR it with ______.
4. EX-ORing an operand with itself results in _______.
5. True or false. The CPFSLT instruction alters the contents of its operands.
6. What value must MYREG have in order for the following code to skip the BRA instruction?
   MOVLW 0x99
   BACK CPFSLT MYREG
   BRA BACK
7. Find the contents of register WREG after execution of the following code:
   MOVLW 0
   IORLW 0x99
   XORLW 0xFF
SECTION 5.4: ROTATE INSTRUCTION AND DATA SERIALIZATION

In many applications there is a need to perform a bitwise rotation of an operand. In the PIC18 the rotation instructions RRCF, RRNCF, RLCF, and RLNCF are designed specifically for that purpose. They allow a program to rotate the file register right or left. We explore the rotate instructions next because they are widely used in many different applications. There are two types of rotations. One is a simple rotation of the bits of the file register, and the other is a rotation through the carry. Each is explained below.

Rotating the bits of fileReg right or left

RRNCF fileReg,d ;rotate fileReg right (no carry)

In rotate right, the 8 bits of the fileReg are rotated right one bit, and bit D0 exits from the least-significant bit and enters into D7 (most-significant bit). After the rotation the result can be in fileReg or WREG, depending on the d bit. See the code and diagram.

MREG EQU 0x20
MOVLW 0x36 ;WREG = 0011 0110
MOVF MYREG
RRNCF MYREG,F ;MYREG = 0001 1011
RRNCF MYREG,F ;MYREG = 1000 1101
RRNCF MYREG,F ;MYREG = 1100 0110
RRNCF MYREG,F ;MYREG = 0110 0011

RLNCF fileReg,d ;rotate fileReg left (no carry)

In rotate left, the 8 bits of the fileReg are rotated left one bit, and bit D7 exits from the MSB (most-significant bit) and enters into D0 (least-significant bit). After the rotation the result can be in fileReg or WREG, depending on the d bit. See the code and diagram.

MREG EQU 0x20
MOVLW 0x72 ;WREG = 0111 0010
MOVF MYREG
RLNCF MYREG,F ;MYREG = 1110 0100
RLNCF MYREG,F ;MYREG = 1100 1001

Notice in the RRNCF and RLNCF instructions that both the Z and N flags are affected.
Rotating through the carry

There are two more rotate instructions in the PIC18. They involve the carry flag. Each is shown next.

```
RRCF fileReg, d ;rotate fileReg right through carry
```

In RRCF, as bits are rotated from left to right, the carry flag enters the MSB and the LSB exits to the carry flag. In other words, in RRCF the C is moved to the MSB, and the LSB is moved to the C. In reality, the carry flag acts as if it is part of the register, making it a 9-bit register.

```
MREG EQU 0x20
BCF STATUS,C ;make C = 0 (carry is D0 of status)
MOVLW 0x26 ;WREG = 0010 0110
MOVWF MYREG
RRCF MYREG,F ;MYREG = 0001 0011 C = 0
RRCF MYREG,F ;MYREG = 0000 1001 C = 1
RRCF MYREG,F ;MYREG = 1000 0100 C = 1
```

```
RLCF fileReg, d ;rotate fileReg left through carry
```

In RLCF, as bits are shifted from right to left, the carry flag enters the LSB and the MSB exits to the carry flag. In other words, in RLCF the C is moved to the LSB, and the MSB is moved to the C. See the following code and diagram. Again the carry flag acts as if it is part of the register, making it a 9-bit register.

```
MREG EQU 0x20
BSF STATUS,C ;make C = 1 (carry is D0 of status)
MOVLW 0x15 ;WREG = 0001 0101
MOVWF MYREG
RLCF MYREG,F ;MYREG = 0010 1011 C = 0
RLCF MYREG,F ;MYREG = 0101 0110 C = 0
RLCF MYREG,F ;MYREG = 1010 1100 C = 0
RLCF MYREG,F ;MYREG = 0101 1000 C = 1
```
Serializing data

Serializing data is a way of sending a byte of data one bit at a time through a single pin of the microcontroller. There are two ways to transfer a byte of data serially:

1. Using the serial port. In using the serial port, programmers have very limited control over the sequence of data transfer. The details of serial port data transfer are discussed in Chapter 10.
2. The second method of serializing data is to transfer data one bit at a time and control the sequence of data and spaces between them. In many new generations of devices such as LCD, ADC, and ROM, the serial versions are becoming popular because they take less space on a printed circuit board. Next, we discuss how to use rotate instructions in serializing data.

Serializing a byte of data

Serializing data is one of the most widely used applications of the rotate instruction. We can use the rotate instruction to transfer a byte of data serially (one bit at a time). Example 5-28 shows how to transfer an entire byte of data serially via any PIC pin.

Example 5-28

Write a program to transfer value 41H serially (one bit at a time) via pin RB1. Put one high at the start and end of the data. Send the LSB first.

Solution:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>RCNT EQU 0x20</td>
<td>fileReg loc for counter</td>
</tr>
<tr>
<td>MYREG EQU 0x21</td>
<td>fileReg loc for rotate</td>
</tr>
<tr>
<td>BCF TRISB,1</td>
<td>make RB1 an output bit</td>
</tr>
<tr>
<td>MOVFW 0x41</td>
<td>WREG = 41</td>
</tr>
<tr>
<td>MOVWF MYREG</td>
<td>value to be serialized</td>
</tr>
<tr>
<td>BCF STATUS,C</td>
<td>(C = 0)</td>
</tr>
<tr>
<td>MOVFW RCNT</td>
<td>load the counter</td>
</tr>
<tr>
<td>BSF PORTB,1</td>
<td>(RB1 = \text{high})</td>
</tr>
<tr>
<td>AGAIN</td>
<td></td>
</tr>
<tr>
<td>RRCF MYREG,F</td>
<td>rotate right via carry</td>
</tr>
<tr>
<td>BNC OVER</td>
<td></td>
</tr>
<tr>
<td>BSF PORTB,1</td>
<td>set the carry bit to PB1</td>
</tr>
<tr>
<td>BRA NEXT</td>
<td></td>
</tr>
<tr>
<td>OVER BCF PORTB,1</td>
<td></td>
</tr>
<tr>
<td>NEXT DBCF RCNT,F</td>
<td></td>
</tr>
<tr>
<td>BNZ AGAIN</td>
<td></td>
</tr>
<tr>
<td>BSF PORTB,1</td>
<td>(RB1 = \text{high})</td>
</tr>
</tbody>
</table>
Example 5-29 shows how to bring in a byte of data serially (one bit at a time). We will see how to use these concepts for a serial RTC (real-time clock) chip in Chapter 16. Example 5-30 shows how to scan the bits in a byte.

**Example 5-29**

Write a program to bring in a byte of data serially (one bit at a time) via pin RC7 and save it in file register location 0x21. The byte comes in with the LSB first.

**Solution:**

```
RCNT EQU 0x20 ;fileReg loc for counter
MYREG EQU 0x21 ;fileReg loc for incoming byte

BSF TRISC, 7 ;make RC7 an input bit
MOVLW 0x8 ;counter
MOVF RCNT ;load the counter
AGAIN BTFSC PORTC, 7 ;skip if RC7 = 0
     BTFSS PORTC, 7 ;skip if RC7 = 1
     BCF STATUS, C ;otherwise carry = 0
     RCF MYREG, F ;rotate right carry into MYREG
     DECF RCNT, F ;repeat until RCNT = 0
     BNZ AGAIN ;now loc 21H has the byte
```

**Example 5-30**

Write a program that finds the number of 1s in a given byte.

**Solution:**

```
R1 EQU 0x20 ;fileReg loc for number of 1s
COUNT EQU 0x21 ;fileReg loc for counter
VALREG EQU 0x22 ;fileReg loc for the byte

BCF STATUS, C ;C = 0
CLRF R1 ;R1 keeps the number of 1s
MOVLW 0x8 ;counter = 08 to rotate 8 times
MOVF COUNT
MOVLW 0x97 ;find the number of 1s in 97H
MOVF VALREG
AGAIN RLCF VALREG, F ;rotate it through the C once
     BNC NEXT ;check for C
     INCF R1, F ;if C = 1 then add one to R1 reg
     NEXT DECF COUNT, F
     BNZ AGAIN ;go through this 8 times
     ;now loc 0x20 has the number of 1s
```
SWAPF fileReg, d

Another useful instruction is the SWAPF instruction. It works on the file register. It swaps the lower nibble and the higher nibble. In other words, the lower 4 bits are put into the higher 4 bits, and the higher 4 bits are put into the lower 4 bits. See the diagrams below and Example 5-31.

Example 5-31

(a) Find the contents of the MYREG register in the following code.
(b) In the absence of a SWAPF instruction, how would you exchange the nibbles?
Write a simple program to show the process.

Solution:
(a)

```
MYREG EQU 0x20
MOVLW 0x72 ;WREG = 72H
MOVWF MYREG ;MYREG = 72H
SWAPF MYREG,F ;MYREG = 27H
```

(b)

```
MYREG EQU 0x20
MOVLW 0x72 ;WREG = 0111 0010
MOVWF MYREG ;MYREG = 0111 0010
RLNCF MYREG,F ;MYREG = 1100 1001
RLNCF MYREG,F ;MYREG = 1001 0011
RLNCF MYREG,F ;MYREG = 0010 0111
```

Review Questions

1. What is the value of MYREG in the file register after the following code is executed?

```
MYREG EQU 0x40
MOVLW 0x25
MOVWF MYREG
RRNCF MYREG,F
RRNCF MYREG,F
RRNCF MYREG,F
RRNCF MYREG,F
```
2. What is the value of MYREG in the file register after the following code is executed?

```assembly
MYREG EQU 0x40
MOVLW 0x25
MOVWF MYREG
RLNCF MYREG,F
RLNCF MYREG,F
RLNCF MYREG,F
RLNCF MYREG,F
```

3. What is the value of MYREG after the following code is executed?

```assembly
MYREG EQU 0x40
CLRF MYREG
BSF STATUS,C ; C = 1
RRCF MYREG,F
BSF STATUS,C ; C = 1
RRCF MYREG,F
```

4. Does “RLCF W” give an error in the PIC?

5. What is in MYREG after the execution of the following code?

```assembly
MYREG EQU 0x40
MOVLW 0x85
MOVWF MYREG
SWAPF MYREG,F
```

**SECTION 5.5: BCD AND ASCII CONVERSION**

In this section we provide some real-world examples of how to use arithmetic and logic instructions. We will cover their applications in real-world devices in future chapters. For example, many newer microcontrollers have a real-time clock (RTC), where the time and date are kept even when the power is off. These microcontrollers provide the time and date in BCD. To display them, however, they must convert BCD values to ASCII. Next, we show the application of logic and rotate instructions in the conversion of BCD and ASCII.

**ASCII numbers**

On ASCII keyboards, when the key “0” is activated, “011 0000” (30H) is provided to the computer. Similarly, 31H (011 0001) is provided for key “1”, and so on, as shown in Table 5-3.

It must be noted that BCD numbers are universal although ASCII is standard in the United States (and many other countries). Because the keyboard, printers, and monitors all use ASCII, how does data get converted from ASCII to BCD, and vice versa? These are the subjects covered next.
Table 5-3: ASCII and BCD Codes for Digits 0–9

<table>
<thead>
<tr>
<th>Key</th>
<th>ASCII (hex)</th>
<th>Binary</th>
<th>BCD (unpacked)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>30</td>
<td>011 0000</td>
<td>0000 0000</td>
</tr>
<tr>
<td>1</td>
<td>31</td>
<td>011 0001</td>
<td>0000 0001</td>
</tr>
<tr>
<td>2</td>
<td>32</td>
<td>011 0010</td>
<td>0000 0010</td>
</tr>
<tr>
<td>3</td>
<td>33</td>
<td>011 0011</td>
<td>0000 0011</td>
</tr>
<tr>
<td>4</td>
<td>34</td>
<td>011 0100</td>
<td>0000 0100</td>
</tr>
<tr>
<td>5</td>
<td>35</td>
<td>011 0101</td>
<td>0000 0101</td>
</tr>
<tr>
<td>6</td>
<td>36</td>
<td>011 0110</td>
<td>0000 0110</td>
</tr>
<tr>
<td>7</td>
<td>37</td>
<td>011 0111</td>
<td>0000 0111</td>
</tr>
<tr>
<td>8</td>
<td>38</td>
<td>011 1000</td>
<td>0000 1000</td>
</tr>
<tr>
<td>9</td>
<td>39</td>
<td>011 1001</td>
<td>0000 1001</td>
</tr>
</tbody>
</table>

**Packed BCD to ASCII conversion**

In many systems we have what is called a real-time clock (RTC). The RTC provides the time of day (hour, minute, second) and the date (year, month, day) continuously, regardless of whether the power is on or off (see Chapter 16). This data, however, is provided in packed BCD. For this data to be displayed on a device such as an LCD, or to be printed by the printer, it must be in ASCII format.

To convert packed BCD to ASCII, you must first convert it to unpacked BCD. Then the unpacked BCD is tagged with 011 0000 (30H). The following demonstrates converting packed BCD to ASCII. See also Example 5-32.

<table>
<thead>
<tr>
<th>Packed BCD</th>
<th>Unpacked BCD</th>
<th>ASCII</th>
</tr>
</thead>
<tbody>
<tr>
<td>29H</td>
<td>02H &amp; 09H</td>
<td>32H &amp; 39H</td>
</tr>
<tr>
<td>0010 1001</td>
<td>0000 0101 &amp;</td>
<td>0011 1001</td>
</tr>
<tr>
<td></td>
<td>0000 1001</td>
<td></td>
</tr>
</tbody>
</table>

**Example 5-32**

Assume that register WREG has packed BCD. Write a program to convert packed BCD to two ASCII numbers and place them in file register locations 6 and 7.

**Solution:**

```
BCD_VAL EQU 0x29
L_ASC EQU 0x06 ;set aside file register location
H_ASC EQU 0x07 ;set aside file register location

MOVLW BCD_VAL ;WREG = 29H, packed BCD
ANDLW 0x0F    ;mask the upper nibble (W = 09)
IORLW 0x30    ;make it an ASCII, W = 39H ('9')
MOVWF L_ASC  ;save it (L_ASC = 39H ASCII char)
MOVLW BCD_VAL ;W = 29H get BCD data once more
ANDLW 0xF0   ;mask the lower nibble (W = 20H)
SWAPF WREG,W ;swap nibbles (WREG = 02H)
IORLW 0x30    ;make it an ASCII, W = 32H ('2')
MOVWF H_ASC  ;save it (H_ASC = 32H ASCII char)
```
To convert ASCII to packed BCD, you first convert it to unpacked BCD (to get rid of the 3), and then combine it to make packed BCD. For example, for 4 and 7 the keyboard gives 34 and 37, respectively. The goal is to produce 47H or “0100 0111”, which is packed BCD. This process is illustrated next.

<table>
<thead>
<tr>
<th>Key</th>
<th>ASCII</th>
<th>Unpacked BCD</th>
<th>Packed BCD</th>
</tr>
</thead>
<tbody>
<tr>
<td>4</td>
<td>34</td>
<td>00000100</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>37</td>
<td>00000111</td>
<td>01000111 which is 47H</td>
</tr>
</tbody>
</table>

MYBCD EQU 0x20 ;set aside location in file register

MOVLW A'4' ;WREG = 34H, hex for ASCII char 4
ANDLW 0x0F ;mask upper nibble (WREG = 04)
MOVWF MYBCD ;save it in MYBCD loc
SWAPF MYBCD,F ;MYBCD = 40H
MOVLW A'7' ;WREG = 37H, hex for ASCII char 7
ANDLW 0x0F ;mask upper nibble (WREG = 07)
IORWF MYBCD,F ;MYBCD = 47H, a packed BCD

After this conversion, the packed BCD numbers are processed and the result will be in packed BCD format. As we saw earlier in this chapter, a special instruction, “DAW”, requires that the data be in packed BCD format.

Review Questions

1. For the following decimal numbers, give the packed BCD and unpacked BCD representations.
   (a) 15        (b) 99
2. Show the binary and hex formats for “76” and its BCD version.
3. Does the WREG register have BCD data after the following instruction is executed?
   MOVLW D'54'
4. 67H in BCD when converted to ASCII is ___H and ___H.
5. Does the following convert unpacked BCD in the WREG register to ASCII?
   MOVLW 0x09
   ADDLW 0x30

SUMMARY

This chapter discussed arithmetic instructions for both signed and unsigned data in the PIC. Unsigned data uses all 8 bits of the byte for data, making a range of 0 to 255 decimal. Signed data uses 7 bits for data and 1 for the sign bit, making a range of −128 to +127 decimal.

Binary coded decimal (BCD) data represents the digits 0 through 9. Both packed and unpacked BCD formats were discussed. The PIC contains special instructions for arithmetic operations on BCD data.

In coding arithmetic instructions for the PIC, special attention has to be given to the possibility of a carry or overflow condition.
This chapter defined the logic instructions AND, OR, XOR, and complement. In addition, PIC Assembly language instructions for these functions were described. Compare and skip instructions were described as well. These functions are often used for bit manipulation purposes.

The rotate and swap instructions of the PIC are used in many applications such as serial devices. This chapter also described BCD and ASCII formats and conversions.

**PROBLEMS**

**SECTION 5.1: ARITHMETIC INSTRUCTIONS**

1. Find the C, Z, and DC flags for each of the following:
   
   (a) \text{MOVLW 0x3F} \quad \text{ADDLW 0x45}
   
   (c) \text{MOVLW 0xFF} \quad \text{MOVF MYREG} \quad \text{BSF STATUS, C} \quad \text{MOVLW 0} \quad \text{ADDFC MYREG, F}
   
   (e) \text{MOVLW 0xFE} \quad \text{MOVF MYREG} \quad \text{BSF STATUS, C} \quad \text{MOVLW 0} \quad \text{ADDFC MYREG, F}
   
   (b) \text{MOVLW 0x99} \quad \text{ADDLW 0x58}
   
   (d) \text{MOVLW 0xFF} \quad \text{ADDLW 0x1}
   
   (f) \text{BCF STATUS, C} \quad \text{MOVLW 0xFF} \quad \text{MOVF MYREG}

2. Write a program to add all the digits of your ID number and save the result in a file register location. The result must be in BCD.

3. Write a program to add the following numbers and save the result in a file register location.
   
   0x25, 0x59, 0x65

4. Modify Problem 3 to make the result in BCD.

5. Write a program to (a) write the value 25H to file register RAM locations 20H–23H, and (b) add all these RAM locations contents together, and save the result in RAM location 60H.

6. State the steps that the SUB instruction will go through for each of the following.
   
   (a) \text{23H - 12H} \quad (b) \text{43H - 53H} \quad (c) \text{99 - 99}

7. For Problem 6, write a program to perform each operation.

8. True or false. The "daw" instruction works only on the WREG register.

9. Write a program to add 7F9AH to BC48H and save the result in RAM memory locations starting at 40H.

10. Write a program to subtract 7F9AH from BC48H, and save the result in RAM memory locations starting at 40H.

11. Write a program to add BCD 7795H to 9548H and save the BCD result in RAM memory locations starting at 40H.

12. Show how to perform \(77 \times 34\) in the PIC18.
13. Show how to perform 77/3 in the PIC18.
14. True or false. The MULLW instruction works on any register of the PIC18.
15. The MULLW instruction places the result in registers _______ and______.

SECTION 5.2: SIGNED NUMBER CONCEPTS AND ARITHMETIC OPERATIONS

16. Show how the following are represented by the assembler:
   (a) -23   (b) +12   (c) -28
   (d) +6FH   (e) -128   (f) +127

17. The memory addresses in computers are ________ (signed, unsigned) numbers.

18. Write a program for each of the following and indicate the status of the OV flag for each:
   (a) (+15) + (-12)   (b) (-123) + (-127)
   (c) (+25H) + (+34H)   (d) (-127) + (+127)

19. Explain the difference between the C and OV flags and where each one is used.
20. When is the OV flag raised? Explain.
21. Which register holds the OV flag?
22. How do you detect the OV flag in the PIC18? How do you detect the C flag?

SECTION 5.3: LOGIC AND COMPARE INSTRUCTIONS

23. Assume that WREG = F0H. Perform the following operations. Indicate the result and the register where it is stored.
   Note: The operations are independent of each other.
   (a) ANDLW 0x45   (b) IORLW 0x90
   (c) XORLW 0x76   (d) ANDLW 0x90
   (e) XORLW 0x90   (f) IORLW 0x90
   (g) ANDLW 0xFF   (h) IORLW 0x99
   (i) XORLW 0xEE   (j) XORLW 0xAA

24. Find the contents of register WREG after each of the following instructions:
   (a) MOVLLW 0x65   (b) MOVLLW 0x70
      ANDLW 0x76   IORLW 0x6B
   (c) MOVLLW 0x95   (d) MOVLLW 0x5D
      XORLW 0xAA   ANDLW 0x78
   (e) MOVLLW 0x0C5  (f) MOVLLW 0x6A
      IORLW 0x12   XORLW 0x6E
   (g) MOVLLW 0x37   IORLW 0x26

25. True or false. In using the CPFSEQ instruction, we must use WREG as one of the registers.
26. Explain how the CPFSGT instruction works.
27. Does the compare instruction affect the flag bits of the status register?
28. Assume that MYREG = 85H. Indicate if it skips after compare is executed in each of the following cases:

(a) \( \text{MOVLW} \ 0x90 \) \( \text{CPFSGT MYREG} \) \( \text{INCF MYREG,F} \) \( \text{ADDLW} \ 0x2 \)

(b) \( \text{MOVLW} \ 0x70 \) \( \text{CPFSGT MYREG} \) \( \text{INCF MYREG,F} \) \( \text{ADDLW} \ 0x2 \)

(c) \( \text{MOVLW} \ 0x85 \) \( \text{CPFSEQ MYREG} \) \( \text{INCF MYREG,F} \) \( \text{ADDLW} \ 0x2 \)

(d) \( \text{MOVLW} \ 0x5D \) \( \text{CPFSLT MYREG} \) \( \text{INCF MYREG,F} \) \( \text{ADDLW} \ 0x2 \)

29. In Problem 28, indicate the value in MYREG.

SECTION 5.4: ROTATE INSTRUCTION AND DATA SERIALIZATION

30. Find register WREG contents after each of the following is executed:

(a) \( \text{MOVLW} \ 0x56 \) \( \text{MOVWF MYREG} \) \( \text{SWAPF MYREG,F} \) \( \text{RRCF MYREG,F} \) \( \text{RRCF MYREG,F} \) \( \text{BCF STATUS,C} \)

(b) \( \text{MOVLW} \ 0x39 \) \( \text{MOVWF MYREG} \) \( \text{MOVWF MYREG,F} \) \( \text{RLCF MYREG,F} \) \( \text{RLCF MYREG,F} \) \( \text{BCF STATUS,C} \)

(c) \( \text{MOVLW} \ 0x4D \) \( \text{MOVWF MYREG} \) \( \text{SWAPF MYREG,F} \) \( \text{RRCF MYREG,F} \) \( \text{RRCF MYREG,F} \) \( \text{BCF Status,C} \)

(d) \( \text{MOVLW} \ 0x7A \) \( \text{MOVWF MYREG} \) \( \text{SWAPF MYREG,F} \) \( \text{RLCF MYREG,F} \) \( \text{RLCF MYREG,F} \) \( \text{BCF Status,C} \)

31. Show the code to replace the SWAPF code:

(a) using the rotate right instructions
(b) using the rotate left instructions

32. Write a program that finds the number of zeros in an 8-bit data item.
33. Write a program that finds the position of the first high in an 8-bit data item. The data is scanned from D0 to D7. Give the result for 68H.
34. Write a program that finds the position of the first high in an 8-bit data item. The data is scanned from D7 to D0. Give the result for 68H.
35. A stepper motor uses the following sequence of binary numbers to move the motor. How would you generate them?
   1100, 0110, 0011, 1001

SECTION 5.5: BCD AND ASCII CONVERSION

36. Write a program to convert the following packed BCD numbers to ASCII. Place the ASCII codes in data RAM locations starting at 40H.

\( \text{MYBCD}_1 \) EQU 0x76
\( \text{MYBCD}_2 \) EQU 0x87
37. Write a program to convert the following ASCII numbers to packed BCD. Place the BCD data in RAM locations starting at 60H.

```
MYASC_1 EQU A'8'
MYASC_2 EQU A'7'
MYASC_3 EQU A'9'
MYASC_4 EQU A'2'
```

ANSWERS TO REVIEW QUESTIONS

SECTION 5.1: ARITHMETIC INSTRUCTIONS

1. WREG
2. PRODH and PRODL
3. No. It should be “MULWF fileReg, F”
4. 255 and 255.
5. True.
6. No. DAW works on WREG only.
7. WREG
8. We cannot mix the literal value, WREG, and fileReg.
9. “ADDWF fileReg, F”
10. WREG
11. (a) WREG = 00, C = 1, and DC = 1
    (b) WREG = FF, C = 0, and DC = 0
12. 

```
  43H  0100 0011
- 05H  0000 0101 2's complement
  3EH  0011 1110
```
13. fileReg = 95H – 4FH = 0 = 46H, WREG = 95H

SECTION 5.2: SIGNED NUMBER CONCEPTS AND ARITHMETIC OPERATIONS

1. D7
2. 16H is 00010110 in binary and its 2’s complement is 1110 1010 or –16H = EA in hex.
3. –128 to +127
4. +9 = 00001001 and –9 = 11110111 or F7 in hex.
5. An overflow is a carry into the sign bit (D7) but the carry is a carry out of register.

SECTION 5.3: LOGIC AND COMPARE INSTRUCTIONS

1. (a) 02
   (b) FFH
   (c) FDH
2. Zero
3. One
4. All zeros
5. False
6. any value less than 0x99
7. 66H
SECTION 5.4: ROTATE INSTRUCTION AND DATA SERIALIZATION

1. 52H
2. 52H
3. C0H
4. No, because WREG is a SFR
5. 58H

SECTION 5.5: BCD AND ASCII CONVERSION

1. (a) 15H = 0001 0101 packed BCD, 0000 0000,0000 0101 unpacked BCD
   (b) 99H = 1001 1001 packed BCD, 0000 1001,0000 1001 unpacked BCD
2. 3736H = 00110111 00110110B
   and in BCD we have 76H = 0111 0110B
3. No. We need to write it as 54H (with the H) or 01010100B to make it BCD. The value 54 without the “H” is interpreted as 36H by the assembler.
4. 36H, 37H
5. Yes, because WREG = 39H
CHAPTER 6

BANK SWITCHING, TABLE PROCESSING, MACROS, AND MODULES

OBJECTIVES

Upon completion of this chapter, you will be able to:

- List all the addressing modes of the PIC18 microcontroller
- Contrast and compare the addressing modes
- Code PIC Assembly language instructions using each addressing mode
- Access the data RAM file register using various addressing modes
- Code PIC18 instructions to manipulate a look-up table
- Access fixed data residing in the program ROM space
- Discuss how to create macros and modules
- Discuss how to access the entire 4K of RAM space in the PIC18
- List the addresses for all 16 banks of the PIC18
- Discuss how to access all banks of the PIC18
- Discuss bank switching for the PIC18
- Code PIC18 programs for ASCII and BCD data conversion
- Code PIC18 programs to create and test the checksum byte
- List the advantages of macros and modules in programming
The CPU can access data in various ways. The data could be in a register, or in memory, or provided as an immediate value. These various ways of accessing data are called *addressing modes*. In this chapter we discuss PIC18 addressing modes in the context of some examples.

The various addressing modes of a microprocessor are determined when it is designed, and therefore cannot be changed by the programmer. The PIC18 provides a total of four distinct addressing modes. They are as follows:

1. Immediate
2. Direct
3. Register indirect
4. Indexed-ROM

In Section 6.1 we look at immediate and direct addressing modes. In Section 6.2 we cover accessing RAM data memory using the register indirect mode. Section 6.3 explains how to access fixed data and look-up tables stored in program ROM. Section 6.4 discusses the bit-addressability of the file register data RAM space. In Section 6.5 we discuss bank switching and show how to access banks other than the access bank. Checksum generation and BCD-ASCII conversions are discussed in Section 6.6. In Section 6.7, macros and modules are examined and modular programming is described.

**SECTION 6.1: IMMEDIATE AND DIRECT ADDRESSING MODES**

In this section, we examine first the immediate addressing mode and then the direct addressing mode.

**Immediate addressing mode**

In this addressing mode, the operand is a literal constant. In immediate addressing mode, as the name implies, the operand comes immediately after the opcode when the instruction is assembled. Notice that immediate data is called a *literal* in the PIC. This addressing mode can be used to load information into WREG and selected registers, but not to any file register. The immediate addressing mode is also used for arithmetic and logic instructions. Examine the following examples.

```assembly
MOVLW 0x25 ;load 25H into WREG
SUBLW D'62' ;subtract WREG from 62
ANDLW B'01000000' ;AND WREG with 40H
```

We can use the EQU directive to access immediate data as shown below.

```assembly
COUNT EQU 0x30
.
.
MOVLW COUNT ;WREG = 30h
```

Notice that we can also use immediate addressing mode to perform arithmetic and logic operations on WREG only. For example, “ADDLW 0x25” adds value 25 to WREG.
Direct addressing mode

As mentioned in Chapter 2, the 256-byte access bank file register is split into two sections: The lower addresses, 00 to 7FH, are assigned to the general purpose registers, and the upper addresses, F80–FFFF, to the SFR. The access bank is the default bank when the PIC18 is powered up. It is the minimum bank that all PIC18 processors have. The MOVFF instruction also plays a role in choosing the access bank. We will discuss that issue in Section 6.5 when we discuss bank switching.

The entire data RAM file register can be accessed using either direct or register indirect addressing modes. The register indirect addressing mode will be discussed in the next section. In direct addressing mode, the operand data is in a RAM memory location whose address is known, and this address is given as a part of the instruction. Contrast this with immediate addressing mode in which the operand data itself is provided with the instruction. While the letter “L” in the instruction means literal (immediate), the letter “F” in the instruction signifies the address of the file register location. See the example below, and note the letter F in the instructions.

\[
\begin{align*}
\text{MOVLLW 0x56} & ; \text{WREG = 56H (immediate addressing mode)} \\
\text{MOVWF 0x40} & ; \text{copy WREG into fileReg RAM location 40H} \\
\text{MOVFF 0x40,0x50} & ; \text{copy data from loc 40H to 50H.}
\end{align*}
\]

The last two instructions use direct addressing mode. If we dissect the opcode we see that the addresses are embedded in the instruction, as shown in Figure 6-1a.

![Figure 6-1a. MOVFF Direct Addressing Opcode](image)

As shown in Figure 6-1b, the address field is an 8-bit address and can take values from 00–FFH. The A bit for bank switching is discussed in Section 6-4. Of course, it is much easier to use names instead of addresses in the program, and we have seen many examples of them in the last few chapters. It must be noted that file register data RAM does not support immediate addressing mode. In other words, to move data into any file register, we must first move it to WREG, and then move it from WREG to the file register using the MOVWF instruction.

![Figure 6-1b. MOVWF Direct Addressing Opcode](image)
The difference between “INCF fileReg, W” and “INCF fileReg, F”

In direct addressing mode, when an operation is performed on a file register, we have the option of saving the result in the file register itself or in WREG. This option is a major source of errors in PIC programming and its correct use must be emphasized. The following code increments the contents of register file location 20H using the direct addressing mode, but the destination for the increment result is decided by the W or F parameter:

```
MOVLW 0 ;WREG = 0
MOVWF 0x20 ;loc 0x20 = (0), WREG = 0
INCF 0x20,W ;loc 0x20 = (0), WREG = 1
INCF 0x20,W ;loc 0x20 = (0), WREG = 1
INCF 0x20,W ;loc 0x20 = (0), WREG = 1
INCF 0x20,F ;loc 0x20 = (1), WREG = 1
INCF 0x20,F ;loc 0x20 = (2), WREG = 1
INCF 0x20 ;loc 0x20 = (3), WREG = 1
INCF 0x20 ;loc 0x20 = (4), WREG = 1
INCF 0x20,W ;loc 0x20 = (4), WREG = 5
```

Notice in the above code when the second parameter is not stated, it is assumed to be fileReg (F).

**DECFSZ and DECF**

Other instructions that need to be examined are DECFSZ and DECF. We can use either one for looping. In “DECFSZ fileReg, d” the fileReg is decremented, and the next instruction is skipped if the fileReg is zero. DECF does not skip the next instruction. Contrast the two codes complementing PORTB 5 times. The following code uses the DECFSZ instruction for looping:

```
CLRF TRISB ;Port B as output
MOVLW 5 ;WREG = 5
MOVWF MYREG ;counter = 5
CLR PORTB ;clear Port B
B1
COMF PORTB ;complement Port B
DECFSZ MYREG,F ;decrement and skip if MYREG = 0
GOTO B1 ;go back since it is not zero
SETF PORTB ;make PB = FFH
```

while the code below uses the BNZ (branch not zero) instruction:

```
CLRF TRISB ;Port B as output
MOVLW 5 ;WREG = 5
MOVWF MYREG ;counter = 5
CLR PORTB ;clear Port B
B2
COMF PORTB ;complement Port B
DECF MYREG,F ;decrement counter
BNZ B2 ;go back if MYREG is not zero
SETF PORTB ;make PB = FFH
```

Notice that if we use “DECF MYREG, W” instead of “DECF MYREG, F” in the BNZ program, we will never get out of the loop because the values of MYREG
<table>
<thead>
<tr>
<th>Symbol</th>
<th>Name</th>
<th>Address</th>
</tr>
</thead>
<tbody>
<tr>
<td>WREG</td>
<td>Working register</td>
<td>FE8H</td>
</tr>
<tr>
<td>PORTA</td>
<td>Port A</td>
<td>F80H</td>
</tr>
<tr>
<td>PORTB</td>
<td>Port B</td>
<td>F81H</td>
</tr>
<tr>
<td>PORTC</td>
<td>Port C</td>
<td>F82H</td>
</tr>
<tr>
<td>LATA</td>
<td>Output latch, Port A</td>
<td>F89H</td>
</tr>
<tr>
<td>LATB</td>
<td>Output latch, Port B</td>
<td>F8AH</td>
</tr>
<tr>
<td>LATC</td>
<td>Output latch, Port C</td>
<td>F8BH</td>
</tr>
<tr>
<td>TRISA</td>
<td>Data direction, Port A</td>
<td>F92H</td>
</tr>
<tr>
<td>TRISB</td>
<td>Data direction, Port B</td>
<td>F93H</td>
</tr>
<tr>
<td>TRISC</td>
<td>Data direction, Port C</td>
<td>F94H</td>
</tr>
<tr>
<td>INDF0</td>
<td>Indirect addressing register 0</td>
<td>FEFH</td>
</tr>
<tr>
<td>INDF1</td>
<td>Indirect addressing register 1</td>
<td>FE7H</td>
</tr>
<tr>
<td>FSR0L</td>
<td>Indirect data memory address pointer 0 low</td>
<td>FE9H</td>
</tr>
<tr>
<td>FSR0H</td>
<td>Indirect data memory address pointer 0 high</td>
<td>FEAH</td>
</tr>
<tr>
<td>FSR1L</td>
<td>Indirect data memory address pointer 1 low</td>
<td>FE1H</td>
</tr>
<tr>
<td>FSR1H</td>
<td>Indirect data memory address pointer 1 high</td>
<td>FE2H</td>
</tr>
<tr>
<td>PLUSW0</td>
<td>Indirect indexed address register</td>
<td>FEBH</td>
</tr>
<tr>
<td>PREINCO</td>
<td>Preincrement register 0</td>
<td>FECH</td>
</tr>
<tr>
<td>POSTDEC0</td>
<td>Post-decrement register 0</td>
<td>FEDH</td>
</tr>
<tr>
<td>POSTINCO</td>
<td>Post-increment register 0</td>
<td>FEEH</td>
</tr>
<tr>
<td>TBLPTRL</td>
<td>Table pointer, low byte</td>
<td>FF6H</td>
</tr>
<tr>
<td>TBLPTRH</td>
<td>Table pointer, high byte</td>
<td>FF7H</td>
</tr>
<tr>
<td>TBLPTRU</td>
<td>Table pointer, upper byte</td>
<td>FF8H</td>
</tr>
<tr>
<td>TABLAT</td>
<td>Program memory table latch</td>
<td>FF5H</td>
</tr>
<tr>
<td>STATUS</td>
<td>Status flag byte</td>
<td>FD8H</td>
</tr>
</tbody>
</table>

and WREG will remain 5 and 4 respectively forever.

**SFR registers and their addresses**

PIC18 registers for Ports A, B, and so on are part of the group of registers commonly referred to as SFRs (special function registers). There are many special function registers and they are widely used, as we will discuss in future chapters. The SFRs can be accessed by their names (which is much easier) or by their addresses. For example, Port B has address F81H, and Port C the address F82H, as shown in Table 6-1. Notice how the following pairs of instructions mean the same thing:

```plaintext
MOVWF 0xF81 ;is the same as
MOVWF PORTB ;which means copy WREG into Port B

CLRF 0xF82 ;is the same as
CLRF PORTC ;which means clear Port C

BSF 0xFD8,0 ;is the same as
BSF STATUS,C ;which make C = 1
```
Table 6-1 lists selected PIC18 special function registers (SFRs) and their addresses. The following two points should be noted about SFR addresses:

1. The special function registers have addresses between F80H and FFFH. These addresses are below FFFFH, because the PIC18 starts assigning SFR addresses at FFFFH and goes down until all SFRs supported by that chip are assigned. Not all the members of the PIC18 family have the same peripherals; therefore, the number of locations used for SFRs varies among the PIC18 family.

2. Not all the address space of F80H to FFFFH is used by the SFR. The unused locations F80H to FFFFH are reserved and must not be used by the PIC18 programmer.

**Example 6-1**

Write code to send 55H to Port B. Include

(a) The register names.
(b) Their addresses.

**Solution:**

(a)

```
CLRF TRISB ;Port B output
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB ;Port B = 55H
```

(b)

From Table 6-1, TRISB address = F93H and PORTB address = F81H

```
CLRF 0xF93 ;Port B output
MOVLW 0x55H ;WREG = 55H
MOVWF 0xF81 ;Port B = 55H
```

Regarding direct addressing mode in the PIC18, notice the following points:

1. If you examine the .lst file for an Assembly language program, you will see that the SFR register names are replaced with their addresses as listed in Table 6-1.
2. The WREG register is one of the SFR registers and has address FE8H.
3. The direct addressing mode is also called register direct to contrast it with the register indirect addressing mode discussed in the next section.

**Review Problems**

1. Can the programmer of a microcontroller make up new addressing modes?
2. Show the instruction to load 1000 0000 (binary) into register WREG.
3. Why is “MOVLF myvalue, fileReg” invalid?
4. True or false. In PIC18, the PC (program counter) is part of the SFR.
5. True or false. In PIC18, the WREG register is not part of the SFR.
SECTION 6.2: REGISTER INDIRECT ADDRESSING MODE

We can use register direct or register indirect addressing modes to access data stored in the general purpose RAM section of the file register. In the last section we showed how to use direct addressing mode, which is also called register direct. The register indirect addressing mode is a very important addressing mode in the PIC18. This topic will be discussed thoroughly in this section.

Register indirect addressing mode

In the register indirect addressing mode, a register is used as a pointer to the data RAM location. In the PIC18, three registers are used for this purpose: FSR0, FSR1, and FSR2. FSR stands for file select register and must not be confused with SFR (special function register). The FSR is a 12-bit register allowing access to the entire 4096 bytes of data RAM space in the PIC18. We use LFSR (load FSR) to load the RAM address. In other words, when FSRx are used as pointers, they must be loaded first with the RAM addresses as shown below.

```
LFSR 0, 0x30 ; load FSR0 with 0x30
LFSR 1, 0x40 ; load FSR1 with 0x40
LFSR 2, 0x6F ; load FSR2 with 0x6F
```

Because FSR0, FSR1, and FSR2 are 12-bit registers they cannot fit into the SFR address space unless they are split into pieces of an 8-bit size. That is exactly what PIC18 has done. The FSR registers have the low-byte and high-byte parts called FSRxL and FSRxH, as shown in the SFR table of Table 6-1. In Table 6-1 we see FSR0L and FSR0H, representing the low and high parts of the 12-bit FSR0 register. Note that the FSRxH is only 4-bit and the upper 4 bits are not used. Another register associated with the register indirect addressing mode is the INDF (indirect register). Each of the FSR0, FSR1, and FSR2 registers has an INDF register associated with it, and these are called INDF0, INDF1, and INDF2. When we move data into INDFx we are moving data into a RAM location pointed to by the FSR. In the same way, when we read data from the INDF register, we are reading data from a RAM location pointed to by the FSR. This is shown below.

```
LFSR 0, 0x30 ; FSR0 = 30H RAM location pointer
MOVF INDF0 ; copy contents of WREG into RAM
; location whose address is held by
; 12-bit FSR0 register
```

Advantages of register indirect addressing mode

One of the advantages of register indirect addressing mode is that it makes accessing data dynamic rather than static, as with direct addressing mode. Example 6-2 shows three cases of copying $55H$ into RAM locations $40H$ to $45H$. Notice in solution (b) that two instructions are repeated numerous times. We can create a loop with those two instructions as shown in solution (c). Solution (c) is the most efficient and is possible only because of the register indirect addressing mode. In Example 6-2, we must use "\texttt{INCF FSR0L, F}" to increment the pointer.
because there is no such instruction as "INCF FSR0, F". Looping is not possible in direct addressing mode, and that is the main difference between the direct and register indirect addressing modes. For example, trying to send a string of data located in consecutive locations of data RAM is much more efficient and dynamic using register indirect addressing mode than using direct addressing mode. See Example 6-3.

**Example 6-2**

Write a program to copy the value 55H into RAM memory locations 40H to 45H using

(a) Direct addressing mode.
(b) Register indirect addressing mode without a loop.
(c) A loop.

**Solution:**

(a)  
```
MOVLW 0x55 ;load WREG with value 55H
MOVF 0x40 ;copy WREG to RAM location 40H
MOVF 0x41 ;copy WREG to RAM location 41H
MOVF 0x42 ;copy WREG to RAM location 42H
MOVF 0x43 ;copy WREG to RAM location 43H
MOVF 0x44 ;copy WREG to RAM location 44H
```

(b)  
```
MOVLW 55H ;load with value 55H
LFSR 0,0x40 ;load the pointer. FSR0 = 40H
MOVF INDF0 ;copy W to RAM loc FSR0 points to
INCF FSR0L,F ;increment pointer. Now FSR0 = 41H
MOVF INDF0 ;copy W to RAM loc FSR0 points to
INCF FSR0L,F ;increment pointer. Now FSR0 = 42H
MOVF INDF0 ;copy W to RAM loc FSR0 points to
INCF FSR0L,F ;increment pointer. Now FSR0 = 43H
MOVF INDF0 ;copy W to RAM loc FSR0 points to
INCF FSR0L,F ;increment pointer. Now FSR0 = 44H
MOVF INDF0 ;copy W to RAM loc FSR0 points to
```

(c)  
```
COUNT EQU 0x10 ;location 10H for counter
MOVLW 0x5 ;WREG = 5
MOVF COUNT ;load the counter, Count = 5
LFSR 0,0x40 ;load pointer. FSR0 = 40H, RAM address
MOVLW 0x55 ;WREG = 55H value to be copied
B1 MOVF INDF0 ;copy WREG to RAM loc FSR0 points to
INCF FSR0L,F ;increment FSR0 pointer
DECF COUNT,F ;decrement the counter
BNZ B1 ;loop until counter = zero
```

Use the MPLAB simulator to examine RAM contents after the above program is run.

40 = (55)
41 = (55)
42 = (55)
43 = (55)
44 = (55)
Example 6-3

Assume that RAM locations 30–34H have a string of ASCII data, as shown below. Write a program to get each character and send it to Port B one byte at a time. Show the program using:

(a) Direct addressing mode.
(b) Register indirect addressing mode.

\[
\begin{align*}
30 &= ('H') \\
31 &= ('E') \\
32 &= ('L') \\
33 &= ('L') \\
34 &= ('O')
\end{align*}
\]

Solution:

(a) Using direct addressing mode

\[
\begin{align*}
\text{CLRF TRISB} &; \text{make Port B an output} \\
\text{MOVF 0x30, PORTB} &; \text{copy contents of loc 0x30 to PB} \\
\text{MOVF 0x31, PORTB} \\
\text{MOVF 0x32, PORTB} \\
\text{MOVF 0x33, PORTB} \\
\text{MOVF 0x34, PORTB}
\end{align*}
\]

(b) Using register indirect mode

\[
\begin{align*}
\text{COUNTREG EQU 0x20} &; \text{fileReg loc for counter} \\
\text{CNTVAL EQU 5} &; \text{counter value} \\
\text{CLRF TRISB} &; \text{make Port B an output (TRSI = 0 = out)} \\
\text{MOVLW CNTVAL} &; \text{WREG = 5} \\
\text{MOVWF COUNTREG} &; \text{load the counter, Count = 5} \\
\text{LFSR 2, 0x30} &; \text{load pointer. FSR2 = 30H, RAM address} \\
\text{MOVF INDF2,W} &; \text{copy RAM loc FSR2 points at to WREG} \\
\text{MOVWF PORTB} &; \text{copy WREG to PORTB} \\
\text{INCF FSR2L} &; \text{increment FSR2 to point at next loc} \\
\text{DECF COUNTRG,F} &; \text{decrement counter} \\
\text{BNZ B3} &; \text{loop until counter = zero}
\end{align*}
\]

When simulating the above program on the MPLAB, make sure that RAM locations 30H–34H have the message “HELLO”. Notice that “\text{MOVF INDF2, w}” moves data from INDF2 into WREG.

When using the MPLAB simulator with examples in this chapter, you may have noticed that you cannot view INDF0, POSTDEC0, or PLUSW0 in the watch window. This is because these registers are not physically implemented memory locations. Accessing these registers indicates indirect addressing. See Figure 2-4 in Chapter 2 for the registers that are not physically implemented.
Auto-increment option for FSR

Because the FSR is a 12-bit register, it can go from 000 to FFFH, which covers the entire 4K RAM space of the PIC18. Using the “INCF FSRnL, F” instruction to increment the pointer can cause a problem when an address such as 5FFH is incremented. The instruction “INCF FSRnL, F” will not propagate the carry into the FSR1H register. The PIC18 gives us the options of auto-increment and auto-decrement for FSRn to overcome this problem. The syntax used for such cases for the CLRF instruction is shown in Table 6-2.

Table 6-2: PIC18 Auto-Increment/Decrement of FSRn for CLRF Instruction

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLRF INDFn</td>
<td>After clearing fileReg pointed to by FSRn, the FSRn stays the same.</td>
</tr>
<tr>
<td>CLRF POSTINCn</td>
<td>After clearing fileReg pointed to by FSRn, the FSRn is incremented.</td>
</tr>
<tr>
<td>CLRF PREINCn</td>
<td>The FSRn is incremented, then fileReg pointed to by FSRn is cleared.</td>
</tr>
<tr>
<td>CLRF POSTDECs</td>
<td>After clearing fileReg pointed to by FSRn, the FSRn is decremented.</td>
</tr>
<tr>
<td>CLRF PLUSWn</td>
<td>Clears fileReg pointed to by (FSRn + WREG), FSRn &amp; W unchanged.</td>
</tr>
</tbody>
</table>

Note: This table shows the syntax for the CLRF instruction, it works for all such instructions. The auto-decrement or auto-increment affects the entire 12 bits of the FSRn and has no effect on status register. This means that FSR0 going from FFF to 000 will not raise any flag. The option of PLUSWn is widely used for a RAM-based look-up table. See Section 6.4.

Example 6-4

Write a program to clear 16 RAM locations starting at RAM address 60H. Use the following:
(a) INCF FSRnL
(b) Auto-increment

Solution:

(a)

COUNTREG EQU 0x10 ;fileReg loc for counter
CNTVAL EQU D'16' ;counter value
MOVLW CNTVAL ;WREG = 16
MOVF CNTREG ;load the counter, Count = 16
LFSR B2, 1,0x60 ;load pointer. FSR1 = 40H, RAM address
DECF COUNTREG,F ;decrement counter
BNZ B2 ;loop until counter = zero

(b)

COUNTREG EQU 0x10 ;fileReg loc for counter
CNTVAL EQU D'16' ;counter value
MOVLW CNTVAL ;WREG = 16
MOVF CNTREG ;load the counter, Count = 16
LFSR B3, 1,0x60 ;load pointer. SFR0 = 40H, RAM address
CLRF POSTINC1 ;clear RAM, increment FSR1 pointer
DECF COUNTREG,F ;decrement counter
BNZ B3 ;loop until counter = zero
Example 6-5

Write a program to copy a block of 5 bytes of data from RAM locations starting at 30H to RAM locations starting at 60H.

Solution:

```assembly
COUNTREG EQU 0x10 ; fileReg loc for counter
CNTVAL EQU D'5' ; counter value
MOVLW CNTVAL ; WREG = 10
MOVWF COUNTREG ; load the counter, count = 10
LFSR 0, 0x30 ; load pointer. FSR0 = 30H, RAM address
LFSR 1, 0x60 ; load pointer. FSR1 = 60H, RAM address
B3 MOVF POSTINC0,W ; copy RAM to WREG and increment FSR0
MOVWF POSTINC1 ; copy WREG to RAM and increment FSR1
DECFS COUNTREG,F ; decrement counter
BNZ B3 ; loop until counter = zero
```

Before we run the above program.

\[30 = (\text{'H'}) \quad 31 = (\text{'E'}) \quad 32 = (\text{'L'}) \quad 33 = (\text{'L'}) \quad 34 = (\text{'O'})\]

After the program is run, the addresses 60–64H have the same data as 30–34H.

\[30 = (\text{'H'}) \quad 31 = (\text{'E'}) \quad 32 = (\text{'L'}) \quad 33 = (\text{'L'}) \quad 34 = (\text{'O'})\]
\[60 = (\text{'H'}) \quad 61 = (\text{'E'}) \quad 62 = (\text{'L'}) \quad 63 = (\text{'L'}) \quad 64 = (\text{'O'})\]

Example 6-6

Assume that RAM locations 40–43H have the following hex data. Write a program to add them together and place the result in locations 0x06 and 0x07.

\[40 = (\text{7D}) \quad 41 = (\text{EB}) \quad 42 = (\text{C5}) \quad 43 = (\text{5B})\]

Solution:

```assembly
COUNTREG EQU 0x20 ; fileReg loc for counter
L_BYTE EQU 0x06 ; fileReg loc for L_Byte
H_BYTE EQU 0x07 ; fileReg loc for H_Byte
CNTVAL EQU 4 ; counter value
MOVLW CNTVAL ; WREG = 4
MOVWF COUNTREG ; load the counter
LFSR 0, 0x40 ; load pointer. FSR0 = 40H, RAM address
CLRIF WREG ; clear WREG
CLRIF H_BYTE ; clear H_BYTE
B5 ADDWF POSTINC0,W ; add RAM to WREG and increment FSR0
BNC OVER ; if C = 0, go to next
INCF H_BYTE,F ; C = 1, add 1 to high byte
OVER DECF COUNTREG,F ; decrement counter
BNZ B5 ; loop until counter = zero
MOVWF L_BYTE
```

The above is a register indirect version of Example 5-2 in Chapter 5 with a loop. Contrast them to see the difference.
To see an example of how to use all three FSRn registers, study and simulate Example 6-7.

**Example 6-7**

Write a program to add the following multibyte BCD numbers and save the result.

\[ \begin{align*}
12896577H + 23647839H
\end{align*} \]

**Solution:**

COUNTREG EQU 0x20 ;fileReg loc for counter
CNTVAL EQU D'4' ;counter value
MOVLW CNTVAL ;WREG = 4
MOVWF COUNTREG ;load the counter. Count = 4
LFSR 0,0x30 ;load pointer. FSR0 = 30H, RAM address
LFSR 1,0x50 ;load pointer. FSR1 = 50H, RAM address
LFSR 2,0x60 ;load pointer. FSR2 = 60H, RAM address
BCF STATUS,C ;clear carry flag for the LSB
B3 MOVF POSTINC0,W ;copy RAM to WREG and INC FSR0
ADDFC POSTINC1,W ;add RAM to WREG and INC FSR1
DAW ;decimal adjust WREG
MOVWF POSTINC2 ;copy WREG to RAM and INC FSR2
DECF COUNTREG,F ;decrement counter
BNZ B3 ;loop until counter = zero

Before the addition we have:

<table>
<thead>
<tr>
<th>MSByte</th>
<th>LSByte</th>
</tr>
</thead>
<tbody>
<tr>
<td>33 = (12)</td>
<td>32 = (89)</td>
</tr>
<tr>
<td>31 = (65)</td>
<td></td>
</tr>
<tr>
<td>53 = (23)</td>
<td>52 = (64)</td>
</tr>
<tr>
<td>51 = (78)</td>
<td></td>
</tr>
<tr>
<td>30 = (77)</td>
<td>50 = (39)</td>
</tr>
</tbody>
</table>

After the addition we have:

<table>
<thead>
<tr>
<th>MSByte</th>
<th>LSByte</th>
</tr>
</thead>
<tbody>
<tr>
<td>63 = (36)</td>
<td>62 = (54)</td>
</tr>
<tr>
<td>61 = (44)</td>
<td>60 = (16)</td>
</tr>
</tbody>
</table>

Notice that we are using the little endian convention of storing a low byte to a low address and a high byte to a high address. Single-step the program in MPLAB and examine the FSRx and memory contents to gain an insight into register indirect addressing mode.

**Review Questions**

1. The instruction “MOVWF 0x40” uses ______ addressing mode. Why?
2. What address is assigned to register FSR0L?
3. What address is assigned to register FSR0H?
4. The FSRn is a(n) ______-bit register.
5. Which registers are allowed to be used for register indirect addressing mode if the data is in the data RAM file register?
SECTION 6.3: LOOK-UP TABLE AND TABLE PROCESSING

So far, we have seen that the PIC18 has a maximum of 2M of code (program) space and 4K of data RAM space. While we never use any of the data RAM space for storing code, we can use the code space to store fixed data. In this section we discuss how to access fixed data residing in the program ROM space of the PIC18. First we examine how to store fixed data in the program ROM space using the DB (define byte) directive.

DB (define byte) and fixed data in program ROM

The DB data directive is widely used to allocate ROM program (code) memory in byte-sized chunks. In other words, DB is used to define an 8-bit fixed data. When DB is used to define fixed data, the numbers can be in decimal, binary, hex, or ASCII formats. The DB directive is used to define ASCII strings. See Example 6-8. In Example 6-8, notice that we must use single quotes (') for a single character or double quotes (") for a string.

Example 6-8

Assume that we have burned the following fixed data into program ROM of a PIC chip. Give the contents of each ROM location starting at 500H. See Appendix F for the hex values of the ASCII characters.

;MY DATA IN ROM
  ORG 500H ;notice it must be an even address
  DATA1 DB '28' ;DECIMAL (1C in hex)
  DATA2 DB '00110101' ;BINARY (35 in hex)
  DATA3 DB 0x39 ;HEX
  ORG 510H ;notice it must be an even address
  DATA4 DB 'Y' ;single ASCII char
  DATA5 DB '2','0','0','5';ASCII numbers
  ORG 518H ;notice it must be an even address
  DATA6 DB "Hello ALI" ;ASCII string
END

Solution:

<table>
<thead>
<tr>
<th>DATA1</th>
<th>DATA2</th>
<th>DATA3</th>
</tr>
</thead>
<tbody>
<tr>
<td>500 = (1C)</td>
<td>501 = (35)</td>
<td>502 = (39)</td>
</tr>
<tr>
<td>DATA4</td>
<td>DATA5</td>
<td></td>
</tr>
<tr>
<td>510 = (59)</td>
<td>511 = (32)</td>
<td>512 = (30)</td>
</tr>
<tr>
<td>Y</td>
<td>2</td>
<td>0</td>
</tr>
<tr>
<td>DATA6</td>
<td></td>
<td></td>
</tr>
<tr>
<td>518 = (48)</td>
<td>519 = (65)</td>
<td>51A = (6C)</td>
</tr>
<tr>
<td>H</td>
<td>e</td>
<td>1</td>
</tr>
<tr>
<td>51D = (20)</td>
<td>51E = (41)</td>
<td>51F = (4C)</td>
</tr>
<tr>
<td>SPACE</td>
<td>A</td>
<td>L</td>
</tr>
</tbody>
</table>

MPASM also allows the use of DATA in place of DB to define values greater than 255 (0xFF) but not larger than 65535 (0xFFFF).
Reading table elements in the PIC18

Example 6-8 showed how to place fixed data into program ROM. The only problem is that the 2M of program (code) space is under the direct control of the program counter register. This means that we need to have a special function register to point to the data to be fetched from the code space. For this reason we can call it register indirect ROM addressing mode. This is an addressing mode widely used to access data elements located in the program ROM space of the PIC18. This is often called table processing.

There is a group of instructions in the PIC18 designed for table processing. These can be used for both table read and table write. We discuss table read first because it is the most widely used. Table 6-3 shows the instructions for table read of the PIC18. To read the fixed data byte, we need an address pointer that points to the data and a register to store the data when it is brought into the CPU. TBLPTR is a 21-bit register and is used to point to the byte to be fetched. With the 21-bit register TBLPTR, we can cover the entire 2M program (code) space for the PIC18. The only problem is that we do not have an instruction to load the 21-bit address into TBLPTR.

TBLPTR is divided into three 8-bit parts. These are called TBLPTRL (low), TBLPTRH (high), and TBLPTRU (upper), and all are part of the SFRs. Notice that the last 2 bits of TBLPTRU (upper) are not used and are cleared to 0s.

The other SFR register used for the table processing is TABLAT. The TABLAT (TABle LATch) register is used for keeping the byte once it is fetched into the CPU. See Example 6-9. In the next few examples, we load only the TBLPTRL and TBLPTRH registers because the table elements are stored in the first 64K of the PIC18 address space (0000–FFFFH). You must load TBLPTRU as well if the data is residing on ROM addresses of 10000H and beyond.

Auto-increment option for TBLPTR

Because the TBLPTR is 21-bit register it can cover from 000000 to 1FFFFFFH, which is the 2M ROM space of the PIC18. Using the “INC'TBLPTRL’r” instruction to increment the pointer can cause a problem when an address such as 5FFH is incremented. The carry will not propagate into TBLPTRH. The PIC18 gives us the options of TBLRD*+ (table read and increment), TBLRD*– (table read and decrement), and so on, as shown in Table 6-3. See Examples 6-10, 6-11, and 6-12.

Table 6-3: PIC18 Table Read Instructions

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBLRD*</td>
<td>Table read</td>
<td>After read, TBLPTR stays the same</td>
</tr>
<tr>
<td>TBLRD*+</td>
<td>Table read with post-inc.</td>
<td>Reads and increments TBLPTR</td>
</tr>
<tr>
<td>TBLRD*-</td>
<td>Table read with post-dec.</td>
<td>Reads and decrements TBLPTR</td>
</tr>
<tr>
<td>TBLRD*+</td>
<td>Table read with pre-inc.</td>
<td>Increments TBLPTR and then reads</td>
</tr>
</tbody>
</table>

Note: The byte of data is read into the TABLATch register from code space pointed to by TBLPTR.
In this program, assume that the word “USA” is burned into ROM locations starting at 500H, and that the program is burned into ROM locations starting at 0. Analyze how the program works and state where “USA” is stored after this program is run.

Solution:

```
ORG 0000H ;burn into ROM starting at 0
CLRF TRISB ;make PB an output
MOVLW 0x0 ;WREG = 0 look-up table low-byte addr
MOVF TBLPTRL ;look-up table low-byte addr
MOVLW 0x05 ;WREG = 5 look-up table high-byte addr
MOVF TBLPTRH ;look-up table high-byte addr
TBLRD* ;TABLAT = 'U' char pointed to by TABPTR
MOVF TBLAT, PORTB ;send it to Port B
INCF TBLPTRL, F ;TBLPTR = 01 pointing to next (501)
TBLRD* ;TABLAT = 'S' char pointed to by TBLPTR
MOVF TBLAT, PORTB ;send it to Port B
INCF TBLPTRL, F ;TBLPTR = 02 pointing to next (502)
TBLRD* ;TABLAT = 'A' char pointed to by TBLPTR
MOVF TBLAT, PORTB ;send it to Port B
HERE GOTO HERE ;stay here forever

;data is burned into code(program) space starting at 500H
ORG 500H
MYDATA DB "USA"
END ;end of program
```

In the above program ROM locations 500H–502H have the following contents.

500 = ('U') 501 = ('S') 502 = ('A')

We start with TBLPTR = 500H (TBLPTRH = 05 and TBLPTRL = 0). The instruction "TBLRD*" moves the contents of ROM location 500H to TABLAT. Register TABLAT contains 55H, the ASCII value for 'U'. This is moved to Port B. Next, TBLPTRL is incremented to make TBLPTR = 501H. The TBLRD instruction will get the contents of the next ROM location 501H, which holds character 'S'. After this program is run, we send the ASCII values for the characters 'U', 'S', and 'A' to Port B one character at a time. The loop version of this program is given in the next example.
Example 6-10

Assuming that program ROM space starting at 250H contains "USA", write a program to send all the characters to Port B one byte at a time.

Solution:

(a) This method uses a counter

RCOUNT EQU 0x20 ;counter loc in fileReg
CNTVAL EQU 0x3 ;counter value
ORG 0000H ;burn into ROM starting at 0
MOVLW 0x50 ;WREG = 50, low-byte addr
MOVWF TBLPTRL ;look-up table low-byte addr
MOVLW 0x02 ;WREG = 2, high-byte addr
MOVWF TBLPTRH ;look-up table high-byte addr
MOVLW CNTVAL ;WREG = 03, counter value
MOVWF RCOUNT ;load counter
CLRF TRISB ;TRSIB = 00 (Port B as output)

B6 TBLRD* ;read table byte pointed to by TBLPTR
MOVF TBLAT, W ;copy to WREG (Z = 1, if null)
BZ EXIT ;is it null char? exit if yes
MOVF PORTB ;send it to Port B
INCF TBLPTRL, F ;increment pointing to next char
DEC F RCOUNT, F ;dec the counter
BNZ B6 ;repeat if counter not zero
HERE GOTO HERE ;stay here

;data is burned into code(program) space starting at 250H
ORG 0x250
MYDATA DB "USA"
END

(b) This method uses null char for end of string

ORG 0000H ;burn into ROM starting at 0
MOVLW 0x50 ;WREG = 50, low-byte addr
MOVWF TBLPTRL ;look-up table low-byte addr
MOVLW 0x02 ;WREG = 2, high-byte addr
MOVWF TBLPTRH ;look-up table high-byte addr
CLRF TRISB ;TRSIB = 00 (Port B as output)

B7 TBLRD* ;bring in next byte
MOVF TBLAT, W ;copy to WREG (Z = 1, if null)
BZ EXIT ;is it null char? exit if yes
MOVF PORTB ;send it to Port B
INCF TBLPTRL, F ;increment pointing to next char
BRA B7 ;continue
EXIT GOTO EXIT

ORG 0x250
MYDATA DB "USA", 0 ;notice null
END
Repeat Example 6-10, using auto-increment.

Solution:

```
ORG 0000H ;burn into ROM starting at 0
MOVLW 0x50 ;WREG = 50 low-byte addr
MOVF TBLPTRL ;look-up table low-byte addr
MOVLW 0x02 ;WREG = 2, high-byte addr
MOVF TBLPTRH ;look-up table high-byte addr
CLRF TRISB ;TRISB = 00 (Port B as output)
B7 TBLRD*+ ;bring in next byte and inc TBLPTR
MOVLW 0x02 ;WREG = 2, high-byte addr
MOVWF PORTB ;send it to Port B
BRA B7 ;continue
EXIT GOTO EXIT
```

Example 6-12

Assume that ROM space starting at 500H contains the message "The Promise of World Peace". Write a program to bring it into CPU one byte at a time and place the bytes in RAM locations starting at 40H.

Solution:

```
ORG 0000H ;burn into ROM starting at 0
MOVLW 0x00 ;WREG = 00 low-byte addr
MOVF TBLPTRL ;look-up table low-byte addr
MOVLW 0x05 ;WREG = 05, high-byte addr
MOVF TBLPTRH ;look-up table high-byte addr
LFSR 2,0x40 ;load pointer. FSR2 = 40H, RAM address
B8 TBLRD*+ ;read the table, then increment TBLPTR
MOVF TABLAT,W ;copy to WREG (Z = 1 if null)
BZ EXIT ;is it null char? exit if yes
MOVWF POSTINC2 ;copy WREG to RAM and INC FSR2
BRA B8
EXIT GOTO EXIT
```

Look-up table and RETLW instruction

The look-up table is a widely used concept in microcontroller programming. It allows access to elements of a frequently used table with minimum operations. As an example, assume that for a certain application we need $x^2$ values in the range of 0 to 9. We can use a look-up table instead of calculating the values,
which takes some time. In the PIC, to get the table element we first call the look-
up table, then we add a fixed value to the PCL (low-byte portion of the program
counter) to index into the look-up table. Upon return from the table, the RETLW
instruction will provide the desired look-up table element in the WREG register.
This is shown in Examples 6-13 and 6-14.

Example 6-13

Assume that the lower three bits of Port C are connected to three switches. Write a pro-
gram to send the following ASCII characters to Port D based on the status of the
switches.

<p>| | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>'0'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>001</td>
<td>'1'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>010</td>
<td>'2'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>011</td>
<td>'3'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>100</td>
<td>'4'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>101</td>
<td>'5'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>110</td>
<td>'6'</td>
<td></td>
<td></td>
</tr>
<tr>
<td>111</td>
<td>'7'</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Solution:

```
ORG 0
SETF TRISC ;TRISC = FFh (Port C as input)
CLR TRISD ;TRISD = 00 (Port D as output)
B1 MOVF PORTC,W ;read x from Port C into WREG
ANDLW B'00000111' ;mask upper 5 bits
CALL ASCII_TABLE ;get ASCII from look-up table
MOVWF PORTD ;copy it to Port D
BRA B1 ;continue
```

;look-up table for ASCII numbers 0-7

```
ASCII_TABLE
MULLW 0x2 ;align it for even address for 2-byte RETLW opcode
MOVFF PRODL, WREG ;put it into WREG for indexing
ADDWF PCL ;PCL = PCL + WREG
RETLW '0' ;ASCII for 0
RETLW '1' ;ASCII for 1
RETLW '2' ;ASCII for 2
RETLW '3' ;notice that each ASCII value is placed
RETLW '4' ;in the ROM at an even address
RETLW '5'
RETLW '6'
RETLW '7'
END
```

![Program Memory](image-url)
Write a program to get the $x$ value from Port B and send $x^2$ to Port C. Assume that RB3–RB0 has the $x$ value of 0–9. Use a look-up table instead of a multiply instruction.

What is the value of Port C if we have 9 at Port B?

**Solution:**

```assembly
ORG 0
SETF TRISB ; TRISB = FFh (Port B as input)
CLRF TRISC ; TRISC = 00 (Port C as output)
B1 MOVF PORTB,W ; read $x$ from Port B into WREG
ANDLW 0x0F ; mask upper bits
CALL XSQR_TABLE ; get $x^2$ from the look-up table
MOVWF PORTC ; copy it to Port C
BRA B1 ; continue

; look-up table for square of numbers 0-9
XSQR_TABLE
  MULLW 0x2 ; align it for even address
  MOVFF PRODL, WREG ; put it into WREG for indexing
  ADDWF PCL ; PCL = PCL + WREG
  RETLW D'0' ; square of 0
  RETLW D'1' ; square of 1
  RETLW D'4' ; square of 2
  RETLW D'9' ; square of 3
  RETLW D'16' ; square of 4 (10 hex)
  RETLW D'25' ; square of 5 (19 hex)
  RETLW D'36' ; square of 6 (24 hex)
  RETLW D'49' ; square of 7 (31 hex)
  RETLW D'64' ; square of 8 (40 hex)
  RETLW D'81' ; square of 9 (51 hex)
END
```

From the screen shot above, notice that location 001A has the "RETLW D'0'" opcode and operand, the square of 0. Location 001C has 01, the square of 1. Location 001E has 04, the square of 2. Location 0020 has 09, the square of 3. Location 0022 has 10, the square of 4 ($4 \times 4 = 16 = 10H$) and so on. Notice that the odd addresses have the opcode for RETLW, which is 0C hex. If we have 9 at Port B, then Port C will have 51H, which is the hex value of decimal 81 ($9^2 = 81$).
Accessing a look-up table in RAM

The look-up table elements can also be in RAM instead of ROM. Sometimes we need to bring in the elements of the look-up table from RAM because the elements are dynamic and can change. The PIC18 allows us to do that using the FSR as pointer. For example, the instruction “MOVFF PLUSW2, PORTD” will bring elements of the look-up table pointed to by the address location formed by the addition of FSR2 + WREG. In this case, WREG is used as an index into the look-up table. See Examples 6-15 and 6-16.

Example 6-15

Repeat Example 6-13 assuming that the look-up table elements are in data RAM locations starting at address 20H as shown below.

20 = (‘0’)
21 = (‘1’)
22 = (‘2’)
23 = (‘3’)
24 = (‘4’)
25 = (‘5’)
26 = (‘6’)
27 = (‘7’)

Solution:

```asm
ORG 0
SETF TRISC ; TRISC = FFh (Port C as input)
CLR TRISD ; TRISD = 00 (Port D as output)
LFSR 2, 0x20 ; load pointer. FSR2 = 20H, RAM address
B1 MOVF PORTC, W ; read x from Port C into WREG
ANDLW B'00000111' ; mask upper 5 bits
MOVFF PLUSW2, PORTD ; get data pointed to by FSR2 + WREG
BRA B1
END
```

When simulating this program on your MPLAB, make sure the RAM locations 20–27H have the elements of the look-up table. Notice that the “MOVFF PLUSW2, PORTD” instruction will bring the value from its RAM location and put it on PORTD.
### Example 6-16

Write a program to get the $x$ value from Port B and send $x^2 + 2x + 3$ to Port C. Assume PB3–PB0 has the $x$ value of 0–9. Use a look-up table instead of a multiply instruction.

#### Solution:

```assembly
ORG 0
SETF TRISB ; TRISB = FFh (Port B as input)
CLR TRISC ; TRISC = 00 (Port C as output)
B1 MOVF PORTB,W ; read $x$ from Port B into WREG
ANDLW 0x0F ; mask upper bits
CALL XSQR_TABLE ; get $x^2$ from the look-up table
MOVWF PORTC ; copy it to Port C
BRA B1 ; continue

XSQR_TABLE
MULLW 0x2 ; align it for even address
MOVFF PRODL, WREG ; put it into WREG for indexing
ADDWF PCL ; PCL = PCL + WREG
RETLW D'3' ; (0) + 2(0) + 3 = 3
RETLW D'6' ; (1) + 2(1) + 3 = 6
RETLW D'11' ; (2) + 2(2) + 3 = 11
RETLW D'18' ; (3) + 2(3) + 3 = 18
RETLW D'27' ; (4) + 2(4) + 3 = 27
RETLW D'38' ; (5) + 2(5) + 3 = 38
RETLW D'51' ; (6) + 2(6) + 3 = 51
RETLW D'66' ; (7) + 2(7) + 3 = 66
RETLW D'83' ; (8) + 2(8) + 3 = 83
RETLW D'102' ; (9) + 2(9) + 3 = 102
END
```

### Writing table elements in PIC18

In PIC18 we also have the TBLWRT instruction, which allows us to write (store) data into program ROM space. While the TBLRD instruction can be used with any family member of the PIC18 regardless of the type of ROM it has, the TBLWRT can be used only with PIC18 chips that have flash ROM for the program ROM space. The TBLWRT instruction will not work for PIC18 chips with OTP (one-time programmable) or mask ROM. In mask ROM, the information (code and data) is burned into the ROM during the chip fabrication by the Microchip Corp. Because writing to flash ROM involves manipulating the configuration bits it is discussed in Chapter 14.

### Review Questions

1. The instruction "TBLRD*" uses register _____ as address pointer.
2. What register is incremented upon execution of the TBLRD* instruction?
3. What register is incremented upon execution of the TBLRD* instruction?
4. What is the size of TBLPTR? How much ROM space does it cover?
5. What register is incremented upon execution of the TBLRD++ instruction?
6. What is the difference between the TBLRD++ and TBLRD** instructions?
7. True or false. The TBLWT instruction works with all ROM versions of the PIC18 family.

CHAPTER 6: BANK SWITCHING, TABLES, MACROS, AND MODULES 213
SECTION 6.4: BIT-ADDRESSABILITY OF DATA RAM

Many microprocessors such as the 386 or Pentium allow programs to access registers and I/O ports in byte size only. In other words, if you need to check a single bit of an I/O port, you must read the entire byte first and then manipulate the whole byte with some logic instructions to get hold of the desired single bit. This is not the case with the PIC as we saw in Chapter 4. Indeed, one of the most important features of the PIC is its ability to access the file register’s RAM location in bits as well as bytes. This means that all I/O ports, SFRs, and general purpose RAM areas for the PIC18 are bit-addressable because they are part of the file register data RAM. WREG is also bit-addressable because it is part of the SFRs. This is a very powerful feature of the PIC18 family. In this section, we provide more programming examples of the bit-addressable option of the PIC18 family.

Bit-addressable file register data RAM

The entire 4096 bytes of file register data RAM of the PIC18 are bit-addressable. This means that while ROM program space is only byte-addressable, the 4K of data RAM is both byte- and bit-addressable. To distinguish between the byte-addressable and bit-addressable options of data RAM, the PIC18 provides two categories of instructions: bit-oriented and byte-oriented instructions. Bit-oriented instructions are called bit-addressable, while byte-oriented instructions are referred to byte-addressable.

The bit-oriented instructions are given in Table 6-4. Notice that the bit-oriented instructions use only one addressing mode, the direct addressing mode. In the first three sections of this chapter we showed various addressing modes of byte-addressable space of the PIC18, among them register indirect addressing mode for both data RAM and program (code) ROM. Note that there is no register indirect addressing mode for bit-oriented instructions in the PIC18.

File register bit-addressability

As we discussed in Chapter 2, the PIC18 can have up to 4096 bytes of file register data RAM depending on the family member. We can access either the entire 8 bits or any single bit without altering the rest. When accessing a file register in a single-bit manner, we use the syntax “Bit-Oriented-instr fileReg, x” where fileReg is any register in the file register and X is the desired bit number from 0 to 7 for data bits D0 to D7. For example, the instruction “BTG 0x20, 7” will toggle D7 of RAM location 20H. As we mentioned earlier in this chapter, every register, including WREG, is assigned a byte address in the file register and

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>BSF fileReg.bit</td>
<td>Bit Set fileReg (set the bit: bit = 1)</td>
</tr>
<tr>
<td>BCF fileReg.bit</td>
<td>Bit Clear fileReg (clear the bit: bit = 0)</td>
</tr>
<tr>
<td>BTG fileReg.bit</td>
<td>Bit Toggle fileReg (complement the bit)</td>
</tr>
<tr>
<td>BTFSC fileReg.bit</td>
<td>Bit test fileReg, skip if clear (skip next instruction if bit = 0)</td>
</tr>
<tr>
<td>BTFSS fileReg.bit</td>
<td>Bit test fileReg, skip if set (skip next instruction if bit = 1)</td>
</tr>
</tbody>
</table>

Note: fileReg can be any location of file register data RAM.
ports PORTA–PORTE are part of the file register. For example, “BSF PORTB, 5” sets high bit RB5 of Port B. Notice that when code such as “BSF PORTB, 5” is assembled, it becomes “8A81” because Port B has the RAM address of 81H in the access bank. Examine the next few examples in this section to gain insight into the bit addressability of the file register in the PIC18.

Example 6-17

A switch is connected to pin RC7. Write a program to check the status of the switch and perform the following:

(a) If switch = 0, send letter ‘N’ to Port D.
(b) If switch = 1, send letter ‘Y’ to Port D.

Solution:

```
BSF TRISC,7 ; make RC7 an input
CLRF TRISD ; make Port D an output port
AGAIN BTFSS PORTC,7 ; bit test RC7 for HIGH
BRA OVER ; it must be LOW
MOVLW A'Y' ; WREG = 'Y' ASCII letter Y
MOVWF PORTD ; issue WREG to PD
GOTO AGAIN ; we could use BRA instead
OVER MOVLW A'N' ; WREG = 'N' ASCII letter N
MOVWF PORTD ; issue WREG to PORTD
GOTO AGAIN ; we can use BRA too
```

Example 6-18

Write a program to toggle RB1 a total of 200 times. Use file register RAM location 32H to hold your counter value.

Solution:

```
MYREG EQU 0x32 ; set aside loc 0x20 reg
CNTVAL EQU D'200' ; load counter into WREG
MOVLW CNTVAL
MOVWF MYREG ; load the count into MYREG location
BCF TRISB,1 ; TRISB bit = 0, make RB1 an output
AGAIN BTG PORTB,1 ; toggle bit RB1
DECFW MYREG,F ; decrement MYREG
BNZ AGAIN ; continue until counter is zero
```
### Example 6-19

A switch is connected to pin RC7. Write a program to get the status of the switch and perform the following.

(a) If RC7 = 0, increment Port B.
(b) If RC7 = 1, decrement Port B.

**Solution:**

```
BSF TRISC, 7 ; make RC7 an input
CLRF TRISB ; make Port B an output port
AGAIN BTFSS PORTC, 7 ; bit test RC7 for HIGH
BRA OVER ; it must be LOW
INCF PORTB,F ; increment
GOTO AGAIN ; we can use BRA too
OVER DECF PORTB,F ; decrement
GOTO AGAIN ; we can use BRA too
```

### Example 6-20

A switch is connected to pin RB0. Write a program to get the status of the switch and save it in D0 of fileReg location 0x20.

**Solution:**

```
MYBITREG EQU 0x20 ; set aside loc 0x20 reg
BSF TRISB, 0 ; make RB0 an input
AGAIN BTFSS PORTB, 0 ; bit test RB0, skip if set
GOTO OVER ; it must be LOW (BRA is OK too)
BSF MYBITREG, 0 ; set bit D0 = 1
GOTO AGAIN ; we can use BRA too
OVER BCF MYBITREG, 0 ; clear D0 (D0 = 0)
GOTO AGAIN ; we can use BRA too
```

### Example 6-21

Write a program to see if the RAM location 37H contains an even value. If so, send it to Port B. If not, make it even and then send it to Port B.

**Solution:**

```
MYREG EQU 0x37 ; set aside loc 0x37 reg
CLRF TRISB ; make Port B an output port
AGAIN BTFSS MYREG, 0 ; bit test D0, skip if set
GOTO OVER ; it must be LOW
BCF MYREG, 0 ; clear bit D0 = 0
OVER MOVFF MYREG, PORTB ; copy it to Port B
GOTO AGAIN ; we can use BRA too
```
Example 6-22

Write a program for Port B to count up from 0000 to 1111 (binary).

Solution:

```
MYREG EQU 0x20
CLRFB TRISB               ;TRISB = 0, make PB output
CLRFB PORTB              ;Port B = 0
AGAIN INCF PORTB,F       ;increment Port B
BTFSS PORTB,4            ;test D4 bit of Port B
BRA AGAIN
GOTO $                   
```

Notice how it counts up from 0000 to 1111 and, when it becomes 10000, skips the loop.

Example 6-23

Write a program to check the status of the D7 bit of file register location 0x20 and make the following decisions:

(a) If D7 = 0, send “NO” to Port B.
(b) If D7 = 1, send “YES” to Port B.

Solution:

```
MYREG EQU 0x20
CLRFB TRISB               ;make Port B an output port
AGAIN BTFSS MYREG,7       ;bit test for HIGH
BRA OVER                  ;it must be LOW
MOVLW A'Y'                ;WREG = 'Y' ASCII letter Y
MOVWF PORTB               ;issue WREG to Port B
MOVLW A'E'                ;WREG = 'E' ASCII letter Y
MOVWF PORTB               ;issue WREG to Port B
MOVLW A'S'                ;WREG = 'S' ASCII letter Y
MOVWF PORTB               ;issue WREG to Port B
GOTO AGAIN                ;we can use BRA too
OVER MOVLW A'N'           ;WREG = 'N' ASCII letter N
MOVWF PORTB               ;issue WREG to Port B
MOVLW A'O'                ;WREG = 'O' ASCII letter Y
MOVWF PORTB               ;issue WREG to Port B
GOTO AGAIN                ;we can use BRA too
```

Status register bit-addressability

Of the bit-addressable registers, we will concentrate on the familiar status register. The rest will be discussed in future chapters.

Now let’s see how we can use bit-addressability of the status register. As we discussed in Chapter 2, 5 bits in the status register are set aside for the flags C, DC, Z, N, and OV. See Figure 6-2 and Example 6-24.
Example 6-24

While we have instructions such as BC (branch carry) and BZ (branch zero) to check the carry and zero flag bits, show how would you use the status register flag to check the (a) C and (b) Z flags.

Solution:

(a) The C flag is D0 of the status register; therefore, we can use the following instruction to check the C flag:

`BTFSS STATUS, C ;bit test C, skip if C = 1`

(b) The Z flag is D3 of the status register; therefore, we can use the following instruction to check the Z flag:

`BTFSS STATUS, Z ;bit test Z, skip if Z = 1`

Review Questions

1. True or false. All I/O ports of the PIC18 are bit-addressable.
2. True or false. The status register of the PIC18 is bit-addressable.
3. True or false. All file register RAM locations of the PIC are bit-addressable.
4. Indicate which of the following registers are bit-addressable.
   (a) Port A  (b) Port B  (c) WREG  (d) status register  (e) 21-bit PC
5. Of the 4096 bytes of RAM in the PIC18, how many bytes are bit-addressable?
6. How would you check to see whether bit D1 of RAM location 3 is HIGH or LOW?
7. State what each instruction does.
   (a) BSF 0x20,1  (b) BCF 0x32,7  (c) BSF 0x12,2
   (d) BSF PORTB,4  (e) BSF STATUS,1
8. Show how to clear the carry flag.
The PIC18 microcontroller has a maximum of 4K of data RAM space. Although not all members of the family have the entire RAM installed, every member of the family has at least the access bank for the file register. The file register RAM is divided into banks of 256 bytes each, which gives us a total of 16 banks in the PIC18. The minimum bank that every PIC18 has is called the access bank, as we discussed in Chapter 2. The access bank is made of 128 bytes of lower addresses and 128 bytes of higher addresses. While the lower 128 bytes of address space 000–07FH are used for general-purpose RAM, the higher 128 bytes are dedicated to the SFRs (special function registers) residing in address space F80–FFFH. The vast majority of the PIC18 chips we see on the Microchip web site have more than just the access bank. In this section we show how to use bank switching to take advantage of the entire data RAM space of the PIC18.

The A bit and bank switching

All the instructions we have used so far assumed the access bank as the default bank. This was achieved by ignoring the letter A in instructions such as “MOVWF fileReg, A”. In other words, the instruction “MOVWF fileReg” is really “MOVWF fileReg, A” where the A bit can be 0 or 1. If A = 0, then the access bank is the default bank. If A = 1, however, then the instruction will use the bank selector register (BSR) to select the bank instead of using the access bank. If A is not stated in a given instruction, it means A = 0 and the access bank is the default bank. That is what we have done so far for the simple reason of making the PIC18 Assembly language easier to understand and master. Next, we examine the role of the BSR register in bank switching.

The BSR register and bank switching

To use banks other than the access bank, we need to set bit A = 1 in the coding of the instruction. With A = 1, we use the BSR (bank select register) to choose the desired bank. The BSR is an 8-bit register and is part of the SFRs. Of the 8 bits of the BSR, only 4 least-significant bits are used in the PIC18. The upper 4 bits are set to zero and are ignored by the PIC18. The 4-bit BSR gives us 16 banks, and because each bank is 256 bytes, we cover the entire 4096 (16 × 256 = 4096) bytes of the data RAM file register using bank switching. The 4K (4096) bytes of the data RAM are organized as banks 0 to F, where the lowest bank, 0, has an address of 00–FFH, and the highest bank is bank F with the addresses of F00–FFFH. In the PIC18, the last 128 bytes of bank F are always set aside for the SFRs, while general purpose registers always start at address 0 of bank 0. Upon power-on reset, BSR = 0 (0000 binary), which indicates that only the lowest addresses of data RAM, from 000 to 0FFH, can be used for the general-purpose register in addition to the SFRs, which always reside in the last half of bank F. Similarly, if we make BSR = 1 (0001 binary), then PIC18 selects bank 1 using the 100–1FFH addresses in addition to the SFRs, which use only the last half of the bank with addresses of F80–FFFH. To select bank 2, we load BSR with the value 02 (0010 binary), which allows access to the bank addresses 200–2FF in addition to the SFR addresses of F80–FFFH. As we can see, no matter how much data RAM we have...
in the PIC18, the GP register always starts at address 000 and goes up, while the SFRs start at the other end of the 4 KB, at address FFF, and come down. At the present time PIC is using only the highest 128 bytes of bank F (F80H–FFFH address) for the SFRs. In the future they might start to use the rest of bank F and may even use bank E for SFRs, if the special functions embedded into the PIC18 keep increasing. Although the number of bytes in bank F used for SFRs in the PIC18 chip varies depending on the functions embedded into the chip, the SFRs always start at address FFF and go down. This point must be emphasized. For example, in the case of the access bank, the last half of bank F is set aside for SFRs, even though in some family members not all 128 bytes are needed due to the limited number of functions supported by that chip. See Table 6-5 for data RAM file registers of some PIC18 chips. Note that although we can use any of the addressing modes, such as immediate, direct, or register indirect to access the GP register regions, we use only direct addressing mode in accessing the SFR registers. To gain a better understanding of the bank switching, we use the PIC18F458 chip to show some examples.

![Diagram of Data RAM Registers](image)

**Table 6-5: Data RAM Bank for Selected PIC18 Chips**

<table>
<thead>
<tr>
<th>File Register</th>
<th>SFR</th>
<th>Available space for GPR</th>
</tr>
</thead>
<tbody>
<tr>
<td>(Bytes)</td>
<td>(Bytes)</td>
<td>(Bytes)</td>
</tr>
<tr>
<td>PIC18F1220</td>
<td>512</td>
<td>256</td>
</tr>
<tr>
<td>PIC18F452</td>
<td>1792</td>
<td>256</td>
</tr>
<tr>
<td>PIC18F2220</td>
<td>768</td>
<td>256</td>
</tr>
<tr>
<td>PIC18F458</td>
<td>1792</td>
<td>256</td>
</tr>
<tr>
<td>PIC18F8722</td>
<td>4096</td>
<td>158</td>
</tr>
</tbody>
</table>

*Note: The newer versions of the PIC18F458/452 are the PIC18F4580/4520.*

Extracted from http://www.microchip.com
Bank switching and "INCF F, D, A" instruction

The PIC18F458 has a total of 1792 bytes for the data RAM file register. The bank organization for the PIC18F458 is shown in Figure 6-3. All the examples we have seen so far ignored the A bit in the instruction, which means that A = 0. With A = 0, the access bank is the default bank. Now to use banks other than the access bank, two things must be done:

1. Load the BSR with the desired bank number, and

Therefore, instruction "INCF MYREG, F, 1" has a totally different meaning from "INCF MYREG, F, 0". The A = 1 means to use the bank pointed to by BSR. In the following code, we first load the bank number into the BSR register, and then manipulate the contents of RAM location 0x240 (location 40 of bank 2):

```
MYREG EQU 0x40

MOVLB 0x2 ;load 2 into BSR (use bank 2)
MOVLW 0 ;WREG = 0
MOVF MYREG, 1 ;loc 0x240 = (0), WREG = 0, Notice A = 1
INCF MYREG, F, 1 ;loc 0x240 = (1), WREG = 0, Notice A = 1
INCF MYREG, F, 1 ;loc 0x240 = (2), WREG = 0, Notice A = 1
INCF MYREG, F, 1 ;loc 0x240 = (3), WREG = 0
```

Contrast the above program with the one below:

```
MOVLB 0x2 ;load 2 into BSR (use bank 2)
MOVLW 0 ;WREG = 0
MOVF MYREG ;loc 0x40 = (0), WREG = 0
INCF MYREG, F ;loc 0x40 = (1), WREG = 0, Notice A = 0
INCF MYREG, F ;loc 0x40 = (2), WREG = 0, Notice A = 0
INCF MYREG, F ;loc 0x40 = (3), WREG = 0
```

Although we loaded BSR, because the A bit was not indicated, MPASM defaults it to zero, which means to use location 0x40 of the access bank. The A bit in the instruction field is given in Figure 6-4.

```
0010 10DA ffff ffff
D = F, destination is fileReg
D = W, destination is WREG
A = 0, use default access bank
A = 1, use bank pointed to by BSR (Bank Selector Register)
```

Figure 6-4. A Bit in the Instruction Field for INCF
Examine the following code to see the role of the D and A bits:

```assembly
MOVLB 0x2 ;load 2 into BSR (use bank 2)
MOVLW 0 ;WREG = 0
MOVWF 0x20,1 ;loc 0x220 = (0), WREG = 0, D = W, A = 1 means Bank 2
INCF 0x20,W,1 ;loc 0x220 = (0), WREG = 1, D = W, A = 1
INCF 0x20,W,1 ;loc 0x220 = (0), WREG = 1, D = W, A = 1
INCF 0x20,F,1 ;loc 0x220 = (1), WREG = 1, D = F, A = 1
INCF 0x20,F,1 ;loc 0x220 = (2), WREG = 1, D = F, A = 1
INCF 0x20,F,1 ;loc 0x220 = (3), WREG = 1, D = F, A = 1
INCF 0x20,F,1 ;loc 0x220 = (4), WREG = 1, D = F, A = 1
```

Simulate the next few examples with MPLAB and examine the data RAM memory to see how bank switching works.

### Example 6-25

Write a program to copy the value 55H into RAM memory locations 340H to 345H using:

(a) direct addressing mode.
(b) a loop.

**Solution:**

(a) 
```assembly
MOVLB 0x3 ;BANK 3
MOVLW 0x55 ;load WREG with value 55H
MOVWF 0x40,1 ;copy WREG to RAM location 340H
MOVWF 0x41,1 ;copy WREG to RAM location 341H
MOVWF 0x42,1 ;copy WREG to RAM location 342H
MOVWF 0x43,1 ;copy WREG to RAM location 343H
MOVWF 0x44,1 ;copy WREG to RAM location 344H
```

(b) 
```assembly
COUNT EQU 0x10 ;loc 10h
MOVLB 0x3 ;BANK 3
MOVLW 0x5 ;WREG = 5
MOVWF COUNT ;load the counter, count = 5
LFSR 0,0x340 ;load pointer. FSR0 = 40H, RAM address
MOVWF 0x55 ;WREG = 55H value to be copied
B1
MOVWF INDF,0 ;copy WREG to RAM loc FSR0 points to
INCF FSR0L ;increment FSR0L pointer
DECFSF COUNT,F,0 ;decrement the counter
BNZ B1 ;loop until counter = zero
```

The following shows RAM contents after the above program is run:

<table>
<thead>
<tr>
<th>Location</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>340</td>
<td>55</td>
</tr>
<tr>
<td>341</td>
<td>55</td>
</tr>
<tr>
<td>342</td>
<td>55</td>
</tr>
<tr>
<td>343</td>
<td>55</td>
</tr>
<tr>
<td>344</td>
<td>55</td>
</tr>
</tbody>
</table>
Table 6-6 shows the banks for various sizes of the data RAM in the PIC18 chip.

<table>
<thead>
<tr>
<th>Data Memory</th>
<th>Banks</th>
</tr>
</thead>
<tbody>
<tr>
<td>64</td>
<td>0, 15</td>
</tr>
<tr>
<td>128</td>
<td>0, 15</td>
</tr>
<tr>
<td>256</td>
<td>0, 15</td>
</tr>
<tr>
<td>512</td>
<td>0-1, 15</td>
</tr>
<tr>
<td>640</td>
<td>0-2, 15</td>
</tr>
<tr>
<td>768</td>
<td>0-2, 15</td>
</tr>
<tr>
<td>1024</td>
<td>0-3, 15</td>
</tr>
<tr>
<td>1280</td>
<td>0-4, 15</td>
</tr>
<tr>
<td>1536</td>
<td>0-5, 15</td>
</tr>
<tr>
<td>1792</td>
<td>0-6, 15</td>
</tr>
<tr>
<td>2048</td>
<td>0-7, 15</td>
</tr>
<tr>
<td>2304</td>
<td>0-8, 15</td>
</tr>
<tr>
<td>2560</td>
<td>0-9, 15</td>
</tr>
<tr>
<td>2816</td>
<td>0-10, 15</td>
</tr>
<tr>
<td>3072</td>
<td>0-11, 15</td>
</tr>
<tr>
<td>3328</td>
<td>0-12, 15</td>
</tr>
<tr>
<td>3584</td>
<td>0-13, 15</td>
</tr>
<tr>
<td>3840</td>
<td>0-14, 15</td>
</tr>
<tr>
<td>3968</td>
<td>0-15</td>
</tr>
</tbody>
</table>

**MOVFF and banks**

The great thing about the MOVFF instruction is that there is no need to worry about bank switching because it can move data anywhere within the 4K of RAM space. See Figure 6-4. Also see Example 6-26.

**Examining Data RAM space using MPLAB simulator**

The MPLAB simulator is a great tool to examine data RAM contents. We encourage its use to examine and verify the results of programs using data RAM.
Example 6-26

Assume RAM locations 330–334H of the PIC18F458 have the string of ASCII data shown below. Write a program to get each character and send it to Port B one byte at a time. Show the program using

(a) direct addressing mode.
(b) register indirect addressing mode.

\[
\begin{align*}
330 &= ('H') \\
331 &= ('E') \\
332 &= ('L') \\
333 &= ('L') \\
334 &= ('O') \\
\end{align*}
\]

Solution:

(a) Using direct addressing mode

```
CLRF TRISB ; make Port B an output
MOVFF 0x330, PORTB ; copy contents of loc 0x330 to PB
MOVFF 0x331, PORTB
MOVFF 0x332, PORTB
MOVFF 0x333, PORTB
MOVFF 0x334, PORTB
```

(b) Using register indirect addressing mode

```
COUNTER EQU 0x20 ; fileReg loc 20 for counter
CNTVAL EQU 5 ; counter value
CLRF TRISB ; make Port B an output (TRISB = out)
MOVLW CNTVAL ; WREG = 5
MOVF CANTREG ; load the counter, count = 5
B3 MOVF INDF2, W ; load pointer. FSR2 = 330H, RAM address
MOVF PORTB ; copy RAM loc FSR2 points at to WREG
INC FSR2L ; increment FSR2 to point at next loc
DECF CANTREG, F ; decrement counter
BNZ B3 ; loop until counter = zero
```

Figure 6-6. Example 6-26 Data RAM Dump in MPLAB
Example 6-27

Write a program for the PIC18F452 chip to put FFH into 16 RAM locations starting at RAM address 160H. Use:

(a) INCF FSRnL.
(b) auto-increment.

Solution:

(a)
COUNTREG EQU 0x10 ; fileReg loc for counter
CNTVAL EQU D'16' ; counter value
    MOVLL CNTVAL ; WREG = 16
    MOVWF COUNTREG ; load the counter, count = 16
    LFSR 1,0x160 ; load pointer. FSR1 = 60H, RAM address
    MOVLL 0xFF ; load 0xFF

B2
    MOVF INDFL,0 ; move W to RAM loc FSR1 points to
    INCF FSRIL ; increment FSRIL, point to next loc
    DECF COUNTREG,F ; decrement counter
    BNZ COUNTREG,B2 ; loop until counter = zero

(b)
COUNTREG EQU 0x10 ; fileReg loc for counter
CNTVAL EQU D'16' ; counter value
    MOVLL CNTVAL ; WREG = 16
    MOVWF COUNTREG ; load the counter, count = 16
    LFSR 1,0x160 ; load pointer. FSR1 = 160H, RAM address
    MOVLL 0xFF ; load 0xFF

B3
    MOVF POSTINC0,0 ; decrement counter
    DECF COUNTREG,F ; loop until counter = zero

Figure 6-7. Data Example 6-27 Data RAM Dump
Example 6-28

Write a program to copy a block of data from RAM locations 330H–33FH to 360H–36FH.

Solution:
COUNTREG EQU 0x20 ;fileReg loc 20 for counter
CNTVAL EQU 0x0F ;counter value = 15
CLRF TRISB ;make Port B an output (TRSIB = FFH)
MOVLW CNTVAL ;WREG = 15
MOVWF COUNTREG,1 ;load the counter, count = 15
LFSR 1,0x330 ;load pointer. FSR1 = 330H, RAM address
LFSR 2,0x360 ;load pointer. FSR2 = 360H, RAM address
B3 MOVFF INDF1, INDF2 ;increment FSR1 to point at next loc
INCF FSR1L ;increment FSR2 to point at next loc
DECF COUNTREG,F ;decrement counter
BNZ B3 ;loop until counter = zero

Figure 6-8. Before and After Example 6-28 is Run

Review Questions
1. True or false. The PIC18 uses the last bank for the SFRs.
2. True or false. The PIC18 uses a total of 256 bytes for each bank.
3. True or false. The first 128 bytes of RAM in the PIC18 are used for the access bank.
4. Give the address for the upper RAM used for SFR.
5. Show how to put value 99H into RAM location 202H in the PIC18F458.
6. Show how to put value 55H into RAM location 408H in a PIC18 with 4K of RAM.
7. True or false. The MOVFF instruction can copy a byte from any RAM location to any other RAM location in the PIC18.
8. The BSR register is a(n) ____-bit register, but only ____ bits of it are used for bank selection.

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In this section we look at some widely used subroutines such as checksum byte, BCD, and ASCII conversion. We will also examine the use of a stack in the PIC18.

Checksum byte in ROM

To ensure the integrity of ROM contents, every system must perform a checksum calculation. The checksum will detect any corruption of the contents of ROM. One cause of ROM corruption is current surge, either when the system is turned on, or during operation. To ensure data integrity in ROM, the checksum process uses what is called a checksum byte. The checksum byte is an extra byte that is tagged to the end of a series of bytes of data. To calculate the checksum byte of a series of bytes of data, the following steps can be taken:

1. Add the bytes together and drop the carries.
2. Take the 2's complement of the total sum, and that is the checksum byte, which becomes the last byte of the series.

To perform a checksum operation, add all the bytes, including the checksum byte. The result must be zero. If it is not zero, one or more bytes of data have been changed (corrupted). To clarify these important concepts, see Example 6-29.

Checksum program

The checksum generation and testing program is given in subroutine form. We have divided Program 6-1 into three subroutines (or subprograms). These three subroutines perform the following operations:

1. Retrieve the data from code ROM.
2. Calculate the checksum byte.
3. Test the checksum byte for any data error.

Each of these subroutines can be used in other applications. Example 6-29 shows how to manually calculate the checksum for a list of values. Also see Program 6-1.
Example 6-29

Assume that we have 4 bytes of hexadecimal data: 25H, 62H, 3FH, and 52H.

(a) Find the checksum byte.
(b) Perform the checksum operation to ensure data integrity.
(c) If the second byte, 62H, has been changed to 22H, show how the checksum method detects the error.

Solution:

(a) Find the checksum byte.

\[
\begin{align*}
25H \\
+ & \quad 62H \\
+ & \quad 3FH \\
+ & \quad 52H \\
\hline
118H & \quad \text{(Dropping the carry of 1, we have 18H. Its 2's complement is E8H. Therefore checksum byte is E8H.)}
\end{align*}
\]

(b) Perform the checksum operation to ensure data integrity.

\[
\begin{align*}
25H \\
+ & \quad 62H \\
+ & \quad 3FH \\
+ & \quad 52H \\
+ & \quad \_E8H \\
\hline
200H & \quad \text{(Dropping the carries, we see 00, indicating that data is not corrupted.)}
\end{align*}
\]

(c) If the second byte 62H has been changed to 22H, show how the checksum method detects the error.

\[
\begin{align*}
25H \\
+ & \quad 22H \\
+ & \quad 3FH \\
+ & \quad 52H \\
+ & \quad \_E8H \\
\hline
1C0H & \quad \text{(Dropping the carry, we get C0H, which is not 00. This means that data is corrupted.)}
\end{align*}
\]

;PROG 6-1: CALCULATING AND TESTING CHECKSUM BYTE
#include P18F458.inc

RAM_ADDR EQU 40H ;RAM space to place the bytes
COUNTRG EQU 0x20 ;fileReg loc for counter
CNTVAL EQU 4 ;counter value = 4 for adding 4 bytes
CNTVAL1 EQU 5 ;counter value = 5 for adding 5 bytes
;including checksum byte

;----------main program
ORG 0
CALL COPY_DATA
CALL CAL_CHKSUM
CALL TEST_CHKSUM
BRA $
;-----copying data from code ROM address 500H to data RAM loc
COPY_DATA
  MOVLW low(MYBYTE) ;WREG = 00 LOW-byte addr
  MOVF TBLPTRL
  MOVLW hi(MYBYTE) ;WREG = 5, HIGH-byte addr
  MOVF TBLPTRH
  MOVLW upper(MYBYTE) ;WREG = 00 upper-byte addr
  MOVF TBLPTRTRU
  LFSR 0, RAM_ADDR ;FSR0 = RAM_ADDR, place to save
  TBLRD++
  MOVF TABLAT,W ;copy to WREG (Z = 1, if null)
  BZ EXIT ;is it null char? exit if yes
  MOVWF POSTINC0 ;copy WREG to RAM and inc pointer
  BRA C1
EXIT RETURN

;-----calculating checksum byte
CAL_CHKSUM
  MOVLW CNTVAL ;WREG = 4
  MOVF COUNTREG ;load the counter, count = 4
  LFSR 0, RAM_ADDR ;load pointer. FSR0 = 40H
  CLRF WREG
  C2
  ADDWF POSTINC0,W ;add RAM to WREG and increment FSR0
  DECF COUNTREG,F ; decrement counter
  BNZ C2 ; loop until counter = zero
  XORLW 0xFF ;1's comp
  ADDLW 1 ;2'compl
  MOVWF POSTINC0
  RETURN

;----------testing checksum byte
TEST_CHKSUM
  MOVLW CNTVAL1 ;WREG = 5
  MOVF COUNTREG ;load the counter, count = 5
  CLRF TRISB ;PORTB = output
  LFSR 0, RAM_ADDR ;load pointer. FSR0 = 40H
  CLRF WREG
  C3
  ADDWF POSTINC0,W ;add RAM and increment FSR0
  DECF COUNTREG,F ; decrement counter
  BNZ C3 ; loop until counter = zero
  XORLW 0x0 ;EX-OR to see if WREG = zero
  BZ G_1 ;is result zero? then good
  MOVLW 'B'
  MOVWF PORTB ;if not, data is bad
  RETURN
  G_1
  MOVLW 'G'
  MOVWF PORTB ;data is not corrupted
  RETURN

;----------my data in program ROM
ORG 0x500
MYBYTE DB 0x25, 0x62, 0x3F, 0x52, 0x00
END

Note the usage of the keywords low, hi, and upper, to indicate the 21-bit address of the program ROM.
BCD to ASCII conversion program

Many RTCs (real-time clocks) provide time and date in BCD format. To display the BCD data on an LCD or a PC screen, we need to convert it to ASCII. Program 6-2 (a) transfers packed BCD data from program ROM to data RAM, (b) converts packed BCD to ASCII, and (c) sends the ASCII to port B for display. We will use a portion of this program in Chapter 16. The displaying of data on LCD will be shown in Chapter 12. See Chapter 5 for the BCD to ASCII conversion algorithm.

;PROG 6-2: CONVERTING PACKED BCD TO ASCII
#include P18F458.inc

RAM_ADDR EQU 0x40
ASC_RAM EQU 0x50
COUNTRREG EQU 0x20 ;fileReg loc for counter
CNTVAL EQU D'4' ;counter value of BCD bytes
CNTVAL1 EQU D'8' ;counter value of ASCII bytes

;------------main program
ORG 0
CALL COPY_DATA
CALL BCD_ASC_CONV
CALL DISPLAY
BRA $ ;-------copying data from code ROM to data RAM

COPY_DATA
MOVLW low(MYBYTE) ;WREG = 00 LOW-byte addr
MOVWF TBLPTRL ;ROM data LOW-byte addr
MOVLW hi(MYBYTE) ;WREG = 5, HIGH-byte addr
MOVWF TBLPTRH ;ROM data HIGH-byte addr
MOVLW upper(MYBYTE) ;WREG = 00 upper-byte addr
MOVWF TBLPRTRU ;ROM data upper-byte addr
LFSR 0,RAM_ADDR ;FSR0 = RAM_ADDR, place to save
C1 TBLRD'+' ;bring in next byte and inc TBLPTR
MOVF TABLAT,W ;copy to WREG (Z = 1, if null)
BZ EXIT ;is it null char? exit if yes
MOVWF POSTINCO ;copy WREG to RAM and inc pointer
BRA C1
EXIT RETURN

;-------convert packed BCD to ASCII

BCD_ASC_CONV
MOVLW CNTVAL ;get the counter value
MOVWF COUNTRREG ;load the counter
LFSR 0,RAM_ADDR ;FSR0 = RAM_ADDR BCD byte pointer
LFSR 1,ASC_RAM ;FSR1 = ASC_RAM ASCII byte pointer
B2 MOVF INDF0,W ;copy BCD to WREG
ANDLW 0x0F ;mask the upper nibble (W = 09)
IORLW 0x30 ;make it an ASCII
MOVWF POSTINCI ;copy to RAM and increment FSR1
MOVF POSTINCO,W ;note the use of instruction
ANDLW 0xF0 ;mask the lower nibble (W = 20H)
SWAPF WREG
IORLW 0x30 ;make it an ASCII
MOVWF POSTINCI ;copy to RAM and increment FSR1
DECF COUNTRREG,F ;decrement counter
BNZ B2 ;loop until counter = zero

230
RETURN

;-----send ASCII data to port B
DISPLAY
    CLRF TRISB
    MOVLW CNTVAL1
    MOVWF COUNTREG
    LFSR 2,ASC_RAM
    MOVF POSTINC2,W
    MOVWF PORTB
    DECF COUNTREG,F
    CNTVAL1
    BNZ COUNTREG,F
    RETURN

;--------my BCD data in program ROM
ORG 0x500
MYBYTE DB 0x25, 0x67, 0x39, 0x52, 0x00
END

Figure 6-9. Results of Program 6-2 After it Ran.

**Binary (hex) to ASCII conversion program**

Many ADC (analog-to-digital converter) chips provide output data in binary (hex). To display the data on an LCD or PC screen, we need to convert it to ASCII. The code for the binary-to-ASCII conversion is shown in Program 6-3. Notice that the subroutine gets a byte of 8-bit binary (hex) data from Port B and converts it to decimal digits, and the second subroutine converts the decimal digits to ASCII and saves them. We are saving the low digit in the lower address location and the high digit in higher address location. This is referred to as the little-endian convention (i.e., low-byte to low-location and high-byte to high-location). All PIC18 products use the little-endian convention. For the binary-to-ASCII conversion algorithm see Chapter 5.

;PROG 6-3: CONVERTING BIN(HEX) TO ASCII
#include P18F458.INC

NUME EQU 0x00 ;RAM loc for NUME
QU  EQU 0x20 ;RAM loc for quotient
RMND_L EQU 0x30 ;the least significant digit loc
RMND_M EQU 0x31 ;the middle significant digit loc
RMND_H EQU 0x32 ;the most significant digit loc
MYDEN EQU D'10' ; value for divide by 10
COUNTREG EQU 0x10 ; fileReg loc for counter
CNTVAL EQU d'3' ; counter value
UNPBCD_ADDR EQU 0x30
ASCII_RESULT EQU 0x40

;--------main program
ORG 0
SETF TRISB ; make PORTB input
CALL BIN_DEC_CON
CALL DEC_ASCII_CON
BRA $ ;

;------converting BIN(HEX) TO DEC (00-FF TO 000-255)
BIN_DEC_CON
MOVFF PORTB,WREG ; get the binary data from PORTB
MOVWF NUME ; load numerator
MOVLW MYDEN ; WREG = 10, the denominator
CLRF QU ; clear quotient
D_1 INCF QU ; inc quotient for every subtraction
SUBWF NUME ; subtract WREG from NUME value
BC D_1 ; if positive go back
ADDWF NUME ; once too many, first digit
DECF QU ; once too many for quotient
MOVFF NUME,RMND_L ; save the first digit
MOVFF QU,NUME ; repeat the process one more time
CLRF QU ; clear QU
D_2 INCF QU ;
SUBWF NUME ; subtract WREG from NUME value
BC D_2 ; once too many
DECF QU ;
MOVFF NUME,RMND_M ; 2nd digit
MOVFF QU,RMND_H ; 3rd digit
RETURN ;

;------converting unpacked BCD digits to displayable ASCII digits
DEC_ASCII_CON
MOVLW CNTVAL ; WREG = 10
MOVWF COUNTREG ; load the counter, count = 10
LFSR 0,UNPBCD_ADDR ; load pointer FSR0
LFSR 1,ASCII_RESULT ; load pointer FSR1
B3 MOVF POSTINC0, W ; copy RAM to WREG, increment FSR0
ADDLW 0x30 ; make it an ASCII
MOVWF POSTINC1 ; copy WREG and increment FSR1
DECF COUNTREG,F ; decrement counter
BNZ B3 ; loop until counter = zero
RETURN
END ; end of the program
Using memory banks for a stack

The stack in the PIC18 is 31 bytes deep and 21 bits wide. Because the program counter is 21 bits wide, the stack must be 21 bits wide also, as we discussed in Chapter 2. This stack is used mainly to save addresses for call and interrupt subroutines. Unlike other microprocessors, the stack in the PIC18 is not part of the data RAM space. Having access to such a large number of banks, however, makes a traditional stack unnecessary. In traditional CPUs (e.g., the x86) a limited number of registers forced us to push main registers into the stack at the beginning of a called subroutine, before we could use the main registers for data manipulation. In the case of the PIC18, all we have to do is to change the default bank to a new bank when we go into the subroutine. In other words, if we want to store a program's data on a stack in a given subroutine, we can use one of the banks of data RAM space instead of the 31 × 21-bit stack. Besides the fact that the 21-bit wide stack does not lend itself to storing 8-bit data, we must reserve all the 31 × 21-bit stack for calls and interrupts. Notice that we can still use the access bank for the storage of global variables. Global variables are discussed in the next section.

Review Questions

1. For the following ASCII numbers, give the ASCII and packed BCD representations.
   (a) '5', '7'  (b) '9', '4'
2. Show the hex format for “2005” and its BCD version.
3. Does the WREG register have BCD data after the following instruction is executed? Assume that the default the radix for the MPASM is decimal.
   MOVlw 95
4. 33H in BCD when converted to ASCII is ___H and ___H.
5. Find the ASCII value for the binary 11110010 if we want to display it on a computer screen as a 3-digit decimal number.
6. The checksum byte method is used to test data integrity in __(RAM, ROM).
7. Find the checksum byte for the following hex values: 88H, 99H, AAH, BBH, CCH, DDH
8. True or false. If we add all the bytes, including the checksum byte, and the result is FFH, there is no error in the data.
SECTION 6.7: MACROS AND MODULES

In this section we explore macros and modules and their use in Assembly language programming. The format and usage of macros are defined, and many examples of their applications are explored. In addition, this section demonstrates modular programming along with rules for writing modules and linking them together. Some very useful modules will be given, along with methods of passing parameters among various modules. Dividing a program into several modules (in C programming these are called functions) allows us to use modules in other applications. It is common practice to divide a program into several modules, test each module, and put them into a library.

What is macro and how is it issued?

There are applications in Assembly language programming where a group of instructions performs a task that is used repeatedly. For example, moving data into a RAM location is done repeatedly in the same program. It does not make sense to rewrite this code every time it is needed. Therefore, to reduce the time that it takes to write code and reduce the possibility of errors, the concept of macros was born. Macros allow the programmer to write the task (code to perform a specific job) once only, and to invoke it whenever it is needed.

MACRO definition

Every macro definition must have three parts, as follows:

```
name      MACRO    dummy1, dummy2, ..., dummyN

......

......

ENDM
```

The MACRO directive indicates the beginning of the macro definition and the ENDM directive signals the end. What goes in between the MACRO and ENDM directives is called the body of the macro. The name must be unique and must follow Assembly language naming conventions. The dummies are names, or parameters, or even registers that are mentioned in the body of the macro. After the macro has been written, it can be invoked (or called) by its name, and appropriate values are substituted for dummy parameters. Moving literal data into file register data RAM is a widely used service, but there is no instruction for that. We can use a macro to do the job as shown in the following code:

```
MOVLF MACRO K, MYREG
MOVLF K
MOVWF MYREG
ENDM
```

The above is the macro definition. Note that dummy arguments of K and MYREG are mentioned in the body of macro.

The following are three examples of how to use the above macro:
1. MOVLF 0x55, 0x20 ;send value 55H to loc 20H

2. VAL_1 EQU 0x55
   RAM_LOC EQU 0x20
   MOVLF VAL_1, RAM_LOC

3. MOVLF 0x55, PORTB ;send value 55H to Port B

The instruction “MOVLF 0x5, 0x20” invokes the macro. The assembler expands the macro by providing the following code in the .lst file:

```
M  MOVLW 5
M  MOVWF 0x20
```

The M indicates that the code is from the macro.

**LOCAL directive**

In the discussion of macros so far, the examples chosen do not have a label or name in the body of the macro. This is because if a macro is expanded more than once in a program and there are labels in the label field of the body of the macro, these labels must be declared as LOCAL. Otherwise, an assembler error would be generated when the same label was encountered in two or more places. The following rules must be observed in the body of the macro:

1. All labels in the label field must be declared LOCAL.
2. The LOCAL directive must be right after the MACRO directive. In other words, it must be placed even before comments and the body of the macro; otherwise, the assembler gives an error.
3. The LOCAL directive can be used to declare all names and labels at once as follows:
   
   LOCAL name1, name2, name3
   
   or one at a time as:
   
   LOCAL name1
   LOCAL name2
   LOCAL name3

To clarify these points, look at the following macro for time delay:

```
DELAY_1 MACRO V1, TREG
   LOCAL BACK
   MOVLW V1
   MOVWF TREG
   BACK NOP
   NOP
   NOP
   NOP
   DECF TREG, F
   BNZ BACK
ENDM
```
Notice that the "BACK" label is defined as LOCAL right after the MACRO directive. Defining this label anywhere else causes an error. The use of a LOCAL directive allows the assembler to define the labels separately each time it encounters them. Examining the list file shows that when the macro is expanded for the first time, the list file has "??0000", for the second time it has "??0001", and for the third time it has "??0002" in place of the "BACK" label, indicating that the "BACK" label is local. To clarify this concept, see Program 6-4 without the LOCAL directive to see how the assembler will give an error. The following code is another macro for a time delay with a nested loop:

```
DELAY_2 MACRO V1, V2, R1, R2
    LOCAL BACK
    LOCAL AGAIN
    MOVLW V2
    MOVWF R2
    AGAIN MOVLW V1
    MOVWF R1
    YEARS NOP
     NOP
     NOP
     NOP
    DECF R1,F
    BNZ BACK
    DECF R2,F
    BNZ AGAIN
ENDM
```

Now examine Program 6-4 to see how to use a macro in a program.

```
;-----------------------------------------------
;Program 6-4: toggling Port B using macros
#include P18F458.INC

;----------------------sending data to fileReg macro
MOVLF MACRO K, MYREG
    MOVLW K
    MOVWF MYREG
ENDM

;----------------------time delay macro
DELAY_1 MACRO V1, TREG
    LOCAL BACK
    MOVLW V1
    MOVWF TREG
    YEARS NOP
     NOP
     NOP
     NOP
    DECF TREG,F
    BNZ BACK
ENDM

;----------------------program starts
ORG 0
CLRF TRISB ; Port B as an output
```
OVER MOVLF 0x55, PORTB
DELAY_1 0x200, 0x10
MOVLF 0xAA, PORTB
DELAY_1 0x200, 0x10
BRA OVER
END

;-------------------------end of file

INCLUDE directive

Assume that several macros are used in every program. Must they be rewritten every time? The answer is no, if the concept of the INCLUDE directive is known. The INCLUDE directive allows a programmer to write macros and save them in a file, and later bring them into any program file. For example, assume that the following widely used macros were written and then saved under the filename "MYMACRO1.MAC".

Assuming that these macros are saved on a disk under the filename "MYMACRO1.MAC", the INCLUDE directive can be used to bring this file into any "asm" file and then the program can call upon any of the macros as many times as needed. When a file includes all macros, the macros are listed at the beginning of the "lst" file and, as they are expanded, will be part of the program. To understand this, see Program 6-5.

;Program 6-5: toggling Port B using macros
#include P18F458.INC
#include "MYMACRO1.MAC" ;get macros from macro file

;--------------------------program starts
ORG 0
CLRF TRISB ;Port B as an output
OVER MOVLF 0x55, PORTB
DELAY_1 0x200, 0x10
MOVLF 0xAA, PORTB
DELAY_1 0x200, 0x10
BRA OVER
END

;--------------------------end of file

NOEXPAND/EXPAND directive

When viewing the .lst file with macros, we see them fully displayed. The expand directive is set by default, and it shows the macro at every location it is called. This is fine for two or three iterations, but when there are more, it can become cumbersome. Using the noexpand directive, we can turn off the display of macros in the list file.

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; Program 6-4: toggling Port B using macros
#include P18F458.INC

NOEXPAND

;----------sending data to fileReg macro
MOVLF MACRO K, MYREG
MOVWF K
MOVWF MYREG
ENDM

;----------time delay macro
DELAY_1 MACRO V1, TREG
LOCAL BACK
MOVLW V1
MOVWF TREG
BACK NOP
NOP
NOP
DECF TREG, F
BNZ BACK
ENDM

;----------program starts
ORG 0
CLRF TRISB ; Port B as an output
OVER MOVLF 0x55, PORTB
DELAY_1 0x200, 0x10
MOVLF 0xAA, PORTB
DELAY_1 0x200, 0x10
BRA OVER
END

Figure 6-11. List File with NOEXPAND Option for Program 6-4
00001 ;Program 6-4: toggling Port B using macros
00002 #include P18F458.INC
00003 EXPAND
00004 ;-----------------------sending data to fileReg macro
00005 MOVLF MACRO K, MYREG
00006 MOVFW K
00007 MOVWF MYREG
00008 ENDM
00009
00010 ;-----------------------------time delay macro
00011 DELAY_1 MACRO V1, TREG
00012 LOCAL BACK
00013 MOVFLV V1
00014 MOVWF TREG
00015 BACK NOP
00016 NOP
00017 NOP
00018 NOP
00019 DECF TREG,F
00020 BNZ BACK
00021 ENDM
00022
00023 ;--------------------------------program starts
00024 ORG 0
00025 CLRF TRISB ;Port B as an output
00026 MOVLF 0x55, PORTB
00027 DELAY_1 0x200, 0x10
00028 MOVWF PORTB
00029 DELAY_1 0x200, 0x10
00030 MOVWF PORTB
00031 ENDM

Figure 6-12. List File with EXPAND Option for Program 6-4

CHAPTER 6: BANK SWITCHING, TABLES, MACROS, AND MODULES
Macros vs. subroutines

Macros and subroutines are useful in writing assembly programs, but each have limitations. Macros increase code size everytime they are invoked. For example, if you call a 10-instruction macro 10 times, the code size is increased by 100 instructions. Whereas, if you call the same subroutine 10 times, the code size is only that of the subroutine instructions. The only problem with subroutines is that they use stack space when called, and this can cause problems when there are nested calls (a subroutine calling another subroutine). The nested call can lead to a stack overflow and cause the program to crash. The PIC18 has provisions for stack overflow, discussed in the PIC18 Reference Manual.

Modules

It is common practice in writing software packages to break down the project into small modules and distribute the task of writing those modules among several programmers. This not only makes the project more manageable but also has other advantages, such as:

1. Each module can be written, debugged, and tested individually.
2. The failure of one module does not stop the entire project.
3. The task of locating and isolating any problem is easier and less time consuming.
4. One can use the modules to link with high-level languages such as C.
5. Parallel development shortens considerably the time required to complete a project.

Next we explain how to write and link modules to create a single executable program.

Writing modules

In programs given in the last section, a main procedure was written that called many other subroutines. In those examples, if one subroutine did not work properly, the entire program would have to be rewritten and reassembled. A more efficient way to develop software is to treat each subroutine as a separate program (or module) with a separate filename. Then each one can be assembled and tested. After testing each program and making sure that each works, they can all be brought together (linked) to make a single program. To enable these modules to be linked together, certain Assembly language directives must be used. Among these directives, the two most widely used are EXTERN (external) and GLOBAL. The GLOBAL directive is the same as PUBLIC in other Assembly language programs. Each is discussed below.
EXTERN directive

The EXTERN directive is used to notify the assembler and linker that certain names and variables that are not defined in the present module are defined externally somewhere else. In the absence of the EXTERN directive, the assembler would show an error because it cannot find where the names are defined. The EXTERN directive has the following format:

EXTERN name1 ; each name can be in a separate EXTERN
EXTERN name2

EXTERN name1, name2 ; or many can be listed in the same EXTERN

GLOBAL directive

Names or parameters defined as EXTERN (indicating that they are defined outside the present module) must be defined as GLOBAL in the module where they are defined. Defining a name as GLOBAL (PUBLIC) allows the assembler and linker to match it with its EXTERN counterpart(s). The following is the format for the GLOBAL directive:

GLOBAL name1 ; each name can be in a separate directive
GLOBAL name2

GLOBAL name1, name2 ; or many can be listed in the same GLOBAL

Program 6-6 should help to clarify these concepts. It demonstrates that for every EXTERN definition there is a GLOBAL directive defined in another module. Notice the entry and exit points of the program. Modules that are called by the main module have their own END directives. See Program 6-6.

;---------------------------------------------------------------------
; PROG 6-6: MAIN.ASM - CALCULATING AND TESTING CHECKSUM BYTE
# include P18F458.INC

RAM_ADDR EQU 40H
COUNTREG EQU 0x20 ; fileReg loc for counter
CNTVAL EQU 4 ; counter value
CNTVAL1 EQU 5 ; counter value

EXTERN CAL_CHKSUM
EXTERN TEST_CHKSUM

PGM CODE
;-----------main program
ORG 0
CALL COPY_DATA ; this subroutine is in this file
CALL CAL_CHKSUM ; this sub is in external file
CALL TEST_CHKSUM ; this sub is in external file
BRA $
;-------copying data from code ROM to data RAM
COPY_DATA
    MOVLW low(MYBYTE) ; WREG = 00 LOW-byte addr.
    MOVWF TBLPTRL ; ROM data LOW-byte addr.
    MOVLW hi(MYBYTE) ; WREG = 5, HIGH-byte addr.
    MOVWF TBLPTRH ; ROM data HIGH-byte addr.
    MOVLW upper(MYBYTE) ; WREG = 00 upper-byte addr.
    MOVWF TBLPTRRU ; ROM data upper-byte addr.
    LFSR 0,RAM_ADDR ; FSR0 = RAM_ADDR, place to save
    MOVF TABLAT,W ; copy to WREG (Z = 1, if null)
    BZ EXIT ; is it null char? exit if yes
    MOVF POSTINC0 ; copy WREG to RAM and inc pointer
    BRA C1
EXIT RETURN

;-------my data in program ROM
ORG 0x500
MYBYTE DB 0x25, 0x62, 0x3F, 0x52, 0x00
END

;PROG 6-6: CALCCSB.ASM - CALCULATING CHECKSUM BYTE
#include P18F458.inc

RAM_ADDR EQU 40H
COUNTRG EQU 0x20 ; fileReg loc for counter
CNTVAL EQU 4 ; counter value
CNTVAL1 EQU 5 ; counter value

GLOBAL CAL_CHKSUM

PGM_CODE ; we use this to inform the linker that
; the code segment has the name PGM
CAL_CHKSUM
    MOVLW CNTVAL ; WREG = 4
    MOVWF COUNTRG ; load the counter
    LFSR 0,RAM_ADDR ; load pointer. FSR0 = 40H
    CLR WREG
    C2 ADDWF POSTINC0,W ; add RAM to WREG and increment FSR0
    DECF COUNTRG,F ; decrement counter
    BNZ C2 ; loop until counter = zero
    XORLW 0xFF ; 1's comp
    ADDLW 1 ; 2'compl
    MOVF POSTINC0
    RETURN
END

;PROG 6-6: TESTCSB.ASM - TESTING CHECKSUM BYTE
#include P18F458.inc

RAM_ADDR EQU 40H
COUNTRG EQU 0x20 ; fileReg loc for counter
CNTVAL EQU 4 ; counter value
CNTVAL1 EQU 5 ; counter value

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GLOBAL TEST_CHKSUM
PGM CODE
TEST_CHKSUM
MOVLW CNTVAL1 ;WREG = 5
MOVWF COUNTREG ;load the counter
CLRF TRISB
LFSR 0,0x40 ;load pointer. FSR0 = 40H
CLRF WREG
C3 ADDWF POSTINC0,W ;add RAM and increment FSR0
DECF COUNTREG,F ;decrement counter
BNZ C3 ;loop until counter = zero
XORLW 0x0 ;EX-OR to see if zero
BZ G_1 ;is result zero? then good
MOVlw 'B'
MOVWF PORTB ;if not, data is bad
RETURN
G_1 MOVlw 'G'
MOVWF PORTB ;data is not corrupted
RETURN
END

Linking modules together

Assuming that each program module in Program 6-6 is assembled separately and saved under the filenames MAIN.O, CALCCSB.O, and TESTCSB.O, the following shows how to link them together with MPLINK in order to generate a single executable file:

> MPLink.exe" "18f458.1kr" "MAIN.O" "CALCCSB.O" "TESTCSB.O" /o"PRG6-6.COF"

Program 6-6 shows how the EXTERN and GLOBAL directives can also be applied to data variables. The linker program resolves external references by matching GLOBAL and EXTERN names. The linker program will search through the files specified in the MPLINK command for the external subroutines.

The MPLAB IDE handles the compiling and linking in one step. This aids in program development by reducing time and errors in typing the command line call.

Review Questions

1. Discuss the benefits of macro programming.
2. List the three parts of a macro.
3. Explain and contrast the macro definition, invoking the macro, and expanding the macro.
4. True or false. A label defined within a macro is automatically understood by the assembler to be local.
5. The______directive is used within a module to indicate that the named variable or subroutine can be used by another module.
6. The______directive is used within a module to indicate that the named variable or subroutine was defined in another module.
SUMMARY

This chapter described the addressing modes of the PIC18. Immediate addressing mode uses a constant for the operand. Direct or register indirect addressing modes can be used to access data stored in data RAM file registers of the PIC18. Register indirect addressing mode uses a register as a pointer to the data. The advantage of this is that it makes addressing dynamic rather than static. Indexed ROM addressing mode is widely used in accessing data elements of look-up table entries located in the program ROM space of the PIC18. The PIC18 allows the reading of fixed data stored in program ROM space, in addition to writing to ROM if the PIC18 is of the flash type.

A group of registers called SFRs (special function registers) can be accessed by their names or their addresses. We also discussed the bit-addressable data RAM locations and ports and showed how to use single-bit instructions to access them directly. The topic of bank switching was discussed, and we showed how to use BSR registers to access all 16 banks of RAM in the PIC18.

We discussed how to break up a program into several or many subroutines and write and test each one separately. Macros and modules were also explored and their advantages were discussed.

PROBLEMS

SECTION 6.1: IMMEDIATE AND DIRECT ADDRESSING MODES

1. Which of the following are invalid uses of immediate addressing mode?
   (a) MOVWLW 0x24  (b) MOVWLW MYREG, 0x30  (c) MOVWLW 0x60

2. Identify the addressing mode for each of the following:
   (a) MOVWF PORTB  (b) MOVWLW 0x50  (c) MOVWF MYREG
   (d) MOVWLW 0  (e) MOVFF MYBREG, YOURREG
   (f) MOVWF YOURREG

3. Indicate the address assigned to each of the following:
   (a) PORTB  (b) WREG  (c) PORTC
   (d) PORTD  (e) PCL  (f) PCH
   (g) PCU  (h) TRISC  (i) TRISB
   (j) STATUS  (k) FSR0L

4. Which bank is used for SFRs?

5. In accessing the SFRs, we must use _____ addressing mode.

6. What does the following instruction do? “MOVWLW 0xF0”

7. What does the following instruction do? “MOVWF PORTC”

8. What does the following instruction do? “MOVF PORTC, W”

9. “CLRF MYREG” is a(n) _______ (valid, invalid) instruction.

10. The byte addresses assigned to the 128 bytes of lower data RAM are ______ to ______.

11. The byte addresses assigned to the SFRs are ______ to ______.

12. Indicate the byte addresses assigned to both of the following. Is there a gap
between them?
(a) RAM locations of access bank  (b) SFR of access bank
13. Write a program to add the following data and place the result in RAM location 20H: The data values are 6, 9, 2, 5, 7

SECTION 6.2: REGISTER INDIRECT ADDRESSING MODE

14. Which registers are allowed to be used as a pointer for register indirect addressing mode when accessing data RAM? Give their names and show how they are loaded.
15. Write a program to copy FFH into RAM locations 50H to 6FH.
16. Write a program to copy 10 bytes of data starting at RAM address 40H to RAM locations starting at 70H.
17. What is the size of the FSRx register?
18. Give the SFR registers related to the FSR0 and FSR1.
19. Write a program to clear RAM locations 0 to 7FH.
20. Write a program to toggle RAM locations 50H to 5FH.
21. Explain the role of the INDFx register.
22. How much RAM space does the FSRx register cover?

SECTION 6.3: LOOK-UP TABLE AND TABLE PROCESSING

23. Compile and state the contents of each ROM location for the following data:

```
ORG 200H
MYDAT_1: DB "Earth"
MYDAT_2: DB "987-65"
MYDAT_3: DB "GAEBH 98"
```

24. Compile and state the contents of each ROM location for the following data:

```
ORG 340H
DAT_1: DB 0x22, 0x56, B'10011001', D'32', 0xF6, B'11110111'
```

25. Which register is allowed to be used as a pointer for register indirect addressing mode when accessing data stored in program ROM? Give the name and show how it is loaded.
26. Explain the role of the TABLAT register.
27. What is the size of the TBLPTR register? How much ROM space does it cover?
28. Give the SFR registers related to the TBLPTR.
29. Write a program to read the following message from ROM and place it in data RAM starting at 50:

```
ORG 0x600
MYDATA DB "1-800-999-9999", 0
```

30. Write a program to find y where \( y = x^2 + 2x + 5 \), and \( x \) is between 0 and 9.
31. Write a program to find \( y \) where \( y = 20x + 5 \), and \( x \) is between 0 and 9.
32. Write a program to read the following message from ROM and place it in data RAM starting at 40:

```
ORG 0x700
MYDATA DB "The earth is but one country", 0
```

33. True or false. The table read instruction works for all PIC18 family members.
34. True or false. The table write instruction works for PIC18 family members.

CHAPTER 6: BANK SWITCHING, TABLES, MACROS, AND MODULES 245
with flash ROM.

35. Assume that the lower four bits of PORTB are connected to four switches. Write a program to send the following ASCII characters to a PORTC, based on the status of the switches:

<table>
<thead>
<tr>
<th>Decimal</th>
<th>Binary</th>
<th>Character</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0000</td>
<td>'0'</td>
</tr>
<tr>
<td>1</td>
<td>0001</td>
<td>'1'</td>
</tr>
<tr>
<td>2</td>
<td>0010</td>
<td>'2'</td>
</tr>
<tr>
<td>3</td>
<td>0011</td>
<td>'3'</td>
</tr>
<tr>
<td>4</td>
<td>0100</td>
<td>'4'</td>
</tr>
<tr>
<td>5</td>
<td>0101</td>
<td>'5'</td>
</tr>
<tr>
<td>6</td>
<td>0110</td>
<td>'6'</td>
</tr>
<tr>
<td>7</td>
<td>0111</td>
<td>'7'</td>
</tr>
<tr>
<td>8</td>
<td>1000</td>
<td>'8'</td>
</tr>
<tr>
<td>9</td>
<td>1001</td>
<td>'9'</td>
</tr>
<tr>
<td>10</td>
<td>1010</td>
<td>'A'</td>
</tr>
<tr>
<td>11</td>
<td>1011</td>
<td>'B'</td>
</tr>
<tr>
<td>12</td>
<td>1100</td>
<td>'C'</td>
</tr>
<tr>
<td>13</td>
<td>1101</td>
<td>'D'</td>
</tr>
<tr>
<td>14</td>
<td>1110</td>
<td>'E'</td>
</tr>
<tr>
<td>15</td>
<td>1111</td>
<td>'F'</td>
</tr>
</tbody>
</table>

SECTION 6.4: BIT-ADDRESSABILITY OF DATA RAM

36. Write a program to generate a square wave with 75% duty cycle on bit RB5.
37. Write a program to generate a square wave with 80% duty cycle on bit RC7.
38. Write a program to monitor RB4. When it goes HIGH, the program will generate a sound (square wave of 50% duty cycle) on pin RB7.
39. Write a program to monitor RC1. When it goes LOW, the program will send the value 55H to RD0.
40. What register does the carry flag belong to?
41. What bit address is assigned to the Z flag?
42. Which of the following instructions are valid? If valid, indicate which bit is altered.
   (a) BSF PORTB, 1
   (b) BSF PORTC, 3
   (c) BCF WREG, 1
   (d) BCF 0x30, 1
   (e) BCF PORTD, 0
   (f) BST STATUS, C
   (g) CLR WREG, 3
   (h) CLR FSR0
43. “BTG PORTB, 0” is a(n) ______ (valid, invalid) instruction.
44. Which of the I/O ports of PORTB, PORTC, and PORTD are bit-addressable?
45. Which of the registers of the PIC18 are bit-addressable?
46. Give an instruction to clear the carry flag.
47. Show how would you check whether the C flag is HIGH.
48. Show how would you check whether the Z flag is HIGH.
49. Give the bit locations in the status register assigned to the flag bits C, Z, DC, and OV.
50. True or false. The bit addresses 0–7 are assigned to each RAM location of 000–FFFH.
51. True or false. SFR registers are not bit-addressable.
52. Write instructions to save the C flag bit in bit 4 of location 10.
53. Write instructions to save the DC flag bit in bit 2 of location 16H.
54. Write instructions to save the Z flag bit in bit 7 of location 12H.
55. Write instructions to see whether the D0 and D1 bits of register WREG are HIGH. If so, divide register WREG by 4.
56. Write a program to see whether the D7 bit of register WREG is HIGH. If so, send a message to the LCD stating that WREG has a negative number.
57. Write a program to set HIGH all the bits of RAM location 20H using the following methods:
   (a) byte addresses  (b) bit addresses
58. Write a program to see whether the WREG register is divisible by 8.
59. Write a program to find the number of zeros in file register location 05.

SECTION 6.5: BANK SWITCHING IN THE PIC18

60. What addressing mode is used to access the SFRs?
61. What addressing mode is used to access the last 128 bytes of RAM in the PIC18?
62. Give the address range of the lower and the upper 128 bytes of the access bank.
63. In the PIC18, the SFRs use the same addresses across all members and they are from _________ to _______.
64. The PIC18 can have a maximum of _____ banks.
65. Explain the difference between these two instructions.
   (a) ADDWF MYREG, F, 1      (b) ADDWF MYREG, F, 0
66. Which addressing modes are used to access various banks?
67. Write a program to put 55H into RAM locations 1C0–1CFH.
68. Write a program to copy the contents of RAM locations 20–2FH to RAM locations 2D0–2DFH.
69. Explain the difference between these two instructions.
   (a) CLRF MYREG,F,1      (b) CLRF MYREG,F,0
70. Explain the difference between these two instructions.
   (a) SETF MYREG,W,1      (b) SETF MYREG,W,0
71. Explain the difference between these two instructions.
   (a) INCF MYREG,F,1      (b) INCF MYREG,F,0

SECTION 6.6: CHECKSUM AND ASCII SUBROUTINES

72. Find the checksum byte for the following ASCII message: "Hello"
73. True or false. If we add all bytes, including the checksum byte, and the result is 00H, there is no error in the data.
74. Write a program to (a) get the data "Hello, my fellow world citizens" from program ROM, (b) calculate the checksum byte, and (c) test the checksum byte for any data error.
75. To display data on LCD or PC monitors, it must be in ________ (BINARY, BCD, ASCII).
76. Assume that the lower 4 bits of PB are connected to four switches. Write a program to send the following ASCII characters to PD based on the status of the switches:

<table>
<thead>
<tr>
<th>PB</th>
<th>ASCII</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>'0'</td>
</tr>
<tr>
<td>0001</td>
<td>'1'</td>
</tr>
<tr>
<td>0010</td>
<td>'2'</td>
</tr>
<tr>
<td>0011</td>
<td>'3'</td>
</tr>
<tr>
<td>0100</td>
<td>'4'</td>
</tr>
<tr>
<td>0101</td>
<td>'5'</td>
</tr>
<tr>
<td>0110</td>
<td>'6'</td>
</tr>
<tr>
<td>0111</td>
<td>'7'</td>
</tr>
<tr>
<td>1000</td>
<td>'8'</td>
</tr>
<tr>
<td>1001</td>
<td>'9'</td>
</tr>
</tbody>
</table>

77. Write a program to convert a series of packed BCD numbers to ASCII. Assume that the packed BCD is located in ROM locations starting at 700H. Place the ASCII codes in RAM locations starting at 40H.

```assembly
ORG 700H
MYDATA DB 76H, 87H, 98H, 43H
```

78. Write a program to convert a series of ASCII numbers to packed BCD. Assume that the ASCII data is located in ROM locations starting at 300H. Place the BCD data in RAM locations starting at 60H.

```assembly
ORG 300H
MYDATA DB "87675649"
```

79. Write a program to get an 8-bit binary number from PD, convert it to ASCII, and save the result in RAM locations 40H, 41H, and 42H. What is the result if PD has 1000 1101 binary as input?

SECTION 6.7: MACROS AND MODULES

80. Give two advantages of macros.

81. Which uses more program ROM space: a macro or a module?

82. Give three reasons to write programs with modules.

83. If a label or parameter is not defined in a given module, it must be declared as

84. If a label or parameter is used by other modules, it must be declared as

in the present module.

85. Repeat Problem 79 using macros and modules.
ANSWERS TO REVIEW QUESTIONS

SECTION 6.1: IMMEDIATE AND DIRECT ADDRESSING MODES

1. No
2. MOVLW B'10000000'
3. PIC does not allow us to move a literal value directly to file register locations.
4. True
5. False

SECTION 6.2: REGISTER INDIRECT ADDRESSING MODE

1. Direct. The memory location address is 0x40.
2. The lower 8 bits of the 12-bit address of the data RAM file register. The address is 0FE9H.
3. The upper 4 bits of the 12-bit address of the data RAM file register. The address is 0FEAH.
4. 12-bit
5. FSR0, FSR1, and FSR2

SECTION 6.3: LOOK-UP TABLE AND TABLE PROCESSING

1. TBLPTR
2. TBLPTR
3. TABLAT
4. 21-bit, 2 MB
5. TBLPTR
6. In TBLRD++ the element is read first, and then TBLPTR is incremented, while in TBLRD+*,
   TBLPTR is incremented first, and then the element is fetched.
7. False. Only with flash ROM.

SECTION 6.4: BIT-ADDRESSABILITY OF DATA RAM

1. True
2. True
3. True
4. a, b, c, and d
5. All of them
6. BTFSS 0x03,1
7. (a) It sets to HIGH bit 1 of RAM location 20H.
   (b) It clears bit 7 of RAM location 32H.
   (c) It sets to HIGH bit 2 of RAM location 12H.
   (d) It sets to HIGH bit 4 of PORTB.
   (e) It sets to HIGH bit 1 of the status register.
8. BCF STATUS,C

SECTION 6.5: BANK SWITCHING IN THE PIC18

1. True
2. True
3. True
4. F80–FFFH
5. MYREG EQU 0x2
   MOVLB 0x02 ;load 2 into BSR (use bank 2)
   MOVLW 0x99 ;WREG = 99h
   MOVWF MYREG,1
6. MYREG EQU 0x08
   MOVLB 0x4 ; load 4 into BSR (use bank 4)
   MOVLO 0x55 ; WREG = 55h
   MOVWF MYREG, 1

7. True
8. 8-bit, 4

SECTION 6.6: CHECKSUM AND ASCII SUBROUTINES
1. The 35H and 37H give 57H in BCD. The 39H and 34H give 94H.
2. The ASCII data is 32H, 30H, 30H, 35H, while 05 and 20H are for BCD.
3. No. To make it BCD, the radix must be set to hex.
4. 33H and 33H
5. 242 or 32H, 34H and 32H
6. ROM
7. 88H + 99H + AAH + BBH + CCH + DDH = 42FH. Dropping the carries we have 2FH, and its 2's complement is D1H.
8. False

SECTION 6.7: MACROS AND MODULES
1. Macro programming can save the programmer time by allowing a set of frequently repeated instructions to be invoked within the program with a single line. This can also make the code easier to read.
2. The three parts of a macro are the MACRO directive, the body, and the ENDM directive.
3. The macro definition is the list of statements the macro will perform. It begins with the MACRO directive and ends with the ENDM directive. The macro is invoked whenever it is called from within an Assembly language program. The macro is expanded when the Assembly program replaces the line invoking the macro with the Assembly language code in the body of the macro.
4. False. A label that is to be local to a macro must be declared local with the LOCAL directive.
5. GLOBAL
6. EXTERN
CHAPTER 7

PIC PROGRAMMING
IN C

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Examine C data types for the PIC18
- Code C18 programs for time delay and I/O operations
- Code C18 programs for I/O bit manipulation
- Code C18 programs for logic and arithmetic operations
- Code C18 programs for ASCII and BCD data conversion
- Code C18 programs for binary (hex) to decimal conversion
- Code C18 programs for data serialization
- Understand C18 C compiler RAM and ROM allocation
Why program the PIC18 in C?

Compilers produce hex files that we download into the ROM of the microcontroller. The size of the hex file produced by the compiler is one of the main concerns of microcontroller programmers for two reasons:

1. Microcontrollers have limited on-chip ROM.
2. The code space for the PIC18 is limited to 2M.

How does the choice of programming language affect the compiled program size? While Assembly language produces a hex file that is much smaller than C, programming in Assembly language is often tedious and time consuming. On the other hand, C programming is less time consuming and much easier to write, but the hex file size produced is much larger than if we used Assembly language. The following are some of the major reasons for writing programs in C instead of Assembly:

1. It is easier and less time consuming to write in C than in Assembly.
2. C is easier to modify and update.
3. You can use code available in function libraries.
4. C code is portable to other microcontrollers with little or no modification.

Several third-party companies develop C compilers for the PIC microcontroller. Our goal is not to recommend one over another, but to provide you with the fundamentals of C programming for the PIC18. You can use the compiler of your choice for the chapter examples and programs. For this book we have chosen Microchip’s C18 C compiler to integrate with MPLAB IDE. Microchip has a student version of the C18 C compiler available for download from their web site. See http://www.MicroDigitalEd.com for tutorials on the C18 C compiler and MPLAB simulator.

C programming for the PIC18 is the main topic of this chapter. In Section 7.1, we discuss data types, and time delays. I/O programming is shown in Section 7.2. The logic operations AND, OR, XOR, inverter, and shift are discussed in Section 7.3. Section 7.4 describes ASCII and BCD conversions and checksums. In Section 7.5, data serialization for the PIC18 is shown. In Section 7.6, we show how the C18 C compiler uses program ROM for data storage. We will examine C18 data RAM allocation in Section 7.7.

SECTION 7.1: DATA TYPES AND TIME DELAYS IN C

In this section we first discuss C data types for the PIC18 and then provide code for time delay functions.

C data types for the PIC18

One of the goals of C18 programmers is to create smaller hex files, so it is worthwhile to re-examine C data types for C18. In other words, a good understanding of C data types for the C18 can help programmers to create smaller hex
files. In this section we focus on the specific C data types that are most useful and widely used for the PIC18 microcontroller. Table 7-1 shows data types and sizes.

### Table 7-1: Some Data Types Widely Used by C18

<table>
<thead>
<tr>
<th>Data Type</th>
<th>Size in Bits</th>
<th>Data Range/Usage</th>
</tr>
</thead>
<tbody>
<tr>
<td>unsigned char</td>
<td>8-bit</td>
<td>0 to 255</td>
</tr>
<tr>
<td>char</td>
<td>8-bit</td>
<td>-128 to +127</td>
</tr>
<tr>
<td>unsigned int</td>
<td>16-bit</td>
<td>0 to 65,535</td>
</tr>
<tr>
<td>int</td>
<td>16-bit</td>
<td>-32,768 to +32,767</td>
</tr>
<tr>
<td>unsigned short</td>
<td>16-bit</td>
<td>0 to 65,535</td>
</tr>
<tr>
<td>short</td>
<td>16-bit</td>
<td>-32,768 to +32,767</td>
</tr>
<tr>
<td>unsigned short long</td>
<td>24-bit</td>
<td>0 to 16,777,215</td>
</tr>
<tr>
<td>short long</td>
<td>24-bit</td>
<td>-8,388,608 to +8,388,607</td>
</tr>
<tr>
<td>unsigned long</td>
<td>32-bit</td>
<td>0 to 4,294,967,295</td>
</tr>
<tr>
<td>long</td>
<td>32-bit</td>
<td>-2,147,483,648 to +2,147,483,648</td>
</tr>
</tbody>
</table>

**Unsigned char**

Because the PIC18 is an 8-bit microcontroller, the character data type is the most natural choice for many applications. The unsigned char is an 8-bit data type that takes a value in the range of 0–255 (00–FFH). It is one of the most widely used data types for the PIC18. In many situations, such as setting a counter value, where there is no need for signed data, we should use the unsigned char instead of the signed char. Remember that C compilers use the signed char as the default unless we put the keyword `unsigned` in front of the char (see Example 7-1). We can also use the unsigned char data type for a string of ASCII characters, including extended ASCII characters. Example 7-2 shows a string of ASCII characters. See Example 7-3 for toggling ports.

In declaring variables, we must pay careful attention to the size of the data and try to use unsigned char instead of int if possible. Because the PIC18 microcontroller has a limited number of registers and data RAM locations, using int in place of char can lead to a larger-size hex file. Such misuse of data types in compilers such as Microsoft Visual C++ for x86 IBM PCs is not a significant issue.

---

**Example 7-1**

Write a C18 program to send values 00–FF to Port B.

**Solution:**

```c
#include <P18F458.h>     //for TRISB and PORTB declarations

void main(void) {
    unsigned char z;
    TRISB = 0;           //make Port B an output
    for(z=0; z<=255; z++)
        PORTB = z;      //NEEDED IF RUNNING IN HARDWARE
    while(1);
}
```

Run the above program on your simulator to see how Port B displays values 00–FFH in binary. Notice that “while(1)” is needed if this program is running in hardware.
Example 7-2

Write a C18 program to send hex values for ASCII characters of 0, 1, 2, 3, 4, 5, A, B, C, and D to Port B.

Solution:
#include <P18F458.h>
void main(void)
{
    unsigned char mynum[] = "012345ABCD"; //data is stored in RAM
    unsigned char z;
    TRISB = 0; //make Port B an output
    for(z=0;z<10;z++)
    {
        PORTB = mynum[z]; //stay here forever
    }
}

Run the above program on your simulator to see how Port B displays values 30H, 31H, 32H, 33H, 34H, 35H, 41H, 42H, 43H, and 44H (the hex values for ASCII 0, 1, 2, etc.). Notice that the last statement “while(1)” is needed only if we run the program in hardware. This is like "GOTO $" or "BRA $" in Assembly language.

Example 7-3

Write a C18 program to toggle all the bits of Port B continuously.

Solution:

// Toggle PB forever
#include <P18F458.h>
void main(void)
{
    TRISB = 0; //make Port B an output
    for(;;) //repeat forever
    {
        PORTB = 0x55; //0x indicates the data is in hex (binary)
        PORTB = 0xAA;
    }
}

Run the above program on your simulator to see how Port B toggles continuously.

Figure 7-1. Examining the Special Function Registers Using MPLAB
Signed char

The signed char is an 8-bit data type that uses the most significant bit (D7 of D7–D0) to represent the – or + value. As a result, we have only 7 bits for the magnitude of the signed number, giving us values from −128 to +127. In situations where + and – are needed to represent a given quantity such as temperature, the use of the signed char data type is necessary.

Again, notice that if we do not use the keyword unsigned, the default is the signed value. For that reason we should stick with the unsigned char unless the data needs to be represented as signed numbers.

Example 7-4

Write a C18 program to send values of −4 to +4 to Port B.

Solution:
//sign numbers
#include <P18F458.h>
void main(void)
{
    char mynum[] = {+1,-1,+2,-2,+3,-3,+4,-4};
    unsigned char z;
    TRISB = 0; //make Port B an output
    for(z=0;z<8;z++)
    {
        PORTB = mynum[z];
        while(1); //stay here forever
    }
}

Run the above program on your simulator to see how PORTB displays values of 1, FFH, 2, FEH, 3, FDH, 4, and FCH (the hex values for +1, −1, +2, −2, etc.). See Chapter 5 for discussion of signed numbers.

Unsigned int

The unsigned int is a 16-bit data type that takes a value in the range of 0 to 65,535 (0000–FFFFH). In the PIC18, unsigned int is used to define 16-bit variables such as memory addresses. It is also used to set counter values of more than 256. Because the PIC18 is an 8-bit microcontroller and the int data type takes two bytes of RAM, we must not use the int data type unless we have to. Because registers and memory accesses are in 8-bit chunks, the misuse of int variables will result in a larger hex file. Such misuse is not a problem in PCs with 512 megabytes of memory, the 32-bit Pentium’s registers and memory accesses, and a bus speed of 133 MHz. For PIC18 programming, however, do not use signed int in places where unsigned char will do the job. Of course, the compiler will not generate an error for this misuse, but the overhead in hex file size will be noticeable. Also, in situations where there is no need for signed data (such as setting counter values), we should use unsigned int instead of signed int. This gives a much wider range for data declaration. Again, remember that the C compiler uses signed int as the default unless we specify the keyword unsigned.
Signed int

Signed int is a 16-bit data type that uses the most significant bit (D15 of D15–D0) to represent the – or + value. As a result, we have only 15 bits for the magnitude of the number, or values from –32,768 to +32,767.

Other data types

The unsigned int is limited to values 0–65,535 (0000–FFFFH). The C18 C compiler supports both short long and long data types, if we want values greater than 16-bit. See Table 7-1. The short long value is 24 bits wide, while the long value is 32 bits wide.

Example 7-5

Write a C18 program to toggle all bits of Port B 50,000 times.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned int z;
    TRISB = 0;       // make Port B an output
    for(z=0;z<=50000;z++)
    {
        PORTB = 0x55;
        PORTB = 0xAA;
    }
    while(1);        // stay here forever
}
```

Run the above program on your simulator to see how Port B toggles continuously. Notice that the maximum value for unsigned int is 65,535.

Example 7-6

Write a C18 program to toggle all bits of Port B 100,000 times.

Solution:

```c
// toggle PB 100,000 times
#include <P18F458.h>
void main(void)
{
    unsigned short long z;
    unsigned int x;
    TRISB = 0;       // make Port B an output
    for(z=0;z<=100000;z++)
    {
        PORTB = 0x55;
        PORTB = 0xAA;
    }
    while(1);        // stay here forever
}
```
Time delay

There are two ways to create a time delay in C18:

1. Using a simple for loop
2. Using the PIC18 timers

In either case, when we write a time delay we must use the oscilloscope to measure the duration of our time delay. Next, we use the for loop to create time delays. The use of the PIC18 timer to create time delays is postponed until Chapter 9.

In creating a time delay using a for loop, we must be mindful of two factors that can affect the accuracy of the delay:

1. The crystal frequency connected to the OSC1–OSC2 input pins is the most important factor in the time delay calculation. The duration of the clock period for the instruction cycle is a function of this crystal frequency.
2. The second factor that affects the time delay is the compiler used to compile the C program. When we program in Assembly language, we can control the exact instructions and their sequences used in the delay subroutine. In the case of C programs, it is the C compiler that converts the C statements and functions to Assembly language instructions. As a result, different compilers produce different code. In other words, if we compile a given C program with different compilers, each compiler produces different-size hex code.

For the above reasons, when we write time delays for C, we must use the oscilloscope to measure the exact duration. Look at Examples 7-7 and 7-8.

Example 7-7

Write a C18 program to toggle all the bits of Port B ports continuously with a 250 ms delay. Assume that the system is PIC18F458 with XTAL = 10 MHz.

Solution:

```c
#include <P18F458.h>
void MSDelay(unsigned int);
void main(void)
{
    TRISB = 0;         //make Port B an output
    while(1)           //repeat forever
    {
        PORTB = 0x55;
        MSDelay(250);
        PORTB = 0xAA;
        MSDelay(250);
    }
}

void MSDelay(unsigned int itime)
{
    unsigned int i; unsigned char j;
    for(i=0;i<itime;i++)
        for(j=0;j<165;j++);
}
```
Example 7-8

Write a C18 program to toggle all the bits of Port C and Port D continuously with a 250 ms delay.

Solution:
// this program is tested for the PIC18F458 with XTAL = 10 MHz
#include <P18F458.h>
void MSDelay(unsigned int);
void main(void)
{
    TRISC = 0;
    TRISD = 0;       // make Ports C and D output
    while(1)         // another way to do it forever
    {
        PORTC = 0x55;
        PORTD = 0x55;
        MSDelay(250);
        PORTC = 0xAA;
        PORTD = 0xAA;
        MSDelay(250);
    }
}
void MSDelay(unsigned int itime)
{
    unsigned int i; unsigned char j;
    for(i=0;i<itime;i++)
    for(j=0;j<165;j++);
}

Figure 7-2. Time Delay Measurement for Example 7-8 Using MPLAB

Review Questions

1. Give the magnitude of the unsigned char and signed char data types.
2. Give the magnitude of the unsigned int and signed int data types.
3. If we are declaring a variable for a person’s age, we should use the ___ data type.
4. True or false. Using a for loop to create a time delay is not recommended if you want your code be portable to other PIC18-based systems.
5. Give two factors that can affect the delay size.
SECTION 7.2: I/O PROGRAMMING IN C

In this section we look at C programming of the I/O ports for the PIC18. We look at both byte and bit I/O programming.

Byte size I/O

As we stated in Chapter 4, ports PORTA–PORTD are byte accessible. We use the PORTA–PORTD labels as defined in the C18 header file. See Example 7-9. Examine the next few examples to get a better understanding of how ports are accessed in C18.

Example 7-9

LEDs are connected to bits in Port B and Port C. Write a C18 program that shows the count from 0 to FFH (0000 0000 to 1111 1111 in binary) on the LEDs.

Solution:

```c
#include <P18F458.h>  //notice how we can define Port C
#define LED PORTC

void main(void)
{
    TRISB = 0;  //make Port B an output
    TRISC = 0;  //make Port C an output
    PORTB = 00; //clear Port B
    LED = 0;    //clear Port C
    for(;;)       //repeat forever
    {
        PORTB++;   //increment Port B
        LED++;     //increment Port C
    }
}
```

Figure 7-3. Example 7-9 Results After 24 Iterations
Example 7-10

Write a C18 program to get a byte of data from Port B, wait 1/2 second, and then send it to Port C.

Solution:

```c
#include <P18F458.h>
void MSDelay(unsigned int);
void main(void)
{
    unsigned char mybyte;
    TRISB = 0xFF;       //Port B as input
    TRISC = 0;          //Port C as output
    while(1)
    {
        mybyte = PORTB;   //get a byte from Port B
        MSDelay(500);
        PORTC = mybyte;   //send it to Port C
    }
}

void MSDelay(unsigned int itime)
{
    unsigned int i;
    unsigned char j;
    for(i=0;i<itime;i++)
        for(j=0;j<165;j++);
}
```

Example 7-11

Write a C18 program to get a byte of data from Port C. If it is less than 100, send it to Port B; otherwise, send it to Port D.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char mybyte;
    TRISC = 0xFF;       //make Port C an input
    TRISB = 0;
    TRISD = 0;          //both Port B and D as output
    while(1)
    {
        mybyte = PORTC;   //get a byte from PORTC
        if(mybyte < 100)
            PORTB = mybyte;   //send it to PORTB if less than 100
        else
            PORTD = mybyte;   //send it to PORTD if more than 100
    }
}
Bit-addressable I/O programming

The I/O ports of PIC18 are bit-addressable. We can access a single bit without disturbing the rest of the port. We use PORTxbits.Rxy to access a single bit of Portx, where x is the port A, B, C, or D, and y is the bit (0–7) of that port. For example, PORTBbits.RB7 indicates PORTB.7. We access the TRISx registers in the same way where TRISBbits.TRISB7 indicates the D7 of the TRISB. Table 7-2 shows the single-bit addresses of a PIC18. Study the next few examples to become familiar with the syntax.

Table 7-2: Single-Bit Addresses of PIC18F458/4580 Ports

<table>
<thead>
<tr>
<th>PORTA</th>
<th>PORTB</th>
<th>PORTC</th>
<th>PORTD</th>
<th>PORTE</th>
<th>Port's Bit</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0</td>
<td>RB0</td>
<td>RC0</td>
<td>RD0</td>
<td>RE0</td>
<td>D0</td>
</tr>
<tr>
<td>RA1</td>
<td>RB1</td>
<td>RC1</td>
<td>RD1</td>
<td>RE1</td>
<td>D1</td>
</tr>
<tr>
<td>RA2</td>
<td>RB2</td>
<td>RC2</td>
<td>RD2</td>
<td>RE2</td>
<td>D2</td>
</tr>
<tr>
<td>RA3</td>
<td>RB3</td>
<td>RC3</td>
<td>RD3</td>
<td></td>
<td>D3</td>
</tr>
<tr>
<td>RA4</td>
<td>RB4</td>
<td>RC4</td>
<td>RD4</td>
<td></td>
<td>D4</td>
</tr>
<tr>
<td>RA5</td>
<td>RB5</td>
<td>RC5</td>
<td>RD5</td>
<td></td>
<td>D5</td>
</tr>
<tr>
<td>RA6</td>
<td>RB6</td>
<td>RC6</td>
<td>RD6</td>
<td></td>
<td>D6</td>
</tr>
<tr>
<td>RA7</td>
<td>RB7</td>
<td>RC7</td>
<td>RD7</td>
<td></td>
<td>D7</td>
</tr>
</tbody>
</table>

PORT bits structure

Figure 7-4 shows the structure for the Port B bits as given by the C18 C compiler. You can find the structure of the ports in the microcontroller header file.

```c
extern volatile near unsigned char PORTB;
extern volatile near union {
    struct {
        unsigned RB0:1;
        unsigned RB1:1;
        unsigned RB2:1;
        unsigned RB3:1;
        unsigned RB4:1;
        unsigned RB5:1;
        unsigned RB6:1;
        unsigned RB7:1;
    }
    struct {
        unsigned INTO:1;
        unsigned INT1:1;
        unsigned CANTX:1;
        unsigned CANRX:1;
        unsigned :1;
        unsigned PGM:1;
        unsigned PGC:1;
        unsigned PGD:1;
    }
} PORTBbits;
```

Figure 7-4. Port B Bit Structure
Example 7-12

Write a C18 program to toggle only bit RB4 continuously without disturbing the rest of the bits of Port B.

Solution:
#include <P18F458.h>
#define mybit PORTBbits.RB4  //declare single bit
void main(void)
{
    TRISBbits.TRISB4=0;  //make RB4 an output
    while(1)
    {
        mybit = 1;  //turn on RB4
        mybit = 0;  //turn off RB4
    }
}

Example 7-13

Write a C18 program to monitor bit PC5. If it is HIGH, send 55H to Port B; otherwise, send AAH to Port D.

Solution:
#include <P18F458.h>
#define mybit PORTCbits.RC5  //notice single-bit declaration
void main(void)
{
    TRISCbits.TRISC5 = 1;  //RC5 as input
    TRISD = 0;  //Ports C and D output
    while(1)
    {
        if(mybit == 1)
            PORTD = 0x55;
        else
            PORTD = 0xAA;
    }
}

Figure 7-5. Example 7-13 Results on MPLAB Simulator
Example 7-14

A door sensor is connected to the RB1 pin, and a buzzer is connected to RC7. Write a C18 program to monitor the door sensor, and when it opens, sound the buzzer. You can sound the buzzer by sending a square wave of a few hundred Hz frequency to it.

Solution:

```c
#include <P18F458.h>
void MSDelay(unsigned int);
#define Dsensor PORTBbits.RB1
#define buzzer PORTCbits.RC7

void main(void)
{
    TRISBbits.TRISB1 = 1;  //PORTB.1 as an input
    TRISCbits.TRISC7 = 0;  //make PORTC.7 an output

    while(Dsensor == 1)
    {
        buzzer = 0;
        MSDelay(200);
        buzzer = 1;
        MSDelay(200);
    }
    while(1);              //stay here forever
}

void MSDelay(unsigned int itime)
{
    unsigned int i;
    unsigned char j;
    for(i=0;i<itime;i++)
        for(j=0;j<165;j++);
}
```

Figure 7-6. MPLAB Logic Analyzer for Example 7-14
Example 7-15

The data pins of an LCD are connected to Port B. The information is latched into the LCD whenever its Enable pin goes from HIGH to LOW. Write a C18 program to send "The Earth is but One Country" to this LCD.

Solution:

```c
#include <P18F458.h>
#define LCDData PORTB  //LCDData declaration
#define En PORTCbits.RC2  //the Enable pin

void main(void)
{
    unsigned char message[] = "The Earth is but One Country";
    unsigned char z;
    TRISB = 0;  //Port B as output
    TRISCbits.TRISC2 = 0;  //PortC.2 as output
    for(z=0; z<28; z++) //send all the 28 characters
    {
        LCDData = message[z];  //a HIGH-
        En=1;  //to-LOW pulse to latch the LCD data
        En=0;
    }
    while(1);  //stay here forever
}
```

Run the above program on your simulator to see how PORTB displays each character of the message. Meanwhile, monitor bit PC.2 after each character is issued.

Figure 7-7. MPLAB Logic Analyzer Results for Example 7-15
Example 7-16

Write a C18 program to toggle all the bits of Port B, Port C, and Port D continuously with a 250 ms delay.

Solution:

```c
#include <P18F458.h>
void MSDelay(unsigned int);  
void main(void) {  
    TRISB = 0;  
    TRISC = 0;  
    TRISD = 0;  
    while(1) //do it forever  
        {  
            PORTB = 0x55;  
            PORTC = 0x55;  
            PORTD = 0x55;  
            MSDelay(250); //250 ms delay  
            PORTB = 0xAA;  
            PORTC = 0xAA;  
            PORTD = 0xAA;  
            MSDelay(250);  
        }  
}  
void MSDelay(unsigned int itime) {  
    unsigned int i;  
    unsigned char j;  
    for(i=0;i<itime;i++)  
        for(j=0;j<165;j++);  
}
```

Example 7-17

Write a C18 program to turn bit 5 of Port B on and off 50,000 times.

Solution:

```c
#include <P18F458.h>
#define MYBIT PORTBbits.RB5  
void main(void) {  
    unsigned int z;  
    TRISBbits.TRISB5 = 0; //make PORTB.5 an output  
    for(z=0;z<50000;z++)  
        {  
            MYBIT = 1;  
            MYBIT = 0;  
        }  
    while(1); //stay here forever  
}
```
Example 7-18

Write a C18 program to get the status of bit RB0, and send it to RC7 continuously.

Solution:

```c
#include <P18F458.h>
define inbit PORTBbits.RB0
#define outbit PORTCbits.RC7
void main(void)
{
  TRISBbits.TRISB0 = 1;  //make RB0 an input
  TRISCbits.TRISC7 = 0;  //make RC7 an output
  while(1)
  {
    outbit = inbit;  //get a bit from RB0
    //and send it to RC7
  }
}
```

Review Questions

1. The address of PORTB is __________.
2. Write a short program that toggles all bits of PORTC.
3. Write a short program that toggles only bit 0 of PORTB.
4. True or false. All bits of PORTB are bit addressable.
5. True or false. All bits of TRISB are bit addressable.
SECTION 7.3: LOGIC OPERATIONS IN C

One of the most important and powerful features of the C language is its ability to perform bit manipulation. Because many books on C do not cover this important topic, it is appropriate to discuss it in this section. This section describes the action of bit-wise logic operators and provides some examples of how they are used.

Table 7-3: Bit-wise Logic Operators for C

<table>
<thead>
<tr>
<th></th>
<th>AND</th>
<th>OR</th>
<th>EX-OR</th>
<th>Inverter</th>
</tr>
</thead>
<tbody>
<tr>
<td>A</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>B</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>A&amp;B</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>A</td>
<td>B</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>A^B</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Y=~B</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Bit-wise operators in C

While every C programmer is familiar with the logical operators AND (&&), OR (||), and NOT (!), many C programmers are less familiar with the bit-wise operators AND (&), OR (|), EX-OR (^), inverter (~), shift right (>>, and shift left (<<). These bit-wise operators are widely used in software engineering for embedded systems and control; consequently, their understanding and mastery are critical in microprocessor-based system design and interfacing. See Table 7-3.

The following shows some examples using the C bit-wise operators:

1. 0x35 & 0x0F = 0x05 /* ANDing */
2. 0x04 | 0x68 = 0x6C /* ORing: */
3. 0x54 ^ 0x78 = 0x2C /* XORing */
4. ~0x55 = 0xAA /* Inverting 55H */

Bit-wise shift operation in C

There are two bit-wise shift operators in C: (1) shift right (>>), and (2) shift left (<<).

Their format in C is as follows:

data >> number of bits to be shifted right

data << number of bits to be shifted left

The following shows some examples of shift operators in C:

1. 0x9A >> 3 = 0x13 /* shifting right 3 times */
2. 0x77 >> 4 = 0x07 /* shifting right 4 times */
3. 0x6 << 4 = 0x60 /* shifting left 4 times */

Study Examples 7-19 through 7-22. These show how the bit-wise operators are used in C.
Example 7-19

Run the following program on your simulator and examine the results.

Solution:

```c
#include <P18F458.h>
void main (void)
{
    TRISB = 0; //make Ports B, C,
    TRISC = 0; //and D output ports
    TRISD = 0;
    PORTB = 0x35 & 0x0F; //ANDing
    PORTC = 0x04 | 0x68; //ORing
    PORTD = 0x54 ^ 0x78; //XORing
    PORTB = ~0x55; //inverting
    PORTC = 0x9A >> 3; //shifting right 3 times
    PORTD = 0x77 >> 4; //shifting right 4 times
    PORTB = 0x6 << 4; //shifting left 4 times
    while(1);
    //stay here forever
}
```

Example 7-20

Write a C18 program to toggle all the bits of Port B and Port C continuously with a 250 ms delay. Use the inverting operator.

Solution:

```c
#include <P18F458.h>
void MSDelay(unsigned int); 
void main(void)
{
    TRISB = 0; //make Ports B and C output
    TRISC = 0;
    PORTB = 0x55;
    PORTC = 0xAA;
    while(1)
    {
        PORTB = ~PORTB;
        PORTC = ~PORTC;
        MSDelay(250);
    }
}

void MSDelay(unsigned int itime)
{
    unsigned int i;
    unsigned char j;
    for(i=0;i<itime;i++)
        for(j=0;j<165;j++)
}
Example 7-21

Rewrite the C18 program to toggle all the bits of Port B, Port C, and Port D continuously with a 250 ms delay. Use the EX-OR operator.

Solution:

```c
#include <P18F4550.h>
void MSDelay(unsigned int);
void main(void)
{
    TRISB = 0;
    TRISC = 0;
    TRISD = 0;   //make Ports B,C, and D output
    PORTB=0x55;
    PORTC=0x55;
    PORTD=0x55;
    while(1)
    {
        PORTB=PORTB^0xFF;
        PORTC=PORTC^0xFF;
        PORTD=PORTD^0xFF;
        MSDelay(250);
    }
}
void MSDelay(unsigned int itime)
{
    unsigned int i;
    unsigned char j;
    for(i=0;i<itime;i++)
        for(j=0;j<165;j++);
}
```

Example 7-22

Rewrite the C18 program to get bit RB0 and send it to RC7 after inverting it.

Solution:

```c
#include <P18F4550.h>
#define inbit PORTBbits.RB0
#define outbit PORTCbits.RC7
void main(void)
{
    TRISBbits.TRISB0 = 1;    //make PORTB.0 an input
    TRISChbits.TRISC7 = 0;   //make PORTC.7 an output
    while(1)
    {
        outbit = ~inbit;    //get a bit from RB0
    }
```
Example 7-23

Write a PIC C18 program to read the RB0 and RB1 bits and issue an ASCII character to PD according to the following table:

<table>
<thead>
<tr>
<th>RB1</th>
<th>RB0</th>
<th>Action</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>send ‘0’ to PORTD (notice ASCII ‘0’ is 0x30)</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>send ‘1’ to PORTD</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>send ‘2’ to PORTD</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>send ‘3’ to PORTD</td>
</tr>
</tbody>
</table>

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char z;
    TRISB = 0xFF;     //make Port B an input
    TRISD = 0;        //make Port D an output
    while(1)          //repeat forever
    {
        z = PORTB;      //read PORTB
        z = z & 0x3;    //mask the unused bits
        switch(z)       //make decision
        {
            case(0):
            {
                PORTD = '0'; //issue ASCII 0
                break;
            }
            case(1):
            {
                PORTD = '1'; //issue ASCII 1
                break;
            }
            case(2):
            {
                PORTD = '2'; //issue ASCII 2
                break;
            }
            case(3):
            {
                PORTD = '3'; //issue ASCII 3
                break;
            }
        }
    }
}
```

---

Special Function Registers

<table>
<thead>
<tr>
<th>Address</th>
<th>SFR Name</th>
<th>Hex</th>
<th>Decimal</th>
<th>Binary</th>
<th>Char</th>
</tr>
</thead>
<tbody>
<tr>
<td>0F80</td>
<td>PORTA</td>
<td>00</td>
<td>0</td>
<td>00000000</td>
<td>.</td>
</tr>
<tr>
<td>0F81</td>
<td>PORTB</td>
<td>02</td>
<td>2</td>
<td>00000010</td>
<td></td>
</tr>
<tr>
<td>0F82</td>
<td>PORTC</td>
<td>00</td>
<td>0</td>
<td>00000000</td>
<td>.</td>
</tr>
<tr>
<td>0F83</td>
<td>PORTD</td>
<td>32</td>
<td>50</td>
<td>00110010</td>
<td>2</td>
</tr>
<tr>
<td>0F84</td>
<td>PORTE</td>
<td>00</td>
<td>0</td>
<td>00000000</td>
<td>.</td>
</tr>
</tbody>
</table>
Review Questions

1. Find the content of PORTB after the following C code in each case:
   (a) `PORTB=0x37&0xCA;`
   (b) `PORTB=0x37|0xCA;`
   (c) `PORTB=0x37^0xCA;`

2. To mask certain bits we must AND them with _____.

3. To set high certain bits we must OR them with _____.

4. EX-ORing a value with itself results in _____.

5. Find the contents of PORTC after execution of the following code:
   ```c
   PORTC = 0;
   PORTC = PORTC | 0x99;
   PORTC = ~PORTC;
   ```

SECTION 7.4: DATA CONVERSION PROGRAMS IN C

Recall that BCD numbers were discussed in Chapters 5 and 6. As stated there, many newer microcontrollers have a real-time clock (RTC) where the time and date are kept even when the power is off. Very often the RTC provides the time and date in packed BCD. To display them, however, it must convert them to ASCII. In this section we show the application of logic and rotate instructions in the conversion of BCD and ASCII.

ASCII numbers

On ASCII keyboards, when the key "0" is activated, "011 0000" (30H) is provided to the computer. Similarly, 31H (011 0001) is provided for the key "1", and so on, as shown in Table 7-4.

<table>
<thead>
<tr>
<th>Key</th>
<th>ASCII (hex)</th>
<th>Binary</th>
<th>BCD (unpacked)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>30</td>
<td>011 0000</td>
<td>0000 0000</td>
</tr>
<tr>
<td>1</td>
<td>31</td>
<td>011 0001</td>
<td>0000 0001</td>
</tr>
<tr>
<td>2</td>
<td>32</td>
<td>011 0010</td>
<td>0000 0010</td>
</tr>
<tr>
<td>3</td>
<td>33</td>
<td>011 0011</td>
<td>0000 0011</td>
</tr>
<tr>
<td>4</td>
<td>34</td>
<td>011 0100</td>
<td>0000 0100</td>
</tr>
<tr>
<td>5</td>
<td>35</td>
<td>011 0101</td>
<td>0000 0101</td>
</tr>
<tr>
<td>6</td>
<td>36</td>
<td>011 0110</td>
<td>0000 0110</td>
</tr>
<tr>
<td>7</td>
<td>37</td>
<td>011 0111</td>
<td>0000 0111</td>
</tr>
<tr>
<td>8</td>
<td>38</td>
<td>011 1000</td>
<td>0000 1000</td>
</tr>
<tr>
<td>9</td>
<td>39</td>
<td>011 1001</td>
<td>0000 1001</td>
</tr>
</tbody>
</table>
Packed BCD to ASCII conversion

The RTC provides the time of day (hour, minute, second) and the date (year, month, day) continuously, regardless of whether the power is on or off. This data is provided in packed BCD, however. To convert packed BCD to ASCII, you must first convert it to unpacked BCD. Then the unpacked BCD is tagged with 0110000 (30H). The following demonstrates converting from packed BCD to ASCII. See also Example 7-24.

<table>
<thead>
<tr>
<th>Packed BCD</th>
<th>Unpacked BCD</th>
<th>ASCII</th>
</tr>
</thead>
<tbody>
<tr>
<td>0x29</td>
<td>0x02, 0x09</td>
<td>0x32, 0x39</td>
</tr>
<tr>
<td>00101001</td>
<td>00000010, 00001001</td>
<td>00110010, 00111001</td>
</tr>
</tbody>
</table>

ASCII to packed BCD conversion

To convert ASCII to packed BCD, you first convert it to unpacked BCD (to get rid of the 3), and then combine to make packed BCD. For example, 4 and 7 on the keyboard give 34H and 37H, respectively. The goal is to produce 47H or "01000111", which is packed BCD.

<table>
<thead>
<tr>
<th>Key</th>
<th>ASCII</th>
<th>Unpacked BCD</th>
<th>Packed BCD</th>
</tr>
</thead>
<tbody>
<tr>
<td>4</td>
<td>34</td>
<td>00000100</td>
<td>01000111 or 47H</td>
</tr>
<tr>
<td>7</td>
<td>37</td>
<td>00000111</td>
<td></td>
</tr>
</tbody>
</table>

After this conversion, the packed BCD numbers are processed and the result will be in packed BCD format. Chapter 16 discusses the RTC chip and uses the BCD and ASCII conversion programs shown in Examples 7-24 and 7-25.

Example 7-24

Write a C18 program to convert packed BCD 0x29 to ASCII and display the bytes on PORTB and PORTC.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char x, y, z;
    unsigned char mybyte = 0x29;
    TRISB = 0;   //make Ports B and C output
    TRISC = 0;   //mask upper 4 bits
    x = mybyte & 0x0F;   //make it ASCII
    PORTB |= 0x30;    //shift it to lower 4 bits
    y = mybyte & 0xF0;  //make it ASCII
    PORTC = y | 0x30;  //mask lower 4 bits
}
```
Example 7-25

Write a C18 program to convert ASCII digits of ‘4’ and ‘7’ to packed BCD and display it on PORTB.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char bcdbyte;
    unsigned char w = '4';
    unsigned char z = '7';
    TRISB = 0;    // make Port B an output
    w = w & 0x0F;   // mask 3
    w = w << 4;    // shift left to make upper BCD digit
    z = z & 0x0F;   // mask 3
    bcdbyte = w | z;  // combine to make packed BCD
    PORTB = bcdbyte;
}
```

Figure 7-9. Result of Example 7-24 on MPLAB Simulator.

Figure 7-10. Result of Example 7-25 on MPLAB Simulator.
Checksum byte in ROM

To ensure the integrity of ROM contents, every system must perform the checksum calculation. The checksum will detect any corruption of the contents of ROM. One of the causes of ROM corruption is current surge, either when the system is turned on or during operation. To ensure data integrity in ROM, the checksum process uses what is called a checksum byte. The checksum byte is an extra byte that is tagged to the end of a series of bytes of data. To calculate the checksum byte of a series of bytes of data, the following steps can be taken:

1. Add the bytes together and drop the carries.
2. Take the 2's complement of the total sum. This is the checksum byte, which becomes the last byte of the series.

To perform the checksum operation, add all the bytes, including the checksum byte. The result must be zero. If it is not zero, one or more bytes of data have been changed (corrupted). To clarify these important concepts, see Examples 7-26 through 7-28.

**Example 7-26**

Assume that we have 4 bytes of hexadecimal data: 25H, 62H, 3FH, and 52H.
(a) Find the checksum byte, (b) perform the checksum operation to ensure data integrity, and (c) if the second byte, 62H, has been changed to 22H, show how checksum detects the error.

**Solution:**

(a) Find the checksum byte.

\[
\begin{align*}
25H &+ 62H \\
&+ 3FH \\
&+ 52H \\
\hline
118H \quad & (\text{dropping carry of 1 and taking 2's complement, we get E8H})
\end{align*}
\]

(b) Perform the checksum operation to ensure data integrity.

\[
\begin{align*}
25H &+ 62H \\
&+ 3FH \\
&+ 52H \\
&+ E8H \\
\hline
200H \quad & (\text{dropping the carries we get 00, which means data is not corrupted})
\end{align*}
\]

(c) If the second byte, 62H, has been changed to 22H, show how checksum detects the error.

\[
\begin{align*}
25H &+ 22H \\
&+ 3FH \\
&+ 52H \\
&+ E8H \\
\hline
1C0H \quad & (\text{dropping the carry, we get C0H, which means data is corrupted})
\end{align*}
\]
Example 7-27

Write a C18 program to calculate the checksum byte for the data given in Example 7-26.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char mydata[] = {0x25,0x62,0x3F,0x52};
    unsigned char sum = 0;
    unsigned char x;
    unsigned char checksumbyte;
    TRISB = 0;
    TRISC = 0;  //make Ports B and C output
    for(x=0; x<4; x++)
    {
        PORTC = mydata[x];  //issue each byte to PORTC
        sum = sum + mydata[x];  //add them together
        PORTB = sum;  //issue the sum to PORTB
    }
    checksumbyte = ~sum + 1;  //make 2’s complement (invert +1)
    PORTB = checksumbyte;  //show the checksum byte
}
```

Single-step the above program on the MPLAB simulator and examine the contents of PORTB and PORTC. Notice that each byte is put on PORTC as they are added.

Example 7-28

Write a C18 program to perform step (b) of Example 7-26. If the data is good, send ASCII character ‘G’ to PORTD. Otherwise, send ‘B’ to PORTD.

Solution:

```c
#include <P18F458.h>
void main(void)
{
    unsigned char mydata[] = {0x25,0x62,0x3F,0x52,0xE8};
    unsigned char checksum = 0;
    unsigned char x;
    TRISD = 0;  //make Port D an output
    for(x=0; x<5; x++)
    {
        checksum = checksum + mydata[x];  //add them together
    }
    if(checksum == 0)
        PORTD = 'G';
    else
        PORTD = 'B';
}
```

Change one or two values in the mydata array and simulate the program to see the results.
Binary (hex) to decimal and ASCII conversion in C18

The printf function is part of the standard I/O library in C and can do many things including converting data from binary (hex) to decimal, or vice versa. But printf takes a lot of memory space and increases your hex file substantially. For this reason, in systems based on the PIC18 microcontroller, it is better to write our own conversion function instead of using printf.

One of the most widely used conversions is binary to decimal conversion. In devices such as ADCs (Analog-to-Digital Converters), the data is provided to the microcontroller in binary. In some RTCs, the time and dates are also provided in binary. In order to display binary data, we need to convert it to decimal and then to ASCII. Because the hexadecimal format is a convenient way of representing binary data, we refer to the binary data as hex. The binary data 00–FFH converted to decimal will give us 000 to 255. One way to do that is to divide it by 10 and keep the remainder, as was shown in Chapters 5 and 6. For example, 11111101 or FDH is 253 in decimal. The following is one version of an algorithm for conversion of hex (binary) to decimal:

<table>
<thead>
<tr>
<th>Hex</th>
<th>Quotient</th>
<th>Remainder</th>
</tr>
</thead>
<tbody>
<tr>
<td>FD/0A</td>
<td>19</td>
<td>3 (low digit) LSD</td>
</tr>
<tr>
<td>19/0A</td>
<td>2</td>
<td>5 (middle digit)</td>
</tr>
<tr>
<td></td>
<td></td>
<td>2 (high digit) (MSD)</td>
</tr>
</tbody>
</table>

Example 7-29 shows the C program for that algorithm.

Example 7-29

Write a C18 program to convert 11111101 (FD hex) to decimal and display the digits on PORTB, PORTC, and PORTD.

Solution:

#include <P18F458.h>

void main(void)
{
    unsigned char x, binbyte, d1, d2, d3;
    TRISB = 0;
    TRISC = 0;
    TRISD = 0;      //Ports B, C, and D output
    binbyte = 0xFD; //binary (hex) byte
    x = binbyte / 10; //divide by 10
    d1 = binbyte % 10; //find remainder (LSD)
    d2 = x % 10;    //middle digit
    d3 = x / 10;    //most-significant digit (MSD)
    PORTB = d1;
    PORTC = d2;
    PORTD = d3;
}

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Review Questions

1. For the following decimal numbers, give the packed BCD and unpacked BCD representations:
   (a) 15   (b) 99

2. Show the binary and hex formats for “76” and its packed BCD version.

3. 67H in BCD when converted to ASCII is ____H and ____H.

4. Does the following convert unpacked BCD to ASCII?
   ```c
   mydata = 0x09 + 0x30;
   ```

5. Why is the use of packed BCD preferable to ASCII?

6. Which takes more memory space: packed BCD or ASCII?

7. In Question 6, which is more universal?

8. Find the checksum byte for the following values: 22H, 76H, 5FH, 8CH, 99H.

9. To test data integrity, we add the bytes together, including the checksum byte. The result must be equal to ____ if the data is not corrupted.

10. An ADC provides an output of 0010 0110. How do we display that on the screen?

SECTION 7.5: DATA SERIALIZATION IN C

Serializing data is a way of sending a byte of data one bit at a time through a single pin of a microcontroller. There are two ways to transfer a byte of data serially:

1. Using the serial port. In using the serial port, the programmer has very limited control over the sequence of data transfer. The details of serial port data transfer are discussed in Chapter 10.

2. The second method of serializing data is to transfer data one bit at a time and control the sequence of data and spaces between them. In many new generations of devices such as LCD, ADC, and EEPROM, the serial versions are becoming popular because they take less space on a printed circuit board. Although we can use standards such as I2C and CAN, not all devices support such standards. For this reason we need to be familiar with data serialization using the C language.

Examine the next four examples to see how data serialization is done in C.
Example 7-30

Write a C18 program to send out the value 44H serially one bit at a time via RC0. The LSB should go out first.

Solution:

//Serializing data via RC0 (SHIFTING RIGHT)
#include <P18F458.h>
define PC0 PORTCbits.RCO
void main(void)
{
    unsigned char conbyte = 0x44;
    unsigned char regALSB;
    unsigned char x;
    regALSB = conbyte;
    TRISCbits.TRISCO = 0; //make RC0 an output
    for(x=0;x<8;x++)
    {
        PC0 = regALSB & 0x01;
        regALSB = regALSB >> 1;
    }
}

Figure 7-11. Shifting Bits Out (LSB Going First)

Example 7-31

Write a C18 program to send out the value 44H serially one bit at a time via RC0. The MSB should go out first.

Solution:

//Serializing data via RC0 (SHIFTING LEFT)
#include <P18F458.h>
define PC0 PORTCbits.RCO
void main(void)
{
    unsigned char conbyte = 0x88;
    unsigned char regAMSB;
    unsigned char x;
    regAMSB = conbyte;
    TRISCbits.TRISCO = 0; //make RC0 an output
    for(x=0;x<8;x++)
    {
        PC0 = (regAMSB >> 7) & 0x01;
        regAMSB = regAMSB << 1;
    }
}
Example 7-32

Write a C18 program to bring in a byte of data serially one bit at a time via the RB0 pin. Place the byte on Port D. The LSB should come in first.

Solution:

```c
//Bringing in data via RB0 (SHIFTING RIGHT)
#include <P18F458.h>
define PBO PORTBbits.RB0
void main(void)
{
    unsigned char x;
    unsigned char REGA=0;
    TRISBbits.TRISB0 = 1;   //RB0 as input
    TRISD = 0;              //Port D as output
    for(x=0;x<8;x++)
    {
        REGA = REGA >> 1;
        REGA |= (PBO & 0x01) << 7;
    }
    PORTD = REGA;
}
```

Figure 7-12. Shifting Bits in (Bring in LSB First)

Example 7-33

Write a C18 program to bring in a byte of data serially one bit at a time via the RB0 pin. The MSB should come in first.

Solution:

```c
//Bringing in data via RB0 (SHIFTING LEFT)
#include <P18F458.h>
define PBO PORTBbits.RB0
void main(void)
{
    unsigned char x;
    unsigned char REGA=0;
    TRISBbits.TRISB0 = 1;   //RB0 as input
    TRISD = 0;              //Port D as output
    for(x=0;x<8;x++)
    {
        REGA = REGA << 1;
        REGA |= PBO & 0x01;
    }
    PORTD = REGA;
}
```
SECTION 7.6: PROGRAM ROM ALLOCATION IN C18

Using program (code) space for predefined fixed data is a widely used option in the PIC18, as we saw in Chapter 6. In that chapter we saw how to use Assembly language instructions to access the data stored in the program code space. In this chapter, we explore the same concept using the C18 C compiler. We will also examine the far and near storage qualifier for ROM.

RAM data space vs. code data space

In the PIC18 we have two spaces in which to store data. They are as follows:

1. The 4096 bytes of data RAM space with address range 000–FFFH. As we have seen in previous chapters, many PIC18 chips have much less than 4096 bytes for the file register data RAM. We also have seen how we can read (from) or write (into) this RAM space directly or indirectly.

2. The 2M of code (program) space with addresses of 000000–1FFFFFH. This 2M bytes of on-chip ROM space is used for storing programs (opcodes) and therefore is directly under control of the program counter (PC). As we have seen in the previous chapters, many PIC18 chips have much less than 2M of on-chip program ROM. We have also seen how to access the program ROM for the purpose of data storage using the TBLRD instruction (see Chapter 6). There is one problem with using this program code space for storage of fixed data: The more code space we use for data, the less is left for our program code. For example, if we have a PIC18 chip such as the PIC18F252 with only 4K of on-chip ROM, and we use 1K to store a look-up table, only 3K is left for the program. For some applications, this can be a problem. For this reason Microchip has added EEPROM memory to the PIC18 to be used for data storage. The EEPROM option of PIC18 is discussed in Chapter 14. Next, we will examine how the C18 compiler uses on-chip ROM space, and discuss how it places data into program ROM.
Table 7-5: Program ROM Size for Some PIC18F Family Members

<table>
<thead>
<tr>
<th>On-Chip Code ROM (Bytes)</th>
<th>Code Address Range (Hex)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F2220 4K</td>
<td>00000–00FFF</td>
</tr>
<tr>
<td>PIC18F2410 16K</td>
<td>00000–03FFF</td>
</tr>
<tr>
<td>PIC18F458/4580 32K</td>
<td>00000–07FFF</td>
</tr>
<tr>
<td>PIC18F6680 64K</td>
<td>00000–0FFFF</td>
</tr>
<tr>
<td>PIC18F8722 128K</td>
<td>00000–1FFFF</td>
</tr>
</tbody>
</table>

Allocating program space to data

In all our C18 examples so far, byte-size variables were stored in the data RAM. As we saw in Chapter 6, it is common practice to use the on-chip program ROM for the purpose of storing fixed data such as strings. This is specially useful since we have limited amount of file register data RAM. To make the C18 compiler use the program (code) ROM space for the fixed data, we use the keyword `rom` as shown in the following lines of C code:

```c
rom char mynum[] = "Hello"; //use code space for data
rom char weekdays = 7, month= 12; //use code space for data
```

The following code shows how to use program space for data in C18:

```c
//Program 7-1
#include <P18F458.h>
rom const char mynum[] = "0123456789"; //uses program

void main(void)
{
    unsigned char z;
    TRISB = 0; //make Port B an output
    for(z=0; z<10; z++)
        PORTB=mynum[z];
}
```

Figure 7-14. Fixed Data Placed in Program ROM as Shown in MPLAB
NEAR and FAR for code

As we have discussed earlier, the PIC18 microcontroller has a maximum of 2M of on-chip program ROM space. Not every family member, however, comes with that much on-chip program ROM. Some PIC18 chips come with as little as 4K and some come with 128K of program ROM. To make a more efficient use of the code space, the C18 compiler allows the use of near and far storage qualifiers to indicate in what region the data and code should be placed. The near qualifier is used to indicate that a program memory data variable is located in the first 64K of the program ROM. In order to indicate that a data variable in program ROM can be found anywhere in the 2M ROM space, we must use the far qualifier. See Table 7-6. Also see Program 7-2A. Note that the far storage qualifier is the default for the C18 if we do not specify it in our program.

Table 7-6: NEAR and FAR Usage for ROM

<table>
<thead>
<tr>
<th>Storage qualifier</th>
<th>ROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>near</td>
<td>In program space of 0000–FFFFFH (64 kB)</td>
</tr>
<tr>
<td>far</td>
<td>In program space of 000000–1FFFFFFH (2 MB)</td>
</tr>
</tbody>
</table>

//Program 7-2A
#include <P18F458.h>

near rom const char mydata[] = "HELLO"; //program ROM data
void main(void)
{
    unsigned char z;
    TRISB = 0; //make Port B an output
    for(z=0;z<5;z++)
        PORTB = mydata[z];
}

Figure 7-15. Using Near Storage Qualifier as Shown by MPLAB
In Program 7-2A, if we change the near to far and compile for the PIC18F8722 chip (which has 128K of program ROM), we have the following:

//Program 7-2B
#include <P18F8722.h>
    far rom const char mydata[] = "HELLO"; // program ROM data
void main(void)
{
    unsigned char z;
    TRISB = 0; // make Port B an output
    for(z=0;z<5;z++)
        PORTB = mydata[z];
}

Notice 5 digits for hex address (00000–FFFFF)

Figure 7-16. Far Storage Qualifier as Shown in MPLAB

Pragma and allocating a fixed address to data and code

As we saw in Chapter 6, the MPLAB assembler allows us to place data or code at a specific ROM address using the ORG directive. To do the same thing in the C18 C compiler we use the #pragma section directive, where “section” is a portion of an application (code or data) that can be assigned an specific memory address location. In the case of the on-chip ROM program memory, we have two options: (1) code, and (2) romdata. The #pragma directive is used for the program because it contains executable instructions, while the #pragma romdata directive is used for fixed data such as strings and look-up tables. Next we explore the use of #pragma to allocate ROM addresses for the code and data.

Putting code in a specific ROM address

To place the code (containing executable instructions) at a specific address location of the program ROM, we use the #pragma code directive. Examine the Program 7-3 to see how the C code for the MSDelay function is placed at the ROM address of 0x300.
//Program 7-3
#include <P18F458.h>
#pragma code main = 0x50    //place the main at ROM addr 0x50
void MSDLay(unsigned int);
void main(void)
{
    unsigned char mydata[] = "HELLO";
    unsigned char z;
    TRISB = 0;            //make Port B an output
    for(z=0;z<5;z++)
    {
        PORTB = mydata[z];
        MSDLay(250);
    }
}
#pragma code MSDLay = 0x300    //place delay at ROM addr 0x300
void MSDLay(unsigned int itime)
{
    unsigned int i;
    unsigned char j;
    for(i=0;i<itime;i++)
    for(j=0;j<165;j++);
}

Running the above programs on the MPLAB simulator and examining the program code space, we see that the main and MSDLay functions are located at the ROM addresses 0x50 and 0x300, respectively.

Figure 7-17. Screen Shot for Program 7-3
Putting data in a specific ROM address

To place the data (containing variables and constants, strings, and look-up tables) at a specific address of program ROM, we use the #pragma romdata directive. Examine the Program 7-4 to see how the C code assigns the program ROM address of 0x200 to the string “Hello”.

//Program 7-4
#include <P18F458.h>
#pragma romdata mydata = 0x200  //place mydata at ROM addr 0x200
near rom const char mydata[] = "HELLO";  //ROM data
void main(void)
{
    unsigned char z;
    TRISB = 0;  //make Port B an output
    for(z=0;z<5;z++)
        PORTB = mydata[z];
}

Run the above program on the MPLAB simulator. Examine the program code space to see the string “HELLO” located at the ROM address starting at 0x50.

Figure 7-18. Screen Shot for Program 7-4

Review Questions

1. The PIC18 family has a maximum of ____ of program ROM space.
2. The PIC18F8722 has ____ K of program ROM
3. True or false. The program (code) ROM space can be used for data storage, but the data space cannot be used for code.
4. True or false. Using the program ROM space for data means the data is fixed and static.
5. If we have a message string with a size of over 1000 bytes, then we use ______ (program ROM, data RAM) to store it.
SECTION 7.7: DATA RAM ALLOCATION IN C18

In this chapter, we explore the data RAM file register usage and its allocation by the C18 C compiler. We will also explore the near and far data storage qualifiers for data RAM. In addition, we will examine how to place a given data or stack at a fixed address using the C18 compiler.

RAM data space usage by the C18 C compiler

As we have seen in Chapters 2 through 6, PIC18 family members can have a maximum of 4K of data RAM, but not all the members come with 4K of RAM. As we have seen before, the data RAM can vary from 256 bytes to 4096 depending on the chip. That means all members of the PIC18 family come with at least one bank of RAM, which is called the access bank. In Assembly language programming, the 128 bytes of the data RAM are used for the SFRs and the remaining RAM is used for the scratch pad. The C18 compiler does the same thing by leaving the SFR region undisturbed and allocating the rest of the RAM to the stack and the variables declared by the C program. See Program 7-5.

```
//Program 7-5
#include <P18F458.h>
void main(void)
{
    unsigned char x=5, y=9;   //uses data RAM to store data
    unsigned char z;
    TRISB = 0;                //make Port B an output
    z = x + y;
    PORTB = z;
}
```

Running the above program on the MPLAB simulator, we can examine the data RAM space to locate x, y, and z as shown in Figure 7-19.

Figure 7-19. Screen Shot for Program 7-5
An array needs contiguous RAM locations for the array elements, which means the size of the array is limited to the size of data RAM in a given PIC18 chip. See Program 7-6 below.

//Program 7-6
#include <P18F458.h>
void main(void)
{
    unsigned char mynum[] = "0123456789";  //uses RAM space
    //to store data
    unsigned char z;
    TRISB = 0;  //make Port B an output
    for(z=0; z<10; z++)
        PORTB = mynum[z];
}

Running the above program on the MPLAB simulator and examining the data RAM space, we can locate values 30H, 31H, 32H, ..., 41H, 42H, 43H, 44H, and so on (the hex values for ASCII ‘0’, ‘1’, ‘2’, etc., as shown below).

![Figure 7-20. Screen Shot for Program 7-6](image)

In the case of arrays with a large number of elements, such as 100, examine the code below:

//Program 7-7
#include <P18F458.h>
void main(void)
{
    unsigned char mydata[100];  //100-byte space in RAM
    unsigned char x, z = 0xFF;
    TRISB = 0;  //make Port B an output
    for(x=0; x<100; x++)
    {
        mydata[x] = z;  //save it in RAM
        PORTB = z;  //give a copy to PORTB too
        z--;  //count down
    }
}

Running the above program on the MPLAB simulator, we can locate values FFH, FEH, FDH, and so on in the data RAM file register, as shown in Figure 7-21.
Change the size of the array and the targeted PIC18 chip (e.g., PIC18F252) and monitor the RAM space allocation by the C compiler.

**NEAR and FAR for data**

The C18 compiler has two storage qualifiers for data RAM allocation called near and far. The far and near qualifiers are used to indicate which sections of data RAM are to be used for the storage of declared variables. The keyword *near* will limit the C18 C compiler usage of RAM to the access bank for the data declaration, while the keyword *far* will put the entire data RAM at the disposal of C compilers. See Table 7-7. This means that programs written for the PIC18 chips with a limited data RAM cannot have too many arrays with a large number of elements. Using C18, compile and simulate Programs 7-8a and 7-8b to see the impact of *near* and *far* in RAM allocation. See Table 7-7.

**Table 7-7: NEAR and FAR Usage for Data RAM**

<table>
<thead>
<tr>
<th>Storage qualifier</th>
<th>RAM</th>
</tr>
</thead>
<tbody>
<tr>
<td>near</td>
<td>In access bank</td>
</tr>
<tr>
<td>far</td>
<td>Anywhere in data RAM file register (default)</td>
</tr>
</tbody>
</table>

//Program 7-8a
#include<P18F458.h>
near unsigned char mydata[100];  //100-byte space in RAM
void main(void)
{
    unsigned char x, z = 0;
    TRISB = 0;  //make Port B an output
    for(x=0; x<100; x++)
    {
        z--;  //count down
        mydata[x] = z;  //save it in RAM
        PORTB = z;  //give a copy to PORTB too
    }
}
//Program 7-8b
#include <P18F458.h>
void main(void)
{
    far unsigned char mydata[100];  //100-byte space in RAM
    unsigned char x, z = 0;
    TRISB = 0;  //make Port B an output
    for(x=0; x<100; x++)
    {
        z--;
        mydata[x] = z;  //save it in RAM
        PORTB = z;  //give a copy to PORTB too
    }
}

Putting data in a specific RAM address

As we saw in Chapter 6, the MPLAB assembler allows us to place data at a specific RAM address using a combination of MOVLW and MOVWF instructions. To place data at a specific data RAM address in the C18 C compiler, we use the #pragma directive. In the last section we examined the use of the #pragma directive to set the ROM memory address. The #pragma directive can also be used to set the data RAM address. There are two options for #pragma when it is used for the data RAM: (1) idata, and (2) udata, where idata stands for initialized data and udata stands for uninitialized data. The idata (initialized data) and udata (uninitialized data) options are widely used by the C18 to assign an explicit address in RAM data. For example, the following code uses idata to place the data string "HELLO" at RAM address starting at 0x150:

//Program 7-9 (using idata)
#include <P18F458.h>
#pragma idata mydata = 0x150
unsigned char mydata[] = "HELLO";  //RAM data
void main(void)
{
    unsigned char z;
    TRISB = 0;  //make Port B an output
    for(z=0; z<5; z++)
    {
        PORTB = mydata[z];
    }
}

We can verify the above concept by simulating the program on the MPLAB and examining the RAM at address 0x150.

Figure 7-22. Screen Shot for Program 7-9

CHAPTER 7: PIC PROGRAMMING IN C

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The following is a repeat of an earlier program and shows how to assign a fixed address of 0x200 using udata.

//Program 7-10 (using udata)
#include <P18F458.h>
#pragma udata mycount = 0x200 //assign RAM address 0x200
far unsigned char mycount[100]; //100-byte space in RAM
void main(void)
{
    unsigned char x, z = 0;
    TRISB = 0; //make Port B an output
    for (x = 0; x < 100; x++)
    {
        z--; //count down
        mycount[x] = z; //save it in RAM
        PORTB = z; //give a copy to PORTB too
    }
}

![File Registers](image)

Figure 7-23. Screen Shot for Program 7-10

The following program shows how to assign a fixed address using both udata and idata.

//Program 7-11 (assigning udata and idata to a fixed address)
#pragma idata x = 0x100 //assign fixed RAM address 0x100 to var x
unsigned char x = 5; //both data are initialized data
#pragma idata y = 0x101 //assign fixed RAM address 0x101 to var y
unsigned char y = 9; //both data are initialized data
#pragma udata z = 0x102 //assign fixed RAM address 0x102 to var z
unsigned char z; //it is uninitialized data
#include <P18F458.h>
void main(void)
{
    TRISB = 0; //make Port B an output
    z = x + y;
    PORTB = z;
}

Although assigning fixed addresses to a string of data in the data RAM can be justified, this practice is not recommended for individual variables, because it is the job of the compiler to assign addresses dynamically.
Overlay storage class

In an attempt to use the data space of the PIC18 more efficiently, the C18 compiler introduces the overlay storage class. The overlay conserves memory by allowing two variables to share the same physical address as long as they are not active at the same time. Compare the following two functions.

```c
unsigned char proga(void)
{
    overlay unsigned char x = 0;
    x = x + 1;
    return x;
}

and

unsigned char progb(void)
{
    overlay unsigned char y = 0;
    y = y + 2;
    return y;
}
```

Because the x and y variables are not active at the same time, the C18 C compiler uses the same physical address location in the file register for both of them. If we remove the keyword overlay in the above programs, the C18 will assign two different locations to the x and y variables. The C18 will also use two different physical locations for the variables when the variables are dependent on each other and are both active. Look at the following cases.

```c
unsigned char progc(void)
{
    overlay unsigned char x = 0;
    x = progd();
    return x;
}

and

unsigned char progd(void)
{
    overlay unsigned char y = 0;
    y = y + 2;
    return y;
}
```
In the previous programs, the C18 compiler assigned a separate RAM location to \( x \) and \( y \), even though we used the keyword *overlay*. Because \texttt{progc} calls function \texttt{progd}, the variables are dependent on each other and they are both active at the same item. Note that the C18 C compiler supports all the ANSI C standard storage classes such as auto, extern, static, and so on. The overlay is a new storage class and applies to local variables.

To gain a better understanding of this concept, examine Example 7-34. It shows three different versions of a program that sends the string “HELLO” to Port B. Simulate each program with the C18 compiler and compare data storage methods. Also, compile each program and compare the hex file size to see the impact of the data storage method on the hex file size.
Compare and contrast the following programs and discuss the advantages and disadvantages of each:

(a)  
```c
#include <P18F458.h>
void main(void) {
    TRISB = 0;  //make Port B an output
    PORTB = 'H';
    PORTB = 'E';
    PORTB = 'L';
    PORTB = 'O';
}
```

(b)  
```c
#include <P18F458.h>
void main(void) {
    unsigned char mydata[] = "HELLO";
    unsigned char z;
    TRISB = 0;  //make Port B an output
    for(z=0; z<5; z++)
        PORTB = mydata[z];
}
```

(c)  
```c
#include <P18F458.h>
void main(void) {
    rom unsigned char mydata[] = "HELLO";  //notice keyword rom
    unsigned char z;
    TRISB = 0;  //make Port B an output
    for(z=0; z<5; z++)
        PORTB = mydata[z];
}
```

Solution:

All the programs send out "HELLO" to PORTB one character at a time. They do the same thing in different ways. The first way is short and simple, but the individual characters are embedded into the program. If we change the characters, the whole program changes. This method also mixes the code and data together. The second one uses the RAM data space to store array elements; therefore, the size of the array is limited to file register size. The third one uses a separate area of the program code space for data. This allows the size of the array to be as big as you want provided that you have enough on-chip program ROM. The more program code space you use for data, however, the less space is left for your program code. Both the (b) and (c) programs are easily upgradable if we want to change the string itself or make it longer. That is not the case for program (a).
Review Questions

1. The PIC18 has a maximum of ____ of data RAM.
2. The PIC18F8722 has ____ of data RAM space.
3. True or false. The data space can be used for code.
4. Which space would you use to declare the following values for C18?
   (a) the number of days in a week
   (b) the number of months in a year
   (c) a counter for a delay
5. True or false. The near storage qualifier is used to place the variables in access RAM.

See the following web site for PIC18 C compilers:

http://www.microchip.com

The following web site has a tutorial for MPLAB and C18:

http://www.MicroDigitalEd.com

Running any of the C18 programs on the PIC18F hardware, the following points must be noted:

1. Disable the WatchDog Timer in the configuration bits.
2. Place “while(1);” at the end of the program to prevent the program from executing again. This plays the role of “HERE BRA HERE” in Assembly language.

SUMMARY

This chapter dealt with C18 programming, specifically I/O programming and time delays in C. We also showed the logic operators AND, OR, XOR, and complement. In addition, some applications for these operators were discussed. This chapter described BCD and ASCII formats and conversions in C. We also compared and contrasted the use of code space and RAM data space in C. The widely used technique of data serialization was also discussed.
PROBLEMS

SECTION 7.1: DATA TYPES AND TIME DELAYS IN C

1. Indicate what data type you would use for the following variables:
   (a) the temperature
   (b) the number of days in a week
   (c) the number of days in a year
   (d) the number of months in a year
   (e) the counter to keep the number of people getting on a bus
   (f) the counter to keep the number of people going to a class
   (g) an address of 64K RAM space
   (h) the age of a person
   (i) a string for a message to welcome people to a building

2. Give the hex value that is sent to the port for each of the following C statements:
   (a) PORTB=14;
   (b) PORTB=0x18;
   (c) PORTB='A';
   (d) PORTB=7;
   (e) PORTB=32;
   (f) PORTB=0x45;
   (g) PORTB=255;
   (h) PORTB=0xFF;

3. Give two factors that can affect time delay code size in the PIC18 microcontroller.

4. Of the two factors in Problem 3, which can be set by the system designer?

5. Can the programmer set the number of clock cycles used to execute an instruction? Explain your answer.

6. Explain why various C compilers produce different hex file sizes.

SECTION 7.2: I/O PROGRAMMING IN C

7. What is the difference between the PORTB.bits.RB4 and TRISBbits.TRISB4?

8. Write a C18 program to toggle all bits of PORTB every 200 ms.

9. Write a C18 program to toggle bits RB1 and RB7 every 200 ms.

10. Write a time delay function for 100 ms.

11. Write a C18 program to toggle only bit RB0 every 200 ms.

12. Write a C18 program to count up PORTB from 0–99 continuously.

SECTION 7.3: LOGIC OPERATIONS IN C

13. Indicate the data on the ports for each of the following:
    Note: The operations are independent of each other.
    (a) PORTB=0xF0&0x45;
    (b) PORTB=0xF0&0x56;
    (c) PORTB=0xF0&0x76;
    (d) PORTB=0x90;
    (e) PORTC=0xF0&0x90;
    (f) PORTC=0xF0&0x90;
    (g) PORTC=0xF0&0xF0;
    (h) PORTC=0xF0&0x90;
    (i) PORTC=0xF0&0xF0;
    (j) PORTC=0xF0&0x90;

14. Find the contents of the port after each of the following operations:
    (a) PORTB=0x65&0x76;
    (b) PORTB=0x70&0x6B;
    (c) PORTC=0x95&0xAA;
    (d) PORTC=0x5D&0x78;
    (e) PORTC=0xC5&0x12;
    (f) PORTD=0x6A&0x6E;
    (g) PORTB=0x37&0x26;
15. Find the port value after each of the following is executed:
   (a) PORTB=0x65>>2 ;  
   (b) PORTC=0x39<<2 ;  
   (c) PORTB=0xD4>>3 ;  
   (d) PORTB=0xA7<<2 ;

16. Show the C code to swap 0x95 to make it 0xA5.

17. Write a C program that finds the number of zeros in an 8-bit data item.

18. A stepper motor uses the following sequence of binary numbers to move the motor. How would you generate them in C18?
   1100, 0110, 0011, 1001

SECTION 7.4: DATA CONVERSION PROGRAMS IN C

19. Write a program to convert the following series of packed BCD numbers to ASCII. Assume that the packed BCD is located in data RAM.
   76H, 87H, 98H, 43H

20. Write a program to convert the following series of ASCII numbers to packed BCD. Assume that the ASCII data is located in data RAM.
   "8767"

21. Write a program to get an 8-bit binary number from PORTB, convert it to ASCII, and save the result if the input is packed BCD of 00–0xFF. Assume that PORTB has 1000 1001 binary as input.

SECTION 7.6: PROGRAM ROM ALLOCATION IN C18

22. Indicate what type of memory (data RAM or code ROM space) you would use for the following variables:
   (a) the temperature
   (b) the number of days in a week
   (c) the number of days in a year
   (d) the number of months in a year

23. True or false. When using program ROM for data, the total size of the array should not exceed 256.

24. Why do we use the ROM code space for video game characters and shapes?

25. What is the advantage of using program ROM space for data?

26. What is the drawback of using program ROM space for data?

27. Write a C18 program to send your first and last names to PORTC. Use the program ROM space for the data.

28. What is the difference between far and near storage?

29. What is the difference between #pragma code and #pragma romdata?

30. In Problem 27, show how to place the last name at ROM address 0x200 and the first name at address 0x220.

31. Indicate the size of the program ROM space for each of the following chips:
   (a) PIC18F452/4520  
   (b) PIC18F458/4580  
   (c) PIC18F8722

32. In Problem 31, discuss what impact the ROM space has on your decision on the amount of the memory to be allocated to data.
SECTION 7.7: DATA RAM ALLOCATION IN C

33. Indicate what type of memory (data RAM, or code ROM space) you would use for the following variables:
   (a) the counter to keep the number of people getting on a bus
   (b) the counter to keep the number of people going to a class
   (c) an address of 64K RAM space
   (d) the age of a person
   (e) a string for a message to welcome people to building

34. Indicate the size of the data RAM space for each of the following chips:
   (a) PIC18F452/4520   (b) PIC18F458/4580   (c) PIC18F8722

35. Discuss why the total size of an array should be limited to 256 bytes if possible.

36. Why will we not use the data RAM space for video game characters and shapes?

37. What is the drawback of using RAM data space for fixed data?

38. What is the advantage of using data RAM space for variables?

39. What is the difference between #pragma udata and #pragma idata?

40. In Problem 27, show how to place the names at RAM address 0x300.

41. Explain when we use overlay for variables.

42. True or false. Overlay is used for variables that are not active at the same time.

ANSWERS TO REVIEW QUESTIONS

SECTION 7.1: DATA TYPES AND TIME DELAY IN C

1. 0 to 255 for unsigned char and -128 to +127 for signed char
2. 0 to 65,535 for unsigned int and -32,768 to +32,767 for signed int
3. Unsigned char
4. True
5. (a) Crystal frequency of PIC18 system
   (b) PIC18 machine cycle timing
   (c) compiler used for C

SECTION 7.2: I/O PROGRAMMING IN C

1. F81H
2. void main()
   {
   // Code...
   
3. #define PB0bit PORTBbits.RB0
   void main()
   {
   // Code...
   
4. True
5. True
SECTION 7.3: LOGIC OPERATIONS IN C

1. (a) 02
   (b) FFH
   (c) FDH
2. Zeros
3. One
4. All zeros
5. 66H

SECTION 7.4: DATA CONVERSION PROGRAMS IN C

1. (a) 15H = 0001 0101 packed BCD, 0000 0001,0000 0101 unpacked BCD
   (b) 99H = 1001 1001 packed BCD, 0000 1001,0000 1001 unpacked BCD
2. 3736H = 00110111 00110110B
   and in packed BCD we have 76H = 0111 0110B
3. 36, 37
4. Yes, because mydata = 0x39.
5. Space savings
6. ASCII
7. BCD
8. E4H
9. 0
10. First, convert from binary to decimal, then convert to ASCII, and then send results to the screen and we will see 038.

SECTION 7.6: PROGRAM ROM ALLOCATION IN C18

1. 2M
2. 128
3. True
4. True
5. Program ROM

SECTION 7.7: DATA RAM ALLOCATION IN C18

1. 4K
2. 4096 bytes
3. False
4. (a) ROM space, (b) ROM space, (c) RAM space
5. True
CHAPTER 8

PIC18F HARDWARE CONNECTION AND ROM LOADERS

OBJECTIVES

Upon completion of this chapter, you will be able to:

» Explain the function of the reset pin of the PIC18F microcontroller
» Show the hardware connection of the PIC18F chip
» Show the use of a crystal oscillator for a clock source
» Explain how to design a PIC18F-based system
» Explain the role of brown-out reset voltage (BOR) in system reset
» Explain the role of the CONFIG registers in PIC18-based systems
» Show the design of the PIC Trainer
» Code a test program in Assembly and C for testing the PIC18
» Show how to download programs into the PIC18F system using Microchip PICkit 2
» Explain the Intel hex file characteristics for 32-bit and 16-bit addresses
This chapter describes the process of physically connecting and testing PIC18F-based systems. In the first section we describe the functions of PIC18F458 pins. The configuration registers of the PIC18 and how they are set are explored in Section 8.2. In Section 8.3 we explain the characteristics of Intel hex files that are produced by MPLAB. In Section 8.4 we discuss the various methods of loading a program into the microcontroller. It also shows the hardware connection for a PIC18 Trainer using the PIC18F452/458 (PIC18F4520/4580) chips.

SECTION 8.1: PIC18F458/452 PIN CONNECTION

The PIC18F458 family members come in different packages, such as DIP (dual in-line package), QFP (quad flat package), and LLC (leadless chip carrier). They all have many pins that are dedicated to various functions such as I/O, ADC, timer, and interrupts. Note that Microchip provides an 18-pin version of the PIC18 family with a reduced number of I/O ports for less demanding applications. Because the vast majority of developers use the 40-pin chip, however, we will concentrate on that. Figure 8-1 shows the pins for the PIC18F458.

![PIC18F458 Pin Diagram](image)
Examining Figure 8-1, note that of the 40 pins, a total of 33 are set aside for the five ports A, B, C, D, and E, with their alternate functions. The rest of the pins are designated as Vdd, GND (Vss), OSC1, OSC2, and MCLR (master clear reset). Next, we describe the function of each pin.

**Vdd (Vcc)**

Two pins are used to provide supply voltage to the chip. The typical voltage source is +5V. Some PIC18F family members have lower voltage for Vdd pins in order to reduce the noise and power dissipation of the PIC system. We can choose other options for the Vdd voltage level by setting the bits in the configuration register. The configuration register for Vdd is discussed in the next section.

**Vss (GND)**

Two pins are also used for ground. In chips with 40 pins and more, it is common to have multiple pins for VCC and GND. This will help reduce the noise (ground bounce) in high-frequency systems, as discussed in Appendix C.

**OSC1 and OSC2**

The PIC18F has many options for the clock source. Most often a quartz crystal oscillator is connected to input pins OSC1 and OSC2. The quartz crystal oscillator connected to the OSC1 and OSC2 pins also needs two capacitors. One side of each capacitor is connected to the ground as shown in Figure 8-3. Note that PIC18F microcontrollers can have speeds of 0 Hz to 40 MHz.

We can choose options for the clock frequency by setting bits in the configuration register. The config register for the oscillator is discussed in the next section.

**MCLR**

Pin 1 (in the PIC18F458 40-pin DIP) is the MCLR (master clear reset) pin. It is an input and is active-LOW (normally HIGH). When a LOW pulse is applied to this pin, the microcontroller will reset and terminate all activities. This is often referred to as a power-on reset (POR).

### Program counter value upon reset

Activating a MCLR reset will cause all values in the registers to be lost. Table 8-1 provides a partial list of PIC18F registers and their values after power-on reset. From Table 8-1 we note that the value of the PC (program counter) is 0 upon reset, forcing the CPU to fetch the first opcode from ROM memory location 00000. This means that we must place the first byte of opcode in ROM location 0 because that is where the CPU expects to find the first instruction.

<table>
<thead>
<tr>
<th>Table 8-1: RESET Values for Some PIC18 Registers</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Register</strong></td>
</tr>
<tr>
<td>PC</td>
</tr>
<tr>
<td>WREG</td>
</tr>
<tr>
<td>SP</td>
</tr>
<tr>
<td>TRISA–TRISE</td>
</tr>
</tbody>
</table>
Figures 8-2a and 8-2b show two ways of connecting the MCLR pin to the power-on reset circuitry. Figure 8-2b uses a momentary switch for reset circuitry. The most difficult time for any system is during the power-up. The CPU needs both a stable clock source and a stable voltage level to function properly. The PIC18 chips come with some features that help the reset process. We can choose these features by setting the bits in the configuration register. The configuration register for the reset pin is discussed in the next section. There are other sources of reset in the PIC18 family, and they are discussed in future chapters.

The pins discussed so far must be connected no matter which family member is used. They are the minimum pin connections that every PIC18 must have. See Figure 8-3.

![Figure 8-2a. PIC18F458 Power-On Reset Circuit](image1)

![Figure 8-2b. PIC18F458 Power-On Reset with a Momentary Switch](image2)

![Figure 8-3. Minimum Connection for PIC18F458](image3)
The number of I/O ports varies among the PIC18 family members, as we saw in Chapter 4. The following is another look at them for the PIC18F458.

**Ports A, B, C, D, and E**

As shown in Figure 8-1 (and discussed in Chapter 4), the ports PORTA, PORTB, PORTC, PORTD, and PORTE use a total of 33 pins. All the ports upon RESET are configured as input, because TRISA–TRISE have the value FFH on them. Tables 8-2 through 8-5 provide summaries of features of ports PORTA–PORTE and their alternative functions. We will study the alternative functions of these pins in future chapters, as we discuss the PIC18 features.

**Table 8-2: PORTA/PORTE Alternate Functions**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>RA0</td>
<td>AN0/CVREF</td>
</tr>
<tr>
<td>RA1</td>
<td>AN1</td>
</tr>
<tr>
<td>RA2</td>
<td>AN2/VREF−</td>
</tr>
<tr>
<td>RA3</td>
<td>AN3/VREF+</td>
</tr>
<tr>
<td>RA4</td>
<td>TOCKI</td>
</tr>
<tr>
<td>RA5</td>
<td>AN4/SS/LVDIN</td>
</tr>
<tr>
<td>RA6</td>
<td>OSC2/CLKO</td>
</tr>
<tr>
<td>RE0</td>
<td>AN5/RD</td>
</tr>
<tr>
<td>RE1</td>
<td>AN6/WR/C1OU</td>
</tr>
<tr>
<td>RE2</td>
<td>AN7/CS/C2OUT</td>
</tr>
</tbody>
</table>

**Table 8-3: PORTB Alternate Functions**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>RB0</td>
<td>INT0</td>
</tr>
<tr>
<td>RB1</td>
<td>INT1</td>
</tr>
<tr>
<td>RB2</td>
<td>INT2/CANTX</td>
</tr>
<tr>
<td>RB3</td>
<td>CANRX</td>
</tr>
<tr>
<td>RB4</td>
<td></td>
</tr>
<tr>
<td>RB5</td>
<td>PGM</td>
</tr>
<tr>
<td>RB6</td>
<td>PGC</td>
</tr>
<tr>
<td>RB7</td>
<td>PGD</td>
</tr>
</tbody>
</table>

**Table 8-4: PORTC Alternate Functions**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>RC0</td>
<td>TIOSO/TICKI</td>
</tr>
<tr>
<td>RC1</td>
<td>TIOSI</td>
</tr>
<tr>
<td>RC2</td>
<td>CCP1</td>
</tr>
<tr>
<td>RC3</td>
<td>SCK/SCL</td>
</tr>
<tr>
<td>RC4</td>
<td>SDI/SDA</td>
</tr>
<tr>
<td>RC5</td>
<td>SDO</td>
</tr>
<tr>
<td>RC6</td>
<td>TX/CK</td>
</tr>
<tr>
<td>RC7</td>
<td>RX/DT</td>
</tr>
</tbody>
</table>

**Table 8-5: PORTD Alternate Functions**

<table>
<thead>
<tr>
<th>Bit</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>RD0</td>
<td>PSP0/C1IN+</td>
</tr>
<tr>
<td>RD1</td>
<td>PSP1/C1IN−</td>
</tr>
<tr>
<td>RD2</td>
<td>PSP2/C2IN+</td>
</tr>
<tr>
<td>RD3</td>
<td>PSP3/C2IN−</td>
</tr>
<tr>
<td>RD4</td>
<td>PSP4/ECCP1/P1A</td>
</tr>
<tr>
<td>RD5</td>
<td>PSP5/P1B</td>
</tr>
<tr>
<td>RD6</td>
<td>PSP6/P1C</td>
</tr>
<tr>
<td>RD7</td>
<td>PSP7/P1D</td>
</tr>
</tbody>
</table>

**Review Questions**

1. Which pin is used to reset the PIC18F458 chip?
2. Upon power-up, the program counter (PC) has a value of ____.
3. Upon power-up, the PIC18F458 fetches the first opcode from ROM address location ________.
4. MCLR is an active-_______ (LOW, HIGH) pin.
5. How many Vdd and Gnd pins are in the PIC18F458 chip?
There are some features of the PIC18 that we can choose by programming the bits of the configuration registers. These features will reduce system cost by eliminating any need for external components. The configuration registers are located at the address starting at 300000H, as shown in Figure 8-4. Notice that the address 300000H is outside the range 000000–1FFFFFH, the address space belonging to program ROM. We write 8-bit values into the configuration register one byte at a time using the CONFIG directive in the source code, as we will see soon. In other words, we provide the values and register name in our application program and let the ROM programmer load them into the config register along with the application program itself. The configuration registers can be accessed from the user program using table reads and writes. In this section we examine some of the basic configuration registers such as reset, clock source, and Vdd voltage. MicroChip website provides the complete list of configuration registers for the PIC microcontrollers. For the configuration registers of a given member of the PIC18 family, see the “Configuration Register Settings Addendum” document on the Microchip web site. Table 8-6 gives a short description of the configuration registers. It must be noted that if a configuration register is incorrectly programmed, it can cause the system to fail. An example of this is changing the clock type connected to the microcontroller.

Table 8-6: PIC18F458 Configuration Registers

<table>
<thead>
<tr>
<th>Address (Hex)</th>
<th>Name</th>
<th>General Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>300001</td>
<td>CONFIG1H</td>
<td>Oscillator selection</td>
</tr>
<tr>
<td>300002</td>
<td>CONFIG2L</td>
<td>Brown out</td>
</tr>
<tr>
<td>300003</td>
<td>CONFIG2H</td>
<td>Watchdog enable</td>
</tr>
<tr>
<td>300006</td>
<td>CONFIG4L</td>
<td>Background debugger and ISCP</td>
</tr>
<tr>
<td>300008</td>
<td>CONFIG5L</td>
<td>Code protection</td>
</tr>
<tr>
<td>300009</td>
<td>CONFIG5H</td>
<td>EEPROM and boot block protection</td>
</tr>
<tr>
<td>3000A</td>
<td>CONFIG6L</td>
<td>Write protection</td>
</tr>
<tr>
<td>3000B</td>
<td>CONFIG6H</td>
<td>Write protection</td>
</tr>
<tr>
<td>3000C</td>
<td>CONFIG7L</td>
<td>Read protection</td>
</tr>
<tr>
<td>3000D</td>
<td>CONFIG7H</td>
<td>Boot block read protection</td>
</tr>
<tr>
<td>3FFFFE</td>
<td>DEVID1</td>
<td>Device ID and revision</td>
</tr>
<tr>
<td>3FFFFF</td>
<td>DEVID2</td>
<td>Device ID</td>
</tr>
</tbody>
</table>
CONFIG1H register and oscillator clock source

The CONFIG1H register is located at address 0x300001 and is set aside for the clock oscillator, as shown in Figure 8-5. The following is a description of the options for the CONFIG1H register.

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>R/P-1</th>
<th>U-0</th>
<th>U-0</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>OSCSEN</td>
<td>bit 0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

bit 7-6   **Unimplemented**: Read as '0'

bit 5    **OSCSEN**: Oscillator System Clock Switch Enable bit

1 = Oscillator system clock switch option is disabled (main oscillator is source)
0 = Oscillator system clock switch option is enabled (oscillator switching is enabled)

bit 4-3   **Unimplemented**: Read as '0'

bit 2-0    **FOSC2:FOSC0**: Oscillator Selection bits

111 = RC oscillator w/OSC2 configured as RA6
110 = HS oscillator with PLL enabled/clock frequency = (4 x Fosc)
101 = EC oscillator w/OSC2 configured as RA6
100 = EC oscillator w/OSC2 configured as divide-by-4 clock output
011 = RC oscillator
010 = HS oscillator
001 = XT oscillator
000 = LP oscillator

Legend:

- R = Readable bit
- P = Programmable bit
- U = Unimplemented bit, read as '0'
- -n = Value when device is unprogrammed
- u = Unchanged from programmed state

Figure 8-5. CONFIG1H Register for Frequency Selection

**FOSC2–FOSC0**

The three bits of FOSC2, FOSC1, and FOSC0 are used to select the clock frequency to the CPU. The default choice is RC (111), which uses the on-chip oscillator with the help of an externally connected resistor and capacitor. In this option, all we have to do is to connect the OSC1 pin to the RC circuit. The values of R and C determine the clock speed. Providing clock to the CPU in this manner leaves the OSC2 (bit 6 of PORTA) available to be used as an I/O pin. We can use option 101 (EC: external clock) and provide an external clock source to the pin OSC1 and let A6 be used as an I/O pin. We can do the same thing with option 100 while OSC2 provides us with an Osc/4 frequency. This Osc/4 clock can be used to synchronize all the system activities with the CPU. The most widely used option is to connect the OSC1 and OSC2 pins to a crystal (or ceramic) oscillator, as shown in Figure 8-6. There are four choices for the crystal oscillator option. They are PPLHS, HS, XT, and LP. The main difference among them is the frequency range as shown in Table 8-7. The LP (low power) option uses the lowest power...
while the highest power consumption belongs to the PPLHS (phase lock loop high speed) option. Notice that the higher the frequency, the more power is dissipated by the CPU, as discussed in Appendix C. We use the HS (high speed) option for many of the circuits discussed in this textbook. If we connect pins OSC1–OSC2 to a 10 MHz crystal oscillator and choose the PPLHS option, then the CPU works on 40 MHz because the PPLHS uses phase lock loop to quadruple the clock source provided to the CPU. The PLLHS also has the highest power dissipation. Notice that the RC option (111) is the cheapest while the LP option (000) has the lowest power dissipation.

![Figure 8-6a. OSC1–OSC2 Connection to Crystal Oscillator](image)

![Figure 8-6b. OSC Connection to an External Clock Source](image)

**OSCSEN**

The OSCSEN bit (D5) of CONFIG1H allows the CPU to switch to an internal clock source, which has a fixed frequency of 32 kHz. Switching the clock source from the external oscillator connected to the OSC1 and OSC2 pins to an internal 32 kHz clock source will reduce power dissipation to an absolute minimum in many systems running on battery power. Using this option, along with the LP option for the crystal frequency, can reduce CPU power consumption to the nanowatt range. Notice that this low-frequency 32 kHz clock source is in addition to the external clock source connected to the OSC1 and OSC2 pins. This secondary clock source of 32 kHz is independent of the OSC1–OSC2 clock source and will continue to provide the clock to the CPU in the event that the crystal frequency goes bad. In this textbook we disable this secondary clock source and use the OSC1–OSC2 oscillator as the main clock source.

<table>
<thead>
<tr>
<th>Osc choice</th>
<th>Crystal Freq</th>
<th>C1 range</th>
<th>C2 range</th>
</tr>
</thead>
<tbody>
<tr>
<td>LP</td>
<td>32 kHz</td>
<td>33 pF</td>
<td>33 pF</td>
</tr>
<tr>
<td>LP</td>
<td>200 kHz</td>
<td>15 pF</td>
<td>15 pF</td>
</tr>
<tr>
<td>XT</td>
<td>200 kHz</td>
<td>47–65 pF</td>
<td>47–65 pF</td>
</tr>
<tr>
<td>XT</td>
<td>1 MHz</td>
<td>15 pF</td>
<td>15 pF</td>
</tr>
<tr>
<td>XT</td>
<td>4 MHz</td>
<td>15 pF</td>
<td>15 pF</td>
</tr>
<tr>
<td>HS</td>
<td>4 MHz</td>
<td>15 pF</td>
<td>15 pF</td>
</tr>
<tr>
<td>HS</td>
<td>8 MHz</td>
<td>15–33 pF</td>
<td>15–33 pF</td>
</tr>
<tr>
<td>HS</td>
<td>20 MHz</td>
<td>15–33 pF</td>
<td>15–33 pF</td>
</tr>
<tr>
<td>HS</td>
<td>25 MHz</td>
<td>15–33 pF</td>
<td>15–33 pF</td>
</tr>
</tbody>
</table>

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**OSCillator frequency and instruction clock cycle**

We examined the instruction cycle time in Chapters 2 through 4 and showed how to create time delay subroutines. In PIC18 microcontrollers, the instruction cycle time is based on 1/4 of the clock source provided to the OSC pins. This is examined once more in Example 8-1.

**Example 8-1**

Find the instruction cycle time for the PIC18F458 chip with the following crystal oscillator connected to the OSC1 and OSC2 pins. The option for CONFIG1H is shown for each crystal speed.

(a) 4 MHz, XT  
(b) 10 MHz, HS  
(c) 20 MHz, HS

**Solution:**

All the options in CONFIG1H use 1/4 of the clock source for the instruction cycle time.

(a) 4 MHz/4 = 1 MHz and instruction cycle time is 1/1 MHz = 1 μs
(b) 10 MHz/4 = 2.5 MHz and instruction cycle time is 1/2.5 MHz = 0.4 μs = 400 ns
(c) 20 MHz/4 = 5 MHz and instruction cycle time is 1/5 MHz = 0.2 μs = 200 ns

If we use 10 MHz crystal oscillator speed and choose the HSPLL option (instead of HS), then the CPU has 40 MHz for the clock source. This means that the instruction cycle time is 1/10 MHz = 0.1 μs = 100 ns because 40 MHz/4 = 10 MHz.

---

**Table 8-8: CONFIG1H Options Using CONFIG Directive in MPLAB**

<table>
<thead>
<tr>
<th>Oscillator Selection</th>
<th>Selection</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>OSC = LP</td>
<td>LP</td>
<td>Low Power</td>
</tr>
<tr>
<td>OSC = XT</td>
<td>XT</td>
<td>Crystal</td>
</tr>
<tr>
<td>OSC = HS</td>
<td>HS</td>
<td>High Speed</td>
</tr>
<tr>
<td>OSC = RC</td>
<td>RC</td>
<td>Resistor/Capacitor</td>
</tr>
<tr>
<td>OSC = EC</td>
<td>EC, OSC2 as Clock Out</td>
<td>External Clock</td>
</tr>
<tr>
<td>OSC = ECIO</td>
<td>EC, OSC2 as RA6</td>
<td>External Clock</td>
</tr>
<tr>
<td>OSC = HSPLL</td>
<td>HS-PLL Enabled</td>
<td>High Speed Phase Lock Loop</td>
</tr>
<tr>
<td>OSC = RCIO</td>
<td>EC, OSC2 as RA6</td>
<td>External Clock</td>
</tr>
</tbody>
</table>

**Oscillator Switch Enable**

| OSCS = ON | Enabled |
| OSCS = OFF | Disabled |
CONFIG directive

Table 8-8 shows the syntax choices for the CONFIG1H byte supported by the MPLAB. Whenever we load an application into the PIC18 program ROM, we need to load the CONFIG bytes into the configuration registers as well. This is done by using the CONFIG directive in the source program. In the source code, we use the CONFIG directive to set the CONFIG1H values according to Table 8-8 as shown below:

```
CONFIG OSC = HS ; high-speed oscillator
CONFIG OSCS = OFF ; disable Osc switch
```

Or, we can combine them into a single statement, as follows:

```
CONFIG OSC = HS, OSCS = OFF ; oscillator, no Osc switch
```

CONFIG2L register and reset voltage

CONFIG2L is located at address 0x300002 and is set aside for the purpose of providing stable voltage and clock frequency during reset. See Figure 8-7. The most difficult time for a system is during power-up. The CPU needs both a stable clock source and a stable voltage level to function properly. Two internal timers help us achieve that: they are called the power-up timer (PWRT) and the oscillator start-up timer (OST). These two internal timers help to reduce the delay associated with the frequency and voltage sources during the power-up process. PWRT

```
```

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
<td>BORV1</td>
<td>BORV0</td>
<td>BOREN</td>
<td>PWRTEN</td>
</tr>
</tbody>
</table>

bit 7-4  Unimplemented: Read as '0'
bit 3-2  BORV1: BORV0: Brown-out Reset voltage bits
        11 = VBOR set to 2.0V
        10 = VBOR set to 2.7V
        01 = VBOR set to 4.2V
        00 = VBOR set to 4.5V
bit 1   BOREN: Brown-out Reset Enable bit
        1 = Brown-out Reset enabled
        0 = Brown-out Reset disabled
bit 0   PWRTEN: Power-up Timer Enable bit
        1 = PWRT disabled
        0 = PWRT enabled

Legend:
R = Readable bit     P = Programmable bit     U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed     u = Unchanged from programmed state

Figure 8-7. CONFIG2L Configuration Register for Reset Voltage

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provides a fixed delay during power-up, which keeps the CPU in the reset state until the power supply stabilizes. The OST timer does the same thing for the crystal oscillator. These two on-chip timers eliminate the need for external circuitry for voltage and frequency stabilization during the power-up. CONFIG2L allows us to set the voltage and frequency to keep the CPU in the reset state until both the clock and power supply are stable. Next, we discuss options for the bits for this important configuration register.

Table 8-9: CONFIG2L Selection for PIC18F458

<table>
<thead>
<tr>
<th>Brown-out Voltage</th>
<th>Power-up Timer</th>
<th>Brown-out Reset</th>
</tr>
</thead>
<tbody>
<tr>
<td>BORV = 45</td>
<td>PWRT = ON</td>
<td>BOR = ON</td>
</tr>
<tr>
<td>4.5 V</td>
<td>Enabled</td>
<td>Enabled</td>
</tr>
<tr>
<td>BORV = 42</td>
<td>PWRT = OFF</td>
<td>BOR = OFF</td>
</tr>
<tr>
<td>4.2 V</td>
<td>Disabled</td>
<td>Disabled</td>
</tr>
<tr>
<td>BORV = 27</td>
<td></td>
<td></td>
</tr>
<tr>
<td>2.7 V</td>
<td></td>
<td></td>
</tr>
<tr>
<td>BORV = 20</td>
<td></td>
<td></td>
</tr>
<tr>
<td>2.0 V</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**BORV1:BORV0**

Occasionally, the power source provided to the \( V_{CC} \) (\( V_{dd} \)) pin fluctuates, causing the CPU to malfunction. The PIC18 family has a provision for this which is called brown-out reset voltage. The brown-out reset voltage (BORV) bits in CONFIG2L allow us to set the minimum voltage for \( V_{dd} \). If it falls below that, the CPU will go into the reset state and stop all activities. This is needed because the voltage connected to the \( V_{dd} \) (\( V_{CC} \)) pins can be set according to the oscillator frequency connected to the OSC1 and OSC2 pins. At the high frequency of 40 MHz with \( V_{dd} = 5 \) V, we set BORV to 4.5 V. That means that if \( V_{dd} \) falls below the BORV of 4.5 V, the CPU will go into the reset state and stop execution of programs without losing any data in registers. For a low-power system with a frequency of 2 MHz and below, we can connect the \( V_{dd} \) to 2 V and set BORV to 1.8 V. In such a situation if \( V_{dd} \) falls below 1.8 V, the CPU will go into the reset state, and when \( V_{dd} \) rises above that level it will come out of reset and continue the program execution. Therefore, the BORV1:BORV0 bits of CONFIG2L will be set according to the \( V_{dd} \) voltage supplied to the \( V_{dd} \) pins and the oscillator frequency connected to the OSC1 and OSC2 pins. In this book, we set BORV = 4.5 V because \( V_{dd} = 5 \) V and the crystal oscillator is 10 MHz.

**BOREN**

This will enable the option BORV1:BORV0 discussed above.

**PWRTEN**

This bit will enable the power-up timer (PWRT). The PWRT provides a fixed delay during power-up, which keeps the CPU in the reset state until the power supply is stabilized.

Table 8-9 provides the syntax options for CONFIG2L as supported by the
MPLAB. We use the CONFIG directive to set the values according to Table 8-9 as shown below:

```plaintext
CONFIG BORV=45 ;for Vdd = 5 V, OSC = 10 MHz
CONFIG PWRT = ON ;use power-timer
CONFIG BOR=ON ;enable BORV option
```

Or, we can combine them into a single statement as follows:

```plaintext
CONFIG BORV = 45, PWRT = ON, BOR=ON
```

**CONFIG2H register and watchdog timer**

CONFIG2H is located at address 0x300003 and is set aside for the watchdog timer. In recent years, microcontrollers have come with a piece of hardware called a *watchdog timer*. We can use the watchdog timer to force the microcontroller into the known state of reset when the system is hung up or out of control due to execution of an incorrect sequence of codes. There are many uses for watchdog timers in embedded systems. One application is to use the watchdog timer to prevent a system from going into an infinite loop due to a software bug. Another application of the watchdog timer can be to catch events that cause the system to

<table>
<thead>
<tr>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
<th>R/P-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>bit 7</td>
<td>WDTPS2</td>
<td>WDTPS1</td>
<td>WDTPS0</td>
<td>WDTEN</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

bit 7-4 **Unimplemented**: Read as ‘0’

bit 3-2 **WDTPS2:WDTPS0**: Watchdog Timer Postscale Select bits

111 = 1:128
110 = 1:64
101 = 1:32
100 = 1:16
011 = 1:8
010 = 1:4
001 = 1:2
000 = 1:1

**Note:** The Watchdog Timer postscale select bits configuration used in the PIC 18FXXX devices has changed from the configuration used in the PIC 18CXXX devices.

bit 0 **WDTEN**: Watchdog Timer Enable bit

1 = WDT enabled
0 = WDT disabled (control is placed on the SWDTEN bit)

**Legend:**

R = Readable bit  
P = Programmable bit  
U = Unimplemented bit, read as ‘0’

-n = Value when device is unprogrammed  
u = Unchanged from programmed state

Figure 8-8. CONFIG2H Configuration Register for Watchdog Timer
These problems can happen due to corruption of the program ROM caused by a power surge, an electrically noisy environment, or inadvertent changes to the program counter. In such situations, the watchdog timer will force the system into a known state of reset, from which the system can recover. In some applications, the system can be put to sleep if there is no activity, thereby saving battery power. In such applications, one can use the watchdog timer to monitor the keyboard and, when there is activity on the keyboard, to awaken the system to process the information. Figure 8-8 shows the CONFIG2H register.

**WDTEN**

This bit will enable the watchdog timer.

**WDTPS2:WDTPS0**

The watchdog timer prescalar bits allow programming the WDT for up to 2 minutes. Appendix A discusses SLEEP instruction with the WDT.

For the applications in this textbook, we turn off the watchdog timer. We can turn off the watchdog timer in the MPLAB or use the CONFIG directive in the source code to set CONFIG2H values according to Table 8-10, as shown below:

```plaintext
CONFIG WDT = OFF
```

**CONFIG4L register and background debugger**

CONFIG4L is located at address 0x300006 and is set aside for the purpose of enabling the background debugger, among other things. See Figure 8-9. Table 8-11 shows the selection options for CONFIG4L. The following are its options.

**DEBUG**

If we connect the PIC18 system to an in-circuit debugger, then we lose the RB6 and RB7 pins of PORTB. By disabling the background debugger option in the CONFIG4L byte, we can use the RB6 and RB7 pins for general purpose I/O.

**STVREN**

The D0 bit is used for stack overflow. As we discussed in Chapter 3, the PIC18 has only 31 locations for the stack. By enabling the D0 bit we will cause the system go into a reset state if stack overflows (or underflows).

**LVP**

The D2 bit is set aside for the low-voltage in-circuit serial programming (ICSP)
via pin RB5. We can disable it and use the RB5 pin as an I/O.

Table 8-11 shows the CONFIG4L byte selection syntax used by the MPLAB assembler. For the applications in this textbook, we turn off all the options of debugger, LVP, and stack overflow using the CONFIG directive as follows:

\[
\text{CONFIG DEBUG = OFF, LVP = OFF, STVR = OFF}
\]

The first four CONFIG registers are the minimum number of registers that we need for any PIC18F452 or 458-based system. The rest of the CONFIG registers are dedicated to program and data protection. See the Microchip website.

<table>
<thead>
<tr>
<th>R/P-1</th>
<th>U-0</th>
<th>U-0</th>
<th>U-0</th>
<th>R/P-1</th>
<th>U-0</th>
<th>R/P-1</th>
</tr>
</thead>
<tbody>
<tr>
<td>DEBUG</td>
<td>-</td>
<td>-</td>
<td>-</td>
<td>LVP</td>
<td>-</td>
<td>STVREN</td>
</tr>
</tbody>
</table>

bit 7 \text{DEBUG}: Background Debugger Enable bit

1 = Background Debugger disabled. RB6 and RB7 configured as general purpose I/O
0 = Background Debugger enabled. RB6 and RB7 are dedicated to In-Circuit Debug.

bit 6-3 \text{Unimplemented}: Read as '0'

bit 2 \text{LVP}: Low-Voltage ICSP Enable bit

1 = Low-Voltage ICSP enabled
0 = Low-Voltage ICSP disabled

bit 1 \text{Unimplemented}: Read as '0'

bit 0 \text{STVREN}: Stack Full/Underflow Reset Enable bit

1 = Stack Full/Underflow will cause Reset
0 = Stack Full/Underflow will not cause Reset

Legend:

\begin{tabular}{l}
R = Readable bit \\
C = Clearable bit \\
U = Unimplemented bit, read as '0' \\
-n = Value when device is unprogrammed \\
u = Unchanged from programmed state
\end{tabular}

Figure 8-9. CONFIG4L Configuration Register for Background Debugger

**CONFIGURATION SETTINGS TO BE NOTED**

1. Note that each member of the PIC18 family has its own values for the configuration registers. They are provided in a document called “PIC18 CONFIGURATION SETTINGs ADDENDUM” and can be found at the following web site:


2. Microchip Corp. recommends using the CONFIG directive instead of \_CONFIG for the PIC18 family. Although the \_CONFIG directive (notice, \_CONFIG has two undersigns) works with the PIC18, it is not recommended. According to Microchip, we should not use both of them in the same program.
The LIST directive

The LIST directive is another component used in the source code for a program that we intend to burn into the PIC ROM. The LIST directive informs the MPLAB assembler about some of the options, such as the Intel hex file format, the radix for data format, the printout of the source code, and so on. Table 8-12 provides some of the major options of the LIST directive used in this textbook.

For an example of how to use the LIST directive, look at the following setting.

\[
\text{LIST } P=18F458, F=\text{INTHX32}, \text{ MM=OFF, } R=\text{HEX, } ST=\text{OFF} \quad X=\text{OFF}
\]

It must be noted that some of the options in Table 8-12 can be set by the MPLAB assembler itself. To ensure that they are set when we share source files, however, we use the LIST directive to set them.

<table>
<thead>
<tr>
<th>Option</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>B = nnn</td>
<td>Set tab space. (Default is 8.)</td>
</tr>
<tr>
<td>C = nnn</td>
<td>Set column width for the print-out. (Default is 132.)</td>
</tr>
<tr>
<td>F = format</td>
<td>Set the hex file output. The choices are INHX32, INHX8M, or INHX8S. Default is INHX8M. (See next section on this.)</td>
</tr>
<tr>
<td>MM = {ON/OFF}</td>
<td>Print memory map in list file. (Default is on.)</td>
</tr>
<tr>
<td>N = nnn</td>
<td>Set lines per page in the print-out. (Default is 60.)</td>
</tr>
<tr>
<td>P = type</td>
<td>Set microcontroller type. (Example: P = PIC18F458.)</td>
</tr>
<tr>
<td>R = radix</td>
<td>Set radix, the data format used throughout the source code. The options are hex, dec, and oct. (Default is hex.)</td>
</tr>
<tr>
<td>ST = {ON/OFF}</td>
<td>Print symbol table in list file. (Default is on.)</td>
</tr>
<tr>
<td>X = {ON/OFF}</td>
<td>Turn macro expansion on or off. (Default is on.)</td>
</tr>
</tbody>
</table>

Putting it all together

All the programs we showed in the first seven chapters were intended to be simulated. In order to create a ready-to-burn program, however, we must provide all the configuration register bytes, and set the desired options of the LIST directive in the source code before assembling and linking the program. By doing so, the hex output file provided by the MPLAB assembler can be burned into the program ROM of the PIC18 chip using a ROM burner. We can also send this hex file to anyone, knowing that it will work because all the configuration registers are already set. We can use the following skeleton source code for the programs that we intend to burn into ROM.
; skeleton of a PIC18 Assembly language program
LIST P=PIC18F458, F=INHX32, MM=OFF, N=0, ST=OFF, R=HEX
#include P18F458.INC
CONFIG OSC=HS, OSCS=OFF ; high-speed XTAL as clk src
CONFIG WDT=OFF ; disable watchdog timer

; Brown-out Reset Volt = 4.5 V and Power-up Timer is on
CONFIG BORV=45, PWRT=ON, BOR=ON

; no Background debug, no Reset if stack overflows
; and pin PB5 = I/O
CONFIG DEBUG=OFF, LVP=OFF, STVR=OFF

........
ORG 0
........
........
END

As an example, examine the following program. It will toggle all the bits of PORTB with some delay in between the “on” and “off” states

; Test Program 8-1: Toggling PORTB for the PIC18F458 and
; XTAL = 10 MHz
LIST P=PIC18F458, F=INHX32, N=0, ST=OFF, R=HEX
#include P18F458.INC
CONFIG OSC = HS, OSCS = OFF
CONFIG WDT = OFF
CONFIG BORV = 45, PWRT = ON, BOR = ON
CONFIG DEBUG = OFF, LVP = OFF, STVR = OFF

R1 EQU 0x07
R2 EQU 0x08
R3 EQU 0x09

ORG 0
CLRF TRISB ; make Port B an output port
MOVLW 0x55 ; WREG = 55h
MOVWF PORTB ; put 55h on port B pins
L3

MOVLW D'2'1
MOVWF R1
D1

MOVLW D'250'1
MOVWF R2
D2

MOVLW D'250'1
MOVWF R3
D3

NOP

DECF R3, F
BNZ D3

DECF R2, F
BNZ D2

DECF R1, F
BNZ D1
RETURN
END
Setting the CONFIG registers in the MPLAB C18 C compiler

In Chapter 7 we covered C programming of the PIC18F using the C18 C compiler. Those programs were intended to be simulated. To create a burnable C program, we must ensure that the configuration registers are set. One way to do that is to use #pragma. We can use the following skeleton for C18 code for the programs that we intend to burn into ROM.

; skeleton of a PIC18 C18 C language program
#pragma config OSC = HS, OSCS = OFF
#pragma config BORV = 45, PWRT = ON, BOR = ON
#pragma config WDT = OFF
#pragma config DEBUG = OFF, LVP = OFF, STVR = OFF

void main (void)
{

....... 

....... 

}

As an example, examine the following C18 program. It will toggle all the bits of PORTB with some delay between the “on” and “off” states.

; Test Program 8-2: Toggling PORTB for the PIC18F458 and
; XTAL = 10 MHz
#pragma config OSC = HS, OSCS = OFF
#pragma config PWRT = OFF, BOR = ON, BORV = 45
#pragma config WDT = OFF
#pragma config DEBUG = OFF, LVP = OFF, STVR = OFF
#include <P18F458.h>

void msdelay(unsigned int ms);
void main(void)
{
    TRISB = 0; // make Port B an output
    while(1)
    {
        PORTB = 0x55;
        msdelay(500);
        PORTB = 0xAA;
        msdelay(500);
    }
}

// this delay is for a 10 MHz clock
void msdelay(unsigned int ms) {
    unsigned int x;
    unsigned char z;
    for(x=0;x<ms;x++)
        for(z=0;z<165;z++);
}
Review Questions

1. A given PIC18F458-based system has a crystal frequency of 16 MHz with HS selected for the CONFIG1H. What is the instruction cycle time for the CPU?
2. Which address is used for the CONFIG1H register?
3. True or false. Upon power-up, both voltage and frequency are stable instantly.
4. The LP option for the OSC1–OSC2 frequency works for the frequency range of _____ to _____ kHz.
5. Which configuration register is used to disable the watchdog timer? What is its address?
6. True or false. Upon power-up, the power-up timer keeps the CPU in the reset state until the voltage source is stable.
7. True or false. The configuration registers are located within the ROM program address space.
8. True or false. The brown-out reset voltage (BORV) can be set at a lower voltage for a system with low crystal frequencies.
9. True or false. The higher the clock frequency for the system, the lower the power dissipation.
10. If we have the statement BORV = 42 in a given source code, what is the lowest $V_{dd}$ voltage level at which the CPU goes into the reset state?

SECTION 8.3: EXPLAINING THE INTEL HEX FILE FOR PIC18

Intel hex file is a widely used file format designed to standardize the loading (transferring) of executable machine code into a ROM chip. Therefore, the loaders that come with every ROM burner (programmer) support the Intel hex file format. In many Windows-based assemblers such as MPLAB, the Intel hex file is produced according to the settings you set. In the PIC MPLAB environment, the object file is fed into the linker program to produce the Intel hex file. The hex file is used by the loader of an EPROM programmer such as the PICkit 2 programmer to transfer (load) the file into the ROM chip. The MPLAB assembler can produce three types of Intel hex files. They are (a) INHX8M, (b) INHX32, and (c) INHX8S. See Table 8-13. In this section we will explain each one with some examples.

Table 8-13: Intel Hex File Formats Produced by MPLAB (See http://www.microchip.com)

<table>
<thead>
<tr>
<th>Format Name</th>
<th>Format Type</th>
<th>File Extension</th>
<th>Max. ROM Address</th>
</tr>
</thead>
<tbody>
<tr>
<td>Intel Hex format</td>
<td>INHX8M</td>
<td>.hex</td>
<td>16-bit address</td>
</tr>
<tr>
<td>Intel Hex 32 format</td>
<td>INHX32</td>
<td>.hex</td>
<td>32-bit address</td>
</tr>
<tr>
<td>Intel Split Hex</td>
<td>INHX8S</td>
<td>.hx1 and .hxh</td>
<td>16-bit address for each</td>
</tr>
</tbody>
</table>

Analyzing the Intel hex (INHX8M) file

We choose the Intel hex type of INHX8M, INHX32, or INHX8S by using the LIST directive or setting the options in the MPLAB assembler itself. If we do not choose one, the MPLAB assembler selects INHX32 by default. Next, we will
analyze the hex file belonging to the list file for INHX8M. The INHX8M file is produced by the MPLAB assembler by choosing the INHX8M option in the LIST directive (or setting the MPLAB assembler). The file has the .hex extension. The INHX8M is used for PIC chips with program ROM space of up to 64K in size. To get Intel hex files for chips with more than 64K of program ROM space, we must use the INHX32 option. Figure 8-10 shows the Intel hex file of INHX8M for the test program whose list file was given earlier. Notice that we have chosen INHX8M in the LIST directive. Since the ROM burner (loader) uses the hex file to download the opcode into ROM, the hex file must provide the following: (1) the number of bytes of information to be loaded, (2) the information itself, and (3) the starting address where the information must be placed. Each line of the hex file consists of six parts as follows:

:BBAAAAATTHHHHHHHHHHCC

The following describes each part:
1. "::" Each line starts with a colon.
2. BB, the count byte. This tells the loader how many bytes are in the line. BB can range from 00 to 16 (10 in hex).
3. AAAA is for the address. This is a 16-bit address for the INHX8M. The loader places the first byte of data into this memory address.
4. TT is for type. This field is either 00 or 01. If it is 00, it means that there are more lines to come after this line. If it is 01, it means that this is the last line and the loading should stop after this line.
5. HH.....H is the real information (data or code). There is a maximum of 16 bytes in this part. The loader places this information into successive memory locations of ROM. Because the PIC18 chips have 16-bit-wide program ROM space, the information in this field is presented as low byte followed by the high byte.

```
:100000000936A550E816E811E07EC00F0FCD7020E3C
:10001000076EFA0E86EFA0E96E000000009065F
:0C002000FCE10806F8E10706F4E112001C
:0300010022020ECA
:010006008079
:06000800FC00FE00F40E5
:00000001FF
```

Separating the fields, we get the following:

```
:BB AAAA TT HHHHHHHHHHHHHHHHHHHHHHCC
:10 0000 00 936A550E816E811B07EC00F0FCD7020E3C 10 0010 00 076EFA0E86EFA0E96E000000009065F 0C0200 00 FCE10806F8E10706F4E112001C :03 0001 00 22020E :01 0006 00 80 :06 0008 00 FC00FE00F40E5 00 0000 01
```

Figure 8-10. Intel Hex File Test Program with the INHX8M Option

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6. CC is a single byte. This last byte is the checksum byte of everything in that line. The checksum byte is used for error checking. Checksum bytes are discussed in detail in Chapters 6 and 7. Notice that the checksum byte at the end of each line represents the checksum byte for everything in that line and not just for the data portion.

Now, compare the data portion of the Intel hex file in Figure 8-10 with the information under the OBJ field of the .lst file in Figure 8-11. Notice that they are

<table>
<thead>
<tr>
<th>LOC</th>
<th>OBJ</th>
<th>LINE</th>
</tr>
</thead>
<tbody>
<tr>
<td>00003</td>
<td>LIST P= PIC18F458, F= INHX8M, N=0, ST= OFF, R= HEX</td>
<td></td>
</tr>
<tr>
<td>00004</td>
<td>#include P18F458.INC</td>
<td></td>
</tr>
<tr>
<td>00005</td>
<td>CONFIG OSC= HS, OSCS= OFF</td>
<td></td>
</tr>
<tr>
<td>00006</td>
<td>CONFIG BORV= 45, PWRT= ON, BOR= ON</td>
<td></td>
</tr>
<tr>
<td>00007</td>
<td>CONFIG WDT= OFF</td>
<td></td>
</tr>
<tr>
<td>00008</td>
<td>CONFIG DEBUG= OFF, LVP= OFF, STVR= OFF</td>
<td></td>
</tr>
<tr>
<td>00009</td>
<td></td>
<td></td>
</tr>
<tr>
<td>00010</td>
<td>R1 EQU 0x07</td>
<td></td>
</tr>
<tr>
<td>00011</td>
<td>R2 EQU 0x08</td>
<td></td>
</tr>
<tr>
<td>00012</td>
<td>R3 EQU 0x09</td>
<td></td>
</tr>
<tr>
<td>00014</td>
<td>ORG 0</td>
<td></td>
</tr>
<tr>
<td>00015</td>
<td>CLR TRISB</td>
<td></td>
</tr>
<tr>
<td>00016</td>
<td>MOV LW 0x55</td>
<td></td>
</tr>
<tr>
<td>00017</td>
<td>MOVWF PORTB</td>
<td></td>
</tr>
<tr>
<td>00018</td>
<td>L3 MOVF PORTB, F</td>
<td></td>
</tr>
<tr>
<td>00019</td>
<td>CALL QDELAY</td>
<td></td>
</tr>
<tr>
<td>00020</td>
<td>BRA L3</td>
<td></td>
</tr>
<tr>
<td>00023</td>
<td>;-----------1/4 SECOND DELAY</td>
<td></td>
</tr>
<tr>
<td>00024</td>
<td>QDELAY</td>
<td></td>
</tr>
<tr>
<td>00025</td>
<td>MOV LW D'2'</td>
<td></td>
</tr>
<tr>
<td>00026</td>
<td>MOVWF R1</td>
<td></td>
</tr>
<tr>
<td>00027</td>
<td>D1 MOV LW D'250'</td>
<td></td>
</tr>
<tr>
<td>00028</td>
<td>MOVWF R2</td>
<td></td>
</tr>
<tr>
<td>00029</td>
<td>D2 MOV LW D'250'</td>
<td></td>
</tr>
<tr>
<td>00030</td>
<td>MOVWF R3</td>
<td></td>
</tr>
<tr>
<td>00031</td>
<td>D3 NOP</td>
<td></td>
</tr>
<tr>
<td>00032</td>
<td>NOP</td>
<td></td>
</tr>
<tr>
<td>00033</td>
<td>DECF R3, F</td>
<td></td>
</tr>
<tr>
<td>00034</td>
<td>BNZ D3</td>
<td></td>
</tr>
<tr>
<td>00035</td>
<td>DECF R2, F</td>
<td></td>
</tr>
<tr>
<td>00036</td>
<td>BNZ D2</td>
<td></td>
</tr>
<tr>
<td>00037</td>
<td>DECF R1, F</td>
<td></td>
</tr>
<tr>
<td>00038</td>
<td>BNZ D1</td>
<td></td>
</tr>
<tr>
<td>00039</td>
<td>RETURN</td>
<td></td>
</tr>
<tr>
<td>00040</td>
<td>END</td>
<td></td>
</tr>
</tbody>
</table>

Figure 8-11. List File for Test Program with the INHX8M Option
(Comments and other lines are deleted for space and simplicity.)
identical, as they should be. The extra information is added by the Intel hex file format. You can run the C language version of the test program and verify its operation. Your C compiler will provide you both the .lst file and Intel hex file if you want to explore the Intel hex file concept.

Examine the next three examples to gain insight into the Intel hex file.

**Example 8-2**

From Figure 8-10, analyze the six parts of line 3.

**Solution:**

After the colon (:), we have 0C, which means that 12 bytes of data are in this line. 0020H is the address at which the data starts. Next, 00 means that this is not the last line of the record. Then the data, which is 12 bytes, is as follows: FCE10806F8E10706F4EH200. Finally, the last byte, 1C, is the checksum byte.

**Example 8-3**

Compare the data portion of the Intel hex file of Figure 8-10 with the opcodes in the list file of the test program given in Figure 8-11. Do they match?

**Solution:**

In the first line of Figure 8-10, the data portion starts with 936AH, where the low byte is followed by the high byte. That means it is 6A93, the opcode for the instruction "CLRF TRISB", as shown in the list file of Figure 8-11. The last byte of the data in line 3 is 1200, which is the opcode for the "RETURN" instruction in the list file.

**Example 8-4**

Verify the checksum byte for line 3 of Figure 8-10. Verify also that the information is not corrupted.

**Solution:**

\[0C + 20 + FC + E1 + 08 + 06 + F8 + E1 + 07 + 06 + F4 + E1 + 12 + 00 = 5E4 \text{ in hex. Dropping the carries (5) gives E4H, and its 2's complement is 1CH, which is}} \]

\[\text{the last byte of line 4. If we add all the information in line 4, including the checksum byte, and drop the carries we should get } 0C + 20 + FC + E1 + 08 + 06 + F8 + E1 + 07 + 06 + F4 + E1 + 12 + 00 = 600H.\]
Analyzing the Intel hex file of INHX32

For PIC chips with program ROM space of more than 64K, we must choose the INHX32 option. Figure 8-13 shows the Intel hex file for the test program (Figure 8-14) assembled with the INHX32 option instead of INHX8M. Notice that INHX8M is used for chips with a ROM size of 64K or less, while ROM chips with more than 64K use INHX32. The 32-bit space in the INHX32 is for the address. That means that the INHX32 can be used for ROM chips with address space of 1 byte to 4 gigabytes, because $2^{32} = 4$ gigabytes, as shown in Figure 8-12.

Notice that the MPLAB produces a file with the .hex extension for the INHX32, just like the INHX8M. The INHX32 is very similar to the INHX8M, except that the TT field has extra options to accommodate the 32-bit address of the ROM chip. As with the INHX8M, each line of the hex file consists of six parts.

:BBAAATTHHHH....HHHHCC

The following describes each part.

1. “:” Each line starts with a colon.
2. BB, the count byte. This tells the loader how many bytes are in the line. BB can range from 00 to 16 (10 in hex).
3. AAAA is a A15–A0 address, or all zeros, depending on the TT status.
4. TT is for the record type. This field has four possibilities as follows:
   00 = Data record is in field HHHH
   01 = End of file record. It means this is the last line and the loading should stop after this line.
   02 = Segment address record.
   04 = Linear address record in the HHHH field (A31–A16 portion of A31–A0 is given in the HHHH field). Notice that if TT = 04, then the HHHH has the upper addresses of A31–A16 and AAAA = 0000.
5. HHH....H is the real information (data, address, or code). There is a maximum of 16 bytes in this part. The loader places this information into successive memory locations of ROM. Because the PIC18 chips have 16-bit ROM, the information in this field is presented as low byte followed by the high byte.
6. CC is a single byte. This last byte is the checksum byte of everything in that line. The checksum byte is used for error checking. Checksum bytes were discussed in detail in Chapters 6 and 7. Notice that the checksum byte at the end of each line represents the checksum byte for everything in that line and not just for the data portion.
Figure 8-13. Intel Hex (INHX32) File Test Program as Provided by the Assembler

Example 8-5

From Figure 8-13, analyze the six parts of (a) line 3, and (b) line 4.

Solution:

(a) In line 3, after the colon (:), we have 02 which means that 2 bytes of data are in this line. The AAAA = 0000 and TT = 04 mean that the upper 16 bits of the address are provided by the HHHH field. That is exactly what we see where the 16-bit address of 000128FOH is given in the HHHH field, which is 0001.

(b) In line 4, after the colon (:), we have 10H (which is 16 in decimal) as the number of bytes in this line. The AAAA = 28F0 is the lower 16-bit address where information will be burned. Next, 00 means that this is not the last line of the record. Then the data, which is 16 bytes, is as follows: 020E076EFA0E086EFA0E096E0000000056. Finally, the last byte, 56, is the checksum byte.
Figure 8-14. List File for Test Program with INHX32 Option (Notice ORG address for QDELAY. Some of the comments and lines are deleted for clarity.)

Intel hex split file of INHX8S

The INHX8S option is called the Intel split hex format. When we choose the INHX8S option in the LIST directive, we get two files: .hxl for the low byte and .hxh for the high byte. Because the PIC18 ROM is 16 bits wide, we have even addresses for the low byte and odd addresses for the high byte, as shown in Figure 8-15. The MPLAB assembler gives us this option because in many PIC18 chips with external memory we need to split the memory into odd and even banks.
to create a 16-bit wide ROM space. Note that ROM chips have pins D0–D7; therefore, their organization is Nkx8 (e.g., 64kx8). That means we must burn the hex file into the even-bank ROM and the .hxh file into the ROM with odd addresses. Note that the .hxl and .hxh formats are the same as the INHX8M with the address space limited to 64K for each bank. That means that with split ROM, we can have a maximum of 128K of ROM with 64K for each bank.

![Figure 8-15. The Odd and Even Banks for External Memory of PIC18](image)

**Review Questions**

1. True or false. The Intel hex file does not use the checksum byte method to ensure data integrity.
2. The first byte of a line in the Intel hex file represents ____.
3. The last byte of a line in the Intel hex file represents ____.
4. In the TT field of an Intel hex file, we have 00. What does it indicate?
5. Find the checksum byte for the following values: 22H, 76H, 5FH, 8CH, 99H.
6. In Question 5, add all the values and the checksum byte. What do you get?
7. True or false. In the TT field of the INHX32 file, we have 04, indicating the record is in the upper 16 bits of the 32-bit address.

**SECTION 8.4: PIC18 TRAINER DESIGN AND LOADING**

In this section, we discuss the connection for a simple PIC18-based trainer. We also show various ways of loading a hex file into the PIC microcontroller. Microchip has skillfully designed their microcontrollers for maximum flexibility of loading programs. The three primary ways to load a program are:

1. A device burner loads the program into the microcontroller separate from the system. This is useful on a manufacturing floor where a gang programmer is used to program many chips at one time. Most mainstream device burners support the PIC families: Advin and EETools are two of the more popular companies. Microchip supplies programmers for all their products; the PICkit 2 and PICSTART PLUS are just two examples. See Microchip’s website for a complete list. You can also build your own device programmer. Doing this will
reduce the cost of purchasing a commercial programmer, which is usually expensive. Building a programmer from scratch is beyond the scope of this text; check the Internet for sites devoted to this.

The device programming method is straightforward: The chip is programmed before it is inserted into the circuit. Or, the chip can be removed and reprogrammed if it is in a socket. A ZIF (zero insertion force) socket is even quicker and less damaging than a standard socket. When removing and reinserting, we must observe ESD (electrostatic discharge) procedures. Although PIC devices are rugged, there is always a risk when handling them. Using this method allows all of the device's resources to be utilized in the design. No pins are shared, nor are internal resources of the chip used as in the other two methods. This allows the embedded designer to use the minimum board space for the design.

2. An in-circuit serial programmer (ICSP) allows the developer to program and debug their microcontroller while it is in the system. This is done by two wires with a system setup to accept this configuration. The Microchip ICD 2 is a wonderful device for debugging programs. This method also allows the manufacturer to install the devices unprogrammed on the board. Before shipping to customers, the microcontroller can be programmed with the most recent file.

In-circuit serial programming is excellent for designs that change or require periodic updating. The ICSP uses two pins, RB7 and RB6. These pins can be used as I/O after the device is programmed. The designer must make sure that these pins do not conflict with the programmer. MCLR also needs a 10 kΩ pull-up resistor for the ICSP. The ICD 2 also needs V_{dd} and Gnd. The designer must bring the pins to a header on the board so that the programmer can connect to it. Figure 8-16 shows the pin connections. The designer must weigh the pros and cons of these methods.

---

**Figure 8-16. ICSP Connections**
3. A boot loader is a piece of code burned into the microcontroller’s program ROM. Its purpose is to communicate with the user’s board to load the program. A boot loader can be written to communicate via a serial port, CAN port, USB port, or even a network connection. A boot loader can also be designed to debug a system, similar to the ICD. This method of programming is excellent for the developer who does not always have a device programmer or an ICD available. Microchip has several application notes on writing boot loaders on their website. The main drawback of the boot loader is that it does require a communication port and program code space on the microcontroller. Also, the boot loader has to be programmed into the device before it can be used, usually by one of the two previous ways.

The boot loader method is ideal for the developer who needs to quickly program and test code. This method also allows the update of devices in the field without the need of ICD tools. All one needs is a computer with a port that is compatible with the board. (The serial port is one of the most commonly used and discussed, but a CAN or USB boot loader can also be written.) This method also consumes the largest amount of resources. Code space must be reserved and protected, and external devices are needed to connect and communicate with the PC. Developing projects using this method really helps programmers test their code. For mature designs that do not change, the other two methods are better suited.

Next, we discuss the issues related to the ROM loader for PIC18F-based systems using chips such as the PIC18F458/4580 and PIC18F452/4520. We will also provide guidelines for design of a simple PIC18 Trainer. If you decide to wire-wrap one of these, make sure that you read Appendix B on wire wrapping.

**PIC18F452/458-based Trainer**

In systems based on a PIC18-type microcontroller, you need a ROM burner to burn your program into the microcontroller. For the PIC18F, the ROM burner can erase the Flash ROM in addition to burning a program into it. In the case of the PIC18C, you also need an EPROM erasure tool because it uses UV-EPROM. Before burning the PIC18C, you need to erase its contents, which takes approximately 20 minutes for UV-EPROM. For the PIC18F, this is not required because it has Flash ROM.

**PIC18 Flash ROM size**

While all PIC18 chips share the same features, they come with different amounts of on-chip ROM. Table 8-14 shows the on-chip ROM size for various

<table>
<thead>
<tr>
<th>PIC18F Code ROM (Bytes)</th>
<th>Code Address Range (Hex)</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F2220 4K</td>
<td>00000–00FFF</td>
</tr>
<tr>
<td>PIC18F2410 16K</td>
<td>00000–03FFF</td>
</tr>
<tr>
<td>PIC18F458/4580 32K</td>
<td>00000–07FFF</td>
</tr>
<tr>
<td>PIC18F6680 64K</td>
<td>00000–0FFF</td>
</tr>
<tr>
<td>PIC18F8722 128K</td>
<td>00000–1FFFF</td>
</tr>
</tbody>
</table>

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PIC18 chips. Refer to the web site http://www.microchip.com for further information. Notice that while the PIC18F2220 comes with 4K of on-chip ROM, and the PIC18F2410 comes with 16K, the PIC18F458 has 32K of on-chip ROM. Also notice that the PIC18F458 is a substitute for the PIC18F452 with extra functions such as controlled area network (CAN).

Example 8-6

Find the ROM memory address of each of the following PIC chips:
(a) PIC18F2220 with 4 KB
(b) PIC18F2410 with 16 KB
(c) PIC18F458/4580 with 32 KB

Solution:

(a) With 4K of on-chip ROM memory space, we have 4096 bytes (4 \times 1024 = 4096). This maps to address locations of 0000 to 0FFFH. Notice that 0 is always the first location.

(b) With 16K of on-chip ROM memory space, we have 16,384 bytes (16 \times 1024 = 16,384), which gives 0000–3FFFH.

(c) With 32K we have 32,768 bytes (32 \times 1024 = 32,768). Converting 32,768 to hex, we get 8000H; therefore, the memory space is 0000 to 7FFFH.

PIC18 Trainer connection

We selected the PIC18F458 for a PIC18-based Trainer because it allows you to easily wirewrap an inexpensive but powerful trainer to be used at work and home. Figure 8-17 shows the connection for the PIC18F-based system to be used with the PICkit 2 programmer.

![PIC18F Connection to PICkit 2 with 6-Pin Header](image)

Figure 8-17. PIC18F Connection to PICkit 2 with 6-Pin Header

*Note: This connection using the PICkit 2 header applies to all families of PIC microcontrollers. The only differences are the pin number and designation.*
The PICkit 2 is an inexpensive programmer available from the Microchip website. The www.MicroDigitalEd.com website shows the schematic for PIC18-based Trainer connection.

**Downloading to the PIC18 Trainer**

After we build our PIC18-based system, we can download the program into the Trainer using the PICkit 2’s programmer utility. See Figure 8-18. Microchip is continuously updating MPLAB IDE to support PICkit 2 for programming of all PIC microcontrollers.

![PICkit 2 Microcontroller Programmer](image)

**Figure 8-18. PICkit 2 Programmer Utility**

**Test program for the PIC18 in Assembly and C**

To test your PIC18 hardware connection, we can run a simple test in which all the bits of PORTB toggle continuously with some delay between the “on” and “off” states. See Programs 8-3 and 8-3C. Notice in these programs that the time delay is based on a 10 MHz crystal. In developing your program, you can use the program shells provided in Figures 8-19 and 8-20.
Trainer Test Program in Assembly

;Program 8-3
LIST P=PIC18F458, F=INHX32, N=0, ST=OFF, R=HEX
#include P18F458.INC
CONFIG OSC = HS, OSCS = OFF
CONFIG WDT = OFF
CONFIG BORV = 45, PWRT = ON, BOR = ON
CONFIG DEBUG = OFF, LVP = OFF, STVR = OFF

R1 EQU 0x07
R2 EQU 0x08
R3 EQU 0x09

ORG 0000H ;note starting address
CLRF TRISB ;make Port B an output port
MOVLW 0x55 ;WREG = 55h
MOVWF PORTB ;put 55h on port B pins
L3 COMF PORTB,F ;toggle bits of Port B
CALL QDELAY ;quarter of a second delay
BRA L3 ;continue

;-------------1/4 SECOND DELAY
QDELAY
    MOVLW D'2'
    MOVWF R1
    MOVLW D'250'
    MOVWF R2
    MOVLW D'250'
    MOVWF R3
    NOP
    NOP
    DECF R3, F
    BNZ D3
    DECF R2, F
    BNZ D2
    DECF R1, F
    BNZ D1
    RETURN
END

#include P18F458.INC
CONFIG OSC = HS, OSCS = OFF
CONFIG WDT = OFF
CONFIG BORV = 45, PWRT = ON, BOR = ON
CONFIG DEBUG = OFF, LVP = OFF, STVR = OFF

ORG 0000H ;start of user code space
;begin user code

END ;end of user code

Figure 8-19: Shell of a Simple Assembly Language Program for MPLAB
Notice that the LIST directive is not used because it is one of the default settings in the
MPLAB IDE.

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Trainer Test Program in C

;Test Program 8-3C: Toggling PORTB for the PIC18F458/4580
;(452/4520) with XTAL = 10 MHz

#pragma config OSC = HS, OSCS = OFF
#pragma config PWRT = OFF, BOR = ON, BORV = 45
#pragma config WDT = OFF, LVP = OFF
#pragma config DEBUG = OFF, STVR = OFF

#include <P18F458.h>

void msdelay(unsigned int ms);
void main(void)
{
   TRISB = 0; //make Port B an output
   while(1)
   {
      PORTB = 0x55;
      msdelay(500);
      PORTB = 0xAA;
      msdelay(500);
   }
}

//this is for a 10 MHz clock
void msdelay(unsigned int ms)
{
   unsigned int x;
   unsigned char z;
   for(x=0;x<ms;x++)
      for(z=0;z<165;z++);
}

#pragma config OSC = HS, OSCS = OFF
#pragma config PWRT = OFF, BOR = ON, BORV = 45
#pragma config WDT = OFF, LVP = OFF
#pragma config DEBUG = OFF, STVR = OFF

#include <P18F458.h>
void main(void)
{
}

Figure 8-20: Shell of a Simple C Language Program for MPLAB
Some troubleshooting tips

Running the test program on your PIC18F458-based trainer (or PIC18F452 system) should toggle all the I/O bits with some delay. If your wire-wrapped system does not work, follow these steps to find the problem:

1. With the power off, check your connection for all pins, especially Vdd and GND.
2. Check MCLR (pin 1) using an oscilloscope. When the system is powered up, pin 1 is HIGH. Upon pressing the momentary switch, it goes LOW. Make sure the momentary switch is connected properly.
3. Observe the OSC1 pin on the oscilloscope while the power is on. You should see a crude sine wave. This indicates that the crystal oscillator is operating.
4. If all the above steps pass inspection, check the content of the on-chip ROM. It must be the same as the opcodes provided by the .lst file. Your assembler produces the .lst file, which lists the opcodes and operands on the left side of the assembly instructions. This must match exactly the contents of your on-chip ROM if the proper steps were taken in burning and loading the program into the on-chip ROM.

Review Questions

1. Which method(s) to program the PIC microcontroller is/are the best for manufacturing of large-scale boards?
2. Which method(s) allow(s) for debugging a system?
3. Which method(s) would allow a small company to develop a prototype and test an embedded system for a variety of customers?
4. True or false. The PIC18C has Flash program ROM.
5. Which pin is used for reset in the PIC18F458/4580?
6. What is the status of the reset pin when it is not activated?
7. What kind of ROM is used in the PIC18F458/4580 chip?
8. True or false. The PIC18 can download the file into its ROM only if it is in Intel hex file format.
9. Give two reasons that the PIC18F is preferable over PIC18C chips.

See the following website for the PIC18 Trainer:

http://www.MicroDigitalEd.com
SUMMARY

This chapter began by describing the function of each pin of the PIC18F458. The CONFIG registers of the PIC18F458 were discussed. These CONFIG registers are located at address 300001H and beyond, which is out of the program ROM address range. They are burned into the PIC chip along with the application. We use CONFIG registers to enable features such as low power frequency and watchdog timer. We also explained the Intel hex file formats INHX8M and INHX32. We examined how the INHX32 format uses the 32-bit address of ROM while INHX8M files are used for 16-bit addresses. Then the design of the PIC18-based trainer was shown.

PROBLEMS

SECTION 8.1: PIC18F458/452 PIN CONNECTION

1. The PIC18F458 DIP package is a(n) ____-pin package.
2. Which pins are assigned to $V_{CC}$ and GND?
3. In the PIC18F458, how many pins are designated as I/O port pins?
4. The crystal oscillator is connected to pins ____ and ____ .
5. If PIC18F458 is rated as 40 MHz, what is the maximum frequency that can be connected to it?
6. Indicate the pin number assigned to MCLR in the DIP package.
7. MCLR stands for ________.
8. The MCLR pin is normally _____ (LOW, HIGH) and needs a ______ (LOW, HIGH) signal to be activated.
9. What are the contents of the PC (program counter) upon reset of the PIC18F458?
10. What are the contents of the SP register upon reset of the PIC18F458?
11. What are the contents of the WREG register upon reset of the PIC18F458?
12. What are the contents of the TRIS registers upon reset of the PIC18F458?
13. In PIC18F458, how many pins are set aside for the $V_{dd}$?
14. In PIC18F458, how many pins are set aside for the $V_{ss}$ (Gnd)?
15. Which of the OSC pins are shared with the PORTA bit?
16. OSC1 and OSC2 are ________ (input, output) pins.
17. MCLR is an ________ (input, output) pin.
18. How many pins are designated as PORTA and what are those in the DIP package?
19. How many pins are designated as PORTB and what are those in the DIP package?
20. How many pins are designated as PORTC and what are those in the DIP package?
21. How many pins are designated as PORTD and what are those in the DIP package?
22. Upon reset, all the bits of ports are configured as _____ (input, output).
23. In the PIC18F458, which port has only 3 pins?
24. Which I/O pin of the PIC18F458 does not have an alternate function and can be used solely for I/O?

SECTION 8.2: PIC18 CONFIGURATION REGISTERS

25. True or false. For the PIC18F in reset state, the CPU does not execute any code.
26. True or false. When the system is powered up, the power-up timer (PWRT) and oscillator start-up timer (OST) keep the PIC18 in the reset state until the voltage and frequency are stable.
27. True or false. The power-up timer (PWRT) and oscillator start-up timer (OST) are components that we must add to the PIC18 externally.
28. True or false. The watchdog timer is a component that we must add to the PIC18 externally.
29. True or false. If we do not provide CONFIG values in our source code, PIC18 uses the default values for them.
30. True or false. The CONFIG registers use the same address space as program ROM.
31. Give the ROM address locations for CONFIG1H, CONFIG2L, CONFIG2H, and CONFIG4L.
32. The CONFIG registers are _____ bits wide.
33. Which CONFIG register is used to set the clock frequency for the PIC18F458?
34. Which CONFIG register is used to set the brown-out reset voltage for the PIC18F458?
35. Which CONFIG register is used to disable the watchdog timer for the PIC18F458?
36. If the brown-out reset voltage is set to 4.2 V, what does it mean to the system?
37. Show the CONFIG directive for a PIC18F458 system with the following options:
   (a) OSC1–OSC2 is connected to 20 MHz and it is the only source of the clock for the system.
   (b) The brown-out voltage is set for 4.2 V and the power-up timer is enabled.
   (c) No watchdog timer
   (d) No overflow on stack, no background debugger, and no LVP
38. For CONFIG1H, which option for the OSC frequency provides the lowest power dissipation?
39. Which CONFIG register is used to set the clock source for the PIC18F458?
40. Find the instruction cycle for the following crystal frequencies connected to OSC1 and OSC2. Assume that the HS option is chosen for all of them.
   (a) 12 MHz   (b) 20 MHz   (c) 25 MHz   (d) 30 MHz

SECTION 8.3: EXPLAINING THE INTEL HEX FILE FOR PIC18

41. True or false. The INHX32 option can be set by MPLAB without using the LIST directive.
42. True or false. The INHX32 option can be used for ROM sizes of more than 64 kilobytes.
43. True or false. The INHX8M option can be used for ROM sizes of more than 64 kilobytes.
44. True or false. The INHX8M option can be used for ROM sizes of less than 64 kilobytes.
45. True or false. The INHX32 option can be used for ROM of any size.
46. Analyze the six parts of line 1 of Figure 8-10.
47. Verify the checksum byte for line 1 of Figure 8-10. Verify also that the information is not corrupted.
48. Verify the checksum byte for line 2 of Figure 8-13. Verify also that the information is not corrupted.
49. What is the difference between the INHX8M and INHX32 hex files?
50. Analyze the INHX32 Intel hex file in Figure 8-13.

SECTION 8.4: PIC18 TRAINER DESIGN AND LOADING OPTIONS

51. True or false. Using the PICkit2, we must remove the PIC18F chip from the system and place it into the programmer.
52. True or false. The PICkit2 can only work with Flash chips.
53. Which of the following choices is the cheapest?
   (a) MPLAB ICD2    (b) PICkit2
54. Write a program to get 8-bit data from PORTB and send it to ports PORTC and PORTD.
55. Write a program to get 8-bit data from PORTD and send it to ports PORTB and PORTC.
56. Which pins of PORTB are PGD (program data) and PGC (program clock)?
57. At what program memory location does the PIC18F458 wake up upon reset? What is the implication of that?
58. Write a program to toggle all the bits of PORTB continuously
   (a) using AAH and 55H    (b) using the COMF instruction.
59. What is the address of the last location of program ROM for the PIC18F458?
60. What is the address of the last location of program ROM for the PIC18F8722?
61. What is the address of the last location of program ROM for the PIC18F452?

ANSWERS TO REVIEW QUESTIONS

SECTION 8.1: PIC18F458/452 PIN CONNECTION

1. 1
2. 000000
3. 000000
4. LOW
5. Two pins for Vdd and 2 pins for Gnd

SECTION 8.2: PIC18 CONFIGURATION REGISTERS

1. 16 MHz/4 = 4 MHz and 1/4 MHz = 250 ns
2. 300001 hex
SECTION 8.3: EXPLAINING THE INTEL HEX FILE FOR PIC18

1. False
2. The number of bytes of data in the line
3. The checksum byte of all the bytes in that line
4. 00 means this is not the last line and that more lines of data follow.
5. \(22H + 76H + 5FH + 8CH + 99H = 21CH\). Dropping the carries we have 1CH and its 2's complement, which is E4H.
6. \(22H + 76H + 5FH + 8CH + 99H + E4H = 300H\). Dropping the carries we have 00, which means that the data is not corrupted.
7. True

SECTION 8.4: PIC18 TRAINER DESIGN AND LOADING OPTIONS

1. Device burner
2. In-circuit serial debugger
3. ICSP
4. False
5. Pin 1
6. HIGH
7. Flash
8. True
9. It can be used with ICSP and has a faster development time.
CHAPTER 9

PIC18 TIMER
PROGRAMMING
IN ASSEMBLY AND C

OBJECTIVES

Upon completion of this chapter, you will be able to:

>> List the timers of the PIC18 and their associated registers
>> Describe the various modes of the PIC18 timers
>> Program the PIC18 timers in Assembly and C to generate time delays
>> Program the PIC18 counters in Assembly and C as event counters
The PIC18 has two to five timers depending on the family member. They are referred to as Timers 0, 1, 2, 3, and 4. They can be used either as timers to generate a time delay or as counters to count events happening outside the microcontroller. In Section 9.1 we see how Timers 0 and 1 are used to generate time delays. In Section 9.2 we show how they are used as event counters. In Section 9.3 we use C language to program the PIC18 timers. Timers 2 and 3 are discussed in Section 9.4.

SECTION 9.1: PROGRAMMING Timers 0 And 1

Every timer needs a clock pulse to tick. The clock source can be internal or external. If we use the internal clock source, then 1/4th of the frequency of the crystal oscillator on the OSC1 and OSC2 pins (Fosc/4) is fed into the timer. Therefore, it is used for time delay generation and for that reason is called a timer. By choosing the external clock option, we feed pulses through one of the PIC18’s pins: this is called a counter. In this section we discuss the PIC18 timer and in the next section we program the timer as a counter.

Basic registers of the timer

Many of the PIC18 timers are 16 bits wide. Because the PIC18 has an 8-bit architecture, each 16-bit timer is accessed as two separate registers of low byte (TMRxL) and high byte (TMRxH). Each timer also has the TCON (timer control) register for setting modes of operation. Next, we discuss each timer separately.

Timer0 registers and programming

Timer0 can be used as an 8-bit or a 16-bit timer. The 16-bit register of Timer0 is accessed as low byte and high byte, as shown in Figure 9-1. The low-byte register is called TMROH (Timer0 low byte) and the high-byte register is referred to as TMROH (Timer0 high byte). These registers can be accessed like any other special function registers. For example, the instruction “MOVWF TMROL” moves the value in WREG into TMRO, the low byte of Timer0. These registers can also be read like any other register. For example, “MOVFF TMRO, PORTB” copies TMROH (low byte of Timer0) to PORTB.

![Figure 9-1. Timer 0 High and Low Registers](image)

T0CON (Timer0 control) register

Each timer has a control register, called TCON, to set the various timer operation modes. T0CON is an 8-bit register used for control of Timer0. The bits for T0CON are shown in Figure 9-2.
TMR0ON | T08BIT | T0CS | T0SE | PSA | T0PS2 | T0PS1 | T0PS0
---|---|---|---|---|---|---|---
TMR0ON | D7 | Timer0 ON and OFF control bit
1 = Enable (start) Timer0
0 = Stop Timer0
T08BIT | D6 | Timer0 8-bit/16-bit selector bit
1 = Timer0 is configured as an 8-bit timer/counter.
0 = Timer0 is configured as a 16-bit timer/counter.
T0CS | D5 | Timer0 clock source select bit
1 = External clock from RA4/T0CKI pin
0 = Internal clock (Fosc/4 from XTAL oscillator)
T0SE | D4 | Timer0 source edge select bit
1 = Increment on H-to-L transition on T0CKI pin
0 = Increment on L-to-H transition on T0CKI pin
PSA | D3 | Timer0 prescaler assignment bit
1 = Timer0 clock input bypasses prescaler.
0 = Timer0 clock input comes from prescaler output.
T0PS2:T0PS0 | D2D1D0 | Timer0 prescaler selector
0 0 0 = 1:2 | Prescale value (Fosc / 4 / 2)
0 0 1 = 1:4 | Prescale value (Fosc / 4 / 4)
0 1 0 = 1:8 | Prescale value (Fosc / 4 / 8)
0 1 1 = 1:16 | Prescale value (Fosc / 4 / 16)
1 0 0 = 1:32 | Prescale value (Fosc / 4 / 32)
1 0 1 = 1:64 | Prescale value (Fosc / 4 / 64)
1 1 0 = 1:128 | Prescale value (Fosc / 4 / 128)
1 1 1 = 1:256 | Prescale value (Fosc / 4 / 256)

Figure 9-2. T0CON (Timer0 Control) Register

**T0CS (Timer0 clock source)**

This bit in the T0CON register is used to decide whether the clock source is internal (Fosc/4) or external. If T0CS = 0, then the Fosc/4 is used as clock source. In this case, the timers are often used for time delay generation. See Example 9-1. If T0CS = 1, the clock source is external and comes from the RA4/T0CKI, which is pin 6 on the DIP package of PIC1818F4580/4520. When the clock source comes from an external source, the timer is used as an event counter. We will discuss that option in the next section. See Example 9-2.

**Example 9-1**

Find the value for T0CON if we want to program Timer0 in 16-bit mode, no prescaler. Use PIC18’s Fosc/4 crystal oscillator for the clock source, increment on positive-edge.

**Solution:**

T0CON = 0000 1000 16-bit, Fosc/4 clock source, no prescaler, Timer0 off
Example 9-2

Find the timer’s clock frequency and its period for various PIC18-based systems, with the following crystal frequencies. Assume that no prescaler is used.

(a) 10 MHz  (b) 16 MHz  (c) 4 MHz

Solution:

(a) \( \frac{1}{4} \times 10 \text{ MHz} = 2.5 \text{ MHz} \) and \( T = \frac{1}{2.5} \text{ MHz} = 0.4 \text{ \mu s} \)

(b) \( \frac{1}{4} \times 16 \text{ MHz} = 4 \text{ MHz} \) and \( T = \frac{1}{4} \text{ MHz} = 0.25 \text{ \mu s} \)

(c) \( \frac{1}{4} \times 4 \text{ MHz} = 1 \text{ MHz} \) and \( T = \frac{1}{1} \text{ MHz} = 1 \text{ \mu s} \)

NOTE: PIC18 TIMERS USE 1/4 OF THE CRYSTAL FREQUENCY, IN ADDITION TO PRESCALER.

TMR0IF flag bit

Notice that the TMR0IF bit (Timer0 interrupt flag) is part of the INTCON (interrupt control) register. See Figure 9-3. The other options of the INTCON register are discussed in Chapter 11. As we will see, when the timer reaches its maximum value of FFFFH, it rolls over to 0000, and TMR0IF is set to 1 (see Figure 9-4). Chapter 11 shows how we can use TMR0IF to cause an interrupt. Next, we describe the 16-bit mode operation for Timer0.

The importance of TMR0IF: In 16-bit mode, when TMR0H:TMR0L overflows from FFFF to 0000 this flag is raised. In 8-bit, it is raised when the timer goes from FF to 00. We monitor this flag bit before we reload the TMR0H:TMR0L registers.

The other bits of this register are discussed in Chapter 11.

Figure 9-3. INTCON (Interrupt Control Register) has the TMR0IF Flag

Figure 9-4. Timer0 Overflow Flag
16-bit timer programming

The following are the characteristics and operations of 16-bit mode:

1. It is a 16-bit timer; therefore, it allows values of 0000 to FFFFH to be loaded into the registers TMR0H and TMR0L.

2. After TMR0H and TMR0L are loaded with a 16-bit initial value, the timer must be started. This is done by “BSF TOCON, TMROON” for Timer0.

3. After the timer is started, it starts to count up. It counts up until it reaches its limit of FFFFH. When it rolls over from FFFFH to 0000, it sets HIGH a flag bit called TMR0IF (timer interrupt flag, which is part of the INTCON register). This timer flag can be monitored. When this timer flag is raised, one option would be to stop the timer.

4. After the timer reaches its limit and rolls over, in order to repeat the process, the registers TMR0H and TMR0L must be reloaded with the original value, and the TMR0IF flag must be reset to 0 for the next round.

Steps to program Timer0 in 16-bit mode

To generate a time delay using the Timer0 mode 16, the following steps are taken:

1. Load the value into the TOCON register indicating which mode (8-bit or 16-bit) is to be used and the selected prescaler option.

2. Load register TMR0H followed by register TMR0L with initial count values.

3. Start the timer with the instruction “BSF TOCON, TMROON”.

4. Keep monitoring the timer flag (TMR0IF) to see if it is raised. Get out of the loop when TMR0IF becomes high.

5. Stop the timer with the instruction “BCF TOCON, TMROON”.

6. Clear the TMR0IF flag for the next round.

7. Go back to Step 2 to load TMR0H and TMR0L again.

To clarify the above steps, see Example 9-3. To calculate the exact time delay and the square wave frequency generated on pin PB5, we need to know the XTAL frequency. See Example 9-4 and Example 9-5.

Notice in Figure 9-5 that we should load TMR0H first, and then load TMR0L, because the value for TMR0H is kept in a temporary register and written to TMR0H when TMR0L is loaded. This will prevent any error in counting if the TMROON flag is set HIGH.

---

**Figure 9-5. Timer0 16-bit Block Diagram**

**Note:**
1. T0CS, T0SE, PSA, T0PS2, T0PS0 (TOCON<5:0>),
2. Upon reset, Timer 0 is enabled in 8-bit mode, with clock input from TOCKI, max. prescale.
Example 9-3

In the following program, we are creating a square wave of 50% duty cycle (with equal portions high and low) on the PORTB.5 bit. Timer0 is used to generate the time delay. Analyze the program.

```plaintext
BCF TRISB,5 ; PB5 as an output
MOVLW 0x08 ; Timer0, 16-bit, int clk, no prescale
MOVF T0CON ; load T0CON reg.
HERE MOVLW 0xFF ; TMR0H = FFH, the high byte
MOVWF TMROH ; load Timer0 high byte
MOVLW 0xF2 ; TMR0L = F2H, the low byte
MOVWF TMROL ; load Timer0 low byte
BCF INTCON, TMROIF ; clear timer interrupt flag bit
BTG PORTB,5 ; toggle PB5
BSF TOCON, TMROON ; start Timer0
AGAIN BTFSS INTCON, TMROIF ; monitor Timer0 flag until
BRA AGAIN ; it rolls over
BCF TOCON, TMROON ; stop Timer0
BRA HERE ; load TH, TL again
```

Solution:

In the above program notice the following steps:

1. TOCON is loaded.
2. FFF2H is loaded into TMR0H–TMRO.L.
3. The Timer0 interrupt flag is cleared by the “BCF INTCON, TMROIF” instruction.
4. PORTB.5 is toggled for the high and low portions of the pulse.
5. Timer0 is started by the “BSF TOCON, TMROON” instruction.
6. Timer0 counts up with the passing of each clock, which is provided by the crystal oscillator. As the timer counts up, it goes through the states of FFF3, FFF4, FFF5, FFF6, FFF7, FFF8, FFF9, FFFA, FFFF, and so on until it reaches FFFFH. One more clock rolls it to 0, raising the Timer0 flag (TMROIF = 1). At that point, the “BTFSS INTCON, TMROIF” instruction bypasses the “BRA AGAIN” instruction.
7. Timer0 is stopped by the instruction “BCF TOCON, TMROON”, and the process is repeated.

Notice that to repeat the process, we must reload the TMRO.L and TMROH registers, and start the timer again.
In Example 9-3, calculate the amount of time delay generated by the timer. Assume that XTAL = 10 MHz.

Solution:

The timer works with the Fosc/4 clock; therefore, we have 10 MHz / 4 = 2.5 MHz as the timer frequency. As a result, each clock has a period of $T = 1 / 2.5$ MHz = 0.4 $\mu$s. In other words, Timer0 counts up each 0.4 $\mu$s resulting in delay = number of counts $\times$ 0.4 $\mu$s.

The number of counts for the rollover is FFFFH − FFF2H = 0DH (13 decimal). However, we add one to 13 because of the extra clock needed when it rolls over from FFFF to 0 and raises the TMR0IF flag. This gives $14 \times 0.4 \mu$s = 5.6 $\mu$s for half the pulse. For the entire period the time delay generated by the timer is $T = 2 \times 5.6 \mu$s = 11.2 $\mu$s.

Example 9-5

Calculate the frequency of the square wave generated on pin PORTB.5.

Solution:

To get a more accurate timing, we need to add clock cycles due to the instructions in the loop.

```assembly
BCF TRISB,5
MOVLW 0x08
MOVWF TOCON
BCF INTCON, TMR0IF

HERE MOVLW 0xFF 1
MOVLW TMROH 1
MOVLW -D'48' 1
MOVLW TMROL 1
CALL DELAY 1
BTG PORTB,5 1
BRA HERE 1

;-------delay using Timer0
DELAY BSF TOCON, TMR0ON 1
AGAIN BTFSS INTCON, TMR0IF 1
BRA AGAIN 1
BCF TOCON, TMR0ON 1
BCF INTCON, TMR0IF 1
RETURN 1

T = 2 \times (48 + 13) \times 0.4 \mu$s = 48.8 \mu$s and $F = 20.491$ kHz.
```
We can develop a formula for delay calculations using 16-bit mode of the timer for a crystal frequency of XTAL = 10 MHz. This is given in Figure 9-6. The scientific calculator in the Accessory directory of Microsoft Windows can help you to find the TMR0H, TMR0L values. This calculator supports decimal, hex, and binary calculations. See Examples 9-6 and 9-7.

(a) in hex

(FFFF - YYXX + 1) \times 0.4 \mu s

where YYXX are the TMR0H, TMR0L initial values respectively. Notice that YYXX values are in hex.

(b) in decimal

Convert YYXX values of the TMR0H, TMR0L register to decimal to get a NNNNN decimal number, then (65536 - NNNNN) \times 0.4 \mu s

Figure 9-6. Timer Delay Calculation for XTAL = 10 MHz with No Prescaler

Example 9-6

Find the delay generated by Timer0 in the following code, using both of the methods of Figure 9-6. Do not include the overhead due to instructions.

```
BCF TRISB,5 ; PB5 as an output
MOVLW 0x80 ; Timer0, 16-bit, int clk, no prescale
MOVF T0CON
BCF INTCON,TMR0IF ; clear Timer0 interrupt
HERE MOVLW 0xB8 ; TMR0H = B8, the high byte
MOVF TMR0H
MOVLW 0x3E ; TMR0L = 3E, the low byte
MOVF TMR0L
BSF T0CON, TMROON ; start Timer0
AGAIN BTFSS INTCON, TMR0IF ; monitor Timer0 flag until
BRA AGAIN ; it rolls over
BCF T0CON, TMROON ; stop Timer0
BCF INTCON, TMR0IF ; clear Timer0 interrupt
BTG PORTB,5 ; toggle PB5
BRA HERE ; load TH, TL again
```

Solution:

(a) (FFFF - B83E + 1) = 47C2H = 18,370 in decimal and 18,370 \times 0.4 \mu s = 7.348 ms.

(b) Because TMR0H : TMR0L = B83EH = 47166 (in decimal) we have 65536 - 47166 = 18,370. This means that the timer counts from B83EH to FFFFH. This plus rolling over to 0 goes through a total of 18,370 clock cycles, where each clock is 0.4 \mu s in duration. Therefore, we have 18,370 \times 0.4 \mu s = 7.348 ms as the width of the pulse.
Find the frequency of the square wave generated by the following program if XTAL = 10 MHz. In your calculation do not include the overhead due to instructions in the loop.

```assembly
BCF TRISB, 5       ; PB5 as an output
MOVLW 0x08         ; Timer0, 16-bit, int clk, no prescale
MOVWF T0CON        ; load T0CON reg.
HERE               ;
MOVLW 0x76          ; TMROH = 76H, the high byte
MOVWF TMR0H         ; load Timer0 high byte
MOVLW 0x34          ; TMROL = 34H, the low byte
MOVWF TMR0L         ; load Timer0 low byte
BCF INTCON, TMR0IF  ; clear timer interrupt flag bit
CALL DELAY          ; delay using Timer0
BTG PORTB, 5        ; toggle PB5
BRA HERE            ; load TH, TL again

; _________ delay using Timer0
DELAY BSF T0CON, TMR0ON ; start Timer0
AGAIN BTFSS INTCON, TMR0IF ; monitor Timer0 flag until
BRA AGAIN           ; it rolls over
BCF T0CON, TMR0ON   ; stop Timer0
RETURN
```

**Solution:**
Because $FFFFH - 7634H = 89CBH + 1 = 89CCH$ and $89CCH = 35,276$ clock count, $35,276 \times 0.4 \mu s = 14.11$ ms and frequency $= 1/(14.11 \text{ ms} \times 2) = 35.434$ Hz. In this calculation, the overhead due to all the instructions in the loop is not included.

---

**Finding values to be loaded into the timer**

Assuming that we know the amount of timer delay we need, the question is how to find the values needed for the TMROH and TMROL registers. To calculate the values to be loaded into the TMROL and TMROH registers, look at Examples 9-8 and 9-9, where we use a crystal frequency of 10 MHz for the PIC18 system.

Assuming XTAL = 10 MHz and no prescaler we can use the following steps for finding the TMROH and TMROL registers’ values:

1. Divide the desired time delay by 0.4 μs.
2. Perform $65,536 - n$, where $n$ is the decimal value we got in Step 1.
3. Convert the result of Step 2 to hex, where $yyxx$ is the initial hex value to be loaded into the timer’s registers.
4. Set TMROL = $xx$ and TMROH = $yy$. 

---

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Example 9-8

Assuming that XTAL = 10 MHz, write a program to generate a square wave with a period of 10 ms on pin PORTB.3.

Solution:

For a square wave with \( T = 10 \text{ ms} \) we must have a time delay of \( 5 \text{ ms} \). Because XTAL = 10 MHz, the counter counts up every 0.4 \( \mu \text{s} \). This means that we need \( 5 \text{ ms} / 0.4 \mu \text{s} = 12,500 \) clocks. 65,536 - 12,500 = 53,036 = CF2CH. Therefore, we have TMR0H = CF and TMR0L = 2C.

Example 9-9

Assuming that XTAL = 10 MHz, modify the program in Example 9-8 to generate a square wave of 2 kHz frequency on pin PORTB.3.

Solution:

Look at the following steps.
(a) \( T = 1 / F = 1 / 2 \text{ kHz} = 500 \mu \text{s} \) the period of the square wave.
(b) \( 1/2 \) of it for the high and low portions of the pulse is 250 \( \mu \text{s} \).
(c) 250 \( \mu \text{s} / 0.4 \mu \text{s} = 625 \) and 65,536 - 625 = 64,911, which in hex is FD8FH.
(d) TMR0L = 8FH and TMR0H = FDH, all in hex.
Modify TMROL and TMR0H in Example 9-8 to get the largest time delay possible. Find the delay in ms. In your calculation, exclude the overhead due to the instructions in the loop.

Solution:
To get the largest delay we make TL and TH both 0. This will count up from 0000 to FFFFH and then roll over to zero.

```
BCF TRISB,3           ;PB3 as an output
MOVLW 0x80           ;Timer0,16-bit, int clk, no prescale
MOVF T0CON           ;load T0CON reg.
HERE
CLRF TMR0H           ;TH = TL = 0
CLRF TMROL           ;clear timer interrupt flag bit
CALL DELAY
BTG PORTB,3          ;toggle PB3
BRA HERE             ;load TH, TL again
;------------------------ delay using Timer0
DELAY BSF T0CON,TMROON ;start Timer0
AGAIN BTFSS INTCON,TMROIF ;monitor Timer0 flag until
BRA AGAIN            ;it rolls over
BCF T0CON,TMROON     ;stop Timer0
RETURN
```

Making TMR0H and TMROL both zero means that the timer will count from 0000 to FFFFH, and then roll over to raise the TMROIF flag. As a result, it goes through a total of 65,536 states. Therefore, we have delay = (65,536 - 0) × 0.4 μs = 26.214 ms. That gives us the smallest frequency of 1 / (2 × 26.214 ms) = 1 / (52.428 ms) = 19.073 Hz.

Using the Windows calculator to find TH, TL

The scientific calculator in Microsoft Windows is a handy and easy-to-use tool to find the TMR0H, TMROL values. Assume that we would like to find the TMR0H, TMROL values for a time delay that uses 35,000 clocks of 0.4 μs. The following steps show the calculation:

1. Bring up the scientific calculator in MS Windows and select decimal.
2. Enter 35,000.
3. Select hex. This converts 35,000 to hex, which is 88B8H.
4. Select +/- to give -35,000 decimal (7748H).
5. The lowest two digits (48) of this hex value are for TMROL and the next two (77) are for TMR0H. We ignore all the Fs on the left because our number is 16-bit data.
Prescaler and generating a large time delay

As we have seen in the examples so far, the size of the time delay depends on two factors, (a) the crystal frequency, and (b) the timer’s 16-bit register. Both of these factors are beyond the control of the PIC18 programmer. We saw in Example 9-10 that the largest time delay is achieved by making both TMR0H and TMR0L zero. What if that is not enough? We can use the prescaler option in the T0CON register to increase the delay by reducing the period. The prescaler option of T0CON allows us to divide the instruction clock by a factor of 2 to 256 as was shown in Figure 9-2.

As we have seen so far, with no prescaler enabled, the crystal oscillator frequency is divided by 4 (Fosc/4) and then fed into Timer0. If we enable the prescaler bit in the T0CON register, however, then we can divide the instruction clock (Fosc/4) further before it is fed into Timer0. The lower 3 bits of the T0CON register give the options of the number we can divide by. As shown in Figure 9-2, this number can be 2, 4, 8, 16, 32, 64, and so on. Notice that the lowest number is 2 and the highest number is 256. Examine Examples 9-11 through 9-15 to see how the prescaler options are programmed.

### Example 9-11

Find the value for T0CON if we want to program Timer0 in 16-bit mode with a prescaler of 64 and use internal clock (Fosc/4) for the clock source, positive-edge.

**Solution:**

From Figure 9-2 we have $T0CON = 0000\ 0101$; 16-bit mode, XTAL clock source, prescaler of 64.

### Example 9-12

Find the timer’s clock frequency and its period for various PIC18-based systems, with the following crystal frequencies. Assume that a prescaler of 1:64 is used.

(a) 10 MHz  
(b) 16 MHz

**Solution:**

![Diagram]

(a) $1/4 \times 10\ MHz = 2.5\ MHz$ and $1/64 \times 2.5\ MHz = 39062.5\ Hz$ due to 1:64 prescaler and $T = 1/39062.5Hz = 25.6\ \mu s$

(b) $1/4 \times 16\ MHz = 4\ MHz$ and $1/64 \times 4\ MHz = 62500\ Hz$ due to prescaler and $T = 1/62500\ Hz = 16\ \mu s$
Example 9-13

Examine the following program and find the time delay in seconds. Exclude the overhead due to the instructions in the loop. Assume XTAL = 10 MHz.

```
BCF TRISB, 2 ; PB2 as an output
MOVLW 0x05 ; Timer0, 16-bit, int clk, prescaler 64
MOVF T0CON ; load T0CON reg.
HERE
MOVLW 0x01 ; TMR0H = 01H, the high byte
MOVF TMROH ; load Timer0 high byte
MOVLW 0x08 ; TMR0L = 08H, the low byte
MOVF TMROL ; load Timer0 low byte
BCF INTCON, TMROIF ; clear timer interrupt flag bit
CALL DELAY
BTG PORTB, 2 ; toggle PB2
BRA HERE ; load TH, TL again

; -------- delay using Timer0
DELAY BSF TOCON, TMROON ; start Timer0
AGAIN BTFSS INTCON, TMROIF ; monitor Timer0 flag until
BRA AGAIN ; it rolls over
BCF TOCON, TMROON ; stop Timer0
RETURN
```

Solution:

TMR0H: TMR0L = 0108H = 264 in decimal and 65,536 – 264 = 65,272. Now 65,272 x 64 x 0.4 μs = 1.671 seconds, or from Example 9-12, we have 65,272 x 25.6 μs = 1.671 seconds.

Example 9-14

Assume XTAL = 10 MHz. (a) Find the clock period fed into Timer0 if a prescaler option of 256 is chosen. (b) Show what is the largest time delay we can get using this prescaler option.

Solution:

(a) \( \frac{1}{4} \times 10 \text{ MHz} = 2.5 \text{ MHz} \) and \( \frac{1}{256} \times 2.5 \text{ MHz} = 9765.625 \text{ Hz} \) due to 1:256 prescaler and \( T = \frac{1}{9765.625} \text{ Hz} = 1024 \text{ μs} = 1.024 \text{ ms} \)

(b) To get the largest delay, we make TMR0L and TMR0H both 0. Making TMR0H and TMR0L both zero means that the timer will count from 0000 to FFFFH, and then roll over to raise the TMR0IF flag. As a result, it goes through a total of 65,536 states. Therefore, we have delay = (65,536 – 0) x 1024 μs = 67,108,864 μs = 67.108864 seconds.
Example 9-15

Assuming XTAL = 10 MHz, write a program to generate a square wave of 50 Hz frequency on pin PORTB.7. Use Timer0, 16-bit mode, with prescaler = 128.

Solution:

Look at the following steps:
(a) \( T = \frac{1}{50} \) Hz = 20 ms, the period of the square wave.
(b) 1/2 of it for the high and low portions of the pulse = 10 ms
(c) 10 ms / 0.4 \( \mu \)s / 128 = 195 and 65,536 – 195 = 65,341 in decimal, and in hex it is FF3DH.
(d) TL = 3D and TH = FF (hex)

8-bit mode programming of Timer0

Timer0 can also be used in 8-bit mode. The 8-bit mode allows only values of 00 to FFH to be loaded into the timer’s register TMRL0. After the timer is started, it starts to count up by incrementing the TMROL register. It counts up until it reaches its limit of FFH. When it rolls over from FFH to 00, it sets HIGH the TMROIF. See Figure 9-7.
Steps to program 8-bit mode of Timer0

To generate a time delay using Timer0 in 8-bit mode, take the following steps:

1. Load the T0CON value register indicating 8-bit mode is selected.
2. Load the TMR0L registers with the initial count value.
3. Start the timer.
4. Keep monitoring the timer flag (TMR0IF) to see if it is raised. Get out of the loop when TMR0IF becomes HIGH.
5. Stop the timer with the instruction “BCF TOCON, TMROON”.
6. Clear the TMR0IF flag for the next round.
7. Go back to Step 2 to load TMR0L again.

Notice that when we choose the 8-bit option, only the TMR0L register is used and the TMR0H has a zero value during the count up. To clarify the above steps, see Examples 9-16 and 9-17.

Example 9-16

Assuming that XTAL = 10 MHz, find (a) the frequency of the square wave generated on pin PORTB.0 in the following program, and (b) the smallest frequency achievable in this program, and the TH value to do that.

```
BCF TRISB, 0 ; PB0 as an output
MOVLW 0x48 ; Timer0, 8-bit, int clk, no prescaler
MOVWF T0CON ; load T0CON reg.
BCF INTCON, TMR0IF ; clear timer interrupt flag bit
HERE MOVLW 0x5 ; TMR0L = 5, the low byte
MOVWF TMR0L ; load Timer0 byte
CALL DELAY
BTG PORTB, 0 ; toggle PB0
BRA HERE ; load TL again
```

---delay using Timer0

```
DELAY BSF T0CON, TMROON ; start Timer0
AGAIN BTFSS INTCON, TMR0IF ; monitor Timer0 flag until
BRA AGAIN ; it rolls over
BCF T0CON, TMROON ; stop Timer0
BCF INTCON, TMR0IF ; clear Timer0 interrupt flag bit
RETURN
```

Solution:

(a) Now (256 - 05) = 251 x 0.4 μs = 100.4 μs is the high portion of the pulse. Because it is a 50% duty cycle square wave, the period T is twice that; as a result T = 2 x 100.4 μs = 200.8 μs, and the frequency = 4.98 kHz.

(b) To get the smallest frequency, we need the largest T, and that is achieved when TMR0H = 00. In that case, we have T = 2 x 256 x 0.4 μs = 204.8 μs and the frequency = 1 / 204.8 μs = 4,882.8 Hz.
Example 9-17

Assume XTAL = 10 MHz. (a) Find the clock period fed into Timer0 if the prescaler option of 256 is chosen. (b) Show what is the largest time delay we can get using this prescaler option.

Solution:

(a) \( \frac{1}{4} \times 10 \text{ MHz} = 2.5 \text{ MHz} \) and \( \frac{1}{256} \times 2.5 \text{ MHz} = 9765.625 \text{ Hz} \) due to 1:256 prescaler and \( T = \frac{1}{9765.625 \text{ Hz}} = 1024 \mu s \)

(b) To get the largest delay, we make TMROL = 0. Making TMROL zero means that the timer will count from 00 to FFH, and then roll over to raise the TMR0IFF flag. As a result, it goes through a total of 256 states. Therefore, we have delay = \((256 - 0) \times 1024 \mu s = 262,144 \mu s = 0.262144 \text{ second.} \)

Assemblers and negative values

Because the timer is in 8-bit mode, we can let the assembler calculate the value for TMR0H. For example, in “MOVlw, -D'100'”, the assembler will calculate the -100 = 9C and make WREG = 9C in hex. This makes our job easier. See Examples 9-18 and 9-19.

Example 9-18

Assuming that we are programming the timers for 8-bit mode, find the value (in hex) loaded into TMROL for each of the following cases.

(a) MOVlw -D'200'  
(b) MOVlw -D'60'  
(c) MOVlw -D'12'

MOVWF TMROL  
MOVWF TMROL  
MOVWF TMROL

Solution:

You can use the Windows scientific calculator to verify the results provided by the assembler. In the Windows calculator, select decimal and enter 200. Then select hex, then +/- to get the negative value. The following is what we get.

<table>
<thead>
<tr>
<th>Decimal</th>
<th>2’s complement (TMROL value)</th>
</tr>
</thead>
<tbody>
<tr>
<td>-200</td>
<td>38H</td>
</tr>
<tr>
<td>-60</td>
<td>C4H</td>
</tr>
<tr>
<td>-12</td>
<td>F4H</td>
</tr>
</tbody>
</table>
Example 9-19

Find (a) the frequency of the square wave generated in the following code, and (b) the duty cycle of this wave. Assume XTAL = 10 MHz

```
BCF TRISB,3 ;PB3 as an output
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
MOVLW 0x48 ;Timer0,8-bit, int clk, no prescaler
MOVWF T0CON ;load T0CON reg.
HERE MOVLW -D'150' ;loading negative value
MOVWF TMRL ;load Timer0 byte
BSF PORTB,3 ;PB3 = 1
CALL DELAY
MOVWF TMRL ;reload Timer0 byte
CALL DELAY
BCF PORTB,3 ;PB3 = 0
MOVWF TMRL ;reload Timer0 byte
CALL DELAY
CALL DELAY
BRA HERE ;load TH, TL again

---delay using Timer0
DELAY BSF T0CON,TMR0ON ;start Timer0
AGAIN BTFSS INTCON,TMR0IF ;monitor Timer0 flag until
BRA AGAIN ;it rolls over
BCF T0CON,TMR0ON ;stop Timer0
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
RETURN
```

Solution:

For the TMRL value in 8-bit mode, the conversion is done by the assembler as long as we enter a negative number. This also makes the calculation easy. Because we are using 150 clocks, we have time for the DELAY subroutine = \(150 \times 0.4 \mu s = 60 \mu s\). The high portion of the pulse is twice the size of the low portion (66% duty cycle). Therefore, we have: \(T = \text{high portion} + \text{low portion} = 2 \times 60 \mu s + 60 \mu s = 180 \mu s\) and frequency = 5.555555 kHz.

Another version of this program could be as follows:

```
BCF TRISB,3 ;PB3 as an output
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
MOVLW 0x48 ;Timer0,8-bit, int clk, no prescaler
MOVWF T0CON ;load T0CON reg.
HERE BSF PORTB,3 ;PB3 = 1
CALL DELAY
CALL DELAY
BCF PORTB,3 ;PB3 = 0
CALL DELAY
CALL DELAY
BRA HERE ;load TH, TL again

---delay using Timer0
DELAY MOVLW -D'150' ;loading negative value
MOVWF TMRL ;load Timer0 byte
BSF T0CON,TMR0ON ;start Timer0
AGAIN BTFSS INTCON,TMR0IF ;monitor Timer0 flag until
BRA AGAIN ;it rolls over
BCF T0CON,TMR0ON ;stop Timer0
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
RETURN
```

CHAPTER 9: PIC18 TIMER PROGRAMMING IN ASSEMBLY AND C
Timer1 programming

Timer1 is a 16-bit timer, and its 16-bit register is split into two bytes, referred to as TMR1L (Timer1 low byte) and TMR1H (Timer1 high byte). See Figure 9-8. Timer1 can be programmed in 16-bit mode only and unlike Timer0, it does not support 8-bit mode. Timer1 also has the T1CON (Timer1 control) register in addition to the TMR1IF (Timer1 interrupt flag). The TMR1IF flag bit goes HIGH when TMR1H:TMR1L overflows from FFFF to 0000. Timer1 also has the prescaler option, but it only supports factors of 1:1, 1:2, 1:4, and 1:8. See Figure 9-9 for the Timer1 block diagram and Figure 9-10 for T1CON register options. The PIR1 register contains the TMR1IF flags. See Figure 9-11.

![Figure 9-8. Timer1 High and Low Registers](image)

![Figure 9-9. Timer1 Block Diagram](image)

Examples 9-20 and 9-21 show how to program Timer1. Notice that in many of the time delay calculations, we have ignored the clocks caused by the overhead instructions in the loop. To get a more accurate time delay, and hence frequency, you need to include them. If you use a digital scope and you don’t get exactly the same frequency as we have calculated, it is because of the overhead associated with those instructions.

In this section, we used the PIC18 timer for time delay generation. However, a more powerful and creative way to use these timers is as event counters. We discuss this use of the counter next.
RD16 D7 16-bit read/write enable bit
1 = Timer1 16-bit is accessible in one 16-bit operation.
0 = Timer1 16-bit is accessible in two 8-bit operations.

D6 Not used

T1CKPS2:T1CKPS0 D5 D4 Timer1 prescaler selector
0 0 = 1:1 Prescale value
0 1 = 1:2 Prescale value
1 0 = 1:4 Prescale value
1 1 = 1:8 Prescale value

T1OSCEN D3 Timer1 oscillator enable bit
1 = Timer1 oscillator is enabled.
0 = Timer1 oscillator is shutoff.

T1SYNC D2 Timer1 synchronization (used only when TMR1CS = 1 for counter mode to synchronize external clock input)
If TMR1CS = 0 this bit is not used.

TMR1CS D1 Timer1 clock source select bit
1 = External clock from pin RC0/T1CKI
0 = Internal clock (Fosc/4 from XTAL)

TMR1ON D0 Timer1 ON and OFF control bit
1 = Enable (start) Timer1
0 = Stop Timer1

Figure 9-10. T1CON (Timer 1 Control ) Register

TMR1IF D1 Timer1 Interrupt overflow flag bit
0 = Timer1 did not overflow.
1 = Timer1 has overflowed (FFFF to 0000).

The importance of TMR1IF: When TMR1H:TMR1L overflows from FFFF to 0000, this flag is raised. We monitor this flag bit before we reload the TMR1H:TMR1L registers.

The other bits of this register are discussed in Chapter 11.

Figure 9-11. PIR1 (Interrupt Control Register 1) Contains the TMR1IF Flag

CHAPTER 9: PIC18 TIMER PROGRAMMING IN ASSEMBLY AND C  353
Example 9-20

Find the frequency of the square wave generated by the following program if XTAL = 10 MHz. In your calculation do not include the overhead due to instructions in the loop.

```assembly
BCF TRISB, 5 ; make PB5 an output
MOVLW 0x0 ; Timer1, 16-bit, int clk, no prescale
MOVF T1CON ; load T1CON reg
HERE MOVLW 0x76 ; TMR1H = 76H, the high byte
MOVF TMR1H ; load Timer1 high byte
MOVLW 0x34 ; TMR1L = 34H, the low byte
MOVF TMR1L ; load Timer1 low byte
BCF PIR1, TMR1IF ; clear timer interrupt flag bit
CALL DELAY
BTG PORTB, RB5 ; toggle PB5
BRA HERE ; load TH, TL again
-----delay using Timer1
DELAY BSF T1CON, TMR1ON ; start Timer1
AGAIN BTFSS PIR1, TMR1IF ; monitor Timer1 flag until
BRA AGAIN ; it rolls over
BCF PIR1, TMR1ON ; stop Timer1
RETURN
```

Solution:
Because FFFFH - 7634H = 89CBH + 1 = 89CCH and 89CCH = 35276 clock count 35276 \times 0.4 \mu s = 14.11 ms and frequency = 1 / (14.11 ms \times 2) = 35.434 Hz. In this calculation, the overhead due to all the instructions in the loop is not included. Calculation is the same as Example 9-7.

Example 9-21

Assuming XTAL = 10 MHz, write a program to generate a square wave of 50 Hz frequency on pin PORTB.5. Use Timer1 in 16-bit mode with the maximum prescaler allowed.

Solution:
Because FFFFH - F3CBH = C34H + 1 = C35H and C35H = 3125 clock count 3125 \times 8 \times 0.4 \mu s = 10 ms and frequency = 1 / (2 \times 10 ms) = 50 Hz. In this calculation, the overhead due to all the instructions in the loop is not included.

```assembly
BCF TRISB, 5 ; make PB5 an output
MOVLW 0x30 ; Timer1, 16-bit, int clk, prescale 1:8
MOVF T1CON ; load T1CON reg
HERE MOVLW 0xF3 ; TMR1H = F3H, the high byte
MOVF TMR1H ; load Timer1 high byte
MOVLW 0xCB ; TMR1L = CBH, the low byte
MOVF TMR1L ; load Timer1 low byte
BCF PIR1, TMR1IF ; clear timer interrupt flag bit
CALL DELAY
BTG PORTB, RB5 ; toggle PB5
BRA HERE ; load TH, TL again
-----delay using Timer1
DELAY BSF T1CON, TMR1ON ; start Timer1
AGAIN BTFSS PIR1, TMR1IF ; monitor Timer1 flag until
BRA AGAIN ; it rolls over
BCF PIR1, TMR1ON ; stop Timer1
RETURN
```
Review Questions

1. How many timers do we have in the PIC18F458/4580?
2. True or false. Timer0 can be used only as a 16-bit timer.
3. True or false. Timer1 can be used only as a 16-bit timer.
4. True or false. The T0CON register is a bit-addressable register.
5. Indicate the selection made in the instruction “MOV T0CON, 0x08”.
6. In 16-bit mode, the counter rolls over when the counter goes from ____ to ____.
7. In 8-bit mode, the counter rolls over when the counter goes from ____ to ____.
8. In the instruction “MOVLW -D’200’”, find the hex value for WREG.
9. To get a 2-ms delay, what numbers should be loaded into TMR0H and TMR0L using 16-bit mode? Assume that XTAL = 10 MHz.
10. To get a 100-µs delay, what number should be loaded into the TMR0L register using 8-bit mode? Assume that XTAL = 10 MHz.

SECTION 9.2: COUNTER PROGRAMMING

In the last section, we used the timers of the PIC18 to generate time delays. These timers can also be used as counters to count events happening outside the PIC18. The use of the timer as an event counter is covered in this section. When the timer is used as a timer, the PIC18’s crystal is used as the source of the frequency. When it is used as a counter, however, it is a pulse outside the PIC18 that increments the TH, TL registers. In counter mode, notice that registers such as T0CON, TMR0H, and TMR0L are the same as for the timer discussed in the last section; they even have the same names.

T0CS bit in T0CON register

Recall from the last section that the T0CS bit (Timer0 clock source) in the T0CON register decides the source of the clock for the timer. If T0CS = 0, the timer gets pulses from the crystal oscillator connected to the OSC1 and OSC2 pins (Fosc/4). In contrast, when T0CS = 1, the timer is used as a counter and gets its pulses from outside the PIC18. Therefore, when T0CS = 1, the counter counts up as pulses are fed from pin RA4 (PORTA.4). The pin is called T0CKI (Timer0 clock input). Notice that the pin belongs to Port A. In the case of Timer0, when T0CS = 1, pin RA4 (PORTA.4) provides the clock pulse and the counter counts up for each clock pulse coming from that pin. Similarly, for Timer1, when TMR1CS = 1, each clock pulse coming in from pin RC0 (PORTC.0) makes the counter count up. See Example 9-22.

In Example 9-23, we are using Timer1 as an event counter that counts up as clock pulses are fed into pin 3.5. These clock pulses could represent the number of people passing through an entrance, or the number of wheel rotations, or any other event that can be converted to pulses.
In Example 9-23, the TL data was displayed in binary. In Example 9-24, the TL registers are converted to ASCII to be displayed on an LCD.

As another example of the application of the timer with C/T = 1, we can feed an external square wave of 60 Hz frequency into the timer. The program will generate the second, the minute, and the hour out of this input frequency and display the result on an LCD. This will be a nice digital clock, but not a very accurate one.

**Example 9-22**

Find the value for T0CON if we want to program Timer0 as an 8-bit mode counter, no prescaler. Use an external clock for the clock source and increment on the positive edge.

**Solution:**

T0CON = 0110 1000  8-bit, external clock source, no prescaler.

**Example 9-23**

Assuming that clock pulses are fed into pin T0CKI, write a program for counter 0 in 8-bit mode to count the pulses and display the state of the TMR0L count on PORTB.

**Solution:**

```assembly
BSF TRISA,RA4           ;PORTA.4 as an input for clock
CLRF TRISB             ;PORTB as an output
MOVLW 0x68             ;Timer0, 8-bit, ext clk, no prescaler
MOVF T0CON             ;load T0CON reg
HERE MOVLW 0x0          ;TMR0L = 0
MOVF TMR0L             ;load Timer0
BCF INTCON,TMROIF      ;clear timer interrupt flag bit
BSF T0CON,TMROON       ;start Timer0
AGAIN MOVFF TMR0L,PORTB ;display the count on PORTB
BTFSS INTCON,TMROIF    ;monitor Timer0 flag until
BRA AGAIN              ;it rolls over
BCF T0CON,TMROON       ;stop Timer0
GOTO HERE
```

PORTB is connected to 8 LEDs and input T0CKI to pulse.
Using external crystal for Timer1 clock

Timer1 has two options when it comes to using the external clock source. It uses either the clock fed into the T1CKI pin or the clock from a crystal connected to the T1OSI and T1OSO pins, as shown in Figure 9-9. Generally, a 32-kHz crystal is connected to the T1OSI and T1OSO pins and is used for saving power during SLEEP mode because the SLEEP instruction does not disable Timer1. Notice that this 32-kHz crystal connected to the T1OSCI and T1OSO pins is in addition to the main crystal connected to the OSC1 and OSC2 pins. The PIC18 uses the main crystal to execute CPU instruction clock cycles among other things, and when the CPU goes into SLEEP mode, the main crystal is shut down to save power. The alternate 32-kHz crystal connected to pins T1OSO and T1OSI provides clock to Timer1 during SLEEP mode, while the main crystal is shut down. This allows the use of the timer to implement an on-chip RTC (real-time clock). Chapter 16 shows how to connect an external RTC to the PIC18. Notice that in order to use the alternate external clock source for Timer1, we must choose the external clock source option of TMR1CS = 1, in addition to enabling the T1OSCEN bit (T1OSCEN = 1) in the T1CON register, as shown in Figure 9-10. Study Examples 9-23 through 9-27 to see how timers are used as counters.

Before we finish this section, we need to state an important point. You might think monitoring the TMR0IF and TMR1IF flags is a waste of the microcontroller’s time. You are right. There is a solution to this: the use of interrupts. Using interrupts enables us to do other things with the microcontroller. When a timer Interrupt flag such as TMR0IF is raised it will inform us. This important and powerful feature of the PIC18 is discussed in Chapter 11.
Example 9-24

Assume that a 1-Hz frequency pulse is connected to input for Timer0 (pin T0CKI). Write a program to display counter 0 on PORTB, PORTC, and PORTD in decimal. Set the initial value of TMR0L to –60.

Solution:

To display the TMR0L count on an LCD, we must convert 8-bit binary data to ASCII. See Chapter 5 for data conversion.

NUME EQU 0x00 ;RAM loc for NUME
QU EQU 0x20 ;RAM loc for quotient
RMND_L EQU 0x30 ;the least significant digit loc
RMND_M EQU 0x31 ;the middle significant digit loc
RMND_H EQU 0x32 ;the most significant digit loc
MYDEN EQU D'10' ;value for divide by 10
BSF TRISA, RA4 ;RA4 as an input
MOVLW 0x68 ;Timer0, 8-bit, ext clk, no prescale
MOVWF TOCON ;load TOCON reg
HERE MOVLW 0x0 ;TMR0L = 0
MOVWF TMRL ;load Timer0
BCF INTCON, TMR0IF ;clear timer interrupt flag bit
BSF TOCON, TMR0ON ;start Timer0
AGAIN MOVF TMRL+W ;save the count in WREG
CALL BIN_ASC_CON
BTFSS INTCON, TMR0IF ;monitor Timer0 flag until
BRA AGAIN ;it rolls over
BCF TOCON, TMR0ON ;stop Timer0
GOTO HERE ;converting 8-bit binary to decimal

;CONV_B_D

MOVFF PORTB, WREG ;load numerator
MOVWF NUME ;WREG = 10, the denominator
CLRF QU ;clear quotient
D_1 INCF QU ;inc quotient for every subtract
SUBWF NUME ;subtract WREG from NUME value
BC D_1 ;if positive go back
ADDWF NUME ;once too many, first digit
DECF QU ;once too many for quotient
MOVFF NUME, RMND_L ;save the first digit
MOVFF QU, NUME ;repeat the process one more time
CLRF QU ;clear QU
D_2 INCF QU ;subtract WREG from NUME value
BC D_2 ;once too many
ADDWF NUME ;2nd digit
DECF QU
MOVFF NUME, RMND_M ;3rd digit
MOVFF QU, RMND_H
RETURN

In order to display the data on LCD, the decimal number must be converted to ASCII. See Chapter 6.
Example 9-25

Assume that a 16-Hz frequency pulse is connected to input for Timer0 (pin T0CKI). Write a program to display the counter values of TMROH and TMROL on ports B and D. Set the initial values to 0. Use Timer0, 16-bit mode, and positive-edge clock. Show the program for (a) no prescaler, (b) prescaler of 1:16.

Solution:

(a)

```
BSF TRISA,RA4 ;RA4 as an input
CLRF TRISB ;PORTB as an output
CLRF TRISD ;PORTD as an output
MOVLW 0x28 ;Timer 0,16-bit,ext clk,no prescale
MOVF T0CON ;load T0CON reg
HERE MOVLW 0x0 ;TMROH = 0
MOVF TMROH ;load Timer0 high byte
MOVLW 0x0 ;TMROL = 0
MOVF TMROL ;load Timer0 low byte
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
BSF T0CON,TMROON ;start Timer0
AGAIN MOVFF TMROH,PORTD ;display high byte count
MOVFF TMROL,PORTB ;display low byte count
BTFSS INTCON,TMR0IF ;monitor Timer0 flag until
BRA AGAIN ;it rolls over
BCF T0CON,TMROON ;stop Timer0
GOTO HERE
```

(b)

```
BSF TRISA,RA4 ;RA4 as an input
CLRF TRISB ;PORTB as an output
CLRF TRISD ;PORTD as an output
MOVLW 0x23 ;T0,16-bit,ext clk,prescale of 1:16
MOVF T0CON ;load T0CON reg
HERE MOVLW 0x0 ;TMROH = 0
MOVF TMROH ;load Timer0 High byte
MOVLW 0x0 ;TMROL = 0
MOVF TMROL ;load Timer0 low byte
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
BSF T0CON,TMROON ;start Timer0
AGAIN MOVFF TMROH,PORTD ;display high byte count
MOVFF TMROL,PORTB ;display low byte count
BTFSS INTCON,TMR0IF ;monitor Timer0 flag until
BRA AGAIN ;it rolls over
BCF T0CON,TMROON ;stop Timer0
GOTO HERE
```

PIC18

![16 Hz clock T0CKI](pic18.png)

PB

<table>
<thead>
<tr>
<th>RA4</th>
</tr>
</thead>
<tbody>
<tr>
<td>PD</td>
</tr>
</tbody>
</table>

to LEDs
Example 9-26

Assuming that clock pulses are fed into pin T0CKI and a buzzer is connected to pin PORTB.1, write a program for counter 0 in 8-bit mode to sound the buzzer every 100 pulses.

Solution:
To sound the buzzer every 100 pulses, we set the initial counter value to -100 (9C in hex), then the counter counts up until it reaches FF. Upon overflow, we can count the buzzer by toggling the PORTB.1 pin.

```
BCF TRISB,1 ;RB1 as an output
BSF TRISA,4 ;RA4 as an input for clock-in
MOVLW 0x68 ;Timer0, 8-bit, ext clk, no prescale
MOVWF T0CON
MOVLW -D'100' ;TMR0L = 0
MOVWF TMR0L
BCF INTCON, TMR0IF ;clear timer interrupt flag bit
BSF T0CON, TMR0ON ;start Timer0
AGAIN BTFSS INTCON, TMR0IF ;monitor Timer0 flag until
  BRA AGAIN ;it rolls over
BCF T0CON, TMR0ON ;stop Timer0
OVER BTG PORTB,1 ;sound the buzzer
CALL DELAY ;quarter second delay
GOTO OVER ;forever
```

Bit 1 of PORTB is connected to a buzzer and input T0CKI to a pulse.

![PIC18F458 Circuit Diagram](image-url)
Example 9-27

Assume that a 1-Hz frequency pulse is connected to input for Timer1 (pin PORTC.0). Write a program to display the counter values of TMR1H and TMR1L on ports B and D. Set the initial values to 0. Use Timer1, 16-bit mode, no prescaler, and positive-edge clock.

Solution:

```
BSF TRISC,RC0 ; PC0 as an input
CLRF TRISB ; PORTB as an output
CLRF TRISD ; PORTD as an output
MOVLW 0x02 ; Timer1, 16-bit, ext clk, no prescale
MOVF T1CON ; load T0CON reg
HERE MOVLW 0x0 ; TMR1H = 0, the low byte
MOVF TMR1H ; load Timer1 high byte
MOVLW 0x0 ; TMR1L = 0, the low byte
MOVF TMR1L ; load Timer1 low byte
BCF PIR1,TMR1IF ; clear timer interrupt flag bit
BSF T1CON,TMR1ON ; start Timer1
AGAIN MOVFF TMR1H,PORTD ; display high byte count
MOVFF TMR1L,PORTB ; display low byte count
BTFSS PIR1,TMR1IF ; monitor Timer1 flag until
BRA AGAIN ; it rolls over
BCF PIR1,TMR1ON ; stop Timer1
GOTO HERE
```

![Diagram of PIC18 microcontroller with LEDs and 1 Hz clock input](image)
Review Questions

1. What provides the clock pulses to PIC18 timers if T0CS = 0?
2. What provides the clock pulses to PIC18 timers if T0CS = 1?
3. Does the discussion in Section 9.1 apply to timers if T0CS = 1?
4. To allow RC0 to be used as an input for the Timer1 clock, what must be done, and why?
5. Do we have a choice of counting up on the positive or negative edge of the clock?

SECTION 9.3: PROGRAMMING TIMERS 0 AND 1 IN C

In Chapter 7 we showed some examples of C programming for the PIC18. In this section we show C programming for the PIC18 timers. As we saw in the examples in Chapter 7, the general-purpose registers of the PIC18 are under the control of the C compiler and are not accessed directly by C statements. All of the SFRs, however, are accessible directly using C statements. As an example of accessing the SFRs directly, we saw how to access ports PORTB–PORTD in Chapter 7. Next, we discuss how to access the PIC18 timers directly using the C18 C compiler.

Accessing timer registers in C

In C18 we can access timer registers such as TMR0H, TMR0L, and T0CON directly using the PIC18Fxxx.h header file. This is shown in Example 9-28. Example 9-28 also shows how to access the TMR0ON and TMR0IF flag bits. Notice that all the SFR registers are bit-accessible.

Calculating delay length using timers

As we saw in the last two sections, the delay length depends on two factors: (a) the crystal frequency, (b) the prescaler factor. The third factor in the delay size is the C compiler because various C compilers generate different hex code sizes. Study Examples 9-28 through 9-33 and verify them using an oscilloscope.
Write a C18 program to toggle all the bits of PORTB continuously with some delay. Use Timer0, 16-bit mode, and no prescaler options to generate the delay.

Solution:

```c
#include <pl8f4580.h>
void TDelay(void);
void main(void)
{
  TRISB=0; //PORTB output port
  while(1) //repeat forever
  {
    PORTB=0x55; //toggle all bits of Port B
    TDelay(); //delay size unknown
    PORTB=0xAA; //toggle all bits of Port B
    TDelay();
  }
}

void TDelay()
{
  T0CON=0x08; //Timer0, 16-bit mode, no prescaler
  TMR0H=0x35; //load TH0
  TMR0L=0x00; //load TL0
  T0CONbits.TMR0ON=1; //turn on T0
  while(INTCONbits.TMR0IF==0); //wait for TF0 to roll over
  T0CONbits.TMR0ON=0; //turn off T0
  INTCONbits.TMR0IF=0; //clear TF0
}
Example 9-29

Write a C18 program to toggle only the PORTB.4 bit continuously every 50 ms. Use Timer0, 16-bit mode, the 1:4 prescaler to create the delay. Assume XTAL = 10 MHz.

Solution:

```c
#include <pl8f4580.h>

void TODelay(void);
#define mybit PORTBbits.RB4

void main(void)
{
    TRISBbits.TRISB4=0;
    while(1)
    {
        mybit=!mybit;     //toggle PORTB.4
        TODelay();        //Timer0, mode 1 (16-bit)
    }
}

void TODelay()
{
    T0CON=0x01;       //Timer0, 16-bit mode, 1:4 prescaler
    TMR0H=0x85;       //load TH0
    TMR0L=0xEE;       //load TL0
    T0CONbits.TMR0ON=1;    //turn on Timer0
    while(INTCONbits.TMR0IF==0); //wait for TF0 to roll over
    T0CONbits.TMR0ON=0;    //turn off Timer0
    INTCONbits.TMR0IF=0;   //clear TF0
}

FFFFh–85EEH = 7A11H = 31249 + 1 = 31250

Timer delay = 31250 × 4 × 0.4 μs = 50 ms
```
Write a C18 program to generate a frequency of 2 Hz only on pin PORTB.5. Use Timer0, 8-bit mode to create the delay.

Solution:

```c
#include <pl8f4580.h>
void TOMSDelay(void);
#define mybit PORTBbits.RB5
void main(void)
{
    unsigned char x, y;
    TRISBbits.TRISB5 = 0;
    while(1)
    {
        mybit^=1; //toggle PortB.5
        for(x=0; x<250; x++) //due to for loop overhead
            for(y=0; y<35; y++) //we put 35 and not 39
                TOMSDelay();
    }
}
void TOMSDelay()
{
    TOCON=0x45; //Timer0, 16-bit mode, prescaler 1:64
    TMR0L=-1; //load TL0
    TOCONbits.TMR0ON=1; //turn on T0
    while(INTCONbits.TMR0IF=0); //wait for TF0 to roll over
    TOCONbits.TMR0ON=0; //turn off T0
    INTCONbits.TMR0IF=0; //clear TF0
}

256 - 255 = 1

1 × 64 × 0.4 μs = 25.6 μs

25.6 μs × 250 × 39 = 0.2496 by calculation.

F = 1 / (2 × 0.2496 s) = 1 / 0.4992 s = 2 Hz

The scope output, however, does not give us this result. This is due to overhead of the for loop in C. To correct this problem, we put 35 instead of 39.
Write a C18 program to generate a frequency of 250 Hz on all bits of PORTC. Use Timer0, 16-bit mode, and no prescaler to create the frequency. Assume XTAL = 10 MHz.

Solution:

```c
#include <p18f4580.h>
void T0Delay(void);
void main(void)
{
    unsigned char x;
    TRISC=0;          //PORTC output port
    PORTC=0x55;
    while(1)
    {
        PORTC=~PORTC;  //toggle all bits of Port C
        for(x=0;x<20;x++)
            T0Delay();
    }
}
void T0Delay()
{
    T0CON=0x0;        //Timer 0, 16-bit mode, no prescaler
    TMR0H=0xFF;      //load TH0
    TMR0L=0x06;      //load TL0
    T0CONbits.TMR0ON=1;   //turn on T0
    while(INTCONbits.TMR0IF==0); //wait for TF0 to roll over
    T0CONbits.TMR0ON=0;     //turn off T0
    INTCONbits.TMR0IF=0;    //clear TF0
}

FF06H = 65286 in decimal

65536 – 65286 = 250

250 × 0.4 μs = 0.1 ms and 20 × 0.1 ms = 2 ms

T = 1 / (2 × 2 ms) = 1 / 4 ms = 250 Hz

Another way is:

T = 1 / 250 Hz = 0.004 second and one half is 0.002 second

0.002 second / 0.4 μs = 5000

5000 / 20 = 250 because the for loop is set to 20.
Example 9-32

A switch is connected to pin PORTB.7. Write a C18 program to monitor SW and create the following frequencies on pin PORTB.0:

<table>
<thead>
<tr>
<th>SW</th>
<th>Frequency</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>500 Hz</td>
</tr>
<tr>
<td>1</td>
<td>750 Hz</td>
</tr>
</tbody>
</table>

Use Timer0 with prescaler for both of them.

Solution:

```c
#include <p18f4580.h>
define mybit PORTBbits.RB0
define SW PORTBbits.RB7
void T0PSDelay(unsigned char);
void main(void)
{
TRISBbits.TRISB7=1; // make PB.7 an input
TRISBbits.TRISB0=0; // make PB.0 an output
SW=1;
while(1)
{
mybit^=1; // toggle PB.0
if(SW==0) // check switch
    T0PSDelay(0);
else
    T0PSDelay(1);
}
}
void T0PSDelay(unsigned char c)
{
T0CON=0x05; // Timer 0, 16-bit mode, prescaler 1:64
if(c==0)
{
    TMR0H=0xFF; // load TH0
    TMR0L=0xD9; // load TL0
}
else
{
    TMR0H=0xFF; // load TH0
    TMR0L=0xE6; // load TL0
}
T0CONbits.TMR0ON=1; // turn on T0
while(INTCONbits.TMR0IF==0); // wait for TF0 to roll over
T0CONbits.TMR0ON=0; // turn off T0
INTCONbits.TMR0IF=0; // clear TF0
}

FFD9H = 65497
FFE6H = 65510
65536-65497 = 39
65536-65510 = 26
39 x 64 x 0.4 µs = 998 µs
26 x 64 x 0.4 µs = 666 µs
1 / (998 µs x 2) = 501 Hz
1 / (666 µs x 2) = 751 Hz

Use the scope and modify TH:TL to get an exact frequency.

CHAPTER 9: PIC18 TIMER PROGRAMMING IN ASSEMBLY AND C
Example 9-33

Write a C18 program to create a frequency of 2500 Hz on pin PORTB.1. Use Timer1 to create the delay.

Solution:

```c
#include <pl8f4580.h>
void T1Delay(void);
#define mybit PORTBbits.RB1

void main(void)
{
    TRISBbits.TRISB1 = 0;
    while(1)
    {
        mybit ^= 1;      //toggle PB.1
        T1Delay();
    }
}

void T1Delay()
{
    T1CON=0x00;       //Timer1, 16-bit mode, no prescaler
    TMR1H=0xFE;      //load TH1
    TMR1L=0x0C;      //load TL1
    T1CONbits.TMR1ON=1;    //turn on T1
    while(PIR1bits.TMR1IF==0);    //wait for TF1 to roll over
    T1CONbits.TMR1ON=0;    //turn off T1
    PIR1bits.TMR1IF=0;    //clear TF1
}
```

\[
\frac{1}{2500 \text{ Hz}} = 400 \mu s
\]

\[
400 \mu s / 2 = 200 \mu s
\]

\[
200 \mu s / 0.4 \mu s = 500
\]

\[
65536 - 500 = 65036 = \text{FE0CH}
\]

C programming of Timers 0 and 1 as counters

In Section 9.2 we showed how to use Timers 0 and 1 as event counters. Timers can be used as counters if we provide pulses from outside the chip instead of using the frequency of the crystal oscillator as the clock source. By feeding pulses to the T0CKI (RA4) and T1CKI (RC0) pins, we turn Timer0 and Timer1 into Counter 0 and Counter 1, respectively. Study Examples 9-34 through 9-37 to see how Timers 0 and 1 are programmed as counters using C language.
Example 9-34

Assume that a 1-Hz external clock is being fed into pin TOCKI (RA4). Write a C18 program for Counter0 in 8-bit mode to count up and display the state of the TMR0L count on PORTB. Start the count at 0H.

Solution:

```c
#include <p18f4580.h>

void main(void)
{
    TRISAbits.TRISA4=1; //make RA4/TOCKI an input
    TRISB=0;
    TOCON=0x68; //Counter 0, 8-bit mode, no prescaler
    TMR0L=0; //set count to 0

    while(1) //repeat forever
    {
        do
        {
            TOCONbits.TMROON=1; //turn on T0
            PORTB=TMR0L; //place value on pins
        }
        while(INTCONbits.TMR0IF==0); //wait for TF0 to roll over
        TOCONbits.TMROON=0; //turn off T0
        INTCONbits.TMR0IF=0; //clear TF0
    }
}
```

PORTB is connected to 8 LEDs.
TOCKI (RA4) is connected to a 1-Hz external clock.
Assume that a 1-Hz external clock is being fed into pin T0CKI (RA4). Write a C program for Counter 0 in mode 1 (16-bit) to count the pulses and display the TMR0H and TMR0L registers on PORTD and PORTB, respectively.

Solution:

```c
#include <pl8f4580.h>

void main(void)
{
    TRISAbits.TRISA4=1;  //make RA4 an input for T0CKI
    TRISB=0;              //PORTB output port
    TRISD=0;              //PORTD output port
    TOCON=0x25;           //Timer0, 16-bit mode, prescaler 1:64
    TMR0H=0;              //set count to 0
    TMR0L=0;              //set count to 0

    while(1)              //repeat forever
    {
        do
        {
            TOCONbits.TMR0ON=1;  //turn on T0
            PORTB=TMR0L;         //
            PORTD=TMR0H;         //place value on pins
        }
        while(INTCONbits.TMR0IF==0);//wait for rollover

        TOCONbits.TMR0ON=0;    //turn off T0
        INTCONbits.TMR0IF=0;   //clear TF0
    }
}
```

![Diagram of PIC18 with connections to PB and LED outputs](image-url)
Example 9-36

Assume that a 64-Hz external clock is being fed into pin T0CKI (RA4). Write a C program for Counter 0 in 8-bit mode to display the count in ASCII. The 8-bit binary count must be converted to ASCII. Display the ASCII digits (in binary) on PORTB, PORTC, and PORTD, where PORTB has the least significant digit. Set the initial value of TMR0L to 0.

Solution:

To display the TMR0L count, we must convert 8-bit binary data to ASCII. See Chapter 7 for data conversion. The ASCII values will be shown in binary. For example, ‘9’ will show as 0011001 on the ports.

```c
#include <p18f4580.h>
void BinToASCII(unsigned char);
void main()
{
    unsigned char value;
    TRISAbits.TRISA4=1;    //make RA4 an input
    TRISB=0;               //make PORTB an output
    TRISC=0;               //make PORTC an output
    TRISD=0;               //make PORTD an output
    TMR0L=0;
    T0CON=0x65;            //Counter 0, 8-bit mode, prescaler 1:64

    while(1)
    {
        do
        {
            T0CONbits.TMR0ON=1;  //turn on T0
            value=TMR0L;
            BinToASCII(value);
        }
        while(INTCONbits.TMR0IF==0);   //wait for TF0 to roll over
        T0CONbits.TMR0ON=0;   //turn off T0
        INTCONbits.TMR0IF=0;  //clear TF0
    }

    void BinToASCII(unsigned char value)     //see Chapter 7
    {
        unsigned char x,d1,d2,d3;
        x=value/10;
        d1=value%10;
        d2=x%10;
        d3=x/10;
        PORTB=0x30 | d1;
        PORTC=0x30 | d2;
        PORTD=0x30 | d3;
    }
}
```

CHAPTER 9: PIC18 TIMER PROGRAMMING IN ASSEMBLY AND C 371
Example 9-37

Assume that a 60-Hz external clock is being fed into pin T0CKI (RA4). Write a C program for Counter 0 in 8-bit mode to display the seconds and minutes on PORTB and PORTD, respectively.

Solution:

```c
#include <p18f4580.h>

void ToTime(unsigned char);

void main()
{
    unsigned char sec;
    TRISB=TRISD=0; //PORTB,D outputs
    T0CON=0x68; //Timer 0, no prescaler
    TMR0L=-60; //sec = 60 pulses
    while(1)
    {
        do
        {
            T0CONbits.TMR0ON=1; //turn on T0
            sec=TMR0L;
            ToTime(sec);
        }
        while(INTCONbits.TMR0IF==0); //wait for TF0 to roll over
        T0CONbits.TMR0ON=0; //turn off T0
        INTCONbits.TMR0IF=0; //clear TF0
    }
}

void ToTime(unsigned char value)
{
    unsigned char sec, min;
    min = value / 60;
    sec = value % 60;
    PORTB = sec;
    PORTD = min;
}
```

By using 60 Hz, we can generate seconds, minutes, and hours.
SECTION 9.4: PROGRAMMING TIMERS 2 AND 3

In this section we examine Timers 2 and 3 of the PIC18 family and show how to program them in both Assembly and C.

Timer2 programming

Timer2 is an 8-bit timer. The 8-bit register of Timer2 is called TMR2. Timer2 also has an 8-bit register called the period register (PR2). We can set the PR2 register to a fixed value and Timer2 will increment from 00 until it matches the value in PR2. At that point, the equal signal will raise the TMR2IF flag and reset TMR2 to 00. The clock source for Timer2 is Fosc/4 with the options of both prescaler and postscaler, as shown in Figure 9-12. Notice from Figure 9-12 that there is no external clock source for Timer2. In other words, it cannot be used as a counter. Examine the next few examples to learn the programming syntax for Timer2. See Figures 9-12 and 9-13.

![Timer2 Block Diagram](image)

**Figure 9-12 Timer2 Block Diagram**

<table>
<thead>
<tr>
<th>TMR2IF</th>
<th>TMR1IF</th>
</tr>
</thead>
</table>

**TMR2IF**

- Timer2 Interrupt overflow flag bit
  - 0 = TMR2 value is not equal to PR2 register.
  - 1 = TMR2 value is equal to PR2 register.

The other bits of this register are discussed in Chapter 11. The location of TMRxIF in the PIR register can vary in future products.

**Figure 9-13. PIR1 (Peripheral Interrupt Flag Register 1) Contains TMR2IF Flag**
### Example 9-38

Assuming that XTAL = 10 MHz, write a program to turn on pin PORTB4 when TMR2 reaches value 100 (decimal).

**Solution:**

Because XTAL = 10 MHz, TMR2 counts up every 0.4 μs. Therefore, when you have TMR2H = PR2 = 100, PORTB4 will be turned on.

```assembly
BCF TRISB, 4 ; make PORTB4 an output
BCF PORTB, 4 ; turn off PORTB4
MOVLW 0x0 ; Timer2, no prescale or postscale
MOVWF T2CON ; load T2CON reg
MOVLW 0x0 ; TMR2 = 0
MOVF TMR2 ; load TMR2
MOVLW D'100' ; PR2 = 100, the period register
MOVWF PR2 ; load PR2
BCF PIR1, TMR2IF ; clear timer interrupt flag
BSF T2CON, TMR2ON ; start Timer2
AGAIN BTFSS PIR1, TMR2IF ; monitor Timer2 flag
BRA AGAIN ;
BSF PORTB, 4 ; turn on PORTB4
BCF T2CON, TMR2ON ; stop Timer2
HERE BRA HERE
```
Example 9-39

Using the prescaler and postscaler, find the longest time delay that we can create using Timer2. Assume that XTAL = 10 MHz.

Solution:
We can create the longest time delay by making PR2 = 255. When TMR2 reaches value 255 (decimal), it toggles a pin.

```c
BCF TRISB,4 ; make PORTB4 an output
BCF PORTB,4 ; turn off PORTB4
MOVlw B'01110111' ; Timer2, prescale = 16, postscale = 16
MOVWF T2CON ; load T2CON reg
MOVlw 0x0 ; TMR2 = 0
MOVWF TMR2 ; load Timer2
HERE MOVlw D'255' ; PR2 = 255, the period register
MOVWF PR2 ; load PR2
BCF PIR1,TMR2IF ; clear timer interrupt flag bit
BSF T2CON,TMR2ON ; start Timer2
AGAIN BTFSS PIR1,TMR2IF ; monitor Timer2 flag
BRA AGAIN ;
BTG PORTB,4 ; turn on PORTB4
BCF T2CON,TMR2ON ; stop Timer2
BRA HERE
```

Because XTAL = 10 MHz, TMR2 counts up every 0.4 µs. Therefore, when you have TMR2H = PR2 = 255, RB4 will be turned on and off every 52 ms because 255 × 0.4 µs × 16 × 16 = 26.112 ms.

Example 9-40

Assuming that XTAL = 10 MHz, write a C18 program to turn on pin PORTB4 when TMR2 reaches value 100. This is a repeat of Example 9-13 in C18.

Solution:
#include <p18f4580.h>
#define mybit PORTBbits.RB4
void main(void)
{
    TRISBbits.TRISB4=0;  // PORTB4 as output
    T2CON=0x0;           // Timer2, no prescaler/postscaler
    TMR2=0x00;           // TMR2 = 0
    mybit=0;             // PB.4 = 0
    PR2=100;             // load period register 2
    T2CONbits.TMR2ON=1;  // turn on T0
    while(PIR1bits.TMR2IF==0);  // wait for TMR2IF to be raised
    mybit=1;             // PB.4 = 0
    T2CONbits.TMR2ON=0;  // turn off T2
    PIR1bits.TMR2IF=0;   // clear TF0
    while(1);            // stay here
}
Example 9-41

Using the prescaler and postscaler, find the longest time delay that we can create using Timer2. Assume that XTAL = 10 MHz. This is a repeat of Example 9-39 in C18.

Solution:

```c
#include <pl8f4580.h>
define mybit PORTBbits.RB4

void main(void) {
    TRISBbits.TRISB4=0;
    T2CON=0x7B; //Timer2, prescaler = 16, postscaler = 16
    TMR2=0x00; //TMR2 = 0
    while(1) {
        PR2=255; //load period register 2
        T2CONbits.TMR2ON=1; //turn on T2
        while(PIRlbits.TMR2IF==0); //wait for TMR2IF to be raised
        mybit=~mybit; //toggle PORTB4
        T2CONbits.TMR2ON=0; //turn off T2
        PIRlbits.TMR2IF=0; //clear TF0
    }
}
```

Because XTAL = 10 MHz, TMR2 counts up every 0.4 μs. Therefore, when TMR2H = PR2 = 255, PORTB4 will be turned on and off every 52 ms because 255 × 0.4 μs × 16 × 16 = 26.112 ms.

Timer3 Programming

Timer3 is a 16-bit timer that can be used as a timer or counter. Its 16-bit register is split into two bytes referred to as TMR3L and TMR3H. (See Figure 9-17.) Timer3 can be programmed in 16-bit mode only and does not support 8-bit mode. We select various options of Timer3 using the T3CON (Timer3 Control) register, as shown in Figure 9-15. Timer3 has the prescaler options of 1:1, 1:2, 1:4, and 1:8, as shown in Figure 9-15. Figure 9-15 also shows the bits related to the CCP (compare/capture pulse-width-modulation) feature of the PIC18. CCP is a widely used feature of the PIC18, and we will discuss it in Chapters 15 and 17. In Chapter 15, we will see how to use the CCP feature along with the interrupt to measure the pulse width. Pulse width modulation (PWM) is an important concept used in DC motor control. We will examine it in detail in Chapter 17. Because Timer3 is a 16-bit timer, the TMR3IF flag bit goes HIGH when TMR3H:TMR3L overflows from FFFF to 0000. The TMR3IF (Timer3 Interrupt flag) is part of the PIR2 register, as shown in Figure 9-16.

Examine the next few examples to learn the programming syntax for Timer3.

376
RD16  D7  16-bit read/write enable bit
0 = Timer3 16-bit is accessible in one 16-bit operation.
1 = Timer3 16-bit is accessible in two 8-bit operations.

T3CCP2:T3CCP1  D6 D3  Timer3 and Timer1 to CPPx Enable bits
0 0 = Timer1 is the clock source for compare/capture of the CCP module.
0 1 = Timer3 is the clock source for compare/capture of the CCP2.
1 x = Timer3 is the clock source for compare/capture of the CCP1.

T3CKPS1:T3CKPS0  D5 D4  Timer3 Input Clock Prescaler Selector
0 0 = 1:1  Prescale value
0 1 = 1:2  Prescale value
1 0 = 1:4  Prescale value
1 1 = 1:8  Prescale value

T1SYNC  D2  Timer3 external clock input synchronization control bit
Used only when TMR3CS = 1 and clock comes from an external source. If TMR3CS = 0, this bit is not used.
1 = Do not synchronize external clock input
0 = Synchronize external clock input

TMR3CS  D1  Timer3 clock source select bit
1 = External clock from pin TIOSI or T1CKI
0 = Internal clock (Fosc/4)

TMR3ON  D0  Timer3 On and Off control bit
1 = Enable (start) Timer1
0 = Stop Timer1

Figure 9-15. T3CON (Timer3 Control) Register

<table>
<thead>
<tr>
<th>TMR3IF</th>
</tr>
</thead>
</table>

TMR3IF  Timer3 interrupt overflow flag bit
0 = Timer3 did not overflow.
1 = Timer3 has overflowed (FFFF to 0000).

The importance of TMR3IF: In 16-bit mode, when TMR3H:TMR3L overflows from FFFF to 0000, this flag is raised.

The location of TMRxIF in the PIR register can vary in future products.

Figure 9-16. PIR2 (Peripheral Interrupt Flag Register 2) Contains the TMR3IF Flag
Example 9-42

Find the frequency of the square wave generated by the following program if \( XTAL = 10 \) MHz. In your calculation do not include overhead due to instructions in the loop.

```
BCF TRISB,5           ;PB5 as an output
MOVLW 0x0             ;Timer3, 16-bit, int clk, no prescale
MOVWF T3CON           ;load T3CON reg
HERE MOVLW 0x76        ;TMR3H = 76H, the high byte
MOVWF TMR3H            ;load Timer3 high byte
MOVLW 0x34             ;TMR3L = 34H, the low byte
MOVWF TMR3L            ;load Timer3 low byte
BCF PIR2,TMR3IF        ;clear timer interrupt flag bit
CALL DELAY
BTG PORTB,RB5          ;toggle PB5
BRA HERE               ;load TH, TL again
```

;—— delay using Timer3

DELAY BSF T3CON,TMR3ON ;start Timer3
AGAIN BTFSS PIR2,TMR3IF ;monitor Timer3 flag until
BRA AGAIN              ;it rolls over
BCF T3CON,TMR3ON       ;stop Timer3
RETURN

Solution:
Because \( FFFFH - 7634H = 89CBH + 1 = 89CCH \) and \( 89CCH = 35276 \) clock count, \( 35276 \times 0.4 \mu s = 14.11 \) ms and frequency = \( 1 / (2 \times 14.11 \) ms) = \( 1 / 28.22 \) ms = 35.434 Hz. In this calculation, the overhead due to all the instructions in the loop is not included. Notice that the calculation is the same as in Example 9-20.
Assume XTAL = 10 MHz, write a program to generate a square wave of 50 Hz frequency on pin PORTB5. Use Timer3, 16-bit mode, with the maximum prescaler allowed.

Solution:

Because FFFFH - 9E58H = 61A7H + 1 = 61A8H and 61A8H = 25,000 clock count, 25000 x 0.4 µs = 1 ms and frequency = 1/2 (1 ms) = 50 Hz. In this calculation, the overhead due to all the instructions in the loop is not included.

```assembly
BCF TRISB,5 ;PB5 as an output
MOVLW 0x0 ;Timer3,16-bit,int clk,no prescale
MOVWF T3CON ;load T3CON reg.
HERE MOVLW 0x9E ;TMR3H = 9EH, the high byte
MOVWF TMR3H ;load Timer3 high byte
MOVLW 0x58 ;TMR3L = 58H, the low byte
MOVWF TMR3L ;load Timer3 low byte
BCF PIR2,TMR3IF ;clear Timer3 interrupt flag bit
CALL DELAY
BTG PORTB,RB5 ;toggle PB5
BRA HERE ;----delay using Timer3
DELAY BSF T3CON,TMR3ON ;start Timer3
AGAIN BTFSS PIR2,TMR3IF ;monitor timer flag until
BRA AGAIN ;it rolls over
BCF T3CON,TMR3ON ;stop timer
RETURN
```

\[ T = \frac{1}{50 \text{ Hz}} = 2 \text{ ms} \]

\[ \frac{1}{2} \times 2 \text{ ms} = 1 \text{ ms} \quad \text{for high and low portions} \]

\[ \frac{1}{2} \times 2 \times 2 \text{ ms} = 2 \text{ ms} \]

\[ 1 \text{ ms} / 0.4 \mu s = 25000 \quad \text{number of clock counts} \]

\[ 65536 - 25000 = 40536 = 9E58H \]
Example 9-44

Assume that a 1-Hz frequency pulse is connected to the input for Timer3 (pin RC0). Write a program to display the counter values TMR3H and TMR3L on ports B and D. Set the initial values of to 0. Use no prescaler.

Solution:

```assembly
BSF TRISC,0 ;PORTC.0 as input T1CLKI
CLRF TRISB ;PORTB as an output
CLRF TRISD ;PORTD as an output
MOVLW 0x02 ;Timer3,16-bit,ext clk,no prescale
MOVF T3CON ;load T0CON reg
HERE MOVFW 0x0 ;TMR3H = 0, the low byte
MOVWF TMR3H ;load Timer3 high byte
MOVFW 0x0 ;TMR1L = 0, the low byte
MOVWF TMR3L ;load Timer3 low byte
BCF PIR2,TMR3IF ;clear timer interrupt flag bit
BSF T3CON,TMR3ON ;start timer
AGAIN MOVFF TMR3L,PORTB ;display low byte count
MOVFF TMR3H,PORTD ;display high byte count
BTFSS PIR2,TMR3IF ;monitor Timer3 flag until
BRA AGAIN ;it rolls over
BCF T3CON,TMR3ON ;stop Timer3
GOTO HERE
```

[Diagram: PIC18 microcontroller with inputs T1CLKI and RC0 connected to a 1 Hz clock, and outputs PB and PD connected to LEDs]
Write a C18 program to create a frequency of 2500 Hz on pin PORTB.1. Use Timer3 to create the delay.

Solution:

```
#include <pl8f4580.h>
void T3Delay(void);
#define mybit PORTBbits.RB1

void main(void)
{
   TRISBbits.TRISB1=0; //PB1 as an output
   T3CON=0x00; //Timer3, 16-bit mode, no prescaler
   while(1)
   {
      mybit=~mybit;  //toggle PB.1
      T3Delay();
   }
}

void T3Delay()
{
   TMR3H=0xFE; //load TH3
   TMR3L=0x0C; //load TL3
   T3CONbits.TMR3ON=1; //turn on T3
   while(PIR2bits.TMR3IF==0); //wait for TF3 to roll over
   T3CONbits.TMR3ON=0; //turn off T3
   PIR2bits.TMR3IF=0; //clear TF3
}
```

$1 / 2500\text{ Hz} = 400\ \mu\text{s}$

$400\ \mu\text{s} / 2 = 200\ \mu\text{s}$

$200\ \mu\text{s} / 0.4\ \mu\text{s} = 500$

$65536 - 500 = 65036 = \text{FE0CH}$
Example 9-46

Assume that a 1-Hz external clock is being fed into pin T3 (RC0). Write a C18 program for Timer3 to be used as a counter. It should count the pulses and display the TMR3H and TMR3L registers on PORTD and PORTB, respectively.

Solution:

```c
#include <p18f4580.h>

void main(void)
{
    TRISCbits.TRISC0=1;       //make RC0 an input for T1CKI
    TRISB = 0;                //make PORTB an output
    TRISD = 0;                //make PORTD an output
    T3CON=0x02;               //Timer1, 16-bit mode, no prescaler
    TMR3H=0;                  //set count to 0
    TMR3L=0;                  //set count to 0
    while(1)
    {
        do
        {
            T3CONbits.TMR3ON=1;   //turn on T3
            PORTB=TMR3L;          //place value on pins
            PORTD=TMR3H;
        }
        while(PIR2bits.TMR3IF==0);  //wait for TF3 to roll over
        T3CONbits.TMR3ON=0;      //turn off T3
        PIR2bits.TMR3IF=0;      //clear TF3
    }
}
```

![Diagram](image.png)
Review Questions

1. What provides the clock pulses to Timer2?
2. Indicate the selection made if T2CON = 0x00.
3. True or false. Timer2 cannot be used for a counter.
4. In Timer3, the counter rolls over when it goes from ____ to ____.
5. If we set PR2 = 200, state when TMR2IF is raised.
6. TMR2IF and TMR3IF are part of registers ____.

SUMMARY

The PIC18 can have up to four or more timers/counters, depending on the family member. When used as timers, they can generate time delays. When used as counters, they can serve as event counters. This chapter showed how to program the timers/counters for various modes.

Generally, the timers are accessed as two 8-bit registers, TMRLx and TMRHx. They can be used as a 16-bit timer, or as an 8-bit timer.

Each timer has its own TCON (Timer Control) register, allowing us to choose various operational modes. Among the modes are the prescaler and timer/counter options. When the timer is used as a timer, the PIC18’s crystal is used as the source of the frequency (Fosc/4); however, when it is used as a counter, it is a pulse outside of the PIC18 that increments the TMRxH, TMRxL registers.

PROBLEMS

SECTION 9.1: PROGRAMMING TIMERS 0 AND 1

1. How many timers are in the PIC18F458?
2. Timer0 of the PIC18 is ____-bit, accessed as ____ and _____.
3. Timer1 of the PIC18 is ____-bit, accessed as ____ and _____.
4. Timer0 supports the highest prescaler value of _____.
5. Timer1 supports the highest prescaler value of _____.
6. The T0CON register is a(n) ____-bit register.
7. What is the job of the T0CON register?
8. True or false. T1CON is a bit-addressable register.
9. Find the T1CON value for 16-bit mode, no prescaler, timer oscillator off, with the clock coming from the PIC18’s crystal.
10. Find the frequency and period used by the timer if the crystal attached to the PIC18 has the following values:
    (a) XTAL = 10 MHz
    (b) XTAL = 20 MHz
    (c) XTAL = 24 MHz
    (d) XTAL = 30 MHz
11. Indicate which register holds the TMRxIF (Timer Interrupt Flag) bit for each of the following timers:
    (a) Timer0
    (b) Timer1
12. Indicate the rollover value (in hex and decimal) of the timer for each of the following modes:
   (a) 16-bit
   (b) 8-bit
13. Indicate when the TMR0IF flag is raised for each of the following modes:
   (a) 16-bit
   (b) 8-bit
14. True or false. Both Timer0 and Timer1 have their own timer interrupt flags.
15. True or false. Both Timer0 and Timer1 have their own timer start flags.
16. Assume that XTAL = 10 MHz. Find the TMR0H,TMR0L value needed to generate a time delay of 2 ms. Use 16-bit mode, no prescaler mode.
17. Assume that XTAL = 10 MHz. Find the TMR0H,TMR0L value needed to generate a time delay of 5 ms. Use 16-bit mode, and the largest prescaler possible.
18. Assume that XTAL = 10 MHz. Find the TMR1H,TMR1L value needed to generate a time delay of 2.5 ms. Use the largest prescaler possible.
19. Assume that XTAL = 10 MHz. Find the TMR1H,TMR1L value needed to generate a time delay of 0.2 ms. Use 16-bit mode, no prescaler mode.
20. Assume that XTAL = 20 MHz. Find the TMR1H,TMR1L value needed to generate a time delay of 2 ms. Use 16-bit mode, and the largest prescaler possible.
21. Assuming that XTAL = 10 MHz, and we are generating a square wave on pin RB7, find the lowest square wave frequency that we can generate using Timer0 in 16-bit mode.
22. Assuming that XTAL = 10 MHz, and we are generating a square wave on pin RB2, find the highest square wave frequency that we can generate using Timer0 in 16-bit mode.
23. Repeat Problems 21 and 22 for 8-bit mode.
24. In 8-bit mode, assuming that TMR0L = F1H, indicate which states Timer0 goes through until TMR0IF is raised. How many states is that?
25. Program Timer0 to generate a square wave of 1 kHz. Assume that XTAL = 10 MHz.
26. Program Timer1 to generate a square wave of 3 kHz. Assume that XTAL = 10 MHz. Use the largest prescaler possible.
27. State the differences between Timer0 and Timer1.
28. Find the value (in hex) loaded into WREG in each of the following:
   (a) MOVLW -D'12'
   (b) MOVLW -D'22'
   (c) MOVLW -D'34'
   (d) MOVLW -D'92'
   (e) MOVLW -D'120'
   (f) MOVLW -D'104'

SECTION 9.2: COUNTER PROGRAMMING

29. To use a timer as an event counter we must set the ____ bit in the T0CON register to _______ (LOW, HIGH).
30. Can we use both Timer0 and Timer1 as event counters?
31. For Counter 0, which pin is used for the input clock?
32. For Counter 1, which pin is used for the input clock?
33. Program Timer1 to be an event counter. Use 16-bit mode, and display the binary count on PORTB and PORTD continuously. Set the initial count to 20,000.
34. Program Timer0 to be an event counter. Use 8-bit mode and display the binary count on PORTB continuously. Set the initial count to 20.
35. The T1CON register is a(n) ___-bit register.
36. True or false. The T1CON register is not a bit-addressable register.

SECTION 9.3: PROGRAMMING TIMERS 0 AND 1 IN C

37. Program Timer0 in C to generate a square wave of 1 kHz. Assume that XTAL = 10 MHz.
38. Program Timer1 in C to generate a square wave of 1 kHz. Assume that XTAL = 10 MHz.
39. Program Timer0 in C to generate a square wave of 3 kHz. Assume that XTAL = 10 MHz.
40. Program Timer1 in C to generate a square wave of 3 kHz. Assume that XTAL = 10 MHz.
41. Program Timer1 in C to be an event counter. Use 16-bit mode and display the binary count on PORTB and PORTD continuously. Set the initial count to 20,000.
42. Program Timer0 in C to be an event counter. Use 8-bit mode and display the binary count on PORTD continuously. Set the initial count to 20.

SECTION 9.4: PROGRAMMING TIMERS 2 AND 3

43. Indicate the highest size of the prescaler supported for each of the following timers:
   (a) Timer2    (b) Timer3
44. Indicate the rollover value (in hex and decimal) of Timer3.
45. Indicate when the timer flag is raised for each of the following:
   (a) Timer2    (b) Timer3
46. True or false. The PR2 register of Timer2 is an 8-bit register.
47. True or false. Both Timer2 and Timer3 are 16-bit timers.
48. Assume that XTAL = 10 MHz. Find the TMR3H,TMR3L value needed to generate a time delay of 2 ms. Use no prescaler.
49. Assume that XTAL = 10 MHz. Find the TMR3H,TMR3L value needed to generate a time delay of 5 ms. Use the highest prescaler allowed.
50. Program Timer3 to be an event counter. Use 16-bit mode and display the binary count on PORTB and PORTD continuously. Set the initial count to 20,000.
51. Program Timer2 in Assembly to toggle pin RB3 when it counts up from 0 to 200. Assume that XTAL = 10 MHz.
52. Program Timer3 in C to generate a square wave of 3 kHz. Assume that XTAL = 10 MHz.
53. Program Timer2 in C to toggle pin RB3 when it counts up from 0 to 200. Assume that XTAL = 10 MHz.
54. Program Timer3 in C to generate a square wave of 1 kHz. Assume that XTAL = 10 MHz.
ANSWERS TO REVIEW QUESTIONS

SECTION 9.1: PROGRAMMING TIMERS 0 AND 1

1. 4
2. False
3. True
4. True
5. 0000 1000 indicates 16-bit mode, no prescaler, and using XTAL for frequency.
6. FFFFH, 0000
7. FFH, 00
8. –200 is 38H; therefore, WREG = 38H
9. 2 ms / 0.4 us = 5000, 65536 – 5000 = 60536 = EC78H, TMR0H = ECH and TMR0L = 78H.
10. 100 us / 0.4 us = 250, 256 – 250 = 06; therefore TMR0L = 06H.

SECTION 9.2: COUNTER PROGRAMMING

1. The crystal attached to the PIC18
2. The clock source for the timer comes from pin RA4 (PORTA4).
3. Yes
4. We must configure the pin as input to allow the clocks to come in from an external source.
5. Yes

SECTION 9.4: PROGRAMMING TIMERS 2 AND 3

1. The crystal attached to the PIC18 (Fosc/4)
2. Prescale of 1, postscale of 1, and stop Timer2
3. True
4. FFFFH, 0
5. TMR2 counts up until it matches PR2. At that time TMR2IF is raised.
6. PIR1 and PIR2 respectively.
CHAPTER 10

PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY AND C

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Contrast and compare serial versus parallel communication
- List the advantages of serial communication over parallel
- Explain serial communication protocol
- Contrast synchronous versus asynchronous communication
- Contrast half- versus full-duplex transmission
- Explain the process of data framing
- Describe data transfer rate and bps rate
- Define the RS232 standard
- Explain the use of the MAX232 and MAX233 chips
- Interface the PIC18 with an RS232 connector
- Discuss the baud rate of the PIC18
- Describe serial communication features of the PIC18
- Describe the main registers used by serial communication of the PIC18
- Program the PIC18 serial port in Assembly and C
Computers transfer data in two ways: parallel and serial. In parallel data transfers, often eight or more lines (wire conductors) are used to transfer data to a device that is only a few feet away. Devices that use parallel transfers include printers and hard disks; each uses cables with many wire strips. Although a lot of data can be transferred in a short amount of time by using many wires in parallel, the distance cannot be great. To transfer to a device located many meters away, the serial method is used. In serial communication, the data is sent one bit at a time, in contrast to parallel communication, in which the data is sent a byte or more at a time. Serial communication of the PIC18 is the topic of this chapter. The PIC18 has serial communication capability built into it, thereby making possible fast data transfer using only a few wires.

In this chapter we first discuss the basics of serial communication. In Section 10.2, PIC18 interfacing to RS232 connectors via MAX232 line drivers is discussed. Serial port programming of the PIC18 is discussed in Section 10.3. Section 10.4 covers PIC18 C programming for the serial port using the C18 compiler.

SECTION 10.1: BASICS OF SERIAL COMMUNICATION

When a microprocessor communicates with the outside world, it provides the data in byte-sized chunks. In some cases, such as printers, the information is simply grabbed from the 8-bit data bus and presented to the 8-bit data bus of the printer. This can work only if the cable is not too long, because long cables diminish and even distort signals. Furthermore, an 8-bit data path is expensive. For these reasons, serial communication is used for transferring data between two systems located at distances of hundreds of feet to millions of miles apart. Figure 10-1 diagrams serial versus parallel data transfers.

![Serial versus Parallel Data Transfer](image)

The fact that in a single data line is used in serial communication instead of the 8-bit data line of parallel communication makes serial transfer not only much cheaper but also enables two computers located in two different cities to communicate over the telephone.

For serial data communication to work, the byte of data must be converted to serial bits using a parallel-in-serial-out shift register; then it can be transmitted
over a single data line. This also means that at the receiving end there must be a serial-in-parallel-out shift register to receive the serial data and pack them into a byte. Of course, if data is to be transferred on the telephone line, it must be converted from 0s and 1s to audio tones, which are sinusoidal signals. This conversion is performed by a peripheral device called a *modem*, which stands for “modulator/demodulator.”

When the distance is short, the digital signal can be transferred as it is on a simple wire and requires no modulation. This is how IBM PC keyboards transfer data to the motherboard. For long-distance data transfers using communication lines such as a telephone, however, serial data communication requires a modem to *modulate* (convert from 0s and 1s to audio tones) and *demodulate* (convert from audio tones to 0s and 1s).

Serial data communication uses two methods, asynchronous and synchronous. The *synchronous* method transfers a block of data (characters) at a time whereas the *asynchronous* method transfers a single byte at a time. It is possible to write software to use either of these methods, but the programs can be tedious and long. For this reason, special IC chips are made by many manufacturers for serial data communications. These chips are commonly referred to as UART (universal asynchronous receiver-transmitter) and USART (universal synchronous-asynchronous receiver-transmitter). The PIC18 chip has a built-in USART, which is discussed in detail in Section 10.3.

---

**Figure 10-2. Simplex, Half-, and Full-Duplex Transfers**

**Half- and full-duplex transmission**

In data transmission, if the data can be both transmitted and received, it is a *duplex* transmission. This is in contrast to *simplex* transmissions such as with printers, in which the computer only sends data. Duplex transmissions can be half or full duplex, depending on whether or not the data transfer can be simultaneous. If data is transmitted one way at a time, it is referred to as *half duplex*. If the data...
can go both ways at the same time, it is full duplex. Of course, full duplex requires two wire conductors for the data lines (in addition to the signal ground), one for transmission and one for reception, in order to transfer and receive data simultaneously. See Figure 10-2.

**Asynchronous serial communication and data framing**

The data coming in at the receiving end of the data line in a serial data transfer is all 0s and 1s; it is difficult to make sense of the data unless the sender and receiver agree on a set of rules, a protocol, on how the data is packed, how many bits constitute a character, and when the data begins and ends.

**Start and stop bits**

Asynchronous serial data communication is widely used for character-oriented transmissions, while block-oriented data transfers use the synchronous method. In the asynchronous method, each character is placed between start and stop bits. This is called framing. In data framing for asynchronous communications, the data, such as ASCII characters, are packed between a start bit and a stop bit. The start bit is always one bit but the stop bit can be one or two bits. The start bit is always a 0 (low) and the stop bit(s) is 1 (high). For example, look at Figure 10-3 in which the ASCII character “A” (8-bit binary 0100 0001) is framed between the start bit and a single stop bit. Notice that the LSB is sent out first.

**Figure 10-3. Framing ASCII ‘A’ (41H)**

Notice in Figure 10-3 that when there is no transfer, the signal is 1 (high), which is referred to as mark. The 0 (low) is referred to as space. Notice that the transmission begins with a start bit followed by D0, the LSB, then the rest of the bits until the MSB (D7), and finally, the one stop bit indicating the end of the character “A”.

In asynchronous serial communications, peripheral chips and modems can be programmed for data that is 7 or 8 bits wide. This is in addition to the number of stop bits, 1 or 2. While in older systems ASCII characters were 7-bit, in recent years, 8-bit data has become common due to the extended ASCII characters. In some older systems, due to the slowness of the receiving mechanical device, two stop bits were used to give the device sufficient time to organize itself before transmission of the next byte. In modern PCs, however, the use of one stop bit is standard. Assuming that we are transferring a text file of ASCII characters using 1 stop bit, we have a total of 10 bits for each character: 8 bits for the ASCII code, and 1 bit each for the start and stop bits. Therefore, each 8-bit character has an extra 2 bits, which gives 25% overhead.
In some systems, the parity bit of the character byte is included in the data frame in order to maintain data integrity. This means that for each character (7- or 8-bit, depending on the system) we have a single parity bit in addition to start and stop bits. The parity bit is odd or even. In the case of an odd parity bit the number of 1s in the data bits, including the parity bit, is odd. Similarly, in an even parity bit system the total number of bits, including the parity bit, is even. For example, the ASCII character “A”, binary 0100 0001, has 0 for the even parity bit. UART chips allow programming of the parity bit for odd-, even-, and no-parity options.

### Data transfer rate

The rate of data transfer in serial data communication is stated in \textit{bps} (bits per second). Another widely used terminology for bps is \textit{baud rate}. However, the baud and bps rates are not necessarily equal. This is because baud rate is the modem terminology and is defined as the number of signal changes per second. In modems, sometimes a single change of signal transfers several bits of data. As far as the conductor wire is concerned, the baud rate and bps are the same, and for this reason in this book we use the terms bps and baud interchangeably.

The data transfer rate of a given computer system depends on communication ports incorporated into that system. For example, the early IBM PC/XT could transfer data at the rate of 100 to 9600 bps. In recent years, however, Pentium-based PCs transfer data at rates as high as 56K. Note that in asynchronous serial data communication, the baud rate is generally limited to 100,000 bps.

### RS232 standards

To allow compatibility among data communication equipment made by various manufacturers, an interfacing standard called RS232 was set by the Electronics Industries Association (EIA) in 1960. In 1963 it was modified and called RS232A. RS232B and RS232C were issued in 1965 and 1969, respectively. In this book we refer to it simply as RS232. Today, RS232 is the most widely used serial I/O interfacing standard. This standard is used in PCs and numerous types of equipment. Because the standard was set long before the advent of the TTL logic family, however, its input and output voltage levels are not TTL compatible. In RS232, a 1 is represented by \(-3\) to \(-25\) V, while a 0 bit is \(+3\) to \(+25\) V, making \(-3\) to \(+3\) undefined. For this reason, to connect any RS232 to a microcontroller system we must use voltage converters such as MAX232 to convert the TTL logic levels to the RS232 voltage level, and vice versa. MAX232 IC chips are commonly referred to as line drivers. RS232 connection to MAX232 is discussed in Section 10.2.

### RS232 pins

Table 10-1 provides the pins and their labels for the RS232 cable, commonly referred to as the DB-25 connector. In labeling, DB-25P refers to the plug connector (male) and DB-25S is for the socket connector (female). See Figure 10-4.
Because not all the pins are used in PC cables, IBM introduced the DB-9 version of the serial I/O standard, which uses only 9 pins, as shown in Table 10-2. The DB-9 pins are shown in Figure 10-5.

**Data communication classification**

Current terminology classifies data communication equipment as DTE (data terminal equipment) or DCE (data communication equipment). DTE refers to terminals and computers that send and receive data, while DCE refers to communication equipment, such as modems, that are responsible for transferring the data. Notice that all the RS232 pin function definitions of Tables 10-1 and 10-2 are from the DTE point of view.

The simplest connection between a PC and a microcontroller requires a minimum of three pins, TX, RX, and ground, as shown in Figure 10-6. Notice in that figure that the RX and TX pins are interchanged.

**Examining RS232 handshaking signals**

To ensure fast and reliable data transmission between two devices, the data transfer must be coordinated. Just as in the case of the printer, because the receiving device may have no room for the data in serial data communication, there must be a way to inform the sender to stop sending data. Many of the pins of the RS-232 connector are used for handshaking signals. Their description is provided below only as a reference, and they can be bypassed because they are not supported by the PIC18 UART chip.

<table>
<thead>
<tr>
<th>Pin</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Protective ground</td>
</tr>
<tr>
<td>2</td>
<td>Transmitted data (TxD)</td>
</tr>
<tr>
<td>3</td>
<td>Received data (RxD)</td>
</tr>
<tr>
<td>4</td>
<td>Request to send (RTS)</td>
</tr>
<tr>
<td>5</td>
<td>Clear to send (CTS)</td>
</tr>
<tr>
<td>6</td>
<td>Data set ready (DSR)</td>
</tr>
<tr>
<td>7</td>
<td>Signal ground (GND)</td>
</tr>
<tr>
<td>8</td>
<td>Data carrier detect (DCD)</td>
</tr>
<tr>
<td>9</td>
<td>Reserved for data testing</td>
</tr>
<tr>
<td>10</td>
<td>Reserved for data testing</td>
</tr>
<tr>
<td>11</td>
<td>Unassigned</td>
</tr>
<tr>
<td>12</td>
<td>Secondary data carrier detect</td>
</tr>
<tr>
<td>13</td>
<td>Secondary clear to send</td>
</tr>
<tr>
<td>14</td>
<td>Secondary transmitted data</td>
</tr>
<tr>
<td>15</td>
<td>Transmit signal element timing</td>
</tr>
<tr>
<td>16</td>
<td>Secondary received data</td>
</tr>
<tr>
<td>17</td>
<td>Receive signal element timing</td>
</tr>
<tr>
<td>18</td>
<td>Unassigned</td>
</tr>
<tr>
<td>19</td>
<td>Secondary request to send</td>
</tr>
<tr>
<td>20</td>
<td>Data terminal ready (DTR)</td>
</tr>
<tr>
<td>21</td>
<td>Signal quality detector</td>
</tr>
<tr>
<td>22</td>
<td>Ring indicator</td>
</tr>
<tr>
<td>23</td>
<td>Data signal rate select</td>
</tr>
<tr>
<td>24</td>
<td>Transmit signal element timing</td>
</tr>
<tr>
<td>25</td>
<td>Unassigned</td>
</tr>
</tbody>
</table>
1. DTR (data terminal ready). When the terminal (or a PC COM port) is turned on, after going through a self-test, it sends out signal DTR to indicate that it is ready for communication. If there is something wrong with the COM port, this signal will not be activated. This is an active-LOW signal and can be used to inform the modem that the computer is alive and kicking. This is an output pin from DTE (PC COM port) and an input to the modem.

2. DSR (data set ready). When the DCE (modem) is turned on and has gone through the self-test, it asserts DSR to indicate that it is ready to communicate. Thus, it is an output from the modem (DCE) and an input to the PC (DTE). This is an active-LOW signal. If for any reason the modem cannot make a connection to the telephone, this signal remains inactive, indicating to the PC (or terminal) that it cannot accept or send data.

3. RTS (request to send). When the DTE device (such as a PC) has a byte to transmit, it asserts RTS to signal the modem that it has a byte of data to transmit. RTS is an active-LOW output from the DTE and an input to the modem.

4. CTS (clear to send). In response to RTS, when the modem has room to store the data it is to receive, it sends out signal CTS to the DTE (PC) to indicate that it can receive the data now. This input signal to the DTE is used by the DTE to start transmission.

5. DCD (data carrier detect). The modem asserts signal DCD to inform the DTE (PC) that a valid carrier has been detected and that contact between it and the other modem is established. Therefore, DCD is an output from the modem and an input to the PC (DTE).

6. RI (ring indicator). An output from the modem (DCE) and an input to a PC (DTE) indicates that the telephone is ringing. RI goes on and off in synchronization with the ringing sound. Of the six handshake signals, this is the least often used because modems take care of answering the phone. If in a given system the PC is in charge of answering the phone, however, this signal can be used.

---

**Table 10-2: IBM PC DB-9 Signals**

<table>
<thead>
<tr>
<th>Pin</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Data carrier detect (DCD)</td>
</tr>
<tr>
<td>2</td>
<td>Received data (RxD)</td>
</tr>
<tr>
<td>3</td>
<td>Transmitted data (TxD)</td>
</tr>
<tr>
<td>4</td>
<td>Data terminal ready (DTR)</td>
</tr>
<tr>
<td>5</td>
<td>Signal ground (GND)</td>
</tr>
<tr>
<td>6</td>
<td>Data set ready (DSR)</td>
</tr>
<tr>
<td>7</td>
<td>Request to send (RTS)</td>
</tr>
<tr>
<td>8</td>
<td>Clear to send (CTS)</td>
</tr>
<tr>
<td>9</td>
<td>Ring indicator (RI)</td>
</tr>
</tbody>
</table>

**Figure 10-5. DB-9 9-Pin Connector**

**Figure 10-6. Null Modem Connection**
From the above description, PC and modem communication can be summarized as follows: While signals DTR and DSR are used by the PC and modem, respectively, to indicate that they are alive and well, it is RTS and CTS that actually control the flow of data. When the PC wants to send data it asserts RTS, and in response, the modem, if it is ready (has room) to accept the data, sends back CTS. If, for lack of room, the modem does not activate CTS, the PC will deassert DTR and try again. RTS and CTS are also referred to as hardware control flow signals.

This concludes the description of the most important pins of the RS232 handshake signals plus TX, RX, and ground. Ground is also referred to as SG (signal ground).

**IBM PC/compatible COM ports**

IBM PC/compatible computers based on x86 (8086, 286, 386, 486, and all Pentiums) microprocessors used to have two COM ports. Both COM ports were RS232-type connectors. Many PCs used one each of the DB-25 and DB-9 RS232 connectors. The COM ports were designated as COM 1 and COM 2. In recent years, one of these has been replaced with the USB port, and COM 1 is the only serial port available, if any. We can connect the PIC18 serial port to the COM 1 port of a PC for serial communication experiments. In the absence of a COM port, we can use a COM-to-USB converter module.

With this background in serial communication, we are ready to look at the PIC18. In the next section we discuss the physical connection of the PIC18 and RS232 connector, and in Section 10.3 we see how to program the PIC18 serial communication port.

**Review Questions**

1. The transfer of data using parallel lines is ________ (faster, slower) but ________ (more expensive, less expensive).
2. True or false. Sending data to a printer is duplex.
3. True or false. In full duplex we must have two data lines, one for transfer and one for receive.
4. The start and stop bits are used in the ________ (synchronous, asynchronous) method.
5. Assuming that we are transmitting the ASCII letter “E” (0100 0101 in binary) with no parity bit and one stop bit, show the sequence of bits transferred serially.
6. In Question 5, find the overhead due to framing.
7. Calculate the time it takes to transfer 10,000 characters as in Question 5 if we use 9600 bps. What percentage of time is wasted due to overhead?
8. True or false. RS232 is not TTL compatible.
9. What voltage levels are used for binary 0 in RS232?
10. True or false. The PIC18 has a built-in UART.
11. On the back of x86 PCs, we normally have ___ COM port connectors.
12. The PC COM ports are designated by DOS and Windows as ___ and ___.
SECTION 10.2: PIC18 CONNECTION TO RS232

In this section, the details of the physical connections of the PIC18 to RS232 connectors are given. As stated in Section 10.1, the RS232 standard is not TTL compatible; therefore, a line driver such as the MAX232 chip is required to convert RS232 voltage levels to TTL levels, and vice versa. The interfacing of PIC18 with RS232 connectors via the MAX232 chip is the main topic of this section.

RX and TX pins in the PIC18

The PIC18 has two pins that are used specifically for transferring and receiving data serially. These two pins are called TX and RX and are part of the PORTC group (RC6 and RC7) of the 40-pin package. Pin 25 of the PIC18 (RC7) is assigned to TX and pin 26 (RC6) is designated as RX. These pins are TTL compatible; therefore, they require a line driver to make them RS232 compatible. One such line driver is the MAX232 chip. This is discussed next.

MAX232

Because the RS232 is not compatible with today’s microprocessors and microcontrollers, we need a line driver (voltage converter) to convert the RS232’s signals to TTL voltage levels that will be acceptable to the PIC18’s TX and RX pins. One example of such a converter is MAX232 from Maxim Corp. (www.maxim-ic.com). The MAX232 converts from RS232 voltage levels to TTL voltage levels, and vice versa. One advantage of the MAX232 chip is that it uses a +5 V power source, which is the same as the source voltage for the PIC18. In other words, with a single +5 V power supply we can power both the PIC18 and MAX232, with no need for the dual power supplies that are common in many older systems.

![Figure 10-7. (a) Inside MAX232 and (b) its Connection to the PIC18 (Null Modem)](image)

The MAX232 has two sets of line drivers for transferring and receiving data, as shown in Figure 10-7. The line drivers used for TX are called T1 and T2,
while the line drivers for RX are designated as R1 and R2. In many applications only one of each is used. For example, T1 and R1 are used together for TX and RX of the PIC18, and the second set is left unused. Notice in MAX232 that the T1 line driver has a designation of T1in and T1out on pin numbers 11 and 14, respectively. The T1in pin is the TTL side and is connected to TX of the microcontroller, while T1out is the RS232 side that is connected to the RX pin of the RS232 DB connector. The R1 line driver has a designation of R1in and R1out on pin numbers 13 and 12, respectively. The R1in (pin 13) is the RS232 side that is connected to the TX pin of the RS232 DB connector, and R1out (pin 12) is the TTL side that is connected to the RX pin of the microcontroller. See Figure 10-7. Notice the null modem connection where RX for one is TX for the other.

MAX232 requires four capacitors ranging from 1 to 22 μF. The most widely used value for these capacitors is 22 μF.

**MAX233**

To save board space, some designers use the MAX233 chip from Maxim. The MAX233 performs the same job as the MAX232 but eliminates the need for capacitors. However, the MAX233 chip is much more expensive than the MAX232. Notice that MAX233 and MAX232 are not pin compatible. You cannot take a MAX232 out of a board and replace it with a MAX233. See Figure 10-8 for MAX233 with no capacitor used.

![Figure 10-8. (a) Inside MAX233 and (b) Its Connection to the PIC18 (Null Modem)](image)

**Review Questions**

1. True or false. The PC COM port connector is the RS232 type.
2. Which pins of the PIC18 are set aside for serial communication, and what are their functions?
3. What are line drivers such as MAX 232 used for?
4. MAX232 can support _ _ lines for TX and _ _ lines for RX.
5. What is the advantage of the MAX233 over the MAX232 chip?
SECTION 10.3: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY

In this section we discuss the serial communication registers of the PIC18 and show how to program them to transfer and receive data using asynchronous mode. The USART (universal synchronous asynchronous receiver) in the PIC18 has both the synchronous and asynchronous features. The synchronous mode can be used to transfer data between the PIC and external peripherals such as ADC and EEPROMs. The asynchronous mode is the one we will use to connect the PIC18-based system to the IBM PC serial port for the purpose of full-duplex serial data transfer. In this section we examine the asynchronous mode only. In the PIC microcontroller six major registers are associated with the UART that we deal with in this chapter. They are (a) SPBGR (serial port baud rate generator), (b) TXREG (Transfer register), (c) RCREG (Receive register), (d) TXSTA (transmit status and control register), (e) RCSTA (receive status and control register), and (f) PIR1 (peripheral interrupt request register1). We examine each of them and show how they are used in full-duplex serial data communication.

SPBRG register and baud rate in the PIC18

Because IBM PC/compatible computers are so widely used to communicate with PIC18-based systems, we will emphasize serial communications of the PIC18 with the COM port of the PC. Some of the baud rates supported by PC HyperTerminal are listed in Table 10-3. You can examine these baud rates by going to the Microsoft Windows HyperTerminal program and clicking on the Communication Settings option. The PIC18 transfers and receives data serially at many different baud rates. The baud rate in the PIC18 is programmable. This is done with the help of the 8-bit register called SPBRG. For a given crystal frequency, the value loaded into the SPBRG decides the baud rate. The relation between the value loaded into SPBRG and the Fosc (frequency of oscillator connected to the OSC1 and OSC2 pins) is dictated by the following formula:

\[
\text{Desired Baud Rate} = \frac{\text{Fosc}}{64\times \text{X} + 64} = \frac{\text{Fosc}}{64\times \text{X} + 1}
\]

where X is the value we load into the SPBGR register. Assuming that Fosc = 10 MHz, we have the following:

\[
\text{Desired Baud Rate} = \frac{\text{Fosc}}{64\times \text{X} + 1} = 10 \text{ MHz}/64\times \text{X} + 1 = 6250 \text{ Hz}/(\text{X} + 1)
\]

To get the X value for different baud rates we can solve the equation as follows:

\[
\text{X} = \left(\frac{156250}{\text{Desired Baud Rate}}\right) - 1
\]

Table 10-4 shows the X values for the different baud rates if Fosc = 10 MHz. Another way to understand the SPBRG values in Table 10-4 is to look at
them from the perspective of the instruction cycle time. As we discussed in previous chapters, the PIC18 divides the crystal frequency (Fosc) by 4 to get the instruction cycle time frequency. In the case of XTAL = 10 MHz, the instruction cycle frequency is 2.5 MHz. The PIC18’s UART circuitry divides the instruction cycle frequency by 16 once more before it is used by an internal timer to set the baud rate. Therefore, 2.5 MHz divided by 16 gives 156,250 Hz. This is the number we use to find the SPBRG value shown in Table 10-4. Example 10-1 shows how to verify the data in Table 10-4. Table 10-5 shows the SPBRG values with the crystal frequency of 4 MHz (Fosc = 4 MHz).

**Example 10-1**

With Fosc = 10 MHz, find the BGRP value needed to have the following baud rates:

(a) 9600  
(b) 4800  
(c) 2400  
(d) 1200

**Solution:**

Because Fosc = 10 MHz, we have 10 MHz/4 = 2.5 MHz for the instruction cycle frequency. This is divided by 16 once more before it is used by UART. Therefore, we have 2.5 MHz/16 = 156250 Hz and X = (156250 Hz/Desired Baud Rate) – 1:

(a) (156250/9600) − 1 = 16.27 − 1 = 15.27 = 15 = F (hex) is loaded into SPBRG
(b) (156250/4800) − 1 = 32.55 − 1 = 31.55 = 32 = 20 (hex) is loaded into SPBRG
(c) (156250/2400) − 1 = 65.1 − 1 = 64.1 = 64 = 40 (hex) is loaded into SPBRG
(d) (156250/1200) − 1 = 130.2 − 1 = 129.2 = 129 = 81 (hex) is loaded into SPBRG

Notice that dividing the instruction cycle frequency by 16 is the setting upon Reset. We can get a higher baud rate with the same crystal by changing this default setting. This is done by making bit BRGH = 1 in the TXSTA register. This is explained at the end of this section.

![Diagram of the timing process](image)

**Table 10-4: SPBRG Values for Various Baud Rates (Fosc = 10 MHz, BRGH = 0)**

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG (Decimal Value)</th>
<th>SPBRG (Hex Value)</th>
</tr>
</thead>
<tbody>
<tr>
<td>38400</td>
<td>3</td>
<td>3</td>
</tr>
<tr>
<td>19200</td>
<td>7</td>
<td>7</td>
</tr>
<tr>
<td>9600</td>
<td>15</td>
<td>F</td>
</tr>
<tr>
<td>4800</td>
<td>32</td>
<td>20</td>
</tr>
<tr>
<td>2400</td>
<td>64</td>
<td>40</td>
</tr>
<tr>
<td>1200</td>
<td>129</td>
<td>81</td>
</tr>
</tbody>
</table>

*Note: For Fosc = 10 MHz we have SPBRG = (156,250/BaudRate) – 1*
Table 10-5: SPBRG Values for Various Baud Rates (Fosc = 4 MHz, BRGH = 0)

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG (Decimal Value)</th>
<th>SPBRG (Hex Value)</th>
</tr>
</thead>
<tbody>
<tr>
<td>19200</td>
<td>2</td>
<td>2</td>
</tr>
<tr>
<td>9600</td>
<td>5</td>
<td>5</td>
</tr>
<tr>
<td>4800</td>
<td>12</td>
<td>0C</td>
</tr>
<tr>
<td>2400</td>
<td>25</td>
<td>19</td>
</tr>
<tr>
<td>1200</td>
<td>51</td>
<td>33</td>
</tr>
</tbody>
</table>

*Note: For Fosc = 4 MHz we have 4 MHz/4 = 1 MHz for instruction cycle freq. The frequency used by the UART is 1 MHz/16 = 62,500 Hz. That means SPBRG = (62500/Baud Rate) – 1*

**TXREG register**

TXREG is another 8-bit register used for serial communication in the PIC18. For a byte of data to be transferred via the TX pin, it must be placed in the TXREG register. TXREG is a special function register (SFR) and can be accessed like any other register in the PIC18. Look at the following examples of how this register is accessed:

```assembly
MOVLW 0x41 ;WREG=41H, ASCII for letter 'A'
MOVWF TXREG ;copy WREG into TXREG

MOVFF PORTB, TXREG ;copy PORTB contents into TXREG
```

The moment a byte is written into TXREG, it is fetched into a register called TSR (transmit shift register). The TSR frames the 8-bit data with the start and stop bits and the 10-bit data is transferred serially via the TX pin. Notice that while TXREG is accessible by the programmer, TSR is not accessible and is strictly for internal use.

**RCREG register**

Similarly, when the bits are received serially via the RX pin, the PIC18 deframes them by eliminating the stop and start bits, making a byte out of the data received, and then placing it in the RCREG register. The following code will dump the received byte into PORTB:

```assembly
MOVFF RCREG, PORTB ;copy RXREG to PORTB
```

**TXSTA (transmit status and control register)**

The TXSTA register is an 8-bit register used to select the synchronous/asynchronous modes and data framing size, among other things. Figure 10-9 describes various bits of the TXSTA register. In this textbook we use the asynchronous mode with a data size of 8 bits. The BRGH bit is used to select a higher speed for transmission. The default is lower baud rate transmission. We will examine the higher transmission rate at the end of this chapter. Notice that D6 of the TXSTA register determines the framing of data by specifying the number of bits per character. We use an 8-bit data size. There are some applications for the 9-bit in which the ninth bit can be used as an address.
<table>
<thead>
<tr>
<th>CSRC</th>
<th>TX9</th>
<th>TXEN</th>
<th>SYNC</th>
<th>0</th>
<th>BRGH</th>
<th>TRMT</th>
<th>TX9D</th>
</tr>
</thead>
<tbody>
<tr>
<td>D7</td>
<td>D6</td>
<td>D5</td>
<td>D4</td>
<td>0</td>
<td>D3</td>
<td>D1</td>
<td>D2</td>
</tr>
</tbody>
</table>

- **CSRC** D7: Clock Source Select (not used in asynchronous mode, therefore D7 = 0.)
- **TX9** D6: 9-bit Transmit Enable
  - 1 = Select 9-bit transmission
  - 0 = Select 8-bit transmission (We use this option, therefore D6 = 0.)
- **TXEN** D5: Transmit Enable
  - 1 = Transmit Enabled
  - 0 = Transmit Disabled
  - We turn “on” and “off” this bit in order to start or stop data transfer.
- **SYNC** D4: USART mode Select (We use asynchronous mode, therefore D4 = 0.)
  - 1 = Synchronous
  - 0 = Asynchronous
- **BRGH** D2: High Baud Rate Select
  - 0 = Low Speed (Default)
  - 1 = High Speed
  - We can double the baud rate with the same Fosc. See the end of this section for further discussion on this bit.
- **TRMT** D1: Transmit Shift Register (TSR) Status
  - 1 = TSR empty
  - 0 = TSR full

**The importance of the TSR register.** To transfer a byte of data serially, we write it into TXREG. The TSR (transmit shift register) is an internal register whose job is to get the data from the TXREG, frame it with the start and stop bits, and send it out one bit at a time via the TX pin. When the last bit, which is the stop bit, is transmitted, the TRMT flag is raised to indicate that it is empty and ready for the next byte. When TSR fetches the data from TXREG, it clears the TRMT flag to indicate it is full. Notice that TSR is a parallel-in-serial-out shift register and is not accessible to the programmer. We can only write to TXREG. Whenever the TSR is empty, it gets its data from TXREG and clears the TXREG register immediately, so it does not send out the same data twice.

**TXD9** D0: 9th bit of Transmit Data (Because we use the 8-bit option, we make D0 = 0)
- Can be used as an address/data or a parity bit in some applications

---

**RCSTA (receive status and control register)**

The RCSTA register is an 8-bit register used to enable the serial port to receive data, among other things. Figure 10-10 describes various bits of the RCSTA register. In this section we use the 8-bit data frame.
<table>
<thead>
<tr>
<th>SPEN</th>
<th>RX9</th>
<th>SREN</th>
<th>CREN</th>
<th>ADDE</th>
<th>FERR</th>
<th>OERR</th>
<th>RX9D</th>
</tr>
</thead>
<tbody>
<tr>
<td>D7</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<tr>
<td>D6</td>
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<tr>
<td>D5</td>
<td></td>
<td></td>
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<tr>
<td>D4</td>
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<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D3</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D2</td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
</tr>
<tr>
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<td></td>
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<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D1</td>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
<td></td>
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<tr>
<td></td>
<td></td>
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<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>D0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
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<td></td>
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</tr>
</tbody>
</table>

SPEN D7: Serial port enable bit
1 = Serial port enabled, which makes TX and RX pins as serial port pins
0 = Serial port disabled

RX9 D6: 9-bit Receive enable bit
1 = Select 9-bit reception
0 = Select 8-bit reception (We use this option; therefore, D6 = 0.)

SREN D5: Single receive enable bit (not used in asynchronous mode D5 = 0)

CREN D4: Continuous receive enable bit
1 = Enable continuous Receive (in asynchronous mode)
0 = Disable continuous Receive (in asynchronous mode)

ADDEN D3: Address delete enable bit (Because used with the 9-bit data frame D3 = 0)

FERR D2: Framing error bit
1 = Framing error
0 = No Framing error

OERR D1: Overrun error bit
1 = Overrun error
0 = No overrun error

TXD9 D0: 9th bit of Receive data (Because we use the 8-bit option, we make D0 = 0)
Can be used as an address/data or a parity bit in some applications.

Figure 10-10. RCSTA: Receive Status and Control Register

--- | RCIF | TXIF | --- | --- | --- | --- |

RCIF: Receive interrupt flag bit
1 = The UART has received a byte of data and it is sitting in the
RCREG register (receive buffer), waiting to be picked up.
Upon reading the RCREG register, the RCIF is cleared to allow the
next byte to be received.
0 = The RCREG is empty.

TXIF: Transmit interrupt flag bit
0 = The TXREG register is full.
1 = The TXREG (transmit buffer) register is empty.

The importance of TXIF: To transmit a byte of data, we write it into TXREG. Upon
writing a byte into TXREG, the TXIF flag is cleared. When the entire byte is transmitt-
ed via the TX pin, the TXIF flag bit is raised to indicate that it is ready for the next
byte. So, we must monitor this flag before we write a new byte into TXREG, otherwise,
we wipe out the last byte before it is transmitted.

Several bits of this register are used by the timer flag, as we saw in Chapter 9. The
location of the flag bits in the PIR1 register is not fixed and can vary in future PIC18
products.

Figure 10-11. PIR1 (Peripheral Interrupt Register 1)
PIR1 (peripheral interrupt request register 1)

In Chapter 9, we saw how some of the bits of PIR1 are used by the timers. Two of the PIR1 register bits are used by the UART. They are TXIF (transmit interrupt flag) and RCIF (receive interrupt flag). See Figure 10-11. We monitor (poll) the TXIF flag bit to make sure that all the bits of the last byte are transmitted before we write another byte into the TXREG. By the same logic, we monitor the RCIF flag to see if a byte of data has come in yet. In Chapter 11 we will see how these flags are used with interrupts instead of polling. Next we will examine how TXIF flags are used in serial data transfer.

Programming the PIC18 to transfer data serially

In programming the PIC18 to transfer character bytes serially, the following steps must be taken:

1. The TXSTA register is loaded with the value 20H, indicating asynchronous mode with 8-bit data frame, low baud rate, and transmit enabled.
2. Make TX pin of PORTC (RC6) an output for data to come out of the PIC.
3. The SPBRG is loaded with one of the values in Table 10-4 (or Table 10-5 if Fosc = 4 MHz) to set the baud rate for serial data transfer.
4. SPEN bit in the RCSTA register is set HIGH to enable the serial port of the PIC18.
5. The character byte to be transmitted serially is written into the TXREG register.
6. Monitor the TXIF bit of the PIR1 register to make sure UART is ready for next byte.
7. To transfer the next character, go to Step 5.

Example 10-2 shows the program to transfer data serially at 9600 baud. Example 10-3 shows how to transfer “YES” continuously.

**Example 10-2**

Write a program for the PIC18 to transfer the letter ‘G’ serially at 9600 baud, continuously. Assume XTAL = 10 MHz.

**Solution:**

```
MOVLW B'00100000'; enable transmit and choose low baud rate
MOVWF TXSTA ; write to reg
MOVLW D'15'; 9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ; write to reg
BCF TRISC, TX ; make TX pin of PORTC an output pin
BSF RCSTA, SPEN ; enable the entire serial port of PIC18
OVER
MOVLW A'G'; ASCII letter 'G' to be transferred
S1
BTFSS PIR1, TXIF ; wait until the last bit is gone
BRA S1 ; stay in loop
MOVWF TXREG ; load the value to be transferred
BRA OVER ; keep sending letter 'G'
```

Example 10-3

Write a program to transmit the message “YES” serially at 9600 baud, 8-bit data, and 1 stop bit. Do this forever.

Solution:

```
MOVLW B'00100000'; enable transmit and choose low baud
MOVF TXSTA ; write to reg
MOVLW D'15'; 9600 bps (Fosc / (64 * Speed) - 1)
MOVF SPBRG ; write to reg
BCF TRISC, TX ; make TX pin of PORTC an output pin
BSP RCSTA, SPEN ; enable the serial port
OVER
    MOVLW A'Y'; ASCII letter 'Y' to be transferred
    CALL TRANS
MCALL A'E'; ASCII letter 'E' to be transferred
CALL TRANS
    MOVLW A'S'; ASCII letter 'S' to be transferred
CALL TRANS
    MOVLW 0x0; NULL to purge the buffer
CALL TRANS
    BRA OVER ; keep doing it
TRANS ;---- serial data transfer subroutine
S1
    BTFSS PIR1, TXIF ; wait until the last bit is gone
    BRA S1 ; stay in loop
    MOVF TXREG ; load the value to be transmitted
RETURN ; return to caller
```

Importance of the TXIF flag

To understand the importance of the role of TXIF, look at the following sequence of steps that the PIC18 goes through in transmitting a character via TX:

1. The byte character to be transmitted is written into the TXREG register.
2. The TXIF flag is set to 1 internally to indicate that TXREG has a byte and will not accept another byte until this one is transmitted.
3. The TSR (Transmit Shift Register) reads the byte from TXREG and begins to transfer the byte starting with the start bit.
4. The TXIF is cleared to indicate that the last byte is being transmitted and TXREG is ready to accept another byte.
5. The 8-bit character is transferred one bit at a time.
6. By monitoring the TXIF flag, we make sure that we are not overloading the TXREG register. If we write another byte into the TXREG register before the TSR has fetched the last one, the old byte could be lost before it is transmitted.

From the above discussion we conclude that by checking the TXIF flag bit, we know whether or not the PIC18 is ready to transfer another byte. The TXIF flag bit can be checked by the instruction “BTFSS PIR1, TXIF” or we can use an interrupt, as we will see in Chapter 11. In Chapter 11 we will show how to use interrupts to transfer data serially, and avoid tying down the microcontroller with instructions such as “BTFSS PIR1, TXIF”.

CHAPTER 10: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY AND C 403
Programming the PIC18 to receive data serially

In programming the PIC18 to receive character bytes serially, the following steps must be taken:

1. The RCSTA register is loaded with the value 90H, to enable the continuous receive in addition to the 8-bit data size option.
2. The TXSTA register is loaded with the value 00H to choose the low baud rate option.
3. SPBRG is loaded with one of the values in Table 10-4 to set the baud rate (assuming XTAL = 10 MHz).
4. Make the RX pin of PORTC (RC7) an input for data to come into the PIC18.
5. The RCIF flag bit of the PIR1 register is monitored for a HIGH to see if an entire character has been received yet.
6. When RCIF is raised, the RCREG register has the byte. Its contents are moved into a safe place.
7. To receive the next character, go to Step 5.

Example 10-4 shows the coding of the above steps.

Example 10-4

Program the PIC18 to receive bytes of data serially and put them on PORTB. Set the baud rate at 9600, 8-bit data, and 1 stop bit.

Solution:

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOVLW B'10010000'</td>
<td>enable receive and serial port itself</td>
</tr>
<tr>
<td>MOVWF RCSTA</td>
<td>write to reg</td>
</tr>
<tr>
<td>MOVLW D'15'</td>
<td>9600 bps (Fosc / (64 * Speed) - 1)</td>
</tr>
<tr>
<td>MOVWF SPBRG</td>
<td>write to reg</td>
</tr>
<tr>
<td>BSF TRISC, RX</td>
<td>make RX pin of PORTC an input pin</td>
</tr>
<tr>
<td>CLRF TRISB</td>
<td>make port B an output port</td>
</tr>
<tr>
<td>;get a byte from serial port and place it on PORTB</td>
<td></td>
</tr>
<tr>
<td>R1</td>
<td>BTFS PIR1, RCIF</td>
</tr>
<tr>
<td></td>
<td>BRA R1</td>
</tr>
<tr>
<td></td>
<td>MOVFF RCREG, PORTB</td>
</tr>
<tr>
<td></td>
<td>BRA R1</td>
</tr>
</tbody>
</table>

Importance of the RCIF flag bit

In receiving bits via its RX pin, the PIC18 goes through the following steps:

1. It receives the start bit indicating that the next bit is the first bit of the character byte it is about to receive.
2. The 8-bit character is received one bit at time. When the last bit is received, a byte is formed and placed in RCREG.
3. The stop bit is received. It is during receiving the stop bit that the PIC18 makes RCIF = 1, indicating that an entire character byte has been received and must
4. By checking the RCIF flag bit when it is raised, we know that a character has been received and is sitting in the RCREG register. We copy the RCREG contents to a safe place in some other register or memory before it is lost.

5. After the RCREG contents are read (copied) into a safe place, the RCIF flag bit is forced to 0 by the UART itself. This allows the next received character byte to be placed in RCREG, and also prevents the same byte from being picked up multiple times.

From the above discussion we conclude that by checking the RCIF flag bit we know whether or not the PIC18 has received a character byte. If we fail to copy RCREG into a safe place, we risk the loss of the received byte. More importantly, note that the RCIF flag bit is raised by the PIC18, and it is also cleared by the CPU when the data in the RCREG is picked up. Note also that if we copy RCREG into a safe place before the RCIF flag bit is raised, we risk copying garbage. The RCIF flag bit can be checked by the instruction “BTFSS PIR1, RCIF” or by using an interrupt, as we will see in Chapter 11.

**Quadruple the baud rate in the PIC18**

There are two ways to increase the baud rate of data transfer in the PIC18:

1. Use a higher-frequency crystal.
2. Change a bit in the TXSTA register, as shown below.

Option 1 is not feasible in many situations because the system crystal is fixed. Therefore, we will explore option 2. There is a software way to quadruple the baud rate of the PIC18 while the crystal frequency stays the same. This is done with the BRGH bit of the TXSTA register. When the PIC18 is powered up, D2 (BRGH bit) of the TXSTA register is zero. We can set it to high by software and thereby quadruple the baud rate.

To see how the baud rate is quadrupled with this method, we show the role of the BRGH bit (D2 bit of the TXSTA register), which can be 0 or 1. We discuss each case.

**Baud rates for BRGH = 0**

When BRGH = 0, the PIC18 divides Fosc/4 (crystal frequency) by 16 once more and uses that frequency for UART to set the baud rate. In the case of XTAL = 10 MHz we have:

Instruction cycle freq. = 10 MHz / 4 = 2.5 kHz

and

2.5 MHz / 16 = 156,250 Hz because BRGH = 0

This is the frequency used by UART to set the baud rate. This has been the basis of all the examples so far because it is the default when the PIC18 is powered up. The baud rate for BRGH = 0 was listed in Table 10-4 and Table 10-5.
Baud rates for BRGH = 1

With the fixed crystal frequency, we can quadruple the baud rate by making BRGH = 1. When the BRGH bit (D2 of the TXSTA register) is set to 1, Fosc/4 of XTAL is divided by 4 (instead of 16) once more, and that is the frequency used by UART to set the baud rate. In the case of XTAL = 10 MHz, we have:

Instruction cycle freq. = 10 MHz / 4 = 2.5 MHz

and

2.5 MHz / 4 = 625000 Hz because BHRG = 1

This is the frequency used by UART to set the baud rate if BHRH = 1.

Table 10-8 shows that the values loaded into SPBREG are the same for both cases; however, the baud rates are quadrupled when BRGH = 1. Look at Examples 10-5 through 10-7 to clarify the data given in Tables 10-6 and 10-7.

### Table 10-6: SPBRG Values for Various Baud Rates (Fosc = 10MHz and BRGH = 1)

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG (Decimal Value)</th>
<th>SPBRG (Hex Value)</th>
</tr>
</thead>
<tbody>
<tr>
<td>57600</td>
<td>10</td>
<td>0A</td>
</tr>
<tr>
<td>38400</td>
<td>15</td>
<td>0F</td>
</tr>
<tr>
<td>19200</td>
<td>32</td>
<td>20</td>
</tr>
<tr>
<td>9600</td>
<td>64</td>
<td>40</td>
</tr>
<tr>
<td>4800</td>
<td>129</td>
<td>81</td>
</tr>
</tbody>
</table>

*Note: For Fosc = 10 MHz we have SPBRG = (625000/Baud Rate) – 1*

### Example 10-5

Find the SPBRG value (in both decimal and hex) to set the baud rate to each of the following:

(a) 9600 if BRGH = 1
(b) 4800 if BRGH = 1

Assume that XTAL = 10 MHz.

**Solution:**

With XTAL = 10 MHz, Fosc/4 = 2.5 MHz. Because BRGH = 1, we have UART frequency = 2.5 MHz/4 = 625,000 Hz.

(a) \((625,500 / 9600) - 1 = 64\); therefore, SPBRG = 64 or SPBRG = 40H (in hex).

(b) \((625,500 / 4800) - 1 = 129\); therefore, SPBRG = 129 or SPBRG = 81H (in hex).
Table 10-7: SPBRG Values for Various Baud Rates (XTAL = 10 MHz)

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG (Decimal)</th>
<th>BRGH = 0</th>
<th>SPBRG (Decimal)</th>
<th>BRGH = 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>57600</td>
<td>2</td>
<td></td>
<td>10</td>
<td></td>
</tr>
<tr>
<td>38400</td>
<td>3</td>
<td></td>
<td>15</td>
<td></td>
</tr>
<tr>
<td>19200</td>
<td>7</td>
<td></td>
<td>32</td>
<td></td>
</tr>
<tr>
<td>9,600</td>
<td>15</td>
<td></td>
<td>64</td>
<td></td>
</tr>
<tr>
<td>4,800</td>
<td>32</td>
<td></td>
<td>129</td>
<td></td>
</tr>
</tbody>
</table>

SPBRG = (156250/Baud rate) – 1

Table 10-8: SPBRG Values vs. Baud Rates for BRGH = 0 and BRGH = 1 (XTAL = 10 MHz)

<table>
<thead>
<tr>
<th>SPBRG (Decimal)</th>
<th>Baud Rate</th>
<th>BRGH = 0</th>
<th>Baud Rate</th>
</tr>
</thead>
<tbody>
<tr>
<td>15</td>
<td>9600</td>
<td></td>
<td>38400</td>
</tr>
<tr>
<td>32</td>
<td>4800</td>
<td></td>
<td>19200</td>
</tr>
<tr>
<td>64</td>
<td>2400</td>
<td></td>
<td>9600</td>
</tr>
</tbody>
</table>

Table 10-9: SPBRG Values for Various Baud Rates (XTAL = 4 MHz)

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG (Decimal)</th>
<th>BRGH = 0</th>
<th>SPBRG (Decimal)</th>
<th>BRGH = 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>19200</td>
<td>3</td>
<td></td>
<td>12</td>
<td></td>
</tr>
<tr>
<td>9,600</td>
<td>6</td>
<td></td>
<td>25</td>
<td></td>
</tr>
<tr>
<td>4,800</td>
<td>12</td>
<td></td>
<td>51</td>
<td></td>
</tr>
<tr>
<td>2,400</td>
<td>25</td>
<td></td>
<td>103</td>
<td></td>
</tr>
</tbody>
</table>

SPBRG = (62500/Baud rate) – 1

Example 10-6

Write a program for the PIC18 to transfer the letter ‘G’ serially at 57600 baud, continuously. Assume XTAL = 10 MHz. Use the BRGH = 1 mode.

Solution:

MOV L W B'00100100' ;enable transmit and choose high baud rate
MOV W F TXSTA ;write to reg
MOV L W D'10' ;57600 bps (Fosc / (16 * Speed) - 1)
MOV W F SPBRG ;write to reg
BCF TRISC, TX ;make TX pin of PORTC an output pin
BSF RCSTA, SPEN ;enable the entire Serial port of PIC18
OVER MOV L W A'G' ;ASCII letter 'G' to be transferred
S1 BTF S S PIR1, TXIF ;wait until the last bit is gone
BRA S1 ;stay in loop
MOV W F TXREG ;load the value to be transferred
BRA OVER ;keep sending letter 'G'

CHAPTER 10: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY AND C 407
Baud rate error calculation

In calculating the baud rate we have used the integer number for the SPBRG register values because PIC microcontrollers can only use integer values. By dropping the decimal portion of the calculated values we run the risk of introducing error into the baud rate. There are several ways to calculate this error. One way would be to use the following formula.

\[
\text{Error} = \frac{\text{Calculated value for the SPBRG} - \text{Integer part}}{\text{Integer part}}
\]

For example, with the XTAL = 10 MHz and BRGH = 0 we have the following for the 9600 baud rate:

\[
\text{SPBRG value} = \frac{156250}{9600} - 1 = 16.27 - 1 = 15.27 = 15
\]

and the error is

\[
(15.27 - 15)/16 = 1.7\%
\]

Another way to calculate the error rate is as follows:

\[
\text{Error} = \frac{\text{calculated baud rate} - \text{desired baud rate}}{\text{desired baud rate}}
\]

Example 10-7

Assuming XTAL = 10 MHz, calculate the baud rate error for the following:
(a) 2400    (b) 1200    (c) 19200    (d) 57600
Use the BRGH = 0 mode.

Solution:

(a) SPBRG Value = \(\frac{156250}{2400} - 1 = 65.1 - 1 = 64.1 = 64\)

\[
\text{Error} = \frac{64.1 - 64}{65} = 0.15\%
\]

(b) SPBRG Value \(\frac{156250}{1200} - 1 = 130.2 - 1 = 129.2 = 129\)

\[
\text{Error} = \frac{129.2 - 129}{130} = 0.15\%
\]

(c) SPBRG Value \(\frac{156250}{19200} - 1 = 8.138 - 1 = 7.138 = 7\)

\[
\text{Error} = \frac{7.138 - 7}{8} = 1.7\%
\]

(d) SPBRG Value \(\frac{156250}{57600} - 1 = 2.71 - 1 = 1.71 = 1\)

\[
\text{Error} = \frac{1.71 - 1}{2} = 35\%
\]

Such an error rate is too high. Let's round up the number to see what happens.

\[
\text{Error} = \frac{3 - 2.7}{3} = 10\%
\]

This means we use SPBRG = 2 instead of SPBRG = 1.
where the desired baud rate is calculated using \( X = \left( \frac{\text{Fosc}}{\text{Desired Baud rate}} \right) 64 - 1 \) and then the integer \( X \) (value loaded into SPBRG reg) is used for the calculated baud rate as follows:

\[
\text{calculated baud rate} = \frac{\text{Fosc}}{(64(\text{X} + 1))} \quad \text{(for BRGH = 0)}
\]

For \( \text{XTAL} = 10 \text{ MHz} \) and 9600 baud rate, we got \( X = 15 \). Therefore, we get the calculated baud rate of \( 10 \text{ MHz}/(64(15 + 1)) = 9765 \). Now the error is calculated as follows:

\[
\text{Error} = \frac{(9765 - 9600)}{9600} = 1.7\%
\]

which is the same as what we got earlier using the other method. Tables 10-10 and 10-11 provide the error rates for SPBRG values of 10 MHz and 4 MHz crystal frequencies, respectively. Compare Examples 10-7 and 10-8 to see how to calculate the error rates two different ways.

**Example 10-8**

Assuming \( \text{XTAL} = 10 \text{ MHz} \), calculate the baud rate error for the following:

(a) 2400  
(b) 1200

Assume \( \text{BRGH} = 0 \)

**Solution:**

(a) SPBRG Value \( \left( \frac{156250}{2400} \right) - 1 = 65.1 - 1 = 64.1 = 64 \)

and calculated baud rate is \( 156250/(64 + 1) = 2403 \)

\[
\text{Error} = \frac{(2403 - 2400)}{2400} = 0.12\%
\]

(b) SPBRG Value \( \left( \frac{156250}{1200} \right) - 1 = 130.2 - 1 = 129.2 = 129 \)

where the calculated baud rate is \( 156250/(129 + 1) = 1202 \)

\[
\text{Error} = \frac{(1202 - 1200)}{1200} = 0.16\%
\]

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG</th>
<th>Error</th>
<th>SPBRG</th>
<th>Error</th>
</tr>
</thead>
<tbody>
<tr>
<td>38400</td>
<td>3</td>
<td>1.5%</td>
<td>15</td>
<td>1.7%</td>
</tr>
<tr>
<td>19200</td>
<td>7</td>
<td>1.7%</td>
<td>32</td>
<td>1.3%</td>
</tr>
<tr>
<td>9,600</td>
<td>15</td>
<td>1.7%</td>
<td>64</td>
<td>0.15%</td>
</tr>
<tr>
<td>4,800</td>
<td>32</td>
<td>1.3%</td>
<td>129</td>
<td>0.15%</td>
</tr>
</tbody>
</table>

\[
\text{SPBRG} = \left( \frac{156250}{\text{Baud rate}} \right) - 1
\]

<table>
<thead>
<tr>
<th>Baud Rate</th>
<th>SPBRG</th>
<th>Error</th>
<th>SPBRG</th>
<th>Error</th>
</tr>
</thead>
<tbody>
<tr>
<td>19200</td>
<td>2</td>
<td>8.3%</td>
<td>12</td>
<td>0.15%</td>
</tr>
<tr>
<td>9,600</td>
<td>6</td>
<td>8%</td>
<td>25</td>
<td>0.15%</td>
</tr>
<tr>
<td>4,800</td>
<td>12</td>
<td>0.15%</td>
<td>51</td>
<td>0.15%</td>
</tr>
<tr>
<td>2,400</td>
<td>25</td>
<td>0.16%</td>
<td>103</td>
<td>0.16%</td>
</tr>
</tbody>
</table>

\[
\text{SPBRG} = \left( \frac{625000}{\text{Baud rate}} \right) - 1
\]
Examine the next few examples to master the topic of PIC18 serial port programming.

**Example 10-9**

Assume a switch is connected to pin RD7. Write a program to monitor its status and send two messages to the serial port continuously as follows:

- **SW = 0** send “NO”
- **SW = 1** send “YES”

Assume XTAL = 10 MHz, and set the baud rate to 9,600.

**Solution:**

```assembly
BSF TRISD, 7 ; PORTD.7 as in input for SW
MOVLW 0x20 ; enable transmit and choose low baud rate
MOVF TXSTA ; write to reg
MOVLW D'15' ; 9600 bps (Fosc / (64 * Speed) - 1)
MOVF SPBRG ; write to reg
BCF TRISC, TX ; make TX pin of PORTC an output pin
BSF RCSTA, SPEN ; enable the entire serial port of PIC18

OVER BTFSS PORTD, 7
BRA NEXT

MOVLW high(MESS1) ; if SW = 0 display "NO"
MOVF TBLPTRH
MOVLW low(MESS1)
MOVF TBLPTRL

PN TBLRD*+ ; read the character
MOVF TABLAT, W
BZ EXIT ; check for end of line
CALL SENDCOM ; send character to serial port
BRA FN ; repeat

NEXT MOVLW high(MESS2) ; if SW = 1 display "YES"
MOVF TBLPTRH
MOVLW low(MESS2)
MOVF TBLPTRL

LN TBLRD*+ ; read the character
MOVF TABLAT, W
BZ EXIT ; check for end of line
CALL SENDCOM ; send character to serial port
BRA LN ; repeat

EXIT MOVLW 0x20 ; send space
CALL SENDCOM
GOTO OVER

; ------------
SENDCOM
S1 BTFSS PIR1, TXIF ; wait until the last bit is gone
BRA S1 ; stay in loop
MOVF TXREG ; load the value to be transferred
RETURN ; return to caller

; ------------
MESS1 DB "NO", 0
MESS2 DB "YES", 0
```

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Example 10-10

Write a program to send the message “The Earth is but One Country” to the serial port continuously. Assume a SW is connected to pin RB0. Monitor its status and set the baud rate as follows:

SW = 0 9600 baud rate
SW = 1 38400 baud rate

Assume XTAL = 10 MHz.

Solution:

As shown in Table 10-8, we can quadruple the baud rate by changing the BRGH bit of the TXSTA register.

```asm
BSF TRISB,0 ;PORTB.0 as in input for SW
BCF TRISC, TX ;make TX pin of PORTC an output pin
BSF RCSTA, SPEN ;enable the entire serial port of PIC18
MOVLW 0x20 ;transmit at low baud rate
MOVWF TXSTA ;write to reg
MOVLW D'15' ;9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ;write to reg
OVER BTFSB PORTB,0 ;test bit PORTB.0 and skip if LOW
BSF TXSTA,BRGH ;transmit at high rate by making BRGH = 1
MOVLW upper(MESSAGE)
MOVWF TBLPTRU
MOVLW high(MESSAGE)
MOVWF TBLPTRH
MOVLW low(MESSAGE)
MOVWF TBLPTRL
NEXT TBLRD*;+ ;read the character
MOVF TABLAT,W ;place it in WREG
BZ NEXT ;if end of line, start over
CALL SENDCOM ;send char to serial port
BRA NEXT ;repeat for the next character
SENDCOM
S1 BTFSS PIR1, TXIF ;wait until the last bit is gone
BRA S1 ;stay in loop
MOVF TXREG ;load the value to be transmitted
RETURN ;return to caller

MESSAGE DB "The Earth is but One Country",0
```
Transmit and receive

Assume that the PIC18 serial port is connected to the COM port of the IBM PC, and we are using the HyperTerminal program on the PC to send and receive data serially. The ports PORTB and PORTD of the PIC18 are connected to LEDs and switches, respectively. Program 10-1 shows a PIC18 program with the following parts: (a) sends the message "YES" once to the PC screen, (b) gets data on switches and transmits it via the serial port to the PC’s screen, and (c) receives any key press sent by HyperTerminal and puts it on LEDs. The program performs part (a) once, but parts (b) and (c) continuously. It uses the 9600 baud rate for XTAL = 10 MHz.

;Program 10-1 Transmit and Receive
ORG 0
;initialize the serial ports for both transmit and receive
MOVLW B'00100100' ;enable transmit, choose high baud
MOVWF TXSTA ;write to reg
MOVLW B'10010000' ;enable receive, serial port itself
MOVWF RCSTA ;write to reg
MOVWF D'15' ;9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ;write to reg
BSF RCSTA, SPEN ;enable the serial port itself
BCF TRISC, TX ;make TX pin of PORTC an output
BSF TRISC, RX ;make RX pin of PORTC an input
CLRF TRISB ;make port B an output port
SETF TRISD ;make port D an input port
;send the message "YES"
MOVLW 'Y' ;ASCII letter 'Y' to be transferred
CALL TRANS
MOVLW 'E' ;ASCII letter 'E' to be transferred
CALL TRANS
MOVLW 'S' ;ASCII letter 'S' to be transferred
CALL TRANS
;get a byte from switches and transmit data to PC screen
OVER MOVF PORTD,W ;get a byte from SW of PORTD
CALL TRANS ;transmit it via serial port
;as keys are pressed on PC receive data and put it on LEDs
CALL RECV ;receive the byte from serial port
MOVWF PORTB ;display it on LEDs of PORTB
BRA OVER ;keep doing it
;--serial transfer (WREG needs the byte to be transmitted)
TRANS
S1 BTFSS PIR1, TXIF ;wait until the last bit is gone
BRA S1 ;
MOVWF TXREG ;load the value to be transferred
RETURN ;return to caller
;----serial data receive subroutine (WREG = received byte)
Interrupt-based data transfer

By now you might have noticed that it is a waste of the microcontroller’s time to poll the TXIF and RXIF flags. In order to avoid wasting the microcontroller’s time we use interrupts instead of polling. In Chapter 11, we will show how to use interrupts to program the PIC18’s serial communication port.

Review Questions

1. Which register of the PIC18 is used to set the baud rate?
2. If XTAL = 10 MHz, what frequency is used by the UART to set the baud rate (assuming default mode)?
3. Which bit of the TXSTA register is used to set the low or high baud rate?
4. With XTAL = 10 MHz, what value should be loaded into SPBRG to have a 9600 baud rate? Give the answer in both decimal and hex.
5. To transmit a byte of data serially, it must be placed in register ________.
6. TXSTA stands for ________ and it is a(n) ______-bit register.
7. Which register is used to set the data frame size?
8. True or false. TXSTA is a bit-addressable register.
9. When is TXIF raised? When is it cleared?
10. Which register has the BRGH bit, and what is its status when the PIC18 is powered up?
SECTION 10.4: PIC18 SERIAL PORT PROGRAMMING IN C

This section shows C programming of the serial ports for the PIC18 chip.

Transmitting and receiving data in PIC18 C

As we saw in Chapter 7, all the special function registers (SFRs) of the PIC18 are accessible directly in C18 compilers by using the appropriate header file. Examples 10-11 through 10-15 show how to program the serial port in PIC18 C. Connect your PIC18 Trainer to the PC’s COM port and use HyperTerminal to test the operation of these examples. Notice that Examples 10-11 through 10-15 are C versions of the Assembly programs in the last section.

Example 10-11

Write a C program for the PIC18 to transfer the letter ‘G’ serially at 9600 baud, continuously. Use 8-bit data and 1 stop bit. Assume XTAL = 10 MHz.

Solution:

```c
#include <P18F4580.h>

void main(void)
{
    TXSTA=0x20; //choose low baud rate, 8-bit
    SPBRG=15;   //9600 baud rate/ XTAL = 10 MHz
    TXSTAbits.TXEN=1;
    RXSTAbits.SPEN=1;

    while(1)
    {
        TXREG='G'; //place value in buffer
        while(PIR1bits.TXIF==0); //wait until all gone
    }
}
```

Review Questions

1. True or false. All the SFR registers of PIC18 are accessible in the C18 C compiler.
2. True or false. C18 compilers support the bit-addressable registers of the PIC18.
3. True or false. The TXIF flag is cleared the moment we write a character to the TXREG register.
4. Which register is used to set the baud rate?
5. To which register does the BRGH bit belong, and what is its role?
Example 10-12

Write a PIC18 C program to transfer the message "YES" serially at 9600 baud, 8-bit data, and 1 stop bit. Do this continuously.

Solution:
#include <P18F458.h>
void SerTx(unsigned char);
void main(void)
{
    TXSTA=0x20; //choose low baud rate, 8-bit
    SPBRG=15; //9600 baud rate/ XTAL = 10 MHz
    TXSTA.bitS=1;
    RCSTA.bitSPEN=1;
    while(1)
    {
        SerTx('Y');
        SerTx('E');
        SerTx('S');
    }
}
void SerTx(unsigned char c)
{
    while(PIR1.bitTXIF==0); //wait until transmitted
    TXREG=c; //place character in buffer
}

Example 10-13

Program the PIC18 in C to receive bytes of data serially and put them on PORTB. Set the baud rate at 9600, 8-bit data, and 1 stop bit.

Solution:
#include <P18F458.h>
void main (void)
{
    TRISB = 0; //PORTB an output
    RCSTA=0x90; //enable serial port and receiver
    SPBRG=15; //9600 baud rate/ XTAL = 10 MHz
    while(1) //repeat forever
    {
        while(PIR1.bitRCIF==0); //wait to receive
        PORTB=RCREG; //save value
    }
}
Write an C18 program to send two different strings to the serial port. Assuming that SW is connected to pin PORTB.5, monitor its status and make a decision as follows:
SW = 0: send your first name
SW = 1: send your last name
Assume XTAL = 10 MHz, baud rate of 9600, and 8-bit data.

Solution:

```c
#include <P18F458.h>
#define MYSW PORTBbits.RB5     //INPUT SWITCH

void main(void)
{
    unsigned char z;
    unsigned char fname[]="ALI";
    unsigned char lname[]="SMITH";
    TRISBbits.TRISB5 = 1; //an input
    TXSTA=0x20;         //choose low baud rate, 8-bit
    SPBRG=15;          //9600 baud rate/ XTAL = 10 MHz
    TXSTAbits.TXEN=1;
    RCSTAbits.SPEN=1;
    if(MYSW==0)         //check switch
    {
        for(z=0;z<3;z++)     //write name
        {
            while(PIR1bits.TXIF==0); //wait for transmit
            TXREG=fname[z];       //place char in buffer
        }
    }
    else
    {
        for(z=0;z<5;z++)     //write name
        {
            while(PIR1bits.TXIF==0); //wait for transmit
            TXREG=lname[z];       //place value in buffer
        }
        while(1);
    }
```
Write a PIC18 C program to send the two messages “Normal Speed” and “High Speed” to the serial port. Assuming that SW is connected to pin PORTB.0, monitor its status and set the baud rate as follows:

- SW = 0 9600 baud rate
- SW = 1 38400 baud rate

Assume that XTAL = 10 MHz for both cases.

Solution:

```c
#include <P18F458.h>
#define MYSW PORTBbits.RB5 //INPUT SWITCH

void main(void)
{
    unsigned char z;
    unsigned char Mess1[]="Normal Speed";
    unsigned char Mess2[]="High Speed";
    TRISBbits.TRISB5 = 1; //an input
    TXSTA=0x20; //choose low baud rate, 8-bit
    SPBRG=15; //9600 baud rate/ XTAL = 10 MHz
    TXSTAbits.TXEN=1;
    RCSTAbits.SPEN=1;
    if(MYSW==0)
    {
        for(z=0;z<12;z++)
        {
            while(PIR1bits.TXIF==0); //wait for transmit
            TXREG=Mess1[z]; //place value in buffer
        }
    }
    else
    {
        TXSTA=TXSTA|0x4; //for high speed
        for(z=0;z<10;z++)
        {
            while(PIR1bits.TXIF==0); //wait for transmit
            TXREG=Mess2[z]; //place value in buffer
        }
        while(1);
    }
}
```

CHAPTER 10: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY AND C 417
SUMMARY

This chapter began with an introduction to the fundamentals of serial communication. Serial communication, in which data is sent one bit at a time, is used in situations where data is sent over significant distances because in parallel communication, where data is sent a byte or more at a time, great distances can cause distortion of the data. Serial communication has the additional advantage of allowing transmission over phone lines. Serial communication uses two methods: synchronous and asynchronous. In synchronous communication, data is sent in blocks of bytes; in asynchronous, data is sent one byte at a time. Data communication can be simplex (can send but cannot receive), half duplex (can send and receive, but not at the same time), or full duplex (can send and receive at the same time). RS232 is a standard for serial communication connectors.

The PIC18's UART was discussed. We showed how to interface the PIC18 with an RS232 connector and change the baud rate of the PIC18. In addition, we described the serial communication features of the PIC18, and programmed the PIC18 for serial data communication. We also showed how to program the serial port of the PIC18 chip in Assembly and C.

PROBLEMS

SECTION 10.1: BASICS OF SERIAL COMMUNICATION

1. Which is more expensive, parallel or serial data transfer?
2. True or false. 0- and 5-V digital pulses can be transferred on the telephone without being converted (modulated).
3. Show the framing of the letter ASCII ‘Z’ (0101 1010), no parity, 1 stop bit.
4. If there is no data transfer and the line is high, it is called ________ (mark, space).
5. True or false. The stop bit can be 1, 2, or none at all.
6. Calculate the overhead percentage if the data size is 7, 1 stop bit, no parity.
7. True or false. RS232 voltage specification is TTL compatible.
8. What is the function of the MAX 232 chip?
9. True or false. DB-25 and DB-9 are pin compatible for the first 9 pins.
10. How many pins of the RS232 are used by the IBM serial cable, and why?
11. True or false. The longer the cable, the higher the data transfer baud rate.
12. State the absolute minimum number of signals needed to transfer data between two PCs connected serially. What are those?
13. If two PCs are connected through the RS232 without the modem, both are configured as a _______ (DTE, DCE) -to- _______ (DTE, DCE) connection.
14. State the nine most important signals of the RS232.
15. Calculate the total number of bits transferred if 200 pages of ASCII data are sent using asynchronous serial data transfer. Assume a data size of 8 bits, 1 stop bit, and no parity. Assume each page has 80x25 of text characters.
16. In Problem 15, how long will the data transfer take if the baud rate is 9,600?
SECTION 10.2: PIC18 CONNECTION TO RS232

17. The MAX232 DIP package has _____ pins.
18. For the MAX232, indicate the $V_{CC}$ and GND pins.
19. The MAX233 DIP package has _____ pins.
20. For the MAX233, indicate the $V_{CC}$ and GND pins.
21. Is the MAX232 pin compatible with the MAX233?
22. State the advantages and disadvantages of the MAX232 and MAX233.
23. MAX232/233 has _____ line driver(s) for the RX wire.
24. MAX232/233 has _____ line driver(s) for the TX wire.
25. Show the connection of pins TX and RX of the PIC18 to a DB-9 RS232 connector via the second set of line drivers of MAX232.
26. Show the connection of the TX and RX pins of the PIC18 to a DB-9 RS232 connector via the second set of line drivers of MAX233.
27. What is the advantage of the MAX233 over the MAX232 chip?
28. Which pins of the PIC18 are set aside for serial communication, and what are their functions?

SECTION 10.3: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY

29. Which of the following baud rates are supported by the HyperTerminal program in PC?
   (a) 4,800  (b) 3,600  (c) 9,600
   (d) 1,800  (e) 1,200  (f) 19,200
30. Which timer of the PIC18 is used for baud rate programming?
31. Which bit of the TXSTA is used for baud rate speed?
32. What is the role of the TXREG register in serial data transfer?
33. TXREG is a(n) _____-bit register.
34. What is the role of the TXSTA register in serial data transfer?
35. TXSTA is a(n) _____-bit register.
36. For XTAL = 10 MHz, find the SPBRG value (in both decimal and hex) for each of the following baud rates.
   (a) 9,600  (b) 4,800  (c) 1,200
37. What is the baud rate if we use SPBRG = 15 to program the baud rate? Assume XTAL = 10 MHz.
38. Write a PIC18 program to transfer serially the letter ‘Z’ continuously at 1,200 baud rate. Assume XTAL = 10 MHz.
39. Write a PIC18 program to transfer serially the message “The earth is but one country and mankind its citizens” continuously at 57,600 baud rate. Assume XTAL = 10 MHz.
40. When is the TXIF flag bit raised or cleared?
41. When is the RCIF flag bit raised or cleared?
42. To which register do RCIF and TXIF belong? Is that register bit-addressable?
43. What is the role of the SPEN bit in the RCSTA register?
44. In a given situation we cannot accept reception of any serial data. How do you block such a reception with a single instruction?
45. To which register does the BRGH bit belong? State its role in rate of data transfer.

46. Is the BRGH bit HIGH or LOW when the PIC18 is powered up?

47. Find the SPBRG for the following baud rates if XTAL = 16 MHz and BRGH = 0.
   (a) 9600  (b) 19200
   (c) 38400  (d) 57600

48. Find the SPBRG for the following baud rates if XTAL = 16 MHz and BRGH = 1.
   (a) 9600  (b) 19200
   (c) 38400  (d) 57600

49. Find the SPBRG for the following baud rates if XTAL = 20 MHz and BRGH = 0.
   (a) 9600  (b) 19200
   (c) 38400  (d) 57600

50. Find the SPBRG for the following baud rates if XTAL = 20 MHz and BRGH = 1.
    (a) 9600  (b) 19200
    (c) 38400  (d) 57600

51. Find the baud rate error for Problem 47.
52. Find the baud rate error for Problem 48.

SECTION 10.4: PIC18 SERIAL PORT PROGRAMMING IN C

53. Write an PIC18 C program to transfer serially the letter ‘Z’ continuously at 1,200 baud rate.
54. Write an PIC18 C program to transfer serially the message “The earth is but one country and mankind its citizens” continuously at 57,600 baud rate.

ANSWERS TO REVIEW QUESTIONS

SECTION 10.1: BASICS OF SERIAL COMMUNICATION

1. Faster, more expensive
2. False; it is simplex.
3. True
4. Asynchronous
5. With 0100 0101 binary the bits are transmitted in the sequence:
   (a) 0 (start bit) (b) 1 (c) 0 (d) 1 (e) 0 (f) 0 (g) 0 (h) 1 (i) 0 (j) 1 (stop bit)
6. 2 bits (one for the start bit and one for the stop bit). Therefore, for each 8-bit character, a total of 10 bits is transferred.
7. \(10000 \times 10 = 100000\) total bits transmitted. \(100000 / 9600 = 10.4\) seconds; \(2 / 10 = 20\%\).
8. True
9. +3 to +25 V
10. True
11. One
12. COM 1, COM 2
SECTION 10.2: PIC18 CONNECTION TO RS232

1. True
2. Pins RC6 and RC7. Pin RC6 is for TX and pin RC7 for RX.
3. They are used for converting from RS232 voltage levels to TTL voltage levels and vice versa.
4. Two, two
5. It does not need the four capacitors that MAX232 must have.

SECTION 10.3: PIC18 SERIAL PORT PROGRAMMING IN ASSEMBLY

1. SPBRG
2. 156,250 Hz
3. BRGH
4. 15 in decimal (or F in hex) because \( \frac{156,250}{9600} - 1 = 15 \)
5. TXREG
6. Transmit Status and Control Register, 8
7. TXSTA
8. True
9. It is raised during transfer of the stop bit. It is cleared when we write a byte to TXREG to be transmitted.
10. TXSTA; it is low upon power-on reset.

SECTION 10.4: PIC18 SERIAL PORT PROGRAMMING IN C

1. True
2. True
3. True
4. SPBRG
5. TXSTA. It allows us to quadruple the baud rate with the same crystal frequency.
CHAPTER 11

INTERRUPT PROGRAMMING IN ASSEMBLY AND C

OBJECTIVES

Upon completion of this chapter, you will be able to:

» Contrast and compare interrupts versus polling
» Explain the purpose of the ISR (interrupt service routine)
» List all the major interrupts of the PIC18
» Explain the purpose of the interrupt vector table
» Enable or disable PIC18 interrupts
» Program the PIC18 timers using interrupts
» Describe the external hardware interrupts of the PIC18
» Program the PIC18 for interrupt-based serial communication
» Define the interrupt priority of the PIC18
» Program PIC interrupts in C
In this chapter we explore the concept of the interrupt and interrupt pro-
gramming. In Section 11.1, the basics of PIC18 interrupts are discussed. In Section
11.2, interrupts belonging to timers are discussed. External hardware interrupts are
discussed in Section 11.3, while the interrupt related to serial communication is
presented in Section 11.4. In Section 11.5, we cover the interrupt associated with
PORTB. In Section 11.6, we cover interrupt priority. Throughout this chapter, we
provide examples in both Assembly and C.

SECTION 11.1: PIC18 INTERRUPTS

In this section, first we examine the difference between polling and inter-
rupts and then describe the various interrupts of the PIC18.

Interrupts vs. polling

A single microcontroller can serve several devices. There are two methods
by which devices receive service from the microcontroller: interrupts or polling.
In the interrupt method, whenever any device needs the microcontroller’s service,
the device notifies it by sending an interrupt signal. Upon receiving an interrupt
signal, the microcontroller stops whatever it is doing and serves the device. The
program associated with the interrupt is called the interrupt service routine (ISR)
or interrupt handler. In polling, the microcontroller continuously monitors the sta-
tus of a given device; when the status condition is met, it performs the service.
After that, it moves on to monitor the next device until each one is serviced.
Although polling can monitor the status of several devices and serve each of them
as certain conditions are met, it is not an efficient use of the microcontroller. The
advantage of interrupts is that the microcontroller can serve many devices (not all
at the same time, of course); each device can get the attention of the microcon-
troller based on the priority assigned to it. The polling method cannot assign pri-
ority because it checks all devices in a round-robin fashion. More importantly, in
the interrupt method the microcontroller can also ignore (mask) a device request
for service. This also is not possible with the polling method. The most important
reason that the interrupt method is preferable is that the polling method wastes
much of the microcontroller’s time by polling devices that do not need service. So
interrupts are used to avoid tying down the microcontroller. For example, in dis-
cussing timers in Chapter 9 we used the bit test instruction “BTFSS TMR0IF”
and waited until the timer rolled over, and while we were waiting we could not do
anything else. That is a waste of microcontroller time that could have been used to
perform some useful tasks. In the case of the timer, if we use the interrupt method,
the microcontroller can go about doing other tasks, and when the TMR0IF flag is
raised, the timer will interrupt the microcontroller in whatever it is doing.

Interrupt service routine

For every interrupt, there must be an interrupt service routine (ISR), or
interrupt handler. When an interrupt is invoked, the microcontroller runs the inter-
rupt service routine. Generally, in most microprocessors, for every interrupt there
is a fixed location in memory that holds the address of its ISR. The group of mem-
ory locations set aside to hold the addresses of ISRs is called the interrupt vector
table. In the case of the PIC18, there are only two locations for the interrupt vector table, locations 0008 and 0018, as shown in Table 11-1. We will discuss the difference between these two in Section 11.6 when we cover interrupt priority.

Table 11-1: Interrupt Vector Table for the PIC18

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>ROM Location (Hex)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power-on Reset</td>
<td>0000</td>
</tr>
<tr>
<td>High Priority</td>
<td>0008 (Default upon power-on reset)</td>
</tr>
<tr>
<td>Low Priority</td>
<td>0018 (See Section 11.6)</td>
</tr>
</tbody>
</table>

Steps in executing an interrupt

Upon activation of an interrupt, the microcontroller goes through the following steps:

1. It finishes the instruction it is executing and saves the address of the next instruction (program counter) on the stack.
2. It jumps to a fixed location in memory called the interrupt vector table. The interrupt vector table directs the microcontroller to the address of the interrupt service routine (ISR).
3. The microcontroller gets the address of the ISR from the interrupt vector table and jumps to it. It starts to execute the interrupt service subroutine until it reaches the last instruction of the subroutine, which is RETFIE (return from interrupt exit).
4. Upon executing the RETFIE instruction, the microcontroller returns to the place where it was interrupted. First, it gets the program counter (PC) address from the stack by popping the top bytes of the stack into the PC. Then it starts to execute from that address.

Notice from Step 4 the critical role of the stack. For this reason, we must be careful in manipulating the stack contents in the ISR. Specifically, in the ISR, just as in any CALL subroutine, the number of pushes and pops must be equal.

Sources of interrupts in the PIC18

There are many sources of interrupts in the PIC18, depending on which peripheral is incorporated into the chip. The following are some of the most widely used sources of interrupts in the PIC18:

1. There is an interrupt set aside for each of the timers, Timers 0, 1, 2, and so on. See Section 11.2.
2. Three interrupts are set aside for external hardware interrupts. Pins RB0 (PORTB.0), RB1 (PORTB.1), and RB2 (PORTB.2) are for the external hardware interrupts INT0, INT1, and INT2, respectively. See Section 11.3.
3. Serial communication's USART has two interrupts, one for receive and another for transmit. See Section 11.4.
4. The PORTB-Change interrupt. See Section 11.5.

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The PIC18 has many more interrupts than the above list shows. We will cover them throughout the book as we study the peripherals of the PIC18. Notice in Table 11-1 that a limited number of bytes is set aside for high-priority interrupts. For example, a total of 8 bytes, from location 0008 to 00017H, are set aside for high-priority interrupts. Normally, the service routine for an interrupt is too long to fit in the memory space allocated. For that reason, a GOTO instruction is placed in the vector table to point to the address of the ISR. The rest of the bytes allocated to the interrupt are unused. In upcoming sections of this chapter, we will see many examples of interrupt programming that clarify these concepts.

From Table 11-1, also notice that only 8 bytes of ROM space are assigned to the reset pin. They are ROM address locations 0–7. For this reason, in our program we put the GOTO as the first instruction and redirect the processor away from the interrupt vector table, as shown in Figure 11-1. In the next section we will see how this works in the context of some examples.

ORG 0 ;wake-up ROM reset location
GOTO MAIN ;bypass interrupt vector table

;---- the wake-up program
ORG 100H
MAIN: .... ;enable interrupt flags
....
END

Figure 11-1. Redirecting the PIC18 from the Interrupt Vector Table at Power-Up

Figure 11-2. Simplified View of Interrupts (default for power-on reset)

**Enabling and disabling an interrupt**

Upon reset, all interrupts are disabled (masked), meaning that none will be responded to by the microcontroller if they are activated. The interrupts must be enabled (unmasked) by software in order for the microcontroller to respond to them. The D7 bit of the INTCON (Interrupt Control) register is responsible for
enabling and disabling the interrupts globally. Figure 11-3 shows the INTCON register. The GIE bit makes the job of disabling all the interrupts easy. With a single instruction (BCF INTCON,GIE), we can make GIE = 0 during the operation of a critical task. See Figure 11-2.

**Steps in enabling an interrupt**

To enable any one of the interrupts, we take the following steps:

1. Bit D7 (GIE) of the INTCON register must be set to HIGH to allow the interrupts to happen. This is done with the “BSF INTCON, GIE” instruction.
2. If GIE = 1, each interrupt is enabled by setting to HIGH the interrupt enable (IE) flag bit for that interrupt. Because there are a large number of interrupts in the PIC18, we have many registers holding the interrupt enable bit. Figure 11-2 shows that the INTCON has interrupt enable bits for Timer0 (TMROIE) and external interrupt 0 (INT0IE). As we study each of peripherals throughout the book we will examine the registers holding the interrupt enable bits. It must be noted that if GIE = 0, no interrupt will be responded to, even if the corresponding enable bit is high. To understand this important point look at Example 11-1.
3. As shown in Figures 11-2 and 11-3, for some of the peripheral interrupts such as TMR1IF, TMR2IF, and TXIF, we have to enable the PEIE flag in addition to the GIE bit.

<table>
<thead>
<tr>
<th>D7</th>
<th>D0</th>
</tr>
</thead>
<tbody>
<tr>
<td>GIE</td>
<td>TMR0IE</td>
</tr>
</tbody>
</table>

**GIE (Global Interrupt Enable)**

GIE = 0 Disables all interrupts. If GIE = 0, no interrupt is acknowledged, even if they are enabled individually.

If GIE = 1, interrupts are allowed to happen. Each interrupt source is enabled by setting the corresponding interrupt enable bit.

**TMROIE**

- Timer0 interrupt enable
- = 0 Disables Timer0 overflow interrupt
- = 1 Enables Timer0 overflow interrupt

**INT0IE**

- Enables or disables external interrupt 0
- = 0 Disables external interrupt 0
- = 1 Enables external interrupt 0

These bits, along with the GIE, must be set high for an interrupt to be responded to.

Upon activation of the interrupt, the GIE bit is cleared by the PIC18 itself to make sure another interrupt cannot interrupt the microcontroller while it is servicing the current one. At the end of the ISR, the RETFIE instruction will make GIE = 1 to allow another interrupt to come in.

**PEIE (Peripheral Interrupt Enable)**

For many of the peripherals, such as Timers 1, 2, .. and the serial port, we must enable this bit in addition to the GIE bit. (See Figure 11-2.)

**Figure 11-3. INTCON (Interrupt Control) Register**

CHAPTER 11: INTERRUPT PROGRAMMING IN ASSEMBLY AND C
Example 11-1

Show the instructions to (a) enable (unmask) the Timer0 interrupt and external hardware interrupt 0 (INT0), and (b) disable (mask) the Timer0 interrupt, then (c) show how to disable (mask) all the interrupts with a single instruction.

Solution:

(a)  BSF INTCON,TMR0IE ; enable (unmask) Timer0 interrupt  
    BSF INTCON,INT0IE ; enable external interrupt 1 (INT0)  
    BSF INTCON,GIE ; allow interrupts to come in

We can perform the above actions with the following two instructions:

    MOVLW B'10110000' ; GIE = 1, TMR0IF = 1, INTIF0 = 1  
    MOVWF INTCON ; load the INTCON reg

(b)  BCF INTCON,TMR0IE ; mask (disable) Timer0 interrupt

c)  BCF INTCON,GIE ; mask all interrupts globally

Review Questions

1. Of the interrupt and polling methods, which one avoids tying down the microcontroller?
2. Give the name of the interrupts in the INTCON register.
3. Upon power-on reset of the PIC18, what memory area is assigned to the interrupt vector table? Can the programmer change the memory space assigned to the table?
4. What is the content of D7 (GIE) of the INTCON register upon reset, and what does it mean?
5. Show the instruction needed to enable the TMR0 interrupt.
6. What address in the interrupt vector table is assigned to high-priority and low-priority interrupts?
SECTION 11.2: PROGRAMMING TIMER INTERRUPTS

In Chapter 9 we discussed how to use Timers 0, 1, 2, and 3 with the polling method. In this section we use interrupts to program the PIC18 timers. Please review Chapter 9 before you study this section.

Rollover timer flag and interrupt

In Chapter 9 we stated that the timer flag is raised when the timer rolls over. In that chapter, we also showed how to monitor the timer flag with the instruction “BTFSS TMR0IF”. In polling TMR0IF, we have to wait until TMR0IF is raised. The problem with this method is that the microcontroller is tied down waiting for TMR0IF to be raised, and cannot do anything else. Using interrupts avoids tying down the controller. If the timer interrupt in the interrupt register is enabled, TMR0IF is raised whenever the timer rolls over and the microcontroller jumps to the interrupt vector table to service the ISR. In this way, the microcontroller can do other things until it is notified that the timer has rolled over. To use an interrupt in place of polling, first we must enable the interrupt because all the interrupts are masked upon power-on reset. The TMRxIE bit enables the interrupt for a given timer. TMRxIE bits are held by various registers as shown in Table 11-2. In the case of Timer0, the INTCON register (Figure 11-4) contains the TMR0IE bit, and PIE1 (peripheral interrupt enable) holds the TMR1IE bit. See Figure 11-5 and Program 11-1.

Table 11-2: Timer Interrupt Flag Bits and Associated Registers

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>Flag Bit</th>
<th>Register</th>
<th>Enable Bit</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>Timer0</td>
<td>TMR0IF</td>
<td>INTCON</td>
<td>TMR0IE</td>
<td>INTCON</td>
</tr>
<tr>
<td>Timer1</td>
<td>TMR1IF</td>
<td>PIR1</td>
<td>TMR1IE</td>
<td>PIE1</td>
</tr>
<tr>
<td>Timer2</td>
<td>TMR2IF</td>
<td>PIR1</td>
<td>TMR2IE</td>
<td>PIE1</td>
</tr>
<tr>
<td>Timer3</td>
<td>TMR3IF</td>
<td>PIR3</td>
<td>TMR3IE</td>
<td>PIE2</td>
</tr>
</tbody>
</table>

![Figure 11-4. INTCON Register with Timer0 Interrupt Enable and Interrupt Flag](image)

Figure 11-5. The Role of Timer Interrupt Enable Flag (TMRxIE)

Note: The TMRxIP (timer interrupt priority) flag is not shown. TMRxIP is used to force the interrupt to land at vector location 0x0018. See Section 11.6.
Notice the following points about Program 11-1:

1. We must avoid using the memory space allocated to the interrupt vector table. Therefore, we place all the initialization codes in memory starting at an address such as 100H. The GOTO instruction is the first instruction that the PIC18 executes when it is awakened at address 00000 upon power-on reset (POR). The GOTO instruction at address 00000 redirects the controller away from the interrupt vector table.

2. In the MAIN program, we enable (unmask) the Timer0 interrupt with instruction “BSF INTCON, TMR0IE” followed by the instruction “BSF INTCON, GIE” to enable all interrupts globally.

3. In the MAIN program, we initialize the Timer0 register and then enter an infinite loop to keep the CPU busy. This could be a real-world application being executed by the CPU. In this case, the loop gets data from PORTC and sends it to PORTD. While the PORTC data is brought in and issued to PORTD continuously, the TMR0IF flag is raised as soon as Timer0 rolls over, and the microcontroller gets out of the loop and goes to 00008H to execute the ISR associated with Timer0. At this point, the PIC18 clears the GIE bit (D7 of INTCON) to indicate that it is currently serving an interrupt and cannot be interrupted again; in other words, no interrupt inside the interrupt. In Section 11.6, we show how to allow an interrupt inside an interrupt.

4. The ISR for Timer0 is located starting at memory location 00200H because it is too large to fit into address space 08–17H, the address allocated to high-priority interrupts.

5. In the ISR for Timer0, notice that the “BCF INTCON, TMR0IF” instruction is needed before the RETFIE instruction. This will ensure that a single interrupt is serviced once and is not recognized as multiple interrupts.

6. RETFIE must be the last instruction of the ISR. Upon execution of the RETFIE instruction, the PIC18 automatically enables the GIE (D7 of the INTCON register) to indicate that it can accept new interrupts.

Program 11-1: For this program, we assume that PORTC is connected to 8 switches and PORTD to 8 LEDs. This program uses Timer0 to generate a square wave on pin PORTB5, while at the same time data is being transferred from PORTC to PORTD.

;Program 11-1

ORG 0000H
GOTO MAIN ; bypass interrupt vector table

;--on default all interrupts land at address 00008
ORG 0008H ; interrupt vector table

BTFSS INTCON, TMR0IF ; Timer0 interrupt?

RETFIE ; No. Then return to main

GOTO T0_ISR ; Yes. Then go Timer0 ISR

;--main program for initialization and keeping CPU busy

ORG 00100H ; after vector table space

MAIN BCF TRISB, 5 ; PB5 as an output

CLRF TRISD ; make PORTD output

SETF TRISC ; make PORTC input
MOVLW 0x08 ;Timer0, 16-bit, no prescale, internal clk
MOVWF T0CON ;load T0CON reg
MOVLW 0x00 ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0xFF ;TMR0L = F2H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BCF INTCON,TMR0IF; clear timer interrupt flag bit
BSF T0CON,TMROON ;start Timer0
BSF INTCON,TMROIE ;enable Timer 0 interrupt
BSF INTCON,GIE ;enable interrupts globally
;--keeping CPU busy waiting for interrupt
OVER
MOVF PORTC,PORTD ;send data from PORTC to PORTD
BRA OVER ;stay in this loop forever

;-----------------------------ISR for Timer 0
T0_ISR
ORG 200H
MOVLW 0xFF ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0xF2 ;TMR0L = F2H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BTG PORTB,5 ;toggle RB5
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
EXIT RETFIE ;return from interrupt (See Example 11-2)
END

Example 11-2

What is the difference between the RETURN and RETFIE instructions? Explain why we cannot use RETURN instead of RETFIE as the last instruction of an ISR.

Solution:

Both perform the same actions of popping off the top bytes of the stack into the program counter, and making the PIC18 return to where it left off. However, RETFIE also performs the additional task of clearing the GIE flag, indicating that the servicing of the interrupt is over and the PIC18 now can accept a new interrupt. If you use RETURN instead of RETFIE as the last instruction of the interrupt service routine, you simply block any new interrupt after the first interrupt, because the GIE would indicate that the interrupt is still being serviced.

In Program 11-1, the Timer0 ISR (interrupt service routine) was too long to be placed in memory locations allocated to the high interrupt (addresses of 0008–00017H). There was enough space, however, to test to see which interrupt was the cause of landing at the 0008 address. Very often, we go from address 0008 to another address with a larger space to check the source of the interrupt, given the fact that the PIC18 has so many interrupts and we have limited space at address 0008. See Program 11-2.
Program 11-2 uses Timer0 and Timer1 interrupts to generate square waves on pins RB1 and RB7 respectively, while data is being transferred from PORTC to PORTD.

;Program 11-2
ORG 0000H
GOTO MAIN ;bypass interrupt vector table

--on default all interrupts land at address 00008
ORG 0008H ;interrupt vector table
GOTO CHK_INT ;go to an address with more space

--check to see the source of interrupt
ORG 0040H ;we got here from 0008

CHK_INT
BTFSC INTCON,TMR0IF ;Is it Timer0 interrupt?
BRA T0_ISR ;Yes. Then branch to T0_ISR
BTFSC PIR1,TMR1IF ;Is it Timer1 interrupt?
BRA T1_ISR ;Yes. Then branch to T1_ISR
RETFIE ;No. Then return to main

--main program for initialization and keeping CPU busy
ORG 0100H ;somewhere after vector table space

MAIN BCF TRISB,1 ;PB1 as an output
BCF TRISB,7 ;PB7 as an output
CLRF TRISD ;make PORTD output
SETF TRISC ;make PORTC input
MOVLW 0x08 ;Timer0,16-bit,
;no prescale, internal clk
MOVWF T0CON ;load T0CON reg
MOVLW 0xFF ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0xF2 ;TMR0L = F2H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BCF INTCON,TMR0IF;clear Timer0 interrupt flag bit
MOVLW 0x0 ;Timer1,16-bit,
;no prescale, internal clk
MOVWF T1CON ;load T1CON reg
MOVLW 0xFF ;TMR1H = FFH, the high byte
MOVWF TMR1H ;load Timer1 high byte
MOVLW 0xF2 ;TMR1L = F2H, the low byte
MOVWF TMR1L ;load Timer1 low byte
BCF PIR1,TMR1IF ;clear Timer1 interrupt flag bit
BSF INTCON,TMR0IE ;enable Timer0 interrupt
BSF PIE1,TMR1IE ;enable Timer1 interrupt
BSF INTCON,PEIE ;enable peripheral interrupts
BSF INTCON,GIE ;enable interrupts globally
BSF T0CON,TMR0ON ;start Timer0
BSF T1CON,TMR1ON ;start Timer1

--;--keeping CPU busy waiting for interrupt
OVER MOVFF PORTC,PORTD ;send data from PORTC to PORTD
BRA OVER ;stay in this loop forever

;-----------------------------ISR for Timer 0
T0_ISR
ORG 200H
MOV LW 0xFF ; TMR0H = FFH, the high byte
MOVWF TMR0H ; load Timer0 high byte
MOV LW 0xF2 ; TMR0L = F2H, the low byte
MOVWF TMR0L ; load Timer0 low byte
BTG PORTB, 1 ; toggle PB1
BCF INTCON, TMR0IF ; clear timer interrupt flag bit
GOTO CHK_INT

;-------------------------------ISR for Timer1

T1_ISR
ORG 300H
MOV LW 0xFF ; TMR1H = FFH, the high byte
MOVWF TMR1H ; load Timer1 high byte
MOV LW 0xF2 ; TMR1L = F2H, the low byte
MOVWF TMR1L ; load Timer1 low byte
BTG PORTB, 7
BCF PIR1, TMROIF ; clear Timer1 interrupt flag bit
GOTO CHK_INT
END

Notice that the addresses 0040H, 0100H, 00200H, and 0300H that we used in Program 11-2 are all arbitrary and can be changed to any addresses we want. The only addresses that we have no choice on are the power-on reset location of 0000 and high-priority address of 0008 because they were fixed at the time of the PIC18 design.

Program 11-3 has two interrupts: (1) PORTD counts up every time Timer0 overflows. It uses 16-bit mode of Timer0 with the largest prescale possible; (2) 1-Hz pulse is fed into Timer1 where Timer1 is used as counter and counts up. Whenever the count reaches 200, it will toggle the pin PORTB.6.

; Program 11-3
ORG 0000H
GOTO MAIN ; bypass interrupt vector table
;--on default all interrupts land at address 00008
ORG 0008H
GOTO CHK_INT

;------------------------find the interrupt source
ORG 0040H
CHK_INT
BTFSC INTCON, TMROIF ; Is it Timer0 interrupt?
BRA T0_ISR ; Yes. Then branch to T0_ISR
BTFSC PIR1, TMROIF ; Is it Timer1 interrupt?
BRA T1_ISR ; Yes. Then branch to T1_ISR
RETFIE ; No. Then return to main

;--the main program for initialization
ORG 00100H ; after vector table space
MAIN BSF TRISC, T13CKI ; PORTC.0 as an input
CLRF TRISD ; make PORTD output
BCF TRISB, 6 ; make RB6 output

Figure 11-6. For Program 11-3

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MOVCLW 0x08 ; 16-bit, prescale = 256, ; internal clk
MOVWF T0CON ; load T0CON reg
MOVCLW 0x00 ; TMR0H = 00H, the high byte
MOVWF TMROH ; load Timer0 high byte
MOVCLW 0x00 ; TMR0L = 0, the low byte
MOVWF TMROL ; load Timer0 low byte
BCF INTCON, TMROIF ; clear timer interrupt flag bit
MOVCLW 0x06 ; Timer1, no prescale,
; ext. clock
MOVWF T1CON ; load T1CON reg
MOVCLW D'255' ; TMR1H = 255
MOVWF TMRIH ; load Timer1 high byte
MOVCLW -D'200' ; TMR1L = 0
MOVWF TMRL ; load Timer1 low byte
BCF PIR1, TMR1IF ; clear timer interrupt flag bit
BSF TOCON, TMROON ; start Timer0
BSF T1CON, TMROON ; start Timer1
BSF INTCON, TMROIE ; enable Timer0 interrupt
BSF PIE1, TMRIE ; enable Timer1 interrupt
BSF INTCON, PETE ; enable peripheral interrupts
BSF INTCON, GIE ; enable interrupts globally
OVER BRA OVER ; stay in this loop forever

--------ISR for Timer0

T0_ISR
ORG 200H
INCF PORTD ; increment PORTD
MOVCLW 0x00 ; TMR0L = 0, the low byte
MOVWF TMROL ; load Timer0 low byte
MOVCLW 0x00 ; TMR0H = 00, the high byte
MOVWF TMROL ; load Timer0 high byte
BCF INTCON, TMROIF ; clear timer interrupt flag bit
GOTO CHK_INT

ISR for Timer2

T1_ISR
ORG 300H
BTG PORTB, 6 ; toggle PORTC.6
MOVCLW D'255' ; TMR1H = 255
MOVWF TMRIH ; load Timer1 high byte
MOVCLW -D'200' ; TMR1L = 0
MOVWF TMRL ; load Timer1 low byte
BCF PIR1, TMR1IF ; clear Timer1 interrupt flag bit
GOTO CHK_INT
END

Notice in Programs 11-2 and 11-3 that we use the "GOTO CHK_INT" instruction instead of RETFIE as the last instruction of the ISR. This is because we are checking for activation of multiple interrupts.
PIC18 interrupt programming in C using C18 compiler

In Chapter 7, we discussed how the C18 compiler uses "#pragma code" to place code at a specific ROM address. Because the C18 does not place an ISR at the interrupt vector table automatically, we must use Assembly language instruction GOTO at the interrupt vector to transfer control to the ISR. This is done as follows:

```
#pragma code high_vector =0x0008 // High-priority interrupt location
void My_HiVect_Int (void) {
    _asm
    GOTO my_isr
    _endasm
}
#pragma code                   // End of code
```

Now we redirect it from address location 00008 to another program to find the source of the interrupt and finally to the ISR. This is done with the help of the keyword `interrupt` as follows:

```
#pragma interrupt my_isr       //interrupt is reserved keyword
void my_isr (void)            //used for high-priority interrupt
{

    //C18 places RETFIE here automatically due to
    //interrupt keyword
}
```

Note that "pragma", "code", and "interrupts" are reserved keywords while the choice of all other labels is up to us. Examine Programs 11-2C and 11-3C. They are the C versions of Programs 11-2 and 11-3.

Program 11-2C uses Timer0 and Timer1 interrupts to generate square waves on pins RB1 and RB7, respectively, while data is being transferred from PORTC to PORTD. This is a C version of Program 11-2.

```
#include <pl8F458.h>
#define myPB1bit PORTBbits.RB1
#define myPB7bit PORTBbits.RB7

void T0_ISR(void);
void T1_ISR(void);

#pragma interrupt chk_isr       //used for high-priority
                            //interrupt only
void chk_isr (void) {
```

CHAPTER 11: INTERRUPT PROGRAMMING IN ASSEMBLY AND C
if (INTCONbits.TMR0IF==1) //Timer0 causes interrupt?
    T0_ISR();     //Yes. Execute Timer0 ISR
if (PIR1bits.TMR1IF==1) //Or was it Timer1?
    T1_ISR();     //Yes. Execute Timer1 ISR

#pragma code My_HiPrio_Int=0x08 //high-priority interrupt
void My_HiPrio_Int (void)
{
    _asm
        GOTO chk_isr
    _endasm
}
#pragma code
void main(void)
{
    TRISBbits.TRISB1=0;    //RB1 = OUTPUT
    TRISBbits.TRISB7=0;    //RB7 = OUTPUT
    TRISC = 255;            //PORTC = INPUT
    TRISD = 0;             //PORTD = OUTPUT
    TOCON=0x00;            //Timer 0, 16-bit mode, no prescaler
    TMR0H=0x35;            //load TH0
    TMR0L=0x00;            //load TL0
    T1CON=0x88;            //Timer 1, 16-bit mode, no prescaler
    TMR1H=0x35;            //load TH1
    TMR1L=0x00;            //load TL1
    INTCONbits.TMR0IF=0;   //clear TF0
    PIR1bits.TMR1IF=0;     //clear TF1
    INTCONbits.TMR0IE=1;   //enable Timer0 interrupt
    INTCONbits.TMR0IE=1;   //enable Timer1 interrupt
    TOCONbits.TMR0ON=1;    //turn on Timer0
    T1CONbits.TMR1ON=1;    //turn on Timer1
    INTCONbits.PEIE=1;     //enable all peripheral interrupts
    INTCONbits.GIE=1;      //enable all interrupts globally
while(1)     //keep looping until interrupt comes
{
    PORTD=PORTC;  //send data from PORTC to PORTD
}
}

void T0_ISR(void)
{
    myPB1bit=~myPB1bit;     //toggle PORTB.1
    TMR0H=0x35;            //load TH0
    TMR0L=0x00;            //load TL0
    INTCONbits.TMR0IF=0;   //clear TF0
}

void T1_ISR(void)
{
    myPB7bit=~myPB7bit;     //toggle PORTB.7
}
TMR1H=0x35;       //load TH0
TMR1L=0x00;       //load TL0
PIR1bits.TMR1IF=0; //clear TF1
}

Program 11-3C shows the C version of Program 11-3.

Program 11-3C has two interrupts: (1) PORTC counts up every time Timer0 overflows. It uses the 16-bit mode of Timer0 with the largest prescaler possible; (2) a 1-Hz pulse is fed into Timer1 where Timer1 is used as a counter and counts up. Whenever the count reaches 200, it will toggle pin RB6.

//Program 11-3C
#include <p18F458.h>
define myPB6bit PORTBbits.RB6

void chk_isr(void);
void T0_ISR(void);
void T1_ISR(void);
#pragma interrupt chk_isr //for high-priority interrupt only
void chk_isr (void)
{
  if (INTCONbits.TMR0IF==1) //Timer0 causes interrupt?
    T0_ISR();             //Yes. Execute Timer0 program
      if (PIR1bits.TMR1IF==1)//Or was it Timer2?
        T1_ISR();          //Yes. Execute Timer2 program

#pragma code My_HiPrio_Int=0x0008 //high-priority interrupt
void My_HiPrio_Int (void)
{
  _asm
    GOTO chk_isr
  _endasm
}
#pragma code

void main(void)
{
  TRISBbits.TRISB6=0;  //RB6 = OUTPUT
  TRISChbits.TRISCO=1; //PORTC0 = INPUT
  TRISD=0;
  TOCON=0x08;         //Timer0, 16-bit mode,
  //no prescaler
  TMR0H=0;            //load Timer0 high byte
  TMR0L=0;            //load Timer0 low byte
  T1CON=0x06;         //Timer 2, no prescaler
  TMR1H=255;          //load Timer1 high byte
  TMR1L=-200;         //load Timer1 low byte
  INTCONbits.TMR0IF=0; //clear TF0
  PIR1bits.TMR1IF=0;  //clear TF1

INTCONbits.TMR0IE=1; //enable Timer0 interrupt
PIElbits.TMR1IE=1; //enable Timer1 interrupt
T0CONbits.TMR0ON=1; //turn on Timer0
T1CONbits.TMR1ON=1; //turn on Timer1
INTCONbits.PEIE=1; //enable all peripheral interrupts
INTCONbits.GIE=1; //enable all interrupts globally
while(1); //keep looping until interrupt comes
}
void T0_ISR(void)
{
    PORTD++; //count up PORTD
    TMR0H=0; //load Timer0 high byte
    TMR0L=0; //load Timer0 low byte
    INTCONbits.TMR0IF=0; //clear TF0
}
void T1_ISR(void)
{
    myPB6bit=~myPB6bit; //toggle PB.6
    TMR1H=255; //load Timer1 high byte
    TMR1L=-200; //load Timer1 low byte
    PIR1bits.TMR2IF=0; //clear TF1
}

Review Questions

1. True or false. A unique address in the interrupt vector table is assigned to each of Timer0–Timer3.
2. Upon power-on reset, what address in the interrupt vector table is assigned to the high-priority interrupt?
3. Which register does TMR1IE belong to? Show how it is enabled.
4. Assume that Timer1 is programmed in 8-bit mode, TMR1L = F5H, and the TMR1IF bit is enabled. Explain how the interrupt for the timer works.
5. True or false. The last two instructions of the ISR for Timer0 are:
   BCF INTCON,TMR0IF
   RETFIE
The PIC18 has three external hardware interrupts. Pins RB0 (PORTB.0), RB1 (PORTB.1), and RB2 (PORTB.2), designated as INTO, INT1, and INT2 respectively, are used as external hardware interrupts. Upon activation of these pins, the PIC18 gets interrupted in whatever it is doing and jumps to the vector table to perform the interrupt service routine. In this section we study these three external hardware interrupts of the PIC18 with some examples in both Assembly and C.

**External interrupts INTO, INT1, and INT2**

There are three external hardware interrupts in the PIC18: INTO, INT1, and INT2. They are located on pins RB0, RB1, and RB2, respectively. See Figures 11-7 and 11-8. On default, all three hardware interrupts are directed to vector table location 0008H, unless we specify otherwise. They must be enabled before they can take effect. This is done using the INTxIE bit. The registers associated with INTxIE bits are shown in Table 11-3. For example, the instruction “BSF INTCON, INTOIE” enables INTO. The INTO is a *positive-edge-triggered interrupt*, which means, when a low-to-high signal is applied to pin RB0 (PORTB.0), the INTOIF is raised, causing the controller to be interrupted. The raising of INTOIF forces the PIC18 to jump to location 0008H in the vector table to service the ISR. In Table 11-3, notice the INTOIF bits and the registers they belong to. Upon power-on reset, the PIC18 makes INTO, INT1, and INT2 rising (positive) edge-triggered interrupts. To make them falling (negative) edge-triggered interrupts, we must program the INTEDGx bits, as we will see shortly.

Examine Program 11-4 and its C version, Program 11-4C, to gain insight into external hardware interrupts.

<table>
<thead>
<tr>
<th>Interrupt (Pin)</th>
<th>Flag bit</th>
<th>Register</th>
<th>Enable bit</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>INTO (RB0)</td>
<td>INTOIF</td>
<td>INTCON</td>
<td>INTOIE</td>
<td>INTCON</td>
</tr>
<tr>
<td>INT1 (RB1)</td>
<td>INT1IF</td>
<td>INTCON3</td>
<td>INT1IE</td>
<td>INTCON3</td>
</tr>
<tr>
<td>INT2 (RB2)</td>
<td>INT2IF</td>
<td>INTCON3</td>
<td>INTO2IE</td>
<td>INTOCON3</td>
</tr>
</tbody>
</table>

Examine Program 11-4 and its C version, Program 11-4C, to gain insight into external hardware interrupts.
Figure 11-8. INT0–INT2 Hardware Interrupts

Program 11-4 connects a switch to INT0 and an LED to pin RB7. In this program, every time INT0 is activated, it toggles the LED, while at the same time data is being transferred from PORTC to PORTD.

;Program 11-4
ORG 0000H
GOTO MAIN ;bypass interrupt vector table

;--on default all interrupts go to to address 00008
ORG 0008H ;interrupt vector table
BTFSS INTCON,INT0IF ;Did we get here due to INT0?
RETFIE ;No. Then return to main
GOTO INTO_ISR ;Yes. Then go INTO ISR

;--the main program for initialization
ORG 00100H
MAIN BCF TRISB,7 ;PB7 as an output
BSF TRISD,INT0 ;make INTO an input pin
CLRF TRISD ;make PORTD output
SETF TRISC ;make PORTC input
BSF INTCON,INT0IE ;enable INTO interrupt
BSF INTCON,GIE ;enable interrupts globally
OVER MOVFF PORTC,PORTD ;send data from PORTC to PORTD
BRA OVER ;stay in this loop forever

;-----------------------------ISR for INTO
INT0_ISR
ORG 200H
BTG PORTB,7 ;toggle PB7
BCF INTCON,INT0IF ;clear INTO interrupt flag bit
RETFIE ;return from ISR
END
Look at Program 11-4. When a rising edge of the signal is applied to pin INTO, the LED will toggle. In this example, to toggle the LED again, the INTO pulse must be brought back LOW and then forced HIGH to create a rising edge to activate the interrupt.

//Program 11-4C (This is the C version of Program 11-4)
#include <pl8F4580.h>
#define mybit PORTBbits.RB7
void chk_isr(void);
void INTO_ISR(void);
#pragma interrupt chk_isr//used for high-priority int
void chk_isr (void)
{
   if (INTCONbits.INT0IF==1) //INTO caused interrupt?
      INTO_ISR( ); //Yes. Execute INTO program
}
#pragma code My_HiPrio_Int=0x08 //high-priority
   //interrupt location
void My_HiPrio_Int (void)
{
   _asm
      GOTO chk_isr
   _endasm
}
#pragma code
void main(void)
{
   
   TRISBbits.TRISB7=0;    //RB7 = OUTPUT
   TRISBbits.TRISB0=1;    //INT0 = INPUT
   TRISC = 0xFF;          //PORTC = INPUT
   TRISD = 0;             //PORTD = OUTPUT
   INTCONbits.INT0IF=0;   //clear TF1
   INTCONbits.INT0IE=1;   //enable Timer0 interrupt
   INTCONbits.GIE=1;      //enable all interrupts
   while(1)               //keep looping until interrupt comes
   {
      PORTD=PORTC;
   }
}
void INTO_ISR(void)
{
   mybit=~mybit;
   INTOCONbits.INT0IF=0;   //clear INTO flag
}
Negative edge-triggered interrupts

Upon power-on reset, the PIC18 makes INT0, INT1, and INT2 positive (rising) edge-triggered interrupts. To make any of them a negative (falling) edge-triggered interrupt, we must program the corresponding bit called INTEDGx, where x can be 0, 1, or 2. The INTCON2 register holds, among other bits, the INTEDG0, INTEDG1, and INTEDG flag bits as shown in Figure 11-10. INTEDG0, INTEDG1, and INTEDG2 are bits D4, D5, and D6 of the INTCON2 register, respectively, as shown in Figure 11-10. The status of these bits determines the negative or positive edge-triggered mode of the hardware interrupts. Upon reset, INTEDGx bits are all 1s, meaning that the external hardware interrupts are positive edge-triggered. By making the INTEDG0 bit LOW, the external hardware interrupts of INT0 become negative edge-triggered interrupts. For example, the instruction "BSF INTCON2, INTEDG1" makes INTEDG1 a negative edge-triggered interrupt, in which, when a high-to-low signal is applied to pin RBI (PORTB.1), the controller will be interrupted and forced to jump to location 0008H in the vector table to service the ISR (assuming that the GIE and INT0IE bits are enabled). This is shown in Program 11-5. Its C version is shown in Program 11-5C.

<table>
<thead>
<tr>
<th>INTEDG0</th>
<th>INTEDG1</th>
<th>INTEDG2</th>
</tr>
</thead>
</table>

INTEDGx  External Hardware Interrupt Edge trigger bit
0 = Interrupt on negative (falling) edge
1 = Interrupt on positive (rising) edge (Default for power-on reset)

In Program 11-5 we assume that pin RB1 (INT1) is connected to a pulse generator and the pin RB7 is connected to an LED. The program will toggle the LED on the falling edge of the pulse. In other words, the LED is turned on and off at the same rate as the pulses are applied to the INT1 pin.

;Program 11-5
ORG 0000H
GOTO MAIN ;bypass interrupt vector table
;--on default all interrupts go to to address 00008
ORG 0008H ;interrupt vector table
BTFSS INTCON3,INT1IF ;Did we get here due to
 ;INT1 interrupt?
RETFIE ;No. Then return to main
GOTO INT1_ISR ;Yes. Then go INT1 ISR
;--the main program for initialization
ORG 00100H
MAIN BCF TRISB,7 ;PB7 as an output
BSF TRISB,INT1 ;make INT1 an input pin

Figure 11-11. For Program 11-5
BSF INTCON3,INT1IE ;enable INT1 interrupt
BCF INTCON2,INTEDG1 ;make it negative
;edge-triggered
BSF INTCON,GIE ;enable interrupts globally
OVER BRA OVER ;stay in this loop forever
//ISR for INT1
INT1_ISR
ORG 200H
BTG PORTB,7 ;toggle on RB7
BCF INTCON3,INT1IF ;clear INT1 interrupt flag bit
RETFIE
END

Program 11-5C (This is the C version of Program 11-5)
#include <p18F4580.h>
define mybit PORTBbits.RB7
void chk_isr(void);
void INT1_ISR(void);
#pragma code My_HiPrio_Int =0x0008 //high-priority interrupt location
void My_HiPrio_Int (void)
{
_asm
GOTO chk_isr
_endasm

#pragma code
#pragma interrupt chk_isr //used for high-priority
//interrupt only

void chk_isr (void)
{
if (INTCON3bits.INT1IF==1) //INT1 causes interrupt?
INT1_ISR( ) ; //Yes. Execute INT1 program
}
void main(void)
{
TRISBbits.TRISB7=0; //RB7 = OUTPUT
TRISBbits.TRISB1=1; //INT1 = INPUT
INTCON3bits.INT1IF=0; //clear INT1
INTCON3bits.INT1IE=1; //enable INT1 interrupt
INTCON2bits.INTEDG1=0; //make it negative edge
INTCONbits.GIE=1; //enable all interrupts
while(1); //keep looping until interrupt comes

void INT1_ISR(void)
{
mybit=~mybit;
INTCON3bits.INT1IF=0; //clear INT1 flag
}
Sampling the edge-triggered interrupt

Before ending this section, we need to answer the question of how often the edge-triggered interrupt is sampled. In edge-triggered interrupts, the external source must be held HIGH for at least two instruction cycles, and then held LOW for at least two instruction cycles to ensure that the transition is seen by the microcontroller.

The rising edge (or the falling edge) is latched by the PIC18 and is held by the INTxIF bits. The INT0IF, INT1IF, and INT1IF2 bits hold the latched rising (or falling, depending on the INTEDGx bit) edge of pins RB0–RB2. The INT0IF–INT2IF bits function as interrupt-in-service flags. When an interrupt-in-service flag is raised, it indicates to the external world that the interrupt is being serviced and no new interrupt on this INTx pin will be responded to until this service is finished. This is just like the busy signal you get when calling a telephone number that is in use. Regarding the INT0IF–INT2IF one more point must be emphasized. The point is that before the ISRs are finished (that is, before execution of instruction RETFIE), these bits (INT0IF–INT2IF) must be cleared, indicating that the interrupt is finished and the PIC18 is ready to respond to another interrupt on that pin. For another interrupt to be recognized, the pin must go back to a logic LOW state and be brought back HIGH to be considered a positive edge-triggered interrupt.

Minimum pulse duration to detect edge triggered interrupts = 2 instruction cycles
For XTAL = 10 MHz, we have an instruction cycle time of 400 ns = 0.4 μs

0.8 μs

2 Instr. Cycle

0.8 μs

2 Instr. Cycle

Review Questions

1. True or false. Upon reset, all external hardware interrupts INT0–INT2 go to the interrupt vector table address of 0008.
2. For PIC18F458, what pins are assigned to INT0–INT2?
3. Show how to enable the INT1.
4. Assume that the INT0IE bit for the external hardware interrupt INT0 is enabled. Explain how this interrupt works when it is activated.
5. True or false. Upon reset, the external hardware interrupt is negative edge-triggered.
6. How do we make sure that a single interrupt is not recognized as multiple interrupts?
7. True or false. The last two instructions of the ISR for INT0 are:
   BCF INTCON2,INT0IF
   RETFIE
SECTION 11.4: PROGRAMMING THE SERIAL COMMUNICATION INTERRUPTS

In Chapter 10 we studied the serial communication of the PIC18. All examples in that chapter used the polling method. In this section we explore interrupt-based serial communication, which allows the PIC18 to do many things, in addition to sending and receiving data from the serial communication port.

RCIF and TXIF flags and interrupts

As you may recall from Chapter 10, TXIF (transfer interrupt) is raised when the last bit of the framed data, the stop bit, is transferred, indicating that the TXREG register is ready to transfer the next byte. RCIF (received interrupt) is raised when the entire frame of data, including the stop bit, is received. In other words, when the RCREG register has a byte, RCIF is raised to indicate that the received byte needs to be picked up before it is lost (overrun) by new incoming serial data. As far as serial communication is concerned, all the above concepts apply equally when using either polling or an interrupt. The only difference is in how the serial communication needs are served. In the polling method, we wait for the flag (TXIF or RCIF) to be raised; while we wait we cannot do anything else. In the interrupt method, we are notified when the PIC18 has received a byte, or is ready to send the next byte; we can do other things while the serial communication needs are served.

In the PIC18 two interrupts are set aside for serial communication. One interrupt is used for send and the other for receive. If the corresponding interrupt bit of TXIE or RCIE is enabled, when TXIF or RCIF is raised the PIC18 gets interrupted and jumps to memory address location 0008H to execute the ISR.

Table 11-4: Serial Port Interrupt Flag Bits and their Associated Registers

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>Flag bit</th>
<th>Register</th>
<th>Enable bit</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>TXIF (Transmit)</td>
<td>TXIF</td>
<td>PIR1</td>
<td>TXIE</td>
<td>PIE1</td>
</tr>
<tr>
<td>RCIF (Receive)</td>
<td>RCIF</td>
<td>PIR1</td>
<td>RCIE</td>
<td>PIE1</td>
</tr>
</tbody>
</table>

Figure 11-12. PIE1 Register Bits Holding TXIE and RCIE

Figure 11-13: Serial Interrupt Enable Flags
Use of serial COM in the PIC18

In the vast majority of applications, the serial interrupt is used mainly for receiving data and is seldom used for sending data serially. This is like receiving a telephone call, where we need a ring to be notified of an incoming call. If we need to make a phone call there are other ways to remind ourselves and so no need for ringing. In receiving the phone call, however, we must respond immediately no matter what we are doing or we will miss the call. Similarly, we use the serial interrupt to receive incoming data so that it is not lost. Look at Program 11-6. Notice that the last instruction of the ISR is RETFIE and there is no clearing of the TXIF flag, since it is done by writing a byte to TXREG.

In Figure 11-13, notice the role of PEIE (PEripheral Interrupt Enable) in allowing serial communication interrupts and other interrupts to come in. This is in addition to the GIE bit discussed in Section 11-1.

For Program 11-6 we assume an 8-bit switch is connected to PORTD. In this program, the PIC18 reads data from PORTD and writes it to TXREG continuously to be transmitted serially. We assume that XTAL = 10 MHz. The baud rate is set at 9600.

;Program 11-6
ORG 0000H
GOTO MAIN ;bypass interrupt vector table
;--on default all interrupts go to to address 00008
ORG 0008H ;interrupt vector table
BTFSC PIR1,TXIF ;Is interrupt due to transmit?
BRA TX_ISR ;Yes. Then go to ISR
RETFIE ;No. Then return
ORG 0040H
TX_ISR ;service routine for TXIF
MOVWF PORTD,TXREG;load new value, clear TXIF
RETFIE ;then return to main
;--the main program for initialization
ORG 00100H
MAIN SETF TRISD ;make PORTD input
MOVLW 0x20
MOVWF TXSTA ;enable transmit and choose
;low baud
MOVLW D'15'
MOVLW 9600 bps
; (Fosc / (64 * Speed) - 1)
MOVF SPBRG ;write to reg
BCF TRISC, TX ;make TX pin of PORTC an
;output pin
BSF RCSTA, SPEN ;enable the serial port
BSF PIE1,TXIE ;enable TX interrupt
BSF INTCON,PEIE ;enable peripheral interrupts
BSF INTCON,GIE ;enable interrupts globally
OVER BRA OVER ;stay in this loop forever
END
Program 11-7 is a modification of Program 11-6 with receive interrupt. In this program, the PIC18 gets data from PORTD and sends it to TXREG continuously while incoming data from the serial port is sent to PORTB. We assume that XTAL = 10 MHz and the baud rate = 9600. This program can be verified by connecting your PICTrainer to the serial port of the x86 IBM PC and using HyperTerminal to send and receive data between the PIC Trainer and the IBM PC.

Program 11-7

| ORG 0000H |
| GOTO MAIN |
| ;bypass interrupt vector table |

---on default all interrupts go to address 00008

| ORG 00008H |
| ;interrupt vector table |

HI_ISR

| BTFS PIC18, TXIF |
| ;is it TX interrupt? |
| BRA TX_ISR |
| ;Yes. Then branch to TX_ISR |
| BTFS PIC18, RCIF |
| ;is it RC interrupt? |
| BRA RC_ISR |
| ;Yes. Then branch to RC_ISR |
| RETIE |

TX_ISR

| MOVFF PORTD, TXREG |
| ;loading TXREG clears TXIF |
| GOTO HI_ISR |

RC_ISR

| MOVFF RCREG, PORTB |
| ;copy received data to PORTB |
| GOTO HI_ISR |

---the main program for initialization

| ORG 00100H |

MAIN

| CLRF TRISB |
| ;PORTB as an output |
| SETF TRISD |
| ;make PORTD input |
| MOVLW 0x20 |
| ;enable transmit and choose low baud |
| MOVF TXSTA |
| ;write to reg |
| MOVLW d'15' |
| ;9600 bps (Fosc / (64 * Speed) - 1) |
| MOVF SPBRG |
| ;write to reg |
| BCF TRISC, TX |
| ;make TX pin of PORTC an output pin |
| BSF TRISC, RX |
| ;make RCV pin of PORTC an input pin |
| MOVLW 0x90 |
| ;enable receive and serial port |
| MOVF RCSTA |
| ;write to reg |
| BSF PIE1, TXIE |
| ;enable TX interrupt |
| BSF PIE1, RCIE |
| ;enable receive interrupt |
| BSF INTCON, PEIE |
| ;enable peripheral interrupts |
| BSF INTCON, GIE |
| ;enable interrupts globally |

OVER

| BRA OVER |
| ;stay in this loop forever |

//Program 11-7C (This is the C version of Program 11-7)

#include <pl8F458.h>

void chk_isr(void);
void TX_ISR(void);
void RC_ISR(void);

#pragma code My_HiPrio_Int=0x08 //high-priority interrupt
void My_HiPrio_Int (void) {

  _asm
    GOTO chk_isr
  _endasm

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void chk_isr (void)
{
    if (PIR1bits.TXIF==1) // Transmit caused interrupt?
        TX_ISR(); // Yes. Execute Transmit program
    if (PIR1bits.RCIF==1) // Receive caused interrupt?
        RC_ISR(); // Yes. Execute Receive program
}

Review Questions

1. True or false. All interrupts, including the TXIF and RXIF, are directed to a single location in the interrupt vector table.

2. What address in the interrupt vector table is assigned to the serial interrupt?

3. Which register do the TXIF and RXIF flags belong to? Show how they are enabled.

4. Assume that the RCIF bit is enabled. Explain how this interrupt gets activated and its actions upon activation.

5. True or false. Upon reset, the serial interrupts are active and ready to go.

6. True or false. The last two instructions of the ISR for the receive interrupt are:

   BCF  RIR1,RCIF
   RETFIE

7. Answer Question 6 for the transmit interrupt.
SECTION 11.5: PORTB-CHANGE INTERRUPT

The four pins of the PORTB (RB4–RB7) can cause an interrupt when any changes are detected on any one of them. They are referred to as “PORTB-Change interrupt” to distinguish them from the INTO–INT2 interrupts, which are also located on PORTB (RB0–RB2). See Figure 11-15. The PORTB-Change interrupt has a single interrupt flag called RBIF and is located in the INTCON register. This is shown in Figure 11-14. In Figure 11-14, also notice the RBIE bit for enabling the PORTB-Change interrupt. In Section 11.3 we discussed the external hardware interrupts of INTO, INT1, and INT2. Notice the following differences between the PORTB-Change interrupt and INTO–INT2 interrupts:

(a) Each of the INTO–INT2 interrupts has its own pin and is independent of the others. These interrupts use pins PORTB.0 (RB0), PORTB.1 (RB1), and PORTB.2 (RB2), respectively. The PORTB-change interrupt uses all four of the PORTB pins RB4–PB7 and is considered to be a single interrupt even though it can use up to four pins.

(b) While each of the INTO–INT2 interrupts has its own flag, and is independent of the others, there is only a single flag for the PORTB-Change interrupt.

(c) While each of the INTO–INT2 interrupts can be programmed to trigger on the negative or positive edge, the PORTB-Change interrupt causes an interrupt if any of its pins changes status from HIGH to LOW, or LOW to HIGH. See Figure 11-16.

PORTB-Change is widely used in keypad interfacing as we will see in Chapter 12. Another way to use the PORTB-Change interrupt is shown in Program 11-8. In that program, we assume a door sensor is connected to pin RB4 and upon opening or closing the door, the buzzer will sound. See Figure 11-17.

<table>
<thead>
<tr>
<th>D7</th>
</tr>
</thead>
<tbody>
<tr>
<td>GIE</td>
</tr>
<tr>
<td>D0</td>
</tr>
</tbody>
</table>

**GIE (Global Interrupt Enable)**

GIE = 0  Disables all interrupts. If GIE = 0, no interrupt is acknowledged, even if they are enabled individually.

If GIE = 1, interrupts are allowed to happen. Each interrupt source is enabled by setting the corresponding interrupt enable bit.

**RBIE**  PORTB-Change Interrupt Enable

- = 0 Disables PORTB-Change interrupt
- = 1 Enables PORTB-Change interrupt

**RBIF**  PORTB-Change Interrupt Flag.

- = 0 None of the RB4–RB7 pins have changed state
- = 1 At least one of the RB4–RB7 pins have changed state

The RBIE bit, along with the GIE, must be set high for any changes on the pins RB4–RB7 to cause an interrupt. The RB4–RB7 pins must also have been configured as input pins for this interrupt to work. In order to clear the RBIF flag we must read the pins of RB4–RB7 and use the instruction “BCF INTCON,RBIF”.

Figure 11-14. INTCON (Interrupt Control) Register
Figure 11-15. PORTB-Change Interrupt Pins

Figure 11-16. PORTB-Change Interrupt (RBIF)

Figure 11-17. PORTB-Change Interrupt for Program 11-8
For Program 11-8 we have connected a door sensor to pin RB4 and a buzzer to pin RC7. In this program, every time the door is opened, it sounds the buzzer by sending it a square wave frequency.

;Program 11-8
MYREG EQU 0x20  ;set aside a couple of registers
DELRG EQU 0x80  ;for buzzer time delay
ORG 0000H
GOTO MAIN  ;bypass interrupt vector table
;--on default all interrupts go to to address 00008
ORG 0008H  ;interrupt vector table
BTFSS INTCON, RBIF  ;Did we get here due to RBIF?
RETFIE  ;No. Then return to main
GOTO PB_ISR  ;Yes. Then go ISR
;--the main program for initialization
ORG 00100H
MAIN BCF TRISC,7  ;PORTC.7 as an output for buzzer
BSF TRISB,4  ;PORTB.4 as an input for interrupt
BSF INTCON, RBIE  ;enable PORTB-Change interrupt
BSF INTCON, GIE  ;enable interrupts globally
OVER BRA OVER  ;stay in this loop forever
;------------------------ISR for PORTB-Change INT
PB_ISR
ORG 200H
MOVF PORTB, W  ;we must read PORTB
MOVLW D'250'  ;for delay
MOVWF MYREG
BUZZ BTG PORTC, 7  ;toggle PC7 for the buzzer
MOVLW D'255'  ;for delay
MOVWF DELRG
DELAY DECF DELRG, F
BNZ DELAY  ;keep sounding the buzzer
DECF MYREG, F
BNZ BUZZ
BCF INTCON, RBIF  ;and clear RBIF interrupt flag bit
RETFIE
END

Notice for the PORTB-Change interrupt, there is no need to enable the PEIE; however, we still need to enable the GIE bit.

It must be noted again that, while the INT0–INT2 interrupts each have their own interrupt flags, there is only a single interrupt flag (RBIF) for all the four pins of the RB4–RB7. Examine Program 11-9. For this program, we assume that each of the pins RB4 and RB5 is connected to an external switch. Upon activation of the SW, an LED reflects the status. See Figure 11-18.
Figure 11-18: PORTB-Change Interrupt for Program 11-9

For Program 11-9 we have connected SW1 and SW2 to pins RB4 and RB5 respectively. In this program, the activation of SW1 and SW2 will result in changing the state of LED1 and LED2 respectively.

;Program 11-9
ORG 0000H
GOTO MAIN ;bypass interrupt vector table
;--on default all interrupts go to to address 00008
ORG 0008H ;interrupt vector table
BTFSS INTCON, RBIF ;Did we get here due to RBIF?
RETFIE ;No. Then return to main
GOTO PB_ISR ;Yes. Then go ISR
;--the main program for initialization
ORG 0100H
MAIN BCF TRISC, 4 ;PC4 as an output
BCF TRISC, 5 ;PC5 as an output
BSF TRISB, 4 ;PB4 as an input for the interrupt
BSF TRISB, 5 ;PB5 as an input for the interrupt
BSF INTCON, RBIE ;enable PORTB interrupt
BSF INTCON, GIE ;enable interrupts globally
OVER BRA OVER ;stay in this loop forever
;-----------------------------ISR for PORTB_Change
PB_ISR
ORG 200H
MOVFF PORTB, W ;get the status of switches
ANDLW 0x30 ;mask unneeded bits
MOVFF W, PORTC ;update LEDs
BCF INTCON, RBIF ;clear RBIF interrupt flag bit
RETFIE
END
//Program 11-9C (This is the C version of Program 11-9)
#include <p18F458.h>
#define LED1 PORTCbits.RC4
#define LED2 PORTCbits.RC5

void chk_isr(void);
void RBINT_ISR(void);

#pragma code My_HiPrio_Int =0x0008 //high-priority int
void My_HiPrio_Int (void)
{
    _asm
        GOTO chk_isr
    _endasm
}
#pragma code
#pragma interrupt chk_isr //used for high-priority int
void chk_isr (void)
{
    if (INTCONbits.RBIF==1)    //RBIF caused interrupt?
        RBINT_ISR( );       //Yes. Execute ISR program
}
void main(void)
{
    TRISCbits.TRISC4=0;     //RC4 = OUTPUT
    TRISCbits.TRISC5=0;     //RC5 = OUTPUT
    TRISBbits.TRISB4 = 1;   //RB4 = INPUT for interrupt
    TRISBbits.TRISB5 = 1;   //RB5 = INPUT for interrupt
    INTCONbits.RBIF=0;      //clear RBIF
    INTCONbits.RBIE=1;      //enable RB interrupt
    INTCONbits.GIE=1;       //enable all interrupts globally
    while(1);              //keep looping until interrupt comes
}
void RBINT_ISR(void)
{
    LED1=PORTBbits.RB4;
    LED2=PORTBbits.RB5;
    INTCONbits.RBIF=0;      //clear RBIF flag
}

Review Questions

1. True or false. There is a single interrupt for each of the PORTB pins.
2. What address in the interrupt vector table is assigned to the PORTB-Change interrupt?
3. Which register do the RBIF and RBIE flags belong to? Show how RBIE is enabled.
4. Give the last two instructions of the ISR for the PORTB-Change interrupt.
5. True or false. Upon reset, the RBIF interrupt is active and ready to go.
SECTION 11.6: INTERRUPT PRIORITY IN THE PIC18

The next topic that we must deal with is what happens if two interrupts are activated at the same time? Which of these two interrupts is responded to first? Interrupt priority is the main topic of discussion in this section.

Setting interrupt priority

In the PIC18 microcontroller, there are only two levels of interrupt priority: (a) low level, and (b) high level. While address 0008 is assigned to high-priority interrupts, the low-priority interrupts are directed to address 0018 in the interrupt vector table. See Table 11-5. Upon power-on reset, all interrupts are automatically designated as high priority and will go to address 00008H. This is done to make the PIC18 compatible with the earlier generation of PIC microcontrollers such as PIC16xxx. We can make the PIC18 a two-level priority system by way of programming the IPEN (interrupt priority enable) bit in the RCON register. Figure 11-19 shows the IPEN bit of the RCON register. Upon power-on reset, the IPEN bit contains 0, making the PIC18 a single priority level chip, just like the PIC16xxx. To make the PIC18 a two-level priority system, we must first set the IPEN bit to HIGH. It is only after making IPEN = 1 that we can assign a low priority to any of the interrupts by programming the bits called IP (interrupt priority). Figure 11-20 shows IPR1 (interrupt priority register) with the IP bits for TXIP, RCIP, TMR1IP, and TMR2IP. If IPEN = 1, then the IP bit will take effect and will assign a given interrupt a low priority. As a result of assigning a low priority to a given interrupt, it will land at the address 0018 instead of 0008 in the interrupt vector table. The IP (interrupt priority) bit along with the IF (interrupt flag) and IE (interrupt enable) bits will complete all the flags needed to program the interrupts for the PIC18. Table 11-6 shows the three flags and the registers they belong to for some of the interrupts used in this chapter. In Table 11-6, notice the absence of the INT0 priority flag. The INT0 has only one priority and that is high priority. That means all the PIC18 interrupts can be assigned a low or high priority level, except the external hardware interrupt of INT0. Study Figures 11-22 through 11-25 very carefully. When examining these figures, the following point must be noted. By making IPEN = 1, we enable the interrupt priority feature. Now we must also enable two bits to enable the interrupts: (a) We must set GIEH = 1. The GIEH bit is part of the INTCON register (Figure 11-21) and is the same as GIE, which we have used in previous sections. In this regard there is no difference between the priority and no-priority systems. (b) The second bit we must set high is GIEL (part of INTCON). Making GIEL = 1 will enable all the interrupts whose IP = 0. That means all the interrupts that have been given the low priority will be forced to vector location 00018H.

Table 11-5: Interrupt Vector Table for the PIC18

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>ROM Location (Hex)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power-on-Reset</td>
<td>0000</td>
</tr>
<tr>
<td>High-priority Interrupt</td>
<td>0008 (Default upon power-on reset)</td>
</tr>
<tr>
<td>Low-priority Interrupt</td>
<td>0018 (Selected with IP bit)</td>
</tr>
</tbody>
</table>

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IPEN  Interrupt Priority Enable bit
   0 = All the interrupts are directed to the vector location 0008 (default).
   1 = Interrupts can be assigned a low or high priority.

The importance of IPEN: Upon power-on reset, all the interrupts of PIC18 are directed to location 0008, making it a single-priority system, just like PIC16xxx. To prioritize the PIC18 interrupts into low- and high-level priorities, we must make IPEN = 1. When IPEN = 1, we can assign either a low or a high priority to any of the interrupts by manipulating the corresponding bit in the IPR (interrupt priority register) for that interrupt. When interrupt priority is enabled (IPEN = 1), we must set both the GIEH and GIEL bits to high in order to enable the interrupts globally. Notice in Figure 11-21 that GIE is the same as GIEH.

Figure 11-19. RCON Register. IPEN Allows Prioritizing the Interrupt into 2 Levels

<table>
<thead>
<tr>
<th>RCIP</th>
<th>TXIP</th>
<th>TMR2IP</th>
<th>TMR1IP</th>
</tr>
</thead>
</table>
| RCIP | USART (Serial COM) Receive Interrupt Priority bit
   0 = Low priority
   1 = High priority |
| TXIP | USART (Serial COM) Transmit Interrupt Priority bit
   0 = Low priority
   1 = High priority |
| TMR2IP | Timer2 Interrupt Priority bit
   0 = Low priority
   1 = High priority |
| TMR1IP | Timer1 Interrupt Priority bit
   0 = Low priority
   1 = High priority |

Figure 11-20. IPR1 Peripheral Interrupt Priority Register 1

Table 11-6: Interrupt Flag Bits for PIC18 Timers

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>Flag bit (Register)</th>
<th>Enable bit (Register)</th>
<th>Priority (Register)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Timer0</td>
<td>TMR0IF (INTCON)</td>
<td>TMR0IE (INTCON)</td>
<td>TMR0IP (INTCON2)</td>
</tr>
<tr>
<td>Timer1</td>
<td>TMR1IF (PIR1)</td>
<td>TMR1IE (PIE1)</td>
<td>TMR1IP (IPR1)</td>
</tr>
<tr>
<td>Timer2</td>
<td>TMR2IF (PIR1)</td>
<td>TMR2IE (PIE1)</td>
<td>TMR2IP (IPR1)</td>
</tr>
<tr>
<td>Timer3</td>
<td>TMR3IF (PIR3)</td>
<td>TMR3IE (PIE2)</td>
<td>TMR3IP (IPR2)</td>
</tr>
<tr>
<td>INT1</td>
<td>INT1IF (PIR1)</td>
<td>INT1IE (PIE1)</td>
<td>INT1IP (INTCON3)</td>
</tr>
<tr>
<td>INT2</td>
<td>INT2IF (PIR1)</td>
<td>INT2IE (PIE1)</td>
<td>INT2IP (INTCON)</td>
</tr>
<tr>
<td>TXIF</td>
<td>TXIF (PIR1)</td>
<td>TXIE (PIE1)</td>
<td>TXIP (IPR1)</td>
</tr>
<tr>
<td>RCIF</td>
<td>RCIF (PIR1)</td>
<td>RCIE (PIE1)</td>
<td>RCIP (IPR1)</td>
</tr>
<tr>
<td>RB_INT</td>
<td>RBIF (INTCON)</td>
<td>RBIE (INTCON)</td>
<td>RBIP (INTCON2)</td>
</tr>
</tbody>
</table>

Note: INT0 has only the high-level priority.
GIE (Global Interrupt Enable) This is also referred to as GIEH

GIE = 0  Disables all interrupts. If GIE = 0, no interrupt is acknowledged, even if they are enabled individually.

If GIE = 1, interrupts are allowed to happen. Each interrupt source is enabled by setting the corresponding interrupt enable bit.

GIEL (This is called Global Interrupt Enable Low to distinguish it from D7)

While GIH (D7) is used to enable or disable all the interrupts (both low and high priority), the GIEL (D6) is used to enable or disable only the low-priority interrupts. That means GIEL works only when the IPEN (Interrupt Priority Enable) bit is enabled.

If IPEN = 0

\[
\begin{align*}
\text{GIEL} = 0 & \quad \text{Disables all peripheral interrupts (same as PEIE)} \\
\text{GIEL} = 1 & \quad \text{Enables all peripheral interrupts (same as PEIE)}
\end{align*}
\]

If IPEN = 1

\[
\begin{align*}
\text{GIEL} = 0 & \quad \text{Disables the Low-priority Interrupt} \\
\text{GIEL} = 1 & \quad \text{Enables the Low-priority Interrupt}
\end{align*}
\]

Figure 11-21. INTCON (Interrupt Control) Register

---

Interrupt Priority (IP) = 1 is high priority (0x0008)

Figure 11-22. Interrupts with High-Priority (IP) Flag
Figure 11-23. High-Priority Interrupts (Redrawn from PIC18 Manual)

Figure 11-24. Low- and High-Priority Interrupt Selection (Redrawn from PIC18 Manual)
Figure 11-25. Low-Priority Interrupt Selection with IP Flag (Redrawn from PIC18 Manual)
Program 11-10 uses Timer0 and Timer1 interrupts to generate square waves on pins RB1 and RB7 respectively, while data is being transferred from PORTC to PORTD. This is a repeat of Program 11-2, except Timer1 has been assigned to low priority.

; Program 11-10
ORG 0000H
GOTO MAIN ; bypass interrupt vector table

;-- high-priority interrupts go to address 00008
ORG 0008H ; high-priority interrupt vector table
BTFSC INTCON, TMR0IF ; Is it Timer0 interrupt?
BRA T0_ISR ; Yes. Then branch to T0_ISR
RETFIE 0x01 ; No. Then fast return to main

;-- low-priority interrupts go to address 00018
ORG 0018H ; low-priority interrupt vector table
BTFSC PIR1, TMR1IF ; Is it Timer1 interrupt?
BRA T1_ISR ; Yes. Then branch to T1_ISR
RETFIE ; No. Then return to main

;-- main program for initialization and keeping CPU busy
ORG 0100H ; somewhere after vector table space

MAIN
BCF TRISB, 1 ; PB1 as an output
BCF TRISB, 7 ; PB7 as an output
CLRF TRISD ; make PORTD output
SETF TRISC ; make PORTC input
MOVLW 0x08 ; Timer0, 16-bit, no prescale, ; internal clk
MOVWF TOCON ; load TOCON reg
MOVWF 0xFF ; TMR0H = FFH, the high byte
MOVWF TMR0H ; load Timer0 high byte
MOVWF 0x00 ; TMR0L = 00H, the low byte
MOVWF TMR0L ; load Timer0 low byte
BCF INTCON, TMR0IF ; clear Timer0 interrupt flag bit
BSF INTCON, TMR0IE ; enable Timer0 interrupt
MOVLW 0x0 ; Timer1, 16-bit, no prescale, ; internal clk
MOVWF T1CON ; load T1CON reg
MOVWF 0xFF ; TMR1H = FFH, the high byte
MOVWF TMR1H ; load Timer1 high byte
MOVWF 0x00 ; TMR1L = 00H, the low byte
MOVWF TMR1L ; load Timer1 low byte
BCF PIR1, TMR1IF ; clear Timer1 interrupt flag bit
BSF PIE1, TMR1IE ; enable Timer1 interrupt
BCF IP1R1, TMR1IP ; make Timer1 low-priority interrupt
BSF RCON, IPEN ; enable priority levels
BSF INTCON, GIEL
BSF INTCON, GIEH ; enable interrupts globally
BSF T1CON, TMR1ON ; start Timer1
BSF TOCON, TMR0ON ; start Timer0

;-- keeping CPU busy waiting for interrupt
OVER
MOVFF PORTC, PORTD ; send data from PORTC to PORTD
BRA OVER ; stay in this loop forever
ISR for Timer0

T0_ISR

ORG 200H
MOVLW 0xFF ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0x00 ;TMR0L = 00H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BTG PORTB,1 ;toggle RB1
BCF INTCON,TMR0IF ;clear timer interrupt flag bit
RETFIE 0x01

ISR for Timer1

T1_ISR

ORG 300H
MOVLW 0xFF ;TMR1H = FFH, the high byte
MOVWF TMR1H ;load Timer0 high byte
MOVLW 0x00 ;TMR1L = 00H, the low byte
MOVWF TMR1L ;load Timer1 low byte
BTG PORTB,7
BCF PIR1,TMR1IF ;clear Timer1 interrupt flag bit
RETFIE

Program 11-11 has four interrupts. It uses Timer0 and Timer1 interrupts to generate square waves on pins RC0 and RC1 respectively. It also uses the transmit and receive interrupts to send and receive data serially. Data from PORTD is transmitted and the received byte is placed on PORTB. Timer0 and receive ISRs have the high priority while Timer1 and send ISRs are assigned a low priority level.

;Program 11-11
ORG 0000H
GOTO MAIN ;bypass interrupt vector table

;--high-priority interrupts go to address 00008
ORG 00008 ;need to redirect because not enough space
GOTO CHK_HI_PRIO

;--no need to redirect because we have plenty of space
ORG 00018 ;low-priority interrupt vector table
BTFSC PIR1,TMR1IF ;is it Timer1 interrupt?
BRA T1_ISR ;Yes. Then branch to T1_ISR
BTFSC PIR1,TXF ;Did we get here due to TxD?
BRA TX_ISR ;Yes. Then branch to TX_ISR
RETFIE ;No. Then return to main

CHK_HI_PRIO ORG 0x50
BTFSC INTCON,TMR0IF ;Is it Timer0 interrupt?
BRA T0_ISR ;Yes. Then branch to T0_ISR
BTFSC PIR1,RCIF ;Did we get here due to RCV int?
BRA RC_ISR ;Yes. Then branch to RC_ISR
RETFIE 0x01 ;No. Then return to main

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;main program for initialization and keeping CPU busy
ORG 0100H ;somewhere after vector table space

MAIN BCF TRISC,RC0
BCF TRISC,RC1
CLRF TRISB ;make PORTB output
SETF TRISD ;make PORTD input
MOVLW 0x08 ;Timer0, 16-bit, no prescale,
;internal clk
MOVWF T0CON ;load T0CON reg
MOVLW 0xFF ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0x00 ;TMR0L = 00H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BCF INTCON,TMR0IF ;clear Timer0 interrupt flag bit
MOVLW 0x0 ;Timer1, 16-bit, no prescale,
;internal clk
MOVWF T1CON ;load T1CON reg
MOVLW 0xFF ;TMR1H = FFH, the high byte
MOVWF TMR1H ;load Timer1 high byte
MOVLW 0x00 ;TMR1L = 00H, the low byte
MOVWF TMR1L ;load Timer1 low byte
BCF PIR1,TMR1IF ;clear Timer1 interrupt flag bit
MOVLW 0x20 ;enable transmit and choose low baud
MOVWF TXSTA ;write to reg
MOVLW D'15' ;9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ;write to reg
MOVLW 0x90 ;enable receive and serial port
MOVWF RCSTA ;write to reg
BCF TRISC, TX ;make TX pin of PORTC an output pin
BSF TRISC, RX ;make RCV pin of PORTC an input pin
BSF RCON,IPEN
BSF PIE1,RCIE
BSF PIE1,TXIE ;enable TX interrupt
BSF INTCON,TMR0IE ;enable Timer0 interrupt
BSF PIE1,TMR1IE ;enable Timer1 interrupt
BCF IPR1,TMR1IP ;make Timer1 a low-priority interrupt
BCF IPR1,TXIP ;make Transmit a low-priority interrupt
BSF T0CON,TMRON ;start Timer0
BSF T1CON,TMR1ON ;start Timer1
BSF INTCON,GIEL ;enable low-priority interrupts
BSF INTCON,GIEH ;enable high-priority interrupts

;--keeping CPU busy waiting for interrupt
OVER BRA OVER ;stay in this loop forever

;------------------ISR for Timer0
T0_ISR ORG 200H
MOVLW 0xFF ;TMR0H = FFH, the high byte
MOVWF TMR0H ;load Timer0 high byte
MOVLW 0x00 ;TMR0L = 00H, the low byte
MOVWF TMR0L ;load Timer0 low byte
BTG PORTC,0 ;toggle RB1

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BCF INTCON, TMR0IF ; clear timer interrupt flag bit
RETFIE 0x01

; -----------------------------ISR for Timer1
T1_ISR  ORG 300H
MOVLW 0xFF ; TMR1H = FFH, the high byte
MOVWF TMR1H ; load Timer0 high byte
MOVLW 0x00 ; TMR1L = 00H, the low byte
MOVWF TMR1L ; load Timer1 low byte
BTG PORTC, 1
BCF PIR1, TMR1IF ; clear Timer1 interrupt flag bit
RETFIE

; -----------------Transmit ISR
TX_ISR
BCF PIR1, TXIF ; clear TX interrupt flag bit
MOVFF PORTD, TXREG
RETFIE

; ---------
RC_ISR
MOVFF RCREG, PORTB ; copy received data to PORTD
BCF PIR1, RCIF ; clear RCIF
RETFIE 1
END

Example 11-3
For Program 11-11 (or Program 11-11C), discuss what happens: (a) if interrupt RCIF is activated when the PIC18 is serving the Timer1 interrupt, (b) Timer1 is activated when Timer0 is being served, (c) RCIF and TMR0IF and TMR1IF are activated at the same time.

Solution:
In Program 11-11, notice that the Receive (RCIF) and Timer0 (TMR0IF) interrupts are assigned to high priority while the Transmit (TXIF) and Timer1 (TMR1IF) interrupts have low priority. As a result we have the following:
(a) if the RCIF is activated during the execution of the Timer1 ISR, the Receive interrupt comes in and its ISR is executed first because it has a higher priority. After it is finished, the PIC18 goes back and finishes the Timer1 ISR.
(b) if the TMR1IF is activated during the execution of the Timer0 ISR, it is ignored because it has lower priority. After the Timer0 ISR is finished, the PIC18 will execute the Timer1 ISR.
(c) If all three, RCIF, TMR0IF, and TMR1IF, are activated at the same time, the Received ISR and and Timer0 ISR are taken care of first because they are assigned to high priority. Between the Receive and Timer0 interrupts, the Timer0 ISR is served first due to the programming sequence we have set in the interrupt vector table for the high-priority interrupts. That means if these three interrupts are activated at the same time, they are executed in the following sequence: Timer0 ISR, Receive ISR, and Timer1 ISR.
Low-priority interrupt programming in C

As we saw in the last four sections, the C18 compiler uses the reserved keyword `interrupt` to designate an interrupt as high priority. To assign low priority level to a given interrupt, it uses the keyword `interruptlow`. See Table 11-7. This is shown in Program 11-11C, which is a repeat of Program 11-11 in C.

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>ROM Location</th>
<th>C18 keyword</th>
</tr>
</thead>
<tbody>
<tr>
<td>High-priority</td>
<td>0x0008 (Default)</td>
<td><code>interrupt</code></td>
</tr>
<tr>
<td>Low-priority</td>
<td>0x0018 (Selected with IP bit)</td>
<td><code>interruptlow</code></td>
</tr>
</tbody>
</table>

Program 11-11C has four interrupts. It uses Timer0 and Timer 1 interrupts to generate square waves on pins RC1 and RC7, respectively. It also uses the transmit and receive interrupts to send and receive data serially. Data from PORTB is transmitted and the received byte is placed on PORTD. Timer0 and receive ISRs have the high priority while Timer1 and transmit ISRs are assigned low priority level.

```c
#include <p18F458.h>
define myPC0bit PORTCbits.RC0
define myPC1bit PORTCbits.RC1

void chk_isr(void);
void chk_low_isr(void);
void T0_ISR(void);
void T1_ISR(void);
void TXISR(void);
void RC_ISR(void);

#pragma code My_HiPrio_Int =0x0008 //high-priority int
void My_HiPrio_Int (void)
{
    _asm
    GOTO chk_isr
    _endasm
}

#pragma code My_Lo_Prio_Int =0x00018 //low-priority int
void My_Lo_Prio_Int (void)
{
    _asm
        GOTO chk_low_isr
    _endasm
}

#pragma interruptlow chk_low_isr //used for low-priority
void chk_low_isr (void)
{
```

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if(PIR1bits.TMR1IF==1)//Timer1 causes interrupt?
  T1_ISR();  //Yes. Execute Timer1 ISR
if (PIR1bits.TXIF==1) //Transmit causes interrupt?
  TX_ISR( );  //Yes. Execute Transmit ISR

#pragma interrupt chk_isr//used for high-priority interrupt
void chk_isr (void)
{
  if (PIR1bits.TMR1IF==1)//Timer0 causes interrupt?
    T0_ISR( );  //Yes. Execute Timer0 ISR
  if (PIR1bits.RCIF==1) //Receiver causes interrupt?
    RC_ISR( );  //Yes. Execute Receiver ISR
}

void main(void)
{
  TRISCbits.TRISC0=0;    //RC0 = OUTPUT
  TRISCbits.TRISC1=0;    //RC1 = OUTPUT
  TRISD = 255;           //PORTD = INPUT
  TRISB = 0;             //PORTB = OUTPUT
  T0CON=0x08;            //Timer0, 16-bit mode,
                          //no prescaler
  TMR0H=0xFF;            //load TH0
  TMR0L=0x00;            //load TL0
  INTCONbits.TMR0IF=0;   //clear TF1
  T1CON=0x08;            //Timer 1, 16-bit mode, no prescaler
  TMR1H=0xFF;            //load TH1
  TMR1L=0x00;            //load TL1
  PIR1bits.TMR1IF=0;     //clear TF1
  TXSTA=0x20;            //choose low baud rate,8-bit
                          //9600 baud rate/ XTAL = 10 MHz
  SPBRG=15;
  RCSTAbits.CREN=1;
  RCSTAbits.SPEN=1;
  TRISCbits.TRISC6=0;    //TX pin = OUTPUT
  TRISCbits.TRISC7=1;    //RCV pin = INPUT
  RCONbits.IPEN=1;
  PIE1bits.RCIE=1;
  PIE1bits.TXIE=1;      //enable TX interrupt
  INTCONbits.TMR0IE=1;
  PIE1bits.TMR1IE=1;
  IP1bits.TMR1IP=0;      //make Timer1 a low-priority
  IP1bits.TXIP=0;        //make TX a low-priority
  T0CONbits.TMROON=1;   //turn on T0
  T1CONbits.TMR1ON=1;   //turn on T1
  INTCONbits.GIEL=1;    //enable low-priority interrupts
  INTCONbits.GIEH=1;    //enable high-priority interrupts
  while(1);            //keep looping until interrupt comes
ISR for Timer0

```c
void T0_ISR(void)
{
    TMR0H=0xFF;  //load TH0
    TMR0L=0x00;  //load TL0
    myPC0bit=~myPC0bit;  //toggle RB1
    INTCONbits.TMR0IF=0;  //clear TF0
}
```

ISR for Timer1

```c
void T1_ISR(void)
{
    TMR1H=0xFF;  //load TH0
    TMR1L=0x00;  //load TL0
    PIR1bits.TMR1IF=0;  //clear TF1
    myPC1bit=~myPC1bit;  //toggle RB1
}
```

ISR for Transmit

```c
void TX_ISR(void)
{
    TXREG=PORTD;  //clear Tx Interrupt flag
}
```

ISR for Receive

```c
void RC_ISR(void)
{
    PORTB=RCREG;
    RCSTAbits.CREN=0;  //clear CREN to clear any error
    RCSTAbits.CREN=1;  //set CREN for continuous reception
}
```

Interrupt inside an interrupt

What happens if the PIC18 is executing an ISR belonging to an interrupt and another interrupt is activated? In such cases, a high-priority interrupt can interrupt a low-priority interrupt. This is an interrupt inside an interrupt. In the PIC18 a low-priority interrupt can be interrupted by a higher-priority interrupt, but not by another low-priority interrupt. Although all the interrupts are latched and kept internally, no low-priority interrupt can get the immediate attention of the CPU until the PIC18 has finished servicing all the high-priority interrupts. The GIE (which is also called GIEH) and GIEL bits play an important role in the process of the interrupt inside the interrupt. Regarding the interrupt inside an interrupt concept, the following points must be emphasized:

1. When a high-priority interrupt is vectored into address 0008H, the GIE bit is disabled (GIEH = 0), thereby blocking another interrupt (low or high) from coming in. The RETFIE instruction at the end of the ISR will enable the GIE (GIE = 1) automatically, which allows interrupts to come in again. If we want to allow another high-priority interrupt to come in during the execution of the current ISR, then we must make GIE = 1 at the beginning of the current ISR.
2. When a low-priority interrupt is vectored into address 0018H, the GIEL bit is disabled (GIEL = 0), thereby blocking another low-priority interrupt from...
coming in. The RETFIE instruction at the end of the ISR will enable the GIEL (GIEL = 1) automatically, which allows low-priority interrupts to come in again. Notice that the low-priority interrupt cannot block a high-priority interrupt from coming in during the execution of the current low-priority ISR because GIEH is still set to one (GIEH = 1).

3. When two or more interrupts have the same priority level. In this case, they are serviced according to the sequence by which the program checks them in the interrupt vector table. We saw many examples of that in this chapter. See Example 11-3.

**Fast context saving in task switching**

In many applications, such as multitasking real-time operating systems (RTOS), the CPU brings in one task (job or process) at a time and executes it before it moves to the next one. In executing each task, which is often organized as the interrupt service routine, access to all the resources of the CPU is critical in performing the task in a timely manner. In early CPUs, the limited number of registers forced programmers to save the entire contents of the CPU on the stack before execution of the new task. This saving of the CPU contents before switching to a new task is called context saving (or context switching). The use of the stack as a place to save the CPU’s contents is tedious, time consuming, and slow. For this reason some CPUs such as x86 microprocessors have instructions such as PUSH (Push All) and POPA (Pop All), which will push and pop all the main registers onto the stack with a single instruction. Because the PIC18 has numerous general purpose registers, there is no need for using the stack to save the CPU’s general purpose registers. However, each task generally needs the key registers of WREG, BSR, and STATUS. For that reason the PIC18 automatically saves these three registers internally in shadow registers when a high-priority interrupt is activated. This way, the three key registers of the main task are saved internally. To restore the original contents of these three key registers, one must use instruction “RETFIE 0x01” instead of “RETFIE” at the end of the high-priority ISR. The “RETFIE 0x01” is called fast context saving in PIC18 literature. Regarding fast context saving in the PIC18, two important points must be noted:

1. It is not available for the low-priority interrupts, and works only for high-priority interrupts. That means that when a low-priority interrupt is activated, there is no fast context saving and we must save these three registers at the beginning of the low-priority ISR, if they are being used by the low-priority ISR.

2. The shadow registers keeping these three key registers have a depth of one, meaning that there is only one of them. For that reason, the fast context saving works only when a high-priority ISR is activated during the main subroutine. If two or three high-priority interrupts are activated at the same time, only the first ISR can use the fast context saving because the depth of shadow registers is only one. In that case, the second and third ISRs must save these key registers at the beginning of the body of their ISRs. This should not be difficult because we know the sequence by which the ISRs are executed, as we saw in many examples in this chapter.
Interrupt latency

The time from the moment an interrupt is activated to the moment the CPU starts to execute the code at the vector address of 0008H (or 0x0018H) is called the interrupt latency. This latency can be anywhere from 2 to 4 instruction cycle times depending on whether the source of the interrupt is an internal (e.g., timers) or external hardware (e.g., hardware INTx and PORTB-Change) interrupt. The duration of an interrupt latency can also be affected by the type of the instruction in which the CPU was executing when the interrupt comes in. It takes slightly longer in cases where the instruction being executed lasts for two instruction cycles (e.g., MOVFF reg,reg) compared to the instructions that last for only one instruction cycle time (e.g., ADDWL). See PIC18 for the timing data sheet.

Triggering the interrupt by software

Sometimes when we need to test an ISR by way of simulation. This can be done with simple instructions to set the interrupts HIGH and thereby cause the PIC18 to jump to the interrupt vector table. For example, if the TMR1IE bit for Timer 1 is set, an instruction such as “BSF INTCN, TMR1IF” will interrupt the PIC18 in whatever it is doing and force it to jump to the interrupt vector table. In other words, we do not need to wait for Timer 1 to roll over to have an interrupt. We can cause an interrupt with an instruction that raises the interrupt flag.

Review Questions

1. True or false. Upon reset, all interrupts have the same priority.
2. Which bit of what register is used to enable the interrupt priority option in the PIC18? Is it a bit-addressable register?
3. Which register has the TXIP bit? Show how to assign it low priority.
4. Assume that INT0 and INT1 have the same low priority. Explain what happens if both INT0 and INT1 are activated at the same time. Also assume that INT0 is checked first in the program for the interrupt vector table.
5. Explain what happens if a higher-priority interrupt is activated while the PIC18 is serving a lower-priority interrupt (i.e., executing a lower-priority ISR).

SUMMARY

An interrupt is an external or internal event that interrupts the microcontroller to inform it that a device needs its service. Every interrupt has a program associated with it called the ISR, or interrupt service routine. The PIC18 has many sources of interrupts, depending on the family members. Some of the most widely used interrupts are for the timers, external hardware interrupts, and serial communication. When an interrupt is activated, the IF (Interrupt flag) bit is raised.

The PIC18 can be programmed to enable (unmask) or disable (mask) an interrupt, which is done with the help of the GIE (global interrupt enable) and IE (interrupt enable) bits. The PIC18 has two levels of priority, low and high. Upon power-on reset, all the interrupts are designated as high priority and are directed to address 0008 in the interrupt vector table. This default setting can be altered with the help of the IP (interrupt priority) bits. By programming the IP bit, we can make
an interrupt a low priority and force it to land at address 0x00018 in the interrupt vector table. This chapter also showed how to program PIC18 interrupts in both Assembly and C languages.

PROBLEMS

SECTION 11.1: PIC18 INTERRUPTS

1. Which technique, interrupt or polling, avoids tying down the microcontroller?
2. List some of the interrupt sources in the PIC18.
3. In the PIC18 what memory area is assigned to the interrupt vector table?
4. True or false. The PIC18 programmer cannot change the memory address location assigned to the interrupt vector table.
5. What memory address in the interrupt vector table is assigned to low-priority interrupts?
6. What memory address in the interrupt vector table is assigned to high-priority interrupts?
7. Do we have a memory address in the interrupt vector table assigned to the Timer0 interrupt?
8. Do we have a memory address in the interrupt vector table assigned to the INT1 interrupt?
9. To which register does the GIE bit belong?
10. Why do we put a GOTO instruction at address 0?
11. What is the state of the GIE bit upon power-on reset, and what does it mean?
12. Show the instruction to enable the INT0 interrupt.
13. Show the instruction to enable the Timer0 interrupt.
14. The TMR0IE bit belongs to register ______.
15. How many bytes of address space in the interrupt vector table are assigned to the high-priority interrupt?
16. How many bytes of address space in the interrupt vector table are assigned to the low-priority interrupt?
17. To put the entire interrupt service routine in the interrupt vector table for high priority, it must be no more than ____ bytes in size.
18. True or false. The INTCON register is not a bit-addressable register.
19. With a single instruction, show how to disable all the interrupts.
20. With a single instruction, show how to disable the INT0 interrupt.
21. True or false. Upon reset, all interrupts are enabled by the PIC18.
22. In the PIC18, how many bytes of ROM space are assigned to the reset?

SECTION 11.2: PROGRAMMING TIMER INTERRUPTS

23. True or false. For each of Timer 0 and Timer1, there is a unique address in the interrupt vector table.
24. What address in the interrupt vector table is assigned to Timer1?
25. Show how to enable the Timer0 interrupt.
26. Which bit of INTCON belongs to the Timer0 interrupt? Show how it is enabled.
27. Assume that Timer0 is programmed in 8-bit mode, TMR0H = F0H, and the TMR0IE bit is enabled. Explain how the interrupt for the timer works.
28. True or false. The last two instructions of the ISR for Timer1 are:
   \[ \text{BCF } \text{PIR1, TMR1IF} \]
   \[ \text{RETFIE} \]
29. Assume that Timer1 is programmed for 16-bit mode, TMR1H = FFH, TMR1L = F8H, and the TMR1IE bit is enabled. Explain how the interrupt is activated.
30. If Timer 1 is programmed for interrupts in 8-bit mode, explain when the interrupt is activated.
31. Write a program using the Timer0 interrupt to create a square wave of 1 Hz on pin RB7 while data from PORTC is being sent to PORTD. Assume XTAL = 10 MHz.
32. Write a program using the Timer1 interrupt to create a square wave of 3 kHz on pin RB7 while data from PORTC is being sent to PORTD. Assume XTAL = 10 MHz.

SECTION 11.3: PROGRAMMING EXTERNAL HARDWARE INTERRUPTS

33. True or false. An address location is assigned to each of the external hardware interrupts INTO, INT1, and INT2.
34. What address in the interrupt vector table is assigned to INTO, INT1 and INT2? How about the pin numbers on PORTB?
35. To which register does the INT0IE bit belong? Show how it is enabled.
36. To which register does the INT1IE bit belong? Show how it is enabled.
37. Show how to enable all three external hardware interrupts.
38. Assume that the INT0IE bit for external hardware interrupt INTO is enabled and is negative edge-triggered. Explain how this interrupt works when it is activated.
39. True or false. Upon reset, all the external hardware interrupts are negative edge-triggered.
40. In Question 38, how do we make sure that a single interrupt is not recognized as multiple interrupts?
41. The INT0IF bit belongs to the _______ register.
42. The INT1IF bit belongs to the _______ register.
43. True or false. The last two instructions of the ISR for INT1 are:
   \[ \text{BCF } \text{INTCON3, INT1IF} \]
   \[ \text{RETFIE} \]
44. Explain the role of INT0IF and INT0IE in the execution of external interrupt 0.
45. Explain the role of INT1IF and INT1IE in the execution of external interrupt 1.
46. Assume that the INT1IE bit for external hardware interrupt INT1 is enabled and is positive edge-triggered. Explain how this interrupt works when it is activated. How can we make sure that a single interrupt is not interpreted as multiple interrupts?
47. True or false. INTO–INT2 are part of the PEIE group.
48. True or false. Upon power-on reset, all of INTO–INT2 are positive edge-triggered.
49. Explain the difference between positive and negative edge-triggered interrupts.
50. How do we make the hardware interrupt negative edge-triggered? 
51. True or false. INT0–INT2 must be configured as an input pin for a hardware 
interrupt to come in. 
52. Which register holds the INTEDGx bits? 

SECTION 11.4: PROGRAMMING THE SERIAL COMMUNICATION INTERRUPTS 
and 
SECTION 11.5: PORTB-CHANGE INTERRUPT 

53. True or false. Two separate interrupts are assigned to each of the interrupts, 
TXIF and RCIF. 
54. Upon power-on reset, what address in the interrupt vector table is assigned to 
the serial interrupt? How many bytes are assigned to it? 
55. To which register does the TXIF belong? Show how it is enabled. 
56. Assume that the TXIE bit for the serial interrupt is enabled. Explain how this 
interrupt gets activated and also explain its working upon activation. 
57. True or false. Upon reset, serial interrupts are blocked. 
58. True or false. The last two instructions of the ISR for the transmit interrupt are: 

BCF PIR1, TXIF 
RETFIE 

59. State how the RCIF is cleared. 
60. Assuming that the TXIE bit is set when TXIF is raised, what happens subse- 
quently? 
61. Assuming that the RCIE bit is set when RCIF is raised, what happens subse- 
quently? 
62. Write a program using interrupts to get data serially and send it to PORTD 
while at the same time any changes on PORTB.4 will cause the LED connect- 
ed to PORTC.7 to toggle. 
63. Provide the following information for the PORTB-Change interrupt. 
   (a) the flag associated with the PORTB-Change interrupt 
   (b) the register to which these flag belong 
   (c) the difference between the PORTB-Change and INT0–INT2 interrupts 
   (d) the pins that are part of the PORTB-Change interrupt 

SECTION 11.6: INTERRUPT PRIORITY IN THE PIC18 

64. True or false. Upon reset, all interrupts have high priority. 
65. What register enables the interrupt priority in the PIC18? Explain its role. 
66. Which register has the INT0IP bit? Show how to assign it low priority. 
67. Which register has TMR1IP bit? Show how to assign it low priority. 
68. Which register has the INT1IP bit? Show how to assign it low priority. 
69. Assume that INT1IP and INT2IP are both 0s. Explain what happens if both 
INT1IF and INT2IF are activated at the same time. 
70. Assume that TMR0IP and TMR1IP are both 0s. Explain what happens if both 
TMR0IF and TMR1IF are activated at the same time. 
71. If both TMR0IP and TMR1IP are set to HIGH, what happens if both are acti- 
vated at the same time?
72. If both INT1IP and INT2IP in the IP are set to HIGH, what happens if both are activated at the same time?
73. Explain what happens if a low-priority interrupt is activated while the PIC18 is serving a high-priority interrupt.
74. Explain what happens if a high-priority interrupt is activated while the PIC18 is serving a low-priority interrupt.
75. Explain the role of the GIEH bit in masking and unmasking the interrupts.
76. True or false. In PIC18, an interrupt inside an interrupt is not allowed.
77. Explain the role of the GIEL bit in masking and unmasking interrupts.
78. Explain the role of RETFIE in enabling the GIEL bit.
79. Explain the difference between the “RETFIE” and “RETFIE 1” instructions.
80. Explain the concept of fast context saving in PIC.

ANSWERS TO REVIEW QUESTIONS

SECTION 11.1: PIC18 INTERRUPTS

1. Interrupts
2. INT0 and TMR0
3. Address locations 0x0008 to 0x00017. No, it is set when the processor is designed.
4. GIE = 0 means that all interrupts are masked, and as a result no interrupts will be responded to by the PIC18.
5. Assuming GIE = 1, we need “BSF INTCON, TMR0IE”.
6. 0008 for the high-priority interrupts and 0x0018 for the low-priority interrupts.

SECTION 11.2: PROGRAMMING TIMER INTERRUPTS

1. False. There is a single address for all the timers, Timer0, Timer1, and so on.
2. 0008H
3. PIE1 and “BSF PIE1, TMR1IE” will enable the Timer1 interrupt.
4. After Timer1 is started, the timer will count up from F5H to FFH on its own while the PIC18 is executing other tasks. Upon rolling over from FFH to 00, the TMROIF flag is raised, which will interrupt the PIC18 in whatever it is doing and forces it to jump to memory location 0008 to execute the ISR belonging to this interrupt.
5. True

SECTION 11.3: PROGRAMMING EXTERNAL HARDWARE INTERRUPTS

1. True
2. Bits RB0 (PORTB.0), RB1 (PORTB.1), and RB2 (PORTB.2)
3. BSF INTCON3,INT1IE
4. Upon application of a low-to-high pulse to pin RB0, the PIC18 is interrupted in whatever it is doing and jumps to ROM location 0008H to execute the ISR.
5. False
6. When the CPU jumps to ROM location 0008 to execute the ISR, the GIE becomes 0, effectively blocking another interrupt from the same source. The last two instructions of the ISR are “BCF INCON, INT0IF” followed by “RETFIE”. While the first instruction will clear the previous request for interrupt, the second one will make GIE = 1, allowing a new interrupt to come in from the same source. That can happen only if a new low-to-high pulse is applied to the pin.
7. True
SECTION 11.4: PROGRAMMING THE SERIAL COMMUNICATION INTERRUPTS

1. True. There is only one interrupt for all all interrupts including the transfer and receive.
2. 0x0008 for high-priority interrupts and 0x0018 for low-priority interrupts.
3. “BSF PIE1, TXIE” will enable the send interrupt and “BSF PIE1, RCIE” will enable the receive interrupt.
4. The RCIF (received interrupt flag) is raised when the entire frame of data, including the stop bit, is received. As a result the received byte is delivered to the RCREG register and the PIC18 jumps to memory location 0008H to execute the ISR belonging to this interrupt. In the serial COM interrupt service routine, we must save the RCREG content before it is lost by the incoming data.
5. False
6. True
7. BCF RIR1, TXIF
   RETFIE

SECTION 11.5: PORTB-CHANGE INTERRUPT

1. False
2. All interrupts, including the PORTB-Change interrupt, go to location 0008 on default.
3. INTCON, and we enable it with the instruction “BSF INTCON, RBIF”
4. BCF INTCON, RBIF
   RETFIE
5. False

SECTION 11.6: INTERRUPT PRIORITY IN THE PIC18

1. True
2. IPEN bit of the RCON register. Yes, it is bit-addressable.
3. IPR1 and the instruction “BCF IPR1, TXIP” will do it.
4. If both are activated at the same time, INTO is serviced first because it is checked first. After INTO is serviced, INT1 is serviced.
5. We have an interrupt inside an interrupt, meaning that the lower-priority interrupt is put on hold and the higher one is serviced. After servicing this higher-priority interrupt, the PIC18 resumes servicing the lower-priority ISR.
CHAPTER 12

LCD AND KEYBOARD INTERFACE

OBJECTIVES

Upon completion of this chapter, you will be able to:

• Describe the functions of the pins of a typical LCD
• List instruction command codes for programming an LCD
• Interface an LCD to the PIC18
• Program an LCD in Assembly and C
• Explain the basic operation of a keyboard
• Describe the key press and detection mechanisms
• Interface a 4 × 4 keypad to the PIC18 using C and Assembly
This chapter explores some real-world applications of the PIC18. We explain how to interface the PIC18 to devices such as an LCD and a keyboard. In Section 12.1, we show LCD interfacing with the PIC18. In Section 12.2, keyboard interfacing with the PIC18 is shown. We use C and Assembly for both sections.

**SECTION 12.1: LCD INTERFACING**

This section describes the operation modes of LCDs, then describes how to program and interface an LCD to a PIC18 using Assembly and C.

**LCD operation**

In recent years the LCD has been finding widespread use replacing LEDs (seven-segment LEDs or other multisegment LEDs). This is due to the following reasons:

1. The declining prices of LCDs.
2. The ability to display numbers, characters, and graphics. This is in contrast to LEDs, which are limited to numbers and a few characters.
3. Incorporation of a refreshing controller into the LCD, thereby relieving the CPU of the task of refreshing the LCD. In contrast, the LED must be refreshed by the CPU (or in some other way) to keep displaying the data.
4. Ease of programming for characters and graphics.

**LCD pin descriptions**

The LCD discussed in this section has 14 pins. The function of each pin is given in Table 12-1. Figure 12-1 shows the pin positions for various LCDs.

**$V_{CC}$, $V_{SS}$, and $V_{EE}$**

While $V_{CC}$ and $V_{SS}$ provide +5 V and ground, respectively, $V_{EE}$ is used for controlling LCD contrast.

**RS, register select**

There are two very important registers inside the LCD. The RS pin is used for their selection as follows. If $RS = 0$, the instruction command code register is selected, allowing the user to send a command such as clear display, cursor at home, and so on. If $RS = 1$ the data register is selected, allowing the user to send data to be displayed on the LCD.

**R/W, read/write**

R/W input allows the user to write information to the LCD or read information from it. R/W = 1 when reading; R/W = 0 when writing.

<table>
<thead>
<tr>
<th>Table 12-1: Pin Descriptions for LCD</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Pin Symbol</strong></td>
</tr>
<tr>
<td>-------------------------------------</td>
</tr>
<tr>
<td>1</td>
</tr>
<tr>
<td>2</td>
</tr>
<tr>
<td>3</td>
</tr>
<tr>
<td>4</td>
</tr>
<tr>
<td>5</td>
</tr>
<tr>
<td>6</td>
</tr>
<tr>
<td>7</td>
</tr>
<tr>
<td>8</td>
</tr>
<tr>
<td>9</td>
</tr>
<tr>
<td>10</td>
</tr>
<tr>
<td>11</td>
</tr>
<tr>
<td>12</td>
</tr>
<tr>
<td>13</td>
</tr>
<tr>
<td>14</td>
</tr>
</tbody>
</table>
**E, enable**

The enable pin is used by the LCD to latch information presented to its data pins. When data is supplied to data pins, a high-to-low pulse must be applied to the En pin in order for the LCD to latch in the data present at the data pins. This pulse must be a minimum of 450 ns wide. In this book we call this delay the SDELAY (short delay) to distinguish it from other delays.

**D0–D7**

The 8-bit data pins, D0–D7, are used to send information to the LCD or read the contents of the LCD’s internal registers.

To display letters and numbers, we send ASCII codes for the letters A–Z, a–z, and numbers 0–9 to these pins while making RS = 1.

There are also instruction command codes that can be sent to the LCD to clear the display or force the cursor to the home position or blink the cursor. Table 12-2 lists the instruction command codes. To send any of the commands listed in Table 12-2 to the LCD, make pin RS = 0. For data, make RS = 1. Then send a high-to-low pulse to the E pin to enable the internal latch of the LCD. There are two ways to send characters (command/data) to the LCD: (1) use a delay before sending the next one, (2) use the busy flag to see if the LCD is ready for the next one.

<table>
<thead>
<tr>
<th>Code (Hex)</th>
<th>Command to LCD Instruction</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Clear display screen</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>Return home</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>Decrement cursor (shift cursor to left)</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>Increment cursor (shift cursor to right)</td>
<td></td>
</tr>
<tr>
<td>5</td>
<td>Shift display right</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>Shift display left</td>
<td></td>
</tr>
<tr>
<td>8</td>
<td>Display off, cursor off</td>
<td></td>
</tr>
<tr>
<td>A</td>
<td>Display off, cursor on</td>
<td></td>
</tr>
<tr>
<td>C</td>
<td>Display on, cursor off</td>
<td></td>
</tr>
<tr>
<td>E</td>
<td>Display on, cursor blinking</td>
<td></td>
</tr>
<tr>
<td>F</td>
<td>Display on, cursor blinking</td>
<td></td>
</tr>
<tr>
<td>10</td>
<td>Shift cursor position to left</td>
<td></td>
</tr>
<tr>
<td>14</td>
<td>Shift cursor position to right</td>
<td></td>
</tr>
<tr>
<td>18</td>
<td>Shift the entire display to the left</td>
<td></td>
</tr>
<tr>
<td>1C</td>
<td>Shift the entire display to the right</td>
<td></td>
</tr>
<tr>
<td>80</td>
<td>Force cursor to beginning of 1st line</td>
<td></td>
</tr>
<tr>
<td>C0</td>
<td>Force cursor to beginning of 2nd line</td>
<td></td>
</tr>
<tr>
<td>38</td>
<td>2 lines and 5x7 matrix</td>
<td></td>
</tr>
</tbody>
</table>

**Note:** This table is extracted from Table 12-4.

---

Figure 12-1. Pin Positions for Various LCDs from Optrex
Sending commands and data to LCDs with a time delay

Program 12-1 shows how to send characters (command/data) to the LCD without checking the busy flag. Notice that we need to wait 5–10 ms (DELAY) between issuing each character to the LCD. We call this delay simply DELAY. In programming an LCD, we also need a long delay for the power-up process. We call it LDELAY (long delay). SDELAY (short delay) is used to make the En signal wide enough for the LCD’s enable input. See Chapter 3 for delays.

Figure 12-2 shows the LCD connections to the microcontroller.

Figure 12-2. LCD Connections

; Program 12-1: Using delay before sending data/command
LCD_DATA EQU PORTD ; LCD data pins RD0-RD7
LCD_CTRL EQU PORTB ; LCD control pins
RS EQU RB0 ; RS pin of LCD
RW EQU RB1 ; R/W pin of LCD
EN EQU RB2 ; E pin of LCD
CLRF TRISD ; PORTD = Output
CLRF TRISB ; PORTB = Output
BCF LCD_CTRL, EN ; enable idle low
CALL LDELAY ; wait for initialization
MOVLW 0x38 ; init. LCD 2 lines, 5x7 matrix
CALL COMNWRT ; call command subroutine
CALL LDELAY ; initialization hold
MOVLW 0x0E ; display on, cursor on
CALL COMNWRT ; call command subroutine
CALL DELAY ; give LCD some time
MOVLW 0x01 ; clear LCD
CALL COMNWRT ; call command subroutine
CALL DELAY ; give LCD some time
MOVLW 0x06 ; shift cursor right
CALL COMNWRT ; call command subroutine
CALL DELAY ; give LCD some time
MOVLW 0x84 ; cursor at line 1, pos. 4
CALL COMNWRT ; call command subroutine
CALL DELAY ; give LCD some time
MOVLW A'N' ; display letter 'N'
CALL DATAWRT ; call display subroutine
CALL DELAY ; give LCD some time
MOVLW A'O' ; display letter 'O'

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CALL DATAWRT ; call display subroutine
AGAIN BTG LCD_CTRL, 0
BRA AGAIN ; stay here
COMNWRT ; send command to LCD
MOVWF LCD_DATA ; copy WREG to LCD DATA pin
BCF LCD_CTRL, RS ; RS = 0 for command
BCF LCD_CTRL, RW ; R/W = 0 for write
BSF LCD_CTRL, EN ; E = 1 for high pulse
CALL SDELAY ; make a wide En pulse
BCF LCD_CTRL, EN ; E = 0 for H-to-L pulse
RETURN

DATAWRT ; write data to LCD
MOVWF LCD_DATA ; copy WREG to LCD DATA pin
BSF LCD_CTRL, RS ; RS = 1 for data
BCF LCD_CTRL, RW ; R/W = 0 for write
BSF LCD_CTRL, EN ; E = 1 for high pulse
CALL SDELAY ; make a wide En pulse
BCF LCD_CTRL, EN ; E = 0 for H-to-L pulse
RETURN

; look in previous chapters for delay routines
END

Sending command or data to the LCD using busy flag

We use RS = 0 to read the busy flag bit to see if the LCD is ready to receive information. The busy flag is D7, and can be read when R/W = 1 and RS = 0, as follows: if R/W = 1, RS = 0. When D7 = 1 (busy flag = 1), the LCD is busy taking care of internal operations and will not accept any new information. When D7 = 0, the LCD is ready to receive new information.

This is shown in Program 12-2.

; Program 12-2: Check busy flag before sending
; data or command to LCD (See Fig. 12-2)
LCD_DATA EQU PORTD ; LCD data pins RD0-RD7
LCD_CTRL EQU PORTB ; LCD control pins
RS EQU RB0 ; RS pin of LCD
RW EQU RB1 ; R/W pin of LCD
EN EQU RB2 ; E pin of LCD
CLRF TRISD ; PORTD = Output
CLRF TRISB ; PORTB = Output
BCF LCD_CTRL, EN ; enable idle low
CALL LDELAY ; long delay (250 ms) for power-up
MOVLW 0x38 ; init. LCD 2 lines, 5x7 char
CALL COMMAND ; issue command
CALL LDELAY ; initialization hold
MOVLW 0x0E ; LCD on, cursor on
CALL COMMAND ; issue command
CALL READY ; Is LCD ready?
MOVLW 0x01 ; clear LCD command
CALL COMMAND ; issue command
CALL READY ; Is LCD ready?
MOVLW 0x06 ; shift cursor right

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CALL COMMAND ;issue command
CALL READY ;Is LCD ready?
MOVLW 0x86 ;cursor: line 1, pos. 6
CALL COMMAND ;command subroutine
CALL READY ;Is LCD ready?
MOVLW A'N' ;display letter 'N'
CALL DATA_DISPLAY
CALL READY ;Is LCD ready?
MOVLW A'O' ;display letter 'O'
CALL DATA_DISPLAY
HERE BRA HERE ;STAY HERE
--------------------
COMMAND MOVF LCD_DATA ;issue command code
BCF LCD_CTRL,RS ;RS = 0 for command
BCF LCD_CTRL,RW ;R/W = 0 for write
BSF LCD_CTRL,EN ;E = 1 for high pulse
CALL SDELAY ;make a wide En pulse
BCF LCD_CTRL,EN ;E = 0 for H-to-L pulse
RETURN
--------------------
DATA_DISPLAY MOVF LCD_DATA ;copy WREG to LCD DATA pin
BSF LCD_CTRL,RS ;RS = 1 for data
BCF LCD_CTRL,RW ;R/W = 0 for write
BSF LCD_CTRL,EN ;E = 1 for high pulse
CALL SDELAY ;make a wide En pulse
BCF LCD_CTRL,EN ;E = 0 for H-to-L pulse
RETURN
--------------------
READY SETF TRISD ;make PORTD input port for LCD data
BCF LCD_CTRL,RS ;RS = 0 access command reg
BSF LCD_CTRL,RW ;R/W = 1 read command reg
;read command reg and check busy flag
BACK BSF LCD_CTRL,EN ;E = 0 for L-to-H pulse
CALL SDELAY ;make a wide En pulse
BCF LCD_CTRL,EN ;E = 1 L-to-H pulse
BTFSC LCD_DATA,7 ;stay until busy flag = 0
BRA BACK
CLRF TRISD ;make PORTD output port for LCD data
RETURN
;look in previous chapters for delay routines
END

Notice in Program 12-2 that the busy flag is D7 of the command register. To read the command register, we make R/W = 1 and RS = 0, and a L-to-H pulse for the E pin will provide us the command register. After reading the command register, if bit D7 (the busy flag) is HIGH, the LCD is busy and no information (command or data) should be issued to it. Only when D7 = 0 can we send data or commands to the LCD. Notice that no time delays are used in this method because we are checking the busy flag before issuing commands or data to the LCD. Contrast the read and write timing for the LCD in Figures 12-3 and 12-4. Note that the E line is negative edge-triggered for the write while it is positive edge-triggered for the read.
Figure 12-3. LCD Timing for Read (L-to-H for E line)

Figure 12-4. LCD Timing for Write (H-to-L for E line)
**LCD data sheet**

In the LCD, one can put data at any location. The following shows address locations and how they are accessed.

<table>
<thead>
<tr>
<th>RS</th>
<th>R/W</th>
<th>DB7</th>
<th>DB6</th>
<th>DB5</th>
<th>DB4</th>
<th>DB3</th>
<th>DB2</th>
<th>DB1</th>
<th>DB0</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>I</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
</tr>
</tbody>
</table>

where AAAAAAA = 0000000 to 0100111 for line 1 and AAAAAAA = 1000000 to 1100111 for line 2. See Table 12-3.

The upper address range can go as high as 0100111 for the 40-character-wide LCD, while for the 20-character-wide LCD it goes up to 010011 (19 decimal = 10011 binary). Notice that the upper-range 0100111 (binary) = 39 decimal, which corresponds to locations 0 to 39 for the LCDs of 40x2 size.

From the above discussion we can get the addresses of cursor positions for various sizes of LCDs. See Figure 12-5 for the cursor addresses for common types of LCDs. Note that all the addresses are in hex. Table 12-4 provides a detailed list of LCD commands and instructions. (Table 12-2 is extracted from this table.)

### Table 12-3: LCD Addressing

<table>
<thead>
<tr>
<th>Line 1 (min)</th>
<th>DB7</th>
<th>DB6</th>
<th>DB5</th>
<th>DB4</th>
<th>DB3</th>
<th>DB2</th>
<th>DB1</th>
<th>DB0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Line 1 (max)</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td></td>
</tr>
<tr>
<td>Line 2 (min)</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>Line 2 (max)</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td></td>
</tr>
</tbody>
</table>

### Table

<table>
<thead>
<tr>
<th>Size</th>
<th>Address Range</th>
<th>CMDs and Instructions</th>
</tr>
</thead>
<tbody>
<tr>
<td>16 x 2 LCD</td>
<td>80 through 8F</td>
<td>C0 C1 C2 C3 C4 C5 C6</td>
</tr>
<tr>
<td>20 x 1 LCD</td>
<td>80 through 93</td>
<td>C0 C1 C2 C3 through D3</td>
</tr>
<tr>
<td>20 x 2 LCD</td>
<td>80 through 93</td>
<td>C0 C1 C2 C3 through D3</td>
</tr>
<tr>
<td>20 x 4 LCD</td>
<td>80 through 93</td>
<td>C0 C1 C2 C3 through D3</td>
</tr>
<tr>
<td>40 x 2 LCD</td>
<td>80 through A7</td>
<td>C0 C1 C2 C3 through E7</td>
</tr>
</tbody>
</table>

*Note: All data is in hex.*

**Figure 12-5. Cursor Addresses for Some LCDs**
<table>
<thead>
<tr>
<th>Instruction</th>
<th>R/W</th>
<th>DB7</th>
<th>DB6</th>
<th>DB5</th>
<th>DB4</th>
<th>DB3</th>
<th>DB2</th>
<th>DB1</th>
<th>DB0</th>
<th>Description</th>
<th>Execution Time (Max)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Clear Display</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>Clears entire display and sets DD RAM address 0 in address counter</td>
<td>1.64 ms</td>
</tr>
<tr>
<td>Return Home</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>Sets DD RAM address 0 as address counter. Also returns display being shifted to original position. DD RAM contents remain unchanged.</td>
<td>1.64 ms</td>
</tr>
<tr>
<td>Entry Mode Set</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1/D</td>
<td>S</td>
<td>Sets cursor move direction and specifies shift of display. These operations are performed during data write and read.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Display On/Off Control</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>D</td>
<td>C</td>
<td>B</td>
<td>Sets On/Off of entire display (D), cursor On/Off (C), and blink of cursor position character (B).</td>
<td>40 μs</td>
</tr>
<tr>
<td>Cursor or Display Shift</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>S/C</td>
<td>R/L</td>
<td>-</td>
<td>Moves cursor and shifts display without changing DD RAM contents.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Function Set</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>DL</td>
<td>N</td>
<td>F</td>
<td>-</td>
<td>Sets interface data length (DL), number of display lines (L), and character font (F).</td>
<td>40 μs</td>
</tr>
<tr>
<td>Set CG RAM Address</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>AGC</td>
<td></td>
<td></td>
<td></td>
<td>Sets CG RAM address. CG RAM data is sent and received after this setting.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Set DD RAM Address</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td></td>
<td></td>
<td>ADD</td>
<td></td>
<td></td>
<td></td>
<td>Sets DD RAM address. DD RAM data is sent and received after this setting.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Read Busy Flag &amp; Address</td>
<td>0</td>
<td>1</td>
<td>BF</td>
<td>AC</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Reads busy flag (BF) indicating internal operation is being performed and reads address counter contents.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Write Data CG or DD RAM</td>
<td>1</td>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td>Write Data</td>
<td></td>
<td></td>
<td></td>
<td>Writes data into DD or CG RAM.</td>
<td>40 μs</td>
</tr>
<tr>
<td>Read Data CG or DD RAM</td>
<td>1</td>
<td>1</td>
<td>Read Data</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>Reads data from DD or CG RAM.</td>
<td>40 μs</td>
</tr>
</tbody>
</table>

Notes:
1. Execution times are maximum times when fcp or fosc is 250 kHz.
2. Execution time changes when frequency changes.
   (e.g., when fcp or fosc is 270 kHz: 40 μs × 250 / 270 = 37 μs.)
3. Abbreviations:
   - DD RAM: Display data RAM
   - CG RAM: Character generator RAM
   - ACC: CG RAM address
   - ADD: DD RAM address, corresponds to cursor address
   - AC: Address counter used for both DD and CG RAM addresses.
   - 1/D = 1: Increment
   - 1/D = 0: Decrement
   - S = 1: Accompanies display shift
   - S/C = 1: Display shift;
   - S/C = 0: Cursor move
   - R/L = 1: Shift to the right;
   - R/L = 0: Shift to the left
   - DL = 1: 8 bits, DL = 0: 4 bits
   - N = 1: 1 line, N = 0: 1 line
   - F = 1: 5 × 10 dots, F = 0: 5 × 7 dots
   - BF = 1: Internal operation; BF = 0: Can accept instruction

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Optrex is one of the largest manufacturers of LCDs. You can obtain datasheets from their Web site, http://www.optrex.com.

The LCDs can be purchased from the following Web sites:
http://www.digikey.com
http://www.jameco.com
http://www.elexp.com
http://www.bgmicro.com

Sending information to LCD using the TBLRD instruction

Program 12-3 shows how to use the TBLRD instruction to send data and commands to an LCD.

For a PIC18 C version of LCD programming see Program 12-1C and Program 12-2C.

;Program 12-3: Using TableRead
;PORTD = D0-D7, RB0 = RS, RB1 = R/W, RB2 = E pins
LCD_DATA EQU PORTD ;LCD data pins RD0-RD7
LCD_CTRL EQU PORTB ;LCD control pins
RS EQU RB0 ;RS pin of LCD
RW EQU RB1 ;R/W pin of LCD
EN EQU RB2 ;E pin of LCD
CLRFB TRISD ;PORTD = Output
CLRFB TRISB ;PORTB = Output
BCF LCD_CTRL,EN ;enable idle low
CALL LDELAY ;long delay (250 ms) for power-up
MOVLW upper(MYCOM)
MOVF TBLPTRU
MOVLW high(MYCOM)
MOVF TBLPTRH
MOVLW low(MYCOM)
MOVF TBLPTRL
C1
TBLRD*+
MOVF TABLAT,W ;give it to WREG
IORLW 0x0 ;Is it the end of command?
BZ SEND_DAT ;if yes then go to display data
CALL COMNWRT ;call command subroutine
CALL DELAY ;give LCD some time
BRA C1
This C18 program sends letters 'M', 'D', and 'E' to the LCD using delays.

```c
//Program 12-1C: This is the C version of Program 12-1.
#include <P18F4580.h>
#define ld data PORTD  //PORTD = LCD data pins (Fig. 12-2)
#define rs PORTBbits.RB0 //rs = PORTB.0
#define rw PORTBbits.RB1 //rw = PORTB.1
#define en PORTBbits.RB2 //en = PORTB.2

void main()
{
    TRISD = 0;   //both ports B and D as output
    TRISB = 0;   //enable idle low
    MSDelay(250);
    lcdcmd(0x38);  //init. LCD 2 lines, 5x7 matrix
    MSDelay(250);
    lcdcmd(0x0E);  //display on, cursor on
    ...  //rest of the program
}
```

CHAPTER 12: LCD AND KEYBOARD INTERFACING
MSDelay(15);
lcdcmd(0x01);  //clear LCD
MSDelay(15);
lcdcmd(0x06);  //shift cursor right
MSDelay(15);
lcdcmd(0x86);  //line 1, position 6
MSDelay(15);
lcddata('M');  //display letter 'M'
MSDelay(15);
lcddata('D');  //display letter 'D'
MSDelay(15);
lcddata('E');  //display letter 'E'
}

void lcdcmd(unsigned char value)
{
    ldata = value;  //put the value on the pins
    rs = 0;
    rw = 0;
    en = 1;        //strobe the enable pin
    MSDLay(1);
    en = 0;
}

void lcddata(unsigned char value)
{
    ldata = value;  //put the value on the pins
    rs = 1;
    rw = 0;
    en = 1;        //strobe the enable pin
    MSDLay(1);
    en = 0;
}

void MSDLay(unsigned int itime)
{
    unsigned int i, j;
    for(i=0; i<itime; i++)
        for(j=0; j<135; j++);
}

The following is the C version of Program 12-2, using the busy flag method.

#include <P18F458.h>
define ldata PORTD  //PORTD = LCD data pins (Fig. 12-2)
define rs PORTBbits.RB0  //rs = PORTB.0
#define rw PORTBbits.RB1  //rw = PORTB.1
#define en PORTBbits.RB2  //en = PORTB.2
#define busy PORTDbits.RD7  //busy = PORTD.7

void main()
{
    TRISD = 0;  //both ports B and D as output
TRISB = 0; //enable idle low
en = 0; //long delay
MSDelay(250); //long delay
lcdcmd(0x38);
MSDelay(250); //long delay
lcdcmd(0x0E);
lcdready(); //check the LCD busy flag
lcdcmd(0x01);
lcdready(); //check the LCD busy flag
lcdcmd(0x06);
lcdready(); //check the LCD busy flag
lcdcmd(0x86);
lcdreadyO;
lcddata('M');
lcdready(); //check the LCD busy flag
tlcddata('D');
lcdready(); //check the LCD busy flag
tlcddata('E');
}

void lcdcmd(unsigned char value)
{
ldata = value; //put the value on the pins
rs = 0;
rw = 0;
en = 1; //strobe the enable pin
MSDelay(1);
en = 0;
}

void lcddata(unsigned char value)
{
ldata = value; //put the value on the pins
rs = 1;
rw = 0;
en = 1; //strobe the enable pin
MSDelay(1);
en = 0;
}

void lcdready()
{
TRISD = 0xFF; //make PORTD an input
rs = 0;
rw = 1;
do
{
en = 1;
MSDelay(1);
en = 0;
}while(busy==1);
TRISD = 0;
}

void MSDLay( unsigned int itime)
{
unsigned int i, j;
for(i=0;i<itime;i++)
   for(j=0;j<135;j++);
}
//Program 12-3C: C version of Program 12-3 Displaying Data in ROM

#include <P18F458.h>
#define ldata PORTD  //PORTD = LCD data pins (Fig. 12-2)
#define rs PORTBbits.RB0  //rs = PORTB.0
#define rw PORTBbits.RB1  //rw = PORTB.1
#define en PORTBbits.RB2  //en = PORTB.2

#pragma romdata mycom = 0x300  //command at ROM addr 0x300
far rom const char mycom[] = {0x0E, 0x01, 0x06, 0x84};

#pragma romdata mydata = 0x320  //data at ROM addr 0x320
far rom const char mydata[] = "HELLO";

void main()
{
    unsigned char z=0;
    TRISD = 0;  //both ports B and D as output
    TRISB = 0;
    en = 0;  //enable idle low
    MSDelay(250);
    lcdcmd(0x38);
    MSDelay(250);

    //send out the commands
    for(z<4;z++)
    {
        lcdcmd(mycom[z]);
        MSDelay(15);
    }

    //send out the data
    for(z=0;z<5;z++)
    {
        lcddata(mydata[z]);
        MSDelay(15);
    }

    while(1);  //infinite loop
}

void lcdcmd(unsigned char value)
{
    ldata = value;  //put the value on the pins
    rs = 0;
    rw = 0;
    en = 1;  //strobe the enable pin
    MSDelay(1);
    en = 0;
}

void lcddata(unsigned char value)
{
    ldata = value;  //put the value on the pins
    rs = 1;
    rw = 0;
}
en = 1; //strobe the enable pin
MSDelay(1);
en = 0;
}
void MSDLay(unsigned int itime)
{
unsigned int i, j;
for(i=0;i<itime;i++)
    for(j=0;j<135;j++);
}

Review Questions
1. The RS pin is an ________ (input, output) pin for the LCD.
2. The E pin is an ________ (input, output) pin for the LCD.
3. The E pin requires an ________ (H-to-L, L-to-H) pulse to latch in information at the data pins of the LCD.
4. For the LCD to recognize information at the data pins as data, RS must be set to ______ (HIGH, LOW).
5. Give the command codes for line 1, first character, and line 2, first character.

SECTION 12.2: KEYBOARD INTERFACING

Keyboards and LCDs are the most widely used input/output devices and a basic understanding of them is essential. In this section, we first discuss keyboard fundamentals, along with key press detection and key identification mechanisms. Then we show how a keyboard is interfaced to a PIC18.

Interfacing the keyboard to the PIC18

At the lowest level, keyboards are organized in a matrix of rows and columns. The CPU accesses both rows and columns through ports; therefore, with two 8-bit ports, an 8 x 8 matrix of keys can be connected to a microprocessor. When a key is pressed, a row and a column make a contact; otherwise, there is no connection between rows and columns. In IBM PC keyboards, a single microcontroller takes care of hardware and software interfacing of the keyboard. In such systems, programs stored in the ROM of the microcontroller scan the keys continuously, identify which one has been activated, and present it to the motherboard. See Example 12-3. In programming for keypad interfacing we must have two processes: (a) key press detection, and (b) key identification. There are two ways by which the PIC18 can perform key press detection: (1) the interrupt method, and (2) the scanning method. In the PIC18, the PORTB-Change interrupt can be used to implement the interrupt-based key press detection. Next we explain the interrupt method.

Interrupt method of key press detection

Figure 12-6 shows a 4 x 4 matrix keypad connected to PORTB. The rows are connected to PORTB.Low (RB3-RB0) and the columns are connected to PORTB.High (RB7-RB4), which is the PORTB-Change interrupt. As we dis-
**Example 12-3**

From Figure 12-6, identify the row and column of the pressed key for the following:
RB3–RB0 = 1110 for the row, RB7–RB4 = 1011 for the column

**Solution:**

From Figure 12-6, the row and column can be used to identify the key. The row belongs to RB0 and the column belongs to RB6; therefore, key number 2 was pressed.

---

![Matrix Keyboard Connection to Ports](image)

**Figure 12-6. Matrix Keyboard Connection to Ports**

As discussed in Chapter 11, any changes on the RB7–RB4 pins will cause an interrupt indicating a key press. Examine Program 12-4, which goes through the following stages:

1. To make sure that the preceding key has been released, 0s are output to all rows at once, and the columns are read and checked repeatedly until all the columns are HIGH. When all columns are found to be HIGH, the program waits for a short amount of time before it goes to the next stage of waiting for a key to be pressed.

2. To see if any key is pressed, the columns are connected to the PORTB-Change interrupt. Therefore, any key press will cause an interrupt and the microcontroller will execute the ISR. The ISR must do two things: (a) ensure that the first key press detection was not erroneous due to spike noise, and (b) wait 20 ms to prevent the same key press from being interpreted as multiple key presses. See Figure 12-8 for keyboard debounce.

3. To detect which row the key press belongs to, the microcontroller grounds one row at a time, reading the columns each time. If it finds that all columns are HIGH, this means that the key press cannot belong to that row; therefore, it grounds the next row and continues until it finds the row the key press belongs
Figure 12-7. Flowchart for Program 12-4
to. Upon finding the row that the key press belongs to, it sets up the starting address for the look-up table holding the scan codes (or the ASCII value) for that row and goes to the next stage to identify the key.

4. To identify the key press, the microcontroller rotates the column bits, one bit at a time, into the carry flag and checks to see if it is LOW. Upon finding the zero, it pulls out the ASCII code for that key from the look-up table; otherwise, it increments the pointer to point to the next element of the look-up table. Figure 12-7 flowcharts this process.

The look-up table method shown in Program 12-4 can be modified to work with any matrix up to $8 \times 4$. Figure 12-7 provides the flowchart for Program 12-4 for scanning and identifying the pressed key.

![Figure 12-8. Keyboard Debounce](image)

Examine Program 12-4. Notice in that program that the interrupt detects the key press. Then it is the job of the ISR to identify to which key the key press belongs (key identification). Program 12-4C is a C18 version of Program 12-4. In the Assembly version (12-4), the character is placed on PORTD, while in the C18 version (12-4C), it is sent to the serial port to be displayed on the monitor.

**Program 12-4:** This program waits for a key press on PORTB, then places the character on PORTD. We assume the following for this program:
- RB3–RB0 connected to rows
- RB7–RB4 connected to columns

```assembly
D15mH EQU D'100'; 15 ms delay high byte of value
D15mL EQU D'255'; low byte of value
COL EQU 0x08; holds the column found
DR15mH EQU 0x09; registers for 15 ms delay
DR15mL EQU 0x0A;

ORG 0x000000
RESET_ISR GOTO MAIN; jump over interrupt table
ORG 0x000008
HI_ISR BTFSC INTCON,RBIF; Was it a PORTB change?
BRA RBIF_ISR; yes then go to ISR
RETFIE; else return
```

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;--------- program for initialization
MAIN   CLRF TRISD ; make PORTD output port
       BCF INTCON2, RBPU; enable PORTB pull-up resistors
       MOVFW TRISB ; make PORTB low output ports
       MOVWF PORTB ; ground all rows
KEYOPEN CPFFSEQ PORTB ; are all keys open
       GOTO KEYOPEN ; wait until keypad ready
       MOVLW upper(KCODE0)
       MOVWF TBLPTRU ; load upper byte of TBLPTR
       MOVLW high(KCODE0)
       MOVWF TBLPTRH ; load high byte of TBLPTR
       BSF INTCON, RBIE ; enable PORTB change interrupt
       BSF INTCON, GIE ; enable all interrupts globally
LOOP   GOTO LOOP ; wait for key press
;--------- key identification ISR
RBIF_ISR CALL DELAY ; wait for debounce
       MOVFF PORTB, COL ; get the column of key press
       MOVLW 0xFE
       MOVWF PORTB ; ground row 0
       CPFFSEQ PORTB ; Did PORTB change?
       BRA ROW0 ; yes then row 0
       MOVLW 0xFC
       MOVWF PORTB ; ground row 1
       CPFFSEQ PORTB ; Did PORTB change?
       BRA ROW1 ; yes then row 1
       MOVLW 0xFB
       MOVWF PORTB ; ground row 2
       CPFFSEQ PORTB ; Did PORTB change?
       BRA ROW2 ; yes then row 2
       MOVLW 0xF7
       MOVWF PORTB ; ground row 3
       CPFFSEQ PORTB ; Did PORTB change?
       BRA ROW3 ; yes then row 3
       GOTO BAD_RBIF ; no then key press too short
ROW0   MOVLW low(KCODE0) ; set TBLPTR = start of row 0
       BRA FIND ; find the column
ROW1   MOVLW low(KCODE1) ; set TBLPTR = start of row 1
       BRA FIND ; find the column
ROW2   MOVLW low(KCODE2) ; set TBLPTR = start of row 2
       BRA FIND ; find the column
ROW3   MOVLW low(KCODE3) ; set TBLPTR = start of row 3
FIND    MOVWF TBLPTRL ; load low byte of TBLPTR
       MOVLW 0xF0
       XORWF COL ; invert high nibble
       SWAPF COL,F ; bring to low nibble
AGAIN   RRCF COL ; rotate to find column
       BC MATCH ; column found, get the ASCII code
       INCF TBLPTRL ; else point to next col. address

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Program 12-4C shows keypad programming in PIC18 C.

**Program 12-4C:** This C18 program reads the keypad and sends the result to the serial port. We assume the following for this program.

- RB0–RB3 connected to rows
- RB4–RB7 connected to columns
- Serial port is set for 9600 baud (10 MHz XTAL), 8-bit mode, and 1 stop bit.

```c
#include <pl8f458.h>
void SerTX(unsigned char x);
void RBIF_ISR(void);
void MSDelay(unsigned int millisecs);

unsigned char keypad[4][4] = {'0','1','2','3','4','5','6','7','8','9','A','B','C','D','E','F'}; 

#pragma code My_HiPrio_Int =0x0008 //high-priority interrupt
void My_HiPrio_Int (void) {
  _asm
    _GOTO chk_isr
  _endasm
```

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#pragma code

#pragma interrupt chk_isr //which ISR

```c
void chk_isr (void) {
    if (INTCONbits.RBIF==1) //RBIF caused interrupt?
        RBIF_ISR( ); //yes go to RBIF_ISR
}
```

```c
#pragma code

void main() {
    TRISD=0; //make PORTD output port
    INTCON2bits.RBPU=0; //enable PORTB pull-up resistors
    TRISB=0xF0; //PORTB low as output and high as input
    PORTB=0xF0; //clear PORTB low
    while(PORTB!=0xF0); //wait until key not pressed
    TXSTA=0x20; //choose low baud rate, 8-bit
    SPBRG=15; //9600 baud rate, XTAL = 10 MHz
    TXSTABits.TXEN=1; //enable transmit
    RCSTABits.SPEN=1; //enable serial port
    INTCONbits.RBIE=1; //enable PORTB interrupt on change
    INTCONbits.GIE=1; //enable interrupts globally
    while(1); //wait until key press
}
```
void MSDelay(unsigned int millisecs)
{
    unsigned int i, j;
    for(i=0;i<millisecs;i++)
        for(j=0;j<135;j++);
}

Scanning method for key press detection

Another method for key press detection is by scanning. In this method, to
detect a pressed key, the microcontroller grounds all rows by providing 0 to the
output latch, then it reads the columns. If the data read from the columns are equal
to 1111, no key has been pressed and the process continues until a key press is
detected. If one of the column bits has a zero, however, this means that a key press
has occurred. After a key press is detected, the microcontroller will go through the
process of identifying the key. Starting with the top row, the microcontroller
grounds it by providing a LOW to the first row only; then it reads the columns. If
the data read is all 1s, no key in that row is activated and the process is moved to
the next row. It grounds the next row, reads the columns, and checks for any zero.
This process continues until the row is identified. After identification of the row in
which the key has been pressed, the next task is to find out which column the
pressed key belongs to. This should be easy since the microcontroller knows at any
time which row and column are being accessed. Figure 12-9 shows the flowchart
for this method. The program implementation is left to the reader.

Some IC chips, such as National Semiconductor's MM74C923, incorporate
keyboard scanning and decoding all in one chip. Such chips use combinations of
counters and logic gates (no microcontroller) to implement the underlying con-
cepts presented in this section.

Review Questions

1. True or false. To see if any key is pressed, all rows are grounded.
2. If RB7–RB4 = 0111 is the data read from the columns, which column does the
   pressed key belong to?
3. True or false. Key press detection and key identification require two different
   processes.
4. In Figure 12-6, if the rows are RB3–RB0 = 1110 and the columns are
   RB7–RB4 = 1110, which key is pressed?
5. True or false. To identify the pressed key, one row at a time is grounded.
Figure 12-9. Flowchart of Scanning Method for Key Press Detection
SUMMARY

This chapter showed how to interface real-world devices such as LCDs and keypads to the PIC18. First, we described the operation modes of LCDs, then described how to program the LCD by sending data or commands to it via its interface to the PIC18.

Keyboards are one of the most widely used input devices for PIC18 projects. This chapter also described the operation of keyboards, including key press detection and key identification mechanisms. Then the PIC18 was shown interfacing with a keyboard. PIC18 programs were written to return the ASCII code for the pressed key.

PROBLEMS

SECTION 12.1: LCD INTERFACING

1. The LCD discussed in this section has _______ (4, 8) data pins.
2. Describe the function of pins E, R/W, and RS in the LCD.
3. What is the difference between the $V_{CC}$ and $V_{EE}$ pins on the LCD?
4. "Clear LCD" is a ________ (command code, data item) and its value is ___ hex.
5. What is the hex value of the command code for "display on, cursor on"?
6. Give the state of RS, E, and R/W when sending a command code to the LCD.
7. Give the state of RS, E, and R/W when sending data character ‘Z’ to the LCD.
8. Which of the following is needed on the E pin in order for a command code (or data) to be latched in by the LCD?
   (a) H-to-L pulse (b) L-to-H pulse
9. True or false. For the above to work, the value of the command code (data) must already be at the D0–D7 pins.
10. There are two methods of sending streams of characters to the LCD: (1) checking the busy flag, or (2) putting some time delay between sending each character without checking the busy flag. Explain the difference and the advantages and disadvantages of each method. Also explain how we monitor the busy flag.
11. For a 16 x 2 LCD, the location of the last character of line 1 is 8FH (its command code). Show how this value was calculated.
12. For a 16 x 2 LCD, the location of the first character of line 2 is C0H (its command code). Show how this value was calculated.
13. For a 20 x 2 LCD, the location of the last character of line 2 is 93H (its command code). Show how this value was calculated.
14. For a 20 x 2 LCD, the location of the third character of line 2 is C2H (its command code). Show how this value was calculated.
15. For a 40 x 2 LCD, the location of the last character of line 1 is A7H (its command code). Show how this value was calculated.
16. For a 40 x 2 LCD, the location of the last character of line 2 is E7H (its command code). Show how this value was calculated.
17. Show the value (in hex) for the command code for the 10th location, line 1 on a $20 \times 2$ LCD. Show calculations.
18. Show the value (in hex) for the command code for the 20th location, line 2 on a $40 \times 2$ LCD. Show calculations.
19. Rewrite the COMNWRT subroutine. Assume connections RC4 = RS, RC5 = R/W, RC6 = E.
20. Repeat Problem 19 for the data write subroutine. Send the string “Hello” to the LCD by checking the busy flag. Use the instruction TBLRD.

SECTION 12.2: KEYBOARD INTERFACING

21. In reading the columns of a keyboard matrix, if no key is pressed we should get all ____ (1s, 0s).
22. In Program 12-4, to detect the key press, which of the following is performed?
   (a) PORTB-Change interrupt  (b) grounding one row at time
23. In Figure 12-6, to identify the key pressed, which of the following is grounded?
   (a) all rows  (b) one row at time  (c) both (a) and (b)
24. For Figure 12-6, indicate the key press for RB7-RB4 = 0111, RB3-RB0 = 1110.
25. Indicate an advantage and a disadvantage of using an IC chip instead of a microcontroller for keyboard scanning and decoding.
26. What is the best compromise for the answer to Problem 25?

ANSWERS TO REVIEW QUESTIONS

SECTION 12.1: LCD INTERFACING

1. Input
2. Input
3. H-to-L
4. HIGH
5. 80H and C0H

SECTION 12.2: KEYBOARD INTERFACING

1. True
2. Column 3
3. True
4. 0
5. True
CHAPTER 13

ADC, DAC, AND SENSOR INTERFACING

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Discuss the ADC (analog-to-digital converter) section of the PIC18 chip
- Interface temperature sensors to the PIC18
- Explain the process of data acquisition using ADC
- Describe factors to consider in selecting an ADC chip
- Program the PIC18’s ADC in C and Assembly
- Describe the basic operation of a DAC (digital-to-analog converter) chip
- Interface a DAC chip to the PIC18
- Program a DAC chip to produce a sine wave on an oscilloscope
- Program DAC chips in PIC18 C and Assembly
- Explain the function of precision IC temperature sensors
- Describe signal conditioning and its role in data acquisition
This chapter explores more real-world devices such as ADCs (analog-to-digital converters), DACs (digital-to-analog converters), and sensors. We will also explain how to interface the PIC18 to these devices. In Section 13.1, we describe analog-to-digital converter (ADC) chips. We will program the ADC portion of the PIC18 chip in Section 13.2. The characteristics of DAC chips are discussed in Section 13.3. In Section 13.4, we show the interfacing of sensors and discuss the issue of signal conditioning.

SECTION 13.1: ADC CHARACTERISTICS

This section will explore ADC programming in PIC18 chips. First, we describe some general aspects of the ADC itself, then show how to program the ADC portion of the PIC18 in both Assembly and C.

ADC devices

Analog-to-digital converters are among the most widely used devices for data acquisition. Digital computers use binary (discrete) values, but in the physical world everything is analog (continuous). Temperature, pressure (wind or liquid), humidity, and velocity are a few examples of physical quantities that we deal with every day. A physical quantity is converted to electrical (voltage, current) signals using a device called a transducer. Transducers are also referred to as sensors. Sensors for temperature, velocity, pressure, light, and many other natural quantities produce an output that is voltage (or current). Therefore, we need an analog-to-digital converter to translate the analog signals to digital numbers so that the microcontroller can read and process them. See Figures 13-1 and 13-2.

---

Figure 13-1. Microcontroller Connection to Sensor via ADC

Figure 13-2. An 8-bit ADC Block Diagram
Table 13-1: Resolution versus Step Size for ADC (Vref = 5 V)

<table>
<thead>
<tr>
<th>n-bit</th>
<th>Number of steps</th>
<th>Step size (mV)</th>
</tr>
</thead>
<tbody>
<tr>
<td>8</td>
<td>256</td>
<td>5/256 = 19.53</td>
</tr>
<tr>
<td>10</td>
<td>1,024</td>
<td>5/1,024 = 4.88</td>
</tr>
<tr>
<td>12</td>
<td>4,096</td>
<td>5/4,096 = 1.2</td>
</tr>
<tr>
<td>16</td>
<td>65,536</td>
<td>5/65,536 = 0.076</td>
</tr>
</tbody>
</table>

Notes: Vcc = 5 V
Step size (resolution) is the smallest change that can be discerned by an ADC.

Some of the major characteristics of the ADC are as follows:

Resolution
ADC has n-bit resolution, where n can be 8, 10, 12, 16, or even 24 bits. The higher-resolution ADC provides a smaller step size, where step size is the smallest change that can be discerned by an ADC. Some widely used resolutions for ADCs are shown in Table 13-1. Although the resolution of an ADC chip is decided at the time of its design and cannot be changed, we can control the step size with the help of what is called Vref. This is discussed below.

Conversion time
In addition to resolution, conversion time is another major factor in judging an ADC. Conversion time is defined as the time it takes the ADC to convert the analog input to a digital (binary) number. The conversion time is dictated by the clock source connected to the ADC in addition to the method used for data conversion and technology used in the fabrication of the ADC chip such as MOS or TTL technology.

Vref
Vref is an input voltage used for the reference voltage. The voltage connected to this pin, along with the resolution of the ADC chip, dictate the step size. For an 8-bit ADC, the step size is Vref/256 because it is an 8-bit ADC, and 2 to the power of 8 gives us 256 steps. See Table 13-1. For example, if the analog input range needs to be 0 to 4 volts, Vref is connected to 4 volts. That gives 4 V/256 = 15.62 mV for the step size of an 8-bit ADC. In another case, if we need a step size of 10 mV for an 8-bit ADC, then Vref = 2.56 V, because 2.56 V/256 = 10 mV. For

Table 13-2: Vref Relation to Vin Range for an 8-bit ADC

<table>
<thead>
<tr>
<th>Vref (V)</th>
<th>Vin (V)</th>
<th>Step Size (mV)</th>
</tr>
</thead>
<tbody>
<tr>
<td>5.00</td>
<td>0 to 5</td>
<td>5/256 = 19.53</td>
</tr>
<tr>
<td>4.0</td>
<td>0 to 4</td>
<td>4/256 = 15.62</td>
</tr>
<tr>
<td>3.0</td>
<td>0 to 3</td>
<td>3/256 = 11.71</td>
</tr>
<tr>
<td>2.56</td>
<td>0 to 2.56</td>
<td>2.56/256 = 10</td>
</tr>
<tr>
<td>2.0</td>
<td>0 to 2</td>
<td>2/256 = 7.81</td>
</tr>
<tr>
<td>1.28</td>
<td>0 to 1.28</td>
<td>1.28/256 = 5</td>
</tr>
<tr>
<td>1</td>
<td>0 to 1</td>
<td>1/256 = 3.90</td>
</tr>
</tbody>
</table>
Table 13-3: $V_{\text{ref}}$ Relation to $V_{\text{in}}$ Range for an 10-bit ADC

<table>
<thead>
<tr>
<th>$V_{\text{ref}}$ (V)</th>
<th>$V_{\text{in}}$ (V)</th>
<th>Step Size (mV)</th>
</tr>
</thead>
<tbody>
<tr>
<td>5.00</td>
<td>0 to 5</td>
<td>5/1,024 = 4.88</td>
</tr>
<tr>
<td>4.096</td>
<td>0 to 4.096</td>
<td>4.096/1,024 = 4</td>
</tr>
<tr>
<td>3.0</td>
<td>0 to 3</td>
<td>3/1,024 = 2.93</td>
</tr>
<tr>
<td>2.56</td>
<td>0 to 2.56</td>
<td>2.56/1,024 = 2.5</td>
</tr>
<tr>
<td>2.048</td>
<td>0 to 2.048</td>
<td>2.048/1,024 = 2</td>
</tr>
<tr>
<td>1.28</td>
<td>0 to 1.28</td>
<td>1/1,024 = 1.25</td>
</tr>
<tr>
<td>1.024</td>
<td>0 to 1.024</td>
<td>1.024/1,024 = 1</td>
</tr>
</tbody>
</table>

the 10-bit ADC, if the $V_{\text{ref}} = 5$V, then the step size is 4.88 mV as shown in Table 13-1. Tables 13-2 and 13-3 show the relationship between the $V_{\text{ref}}$ and step size for the 8- and 10-bit ADCs, respectively. In some applications, we need the differential reference voltage where $V_{\text{ref}} = V_{\text{ref}}(+)-V_{\text{ref}}(-)$. Often the $V_{\text{ref}}(-)$ pin is connected to ground and the $V_{\text{ref}}(+)$ pin is used as the $V_{\text{ref}}$.

**Digital data output**

In an 8-bit ADC we have an 8-bit digital data output of D0–D7 while in the 10-bit ADC the data output is D0–D9. To calculate the output voltage, we use the following formula:

$$D_{\text{out}} = \frac{V_{\text{in}}}{\text{step size}}$$

where $D_{\text{out}} = \text{digital data output (in decimal)}$, $V_{\text{in}} = \text{analog input voltage}$, and step size (resolution) is the smallest change, which is $V_{\text{ref}}/256$ for an 8-bit ADC. See Example 13-1. This data is brought out of the ADC chip either one bit at a time (serially), or in one chunk, using a parallel line of outputs. This is discussed next.

**Example 13-1**

For an 8-bit ADC, we have $V_{\text{ref}} = 2.56$ V. Calculate the D0–D7 output if the analog input is: (a) 1.7 V, and (b) 2.1 V.

**Solution:**

Because the step size is 2.56/256 = 10 mV, we have the following:

(a) $D_{\text{out}} = 1.7 \text{ V}/10 \text{ mV} = 170$ in decimal, which gives us 10101011 in binary for D7–D0.

(b) $D_{\text{out}} = 2.1 \text{ V}/10 \text{ mV} = 210$ in decimal, which gives us 11010010 in binary for D7–D0.

**Parallel versus serial ADC**

The ADC chips are either parallel or serial. In parallel ADC, we have 8 or more pins dedicated to bringing out the binary data, but in serial ADC we have only one pin for data out. That means that inside the serial ADC, there is a paral-
Figure 13-4. MAX1112 Serial ADC Block Diagram

Figure 13-3. ADC0848 Parallel ADC Block Diagram

The D0–D7 data pins of the 8-bit ADC provide an 8-bit parallel data path between the ADC chip and the CPU. In the case of the 16-bit parallel ADC chip, we need 16 pins for the data path. In order to save pins, many 12- and 16-bit ADCs use pins D0–D7 to send out the upper and lower bytes of the binary data. In recent years, for many applications where space is a critical issue, using such a large number of pins for data is not feasible. For this reason, serial devices such as the serial ADC are becoming widely used. While the serial ADCs use fewer pins and their smaller packages take much less space on the printed circuit board, more CPU time is needed to get the converted data from the ADC because the CPU must get data one bit at a time, instead of in one single read operation as with the parallel ADC. ADC848 is an example of a parallel ADC with 8 pins for the data output, while the MAX1112 is an example of a serial ADC with a single pin for Dout. Figures 13-3 and 13-4 show the block diagram for ADC848 and MAX1112, respectively.
Analog input channels

Many data acquisition applications need more than one ADC. For this reason, we see ADC chips with 2, 4, 8, or even 16 channels on a single chip. Multiplexing of analog inputs is widely used as shown in the ADC848 and MAX1112. In these chips, we have 8 channels of analog inputs, allowing us to monitor multiple quantities such as temperature, pressure, heat, and so on. PIC18 microcontroller chips come with 5 to 15 ADC channels, depending on the family member. The PIC18 ADC feature is discussed in the next section.

Start conversion and end-of-conversion signals

The fact that we have multiple analog input channels and a single digital output register makes it necessary for start conversion (SC) and end-of-conversion (EOC) signals. When SC is activated, the ADC starts converting the analog input value of Vin to an n-bit digital number. The amount of time it takes to convert varies depending on the conversion method as was explained earlier. When the data conversion is complete, the end-of-conversion signal notifies the CPU that the converted data is ready to be picked up.

From the discussion we conclude that the following steps must be followed for data conversion by an ADC chip:

1. Select a channel.
2. Activate the start conversion (SC) signal to start the conversion of analog input.
3. Keep monitoring the end-of-conversion (EOC) signal.
4. After the EOC has been activated, we read data out of the ADC chip.

Review Questions

1. Give two factors that affect the step size calculation.
2. The ADC0848 is a(n) ______-bit converter.
3. True or false. While the ADC0848 has 8 pins for D_OUT, the MAX1112 has only one D_OUT pin.
4. Indicate the number of analog input channels for each of the following ADC chips.
   (a) ADC0848  (b) MAX1112
5. Find the step size for an 8-bit ADC, if V_ref = 1.28 V
6. For question 5, calculate the D0–D7 output if the analog input is: (a) 0.7 V, and (b) 1 V.
SECTION 13.2: ADC PROGRAMMING IN THE PIC18

Because the ADC is widely used in data acquisition, in recent years an increasing number of microcontrollers have an on-chip ADC peripheral, just like timers and USART. An on-chip ADC eliminates the need for an external ADC connection, which leaves more pins for other I/O activities. The vast majority of the PIC18 chips come with 8 channels of ADC, and some PIC18s have as many as 16 channels of ADCs. In this section we discuss the ADC feature of the PIC18452/458 and show how it is programmed in both Assembly and C.

PIC18F452/458 ADC features programming

The ADC peripheral of the PIC18 has the following characteristics:
(a) It is a 10-bit ADC.
(b) It can have 5 to 15 channels of analog input channels, depending on the family member. In PIC18452/458, pins RA0–RA7 of PORTA are used for the 8 analog channels. See Figures 13-5A and 13-5B.
(c) The converted output binary data is held by two special function registers called ADRESL (A/D Result Low) and ADRESH (A/D Result High).

![Figure 13-5A: PIC18F458 Pins with ADC Channels Shown in Bold](image-url)
(d) Because the ADRESH:ADRESL registers give us 16 bits and the ADC data out is only 10 bits wide, 6 bits of the 16 are unused. We have the option of making either the upper 6 bits or the lower 6 bits unused.

(e) We have the option of using Vdd (Vcc), the voltage source of the PIC18 chip itself, as the Vref or connecting it to an external voltage source for the Vref.

(f) The conversion time is dictated by the Fosc of crystal frequency connected to the OSCs pins. While the Fosc for PIC18 can be as high as 40 MHz, the conversion time can not be shorter than 1.6 ms.

(g) It allows the implementation of the differential Vref voltage using the Vref(+) and Vref(−) pins, where Vref = Vref (+) − Vref(−).

Many of the above features can be programmed by way of ADCON0 (A/D control register 0) and ADCON1 (A/D control register 1), as we will see next.
ADCON0 register

The ADCON0 register is used to set the conversion time and select the analog input channel among other things. Figure 13-6 shows the ADCON0 register. In order to reduce the power consumption of the PIC18, the ADC feature is turned off when the microcontroller is powered up. We turn on the ADC with the ADON bit of the ADCON0 register, as shown in Figure 13-6. The other important bit is the GO/DONE bit. We use this bit to start conversion and monitor it to see if conversion has ended. Notice in ADCCON0 that not all family members have all the 8 analog input channels. The conversion time is set with the ADCS bits. While ADCS1 and ADCS0 are held by the ADCON0 register, ADCS2 is part of the ADCON1 register. This is discussed next.

<table>
<thead>
<tr>
<th>ADCS2 (from ADCON1)</th>
<th>ADCS1</th>
<th>ADCS0</th>
<th>Conversion Clock Source</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>Fosc/2</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>Fosc/8</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0</td>
<td>Fosc/32</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
<td>Internal RC used for clock source</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>0</td>
<td>Fosc/4</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>1</td>
<td>Fosc/16</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>Fosc/64</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
<td>Internal RC used for clock source</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CHS2</th>
<th>CHS1</th>
<th>CHS0</th>
<th>CHANNEL SELECTION</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>CHAN0 (AN0)</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>CHAN1 (AN1)</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0</td>
<td>CHAN2 (AN2)</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
<td>CHAN3 (AN3)</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>0</td>
<td>CHAN4 (AN4)</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>1</td>
<td>CHAN5 (AN5) not implemented on 28-pin PIC18</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>CHAN6 (AN6) not implemented on 28-pin PIC18</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
<td>CHAN7 (AN7) not implemented on 28-pin PIC18</td>
</tr>
</tbody>
</table>

GO/DONE A/D conversion status bit.
1 = A/D conversion is in progress. This is used as start conversion, which means after the conversion is complete, it will go LOW to indicate the end of conversion.
0 = A/D conversion is complete and digital data is available in registers ADRESH and ADRESL.

ADON A/D on bit
0 = A/D part of the PIC18 is off and consumes no power. This is the default and we should leave it off for applications in which ADC is not used.
1 = A/D feature is powered up.

Figure 13-6. ADCON0 (A/D Control Register 0)
Another major register of the PIC18’s ADC feature is ADCON1. The ADCON1 register is used to select the Vref voltage among other things. It is shown in Figure 13-7. After the A/D conversion is complete, the result sits in registers ADRESL (A/D Result Low Byte) and ADRESH (A/D Result High Byte). The ADFM bit of the ADCON1 is used for making it right-justified or left-justified because we need only 10 bits of the 16. See Figure 13-8.

<table>
<thead>
<tr>
<th>ADFM</th>
<th>ADCS2</th>
<th>PCFG3</th>
<th>PCFG2</th>
<th>PCFG1</th>
<th>PCFG0</th>
</tr>
</thead>
</table>

**ADFM** A/D Result format select bit
- 1 = Right justified: The 10-bit result is in the ADRESL register and the lower 2 bits of ADRESH. That means the 6 most significant bits of the ADRESH register are all 0s.
- 0 = Left justified: The 10-bit result is in the ADRESL register and the upper 2 bits of ADRESL. That means the 6 least significant bits of the ADRESL register are all 0s.

**ADCS2** A/D Clock Select bit 2. This bit along with the ADCS1 and ADCS0 bits of the ADCON0 register decide the conversion clock for the ADC. The default value for ADCS2 is 0, which means setting the ADCS0 and ADCS1 values of ADCON0 can give us clock conversion of Fosc/2, Fosc/8, and Fosc/32. See the ADCON0 register.

**PCFGs: A/D Port Configuration Control bits:**

<table>
<thead>
<tr>
<th>PCFGs</th>
<th>AN7</th>
<th>AN6</th>
<th>AN5</th>
<th>AN4</th>
<th>AN3</th>
<th>AN2</th>
<th>AN1</th>
<th>AN0</th>
<th>Vref+</th>
<th>Vref-</th>
<th>C/R</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>8/0</td>
</tr>
<tr>
<td>0001</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vref+</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>7/1</td>
</tr>
<tr>
<td>0010</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>5/0</td>
</tr>
<tr>
<td>0011</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>Vref+</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>4/1</td>
</tr>
<tr>
<td>0100</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>D</td>
<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>3/0</td>
</tr>
<tr>
<td>0101</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>Vref+</td>
<td>D</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>Vss</td>
<td>2/1</td>
</tr>
<tr>
<td>0110</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>-</td>
<td>-</td>
<td>0/0</td>
</tr>
<tr>
<td>1000</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>Vref+</td>
<td>Vref-</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>AN2</td>
<td>6/2</td>
</tr>
<tr>
<td>1001</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>A</td>
<td>A</td>
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<td>A</td>
<td>A</td>
<td>Vdd</td>
<td>Vss</td>
<td>6/0</td>
</tr>
<tr>
<td>1010</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>A</td>
<td>Vref+</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>Vss</td>
<td>5/1</td>
</tr>
<tr>
<td>1011</td>
<td>D</td>
<td>D</td>
<td>A</td>
<td>Vref+</td>
<td>Vref-</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>AN2</td>
<td>4/2</td>
</tr>
<tr>
<td>1100</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>Vref+</td>
<td>Vref-</td>
<td>A</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>AN2</td>
<td>3/2</td>
</tr>
<tr>
<td>1110</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>Vref+</td>
<td>Vref-</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>AN2</td>
<td>2/2</td>
</tr>
<tr>
<td>1111</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>D</td>
<td>Vref+</td>
<td>Vref-</td>
<td>A</td>
<td>A</td>
<td>AN3</td>
<td>AN2</td>
<td>1/2</td>
</tr>
</tbody>
</table>

A = Analog input, D = Digital I/O
C/R = # of analog input channels / # of pins used for A/D voltage reference
The default is option 0000, which gives us 8 channels of analog input and uses the Vdd of PIC18 as Vref.

**Figure 13-7. ADCON1 (A/D Control Register 1)**
The port configuration for A/D channels is handled by the PCFG (A/D port configuration) bits. While in chips such as the PIC18452/458, we can have up to 8 channels of analog input, not all applications need that many ADC inputs. The PORTA pins of RA0–RA3 and RA5 and RE0–RE2 of PORTE are used for the analog input channels. With PCFG = 0110, we can use all the pins of the PORTA as the digital I/O. The default is PCFG = 0000, which allows us to use all 8 pins for analog inputs. In that case Vref = Vdd, the same voltage source used by the PIC18 chip itself. In many applications we need Vref other than Vdd. The AN3 pin can be used as an external source of voltage for Vref. For example, option PCFG = 0101 allows us to use two channels for analog inputs, AN3 = Vref, and the other 5 pins for the digital I/O. In this case the Vss (Gnd) pin of the PIC18 is used for the Vref (–). See Examples 13-2 and 13-3.

**Example 13-2**

For a PIC18-based system, we have $V_{ref} = Vdd = 5$ V. Find (a) the step size, and (b) the ADCON1 value if we need 3 channels. Assume that the ADRESH:ADRESL registers are right justified.

**Solution:**

(a) The step size is $5/1,024 = 4.8$ mV.

(b) ADCON1 = 1x000100 because option 100 gives us 3 analog input channels. The x = ADCS2 is decided by the conversion speed.

**Example 13-3**

For a PIC18-based system, we have $V_{ref} = 2.56$ V. Find (a) the step size, and (b) the ADCON1 value if we need 3 channels. Assume that the ADRESH:ADRESL registers are right justified.

**Solution:**

(a) The step size is $2.56/1,024 = 2.5$ mV.

(b) ADCON1 = 1x000011 because option 0011 gives us 3 analog input channels where x = ADCS2 is decided by the conversion speed.
Calculating A/D conversion time

By using the ADCS (A/D clock source) bits of both the ADCON0 and ADCON1 registers we can set the A/D conversion time. The conversion time is defined in terms of Tad, where Tad is the conversion time per bit. To calculate the Tad, we can select a conversion clock source of Fosc/2, Fosc/4, Fosc/8, Fosc/16, Fosc/32, or Fosc/64, where Fosc is the speed of the crystal frequency connected to the PIC18 chip. For the PIC18, the conversion time is 12 times the Tad. Notice that the Tad cannot be faster than 1.6 ms. Look at Examples 13-4 and 13-5 for clarification.

We can also use the the internal RC oscillator for the conversion clock source, instead of the Fosc of the external crystal oscillator. In that case the Tad is typically 4–6 μs and conversion time is 12 × 6 μs = 72 μs.

Another timing factor that we must pay attention to is the acquisition time (Tacq). After an A/D channel is selected, we must allow some time for the sample-and-hold capacitor (C hold) to charge fully to the input voltage level present at the channel. It is only after the elapsing of this acquisition time that the A/D conversion can be started. Although many factors (e.g., Vdd and temperature) affect the duration of Tacq, we can use a typical value of 15 μs. In some newer generations of the PIC18, we have the option of controlling the exact time of Tacq by programming the internal register ADCON2. In the PIC18F452/458, we have only the ADCON0 and ADCON1 registers. See Example 13-6.

Example 13-4

A PIC18 is connected to the 10 MHz crystal oscillator. Calculate the conversion time for all options of ADCS bits in both the ADCON0 and ADCON1 registers.

Solution:

The options for the conversion clock source for both ADCON0 and ADCON1 are as follows:

(a) For Fosc/2, we have 10 MHz / 2 = 5 MHz.
   Tad = 1 / 5 MHz = 200 ns. Invalid because it is faster than 1.6 μs.
(b) For Fosc/4, we have 10 MHz / 4 = 2.5 MHz.
   Tad = 1 / 2.5 MHz = 400 ns. Invalid because it is faster than 1.6 μs.
(c) For Fosc/8, we have 10 MHz / 8 = 1.25 MHz.
   Tad = 1 / 2.5 MHz = 800 ns. Invalid because it is faster than 1.6 μs.
(d) For Fosc/16, we have 10 MHz / 16 = 625 kHz.
   Tad = 1 / 625 kHz = 1.6 μs. The conversion time = 12 × 1.6 μs = 19.2 μs
(e) For Fosc/32, we have 10 MHz / 32 = 312.5 kHz.
   Tad = 1 / 312.5 kHz = 3.2 μs. The conversion time = 12 × 3.2 μs = 38.4 μs
(f) For Fosc/64, we have 10 MHz / 64 = 156.25 kHz.
   Tad = 1 / 156.25 kHz = 6.4 μs. The conversion time = 12 × 6.4 μs = 76.8 μs

Notice that for the Fosc/4, Fosc/16, and Fosc/64 selections, we must use the ADSC2 bit in the ADCON1 register, in addition to the ADCS bits in the ADCON0 register.
Example 13-5

A PIC18 is connected to the 4 MHz crystal oscillator. Calculate the conversion time if we want to use only the ADCS bits of the ADCON0 register.

Solution:

The options for the conversion clock source available in the ADCON0 register are as follows:

(a) For Fosc/2, we have 4 MHz / 2 = 2 MHz.
\[
T_{ad} = \frac{1}{2} \text{MHz} = 400 \text{ ns}. \quad \text{Invalid because it is faster than } 1.6 \mu s.
\]

(b) For Fosc/8, we have 4 MHz / 8 = 500 kHz.
\[
T_{ad} = \frac{1}{8} \text{MHz} = 2 \mu s. \quad \text{The conversion time } = 12 \times 2 \mu s = 24 \mu s.
\]

(c) For Fosc/32, we have 4 MHz / 32 = 125 kHz.
\[
T_{ad} = \frac{1}{32} \text{MHz} = 8 \mu s. \quad \text{The conversion time } = 12 \times 8 \mu s = 96 \mu s.
\]

Example 13-6

Find the values for the ADCON0 and ADCON1 registers for the following options: (a) channel AN0 as analog input, (b) Vref+ = Vdd, Vref− = Vss, (c) Fosc/64, (d) A/D result is right justified, and (e) A/D module is on.

Solution:

From Figure 13-6, we have ADCON0 = 10000x1. With x = 0 we have 10000001.

From Figure 13-7, we have ADCON1 = 11xx1110. With x = 0 we have 11001110.

Steps in programming the A/D converter using polling

To program the A/D converter of the PIC18, the following steps must be taken:

1. Turn on the ADC module of the PIC18 because it is disabled upon power-on reset to save power. We can use the "BSF ADCON0, ADON" instruction.
2. Make the pin for the selected ADC channel an input pin. We use "BSF TRISA, x" or "BSF TRISE, x" where x is the channel number.
3. Select voltage reference and A/C input channels. We use registers ADCON0 and ADCON1.
4. Select the conversion speed. We use registers ADCON0 and ADCON1.
5. Wait for the required acquisition time.
6. Activate the start conversion bit of GO/DONE.
7. Wait for the conversion to be completed by polling the end-of-conversion (GO/DONE) bit.
8. After the GO/DONE bit has gone LOW, read the ADRESL and ADRESH registers to get the digital data output.
9. Go back to step 5.
Programming PIC18F458 ADC in Assembly

The Assembly language Program 13-1 illustrates the steps for ADC conversion shown above. The C version of the program is shown in Program 13-1C.

Program 13-1: This program gets data from channel 0 (RA0) of ADC and displays the result on PORTC and PORTD. This is done every quarter of second.

;Program 13-1
ORG 0000H
CLRF TRISC ;make PORTC an output
CLRF TRISD ;make PORTD an output
BSF TRISA, 0 ;make RA0 an input for analog input
MOVLW 0x81 ;Fosc/64, channel 0, A/D is on
MOVWF ADCON0
MOVLW 0xCE ;right justified, Fosc/64, AN0 = analog
MOVWF ADCON1
OVER CALL DELAY ;wait for Tacq (sample and hold time)
BSF ADCON0, GO ;start conversion
BACK BTFSC ADCON0, DONE ;keep polling end-of-conversion
BRA BACK ;wait for end of conversion
MOVFF ADRESL, PORTC ;give the low byte to PORTC
MOVFF ADRESH, PORTD ;give the high byte to PORTD
CALL QSEC_DELAY
BRA OVER ;keep repeating it
END

---

Figure 13-9. A/D Connection for Program 13-1
Programming PIC18F458 A/D in C

Program 13-1C is the C version of the ADC conversion for Program 13-1.

Program 13-1C: This program gets data from channel 0 (RA0) of ADC and displays the result on PORTC and PORTD. This is done every quarter of second. This is the C version of Program 13-1.

```c
void main(void)
{
    TRISC=0; //make PORTC output port
    TRISD=0; //make PORTD output port
    TRISAbits.TRISA0=0; //RA0 = INPUT for analog input
    ADCON0 = 0x81; //Fosc/64, channel 0, A/D is on
    ADCON1 = 0xCE; //right justified, Fosc/64,
                   //AN0 = analog

    while(1)
    {
        DELAY(1); //give A/D channel time to sample
        ADCON0bits.GO = 1; //start converting
        while(ADCON0bits.DONE == 1);
        PORTC=ADRESL; //display low byte on PORTC
        PORTD=ADRESH; //display high byte on PORTD
        DELAY(250); //wait for one quarter of a
                     //second before trying again
    }
}
```

Programming A/D converter using interrupts

In Chapter 11, we showed how to use interrupts instead of polling to avoid tying down the microcontroller. To program the A/D using the interrupt method, we need to set HIGH the ADIE (A/D interrupt enable) flag. If ADIE = 1, then upon the completion of the conversion, the ADIF (A/D interrupt flag) becomes HIGH, which will force the CPU to jump to read binary outputs. Table 13-4 shows to which registers these two flags belong.

<table>
<thead>
<tr>
<th>Interrupt</th>
<th>Flag bit</th>
<th>Register</th>
<th>Enable bit</th>
<th>Register</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADIF (ADC)</td>
<td>ADIF</td>
<td>PIR1</td>
<td>ADIE</td>
<td>PIE1</td>
</tr>
</tbody>
</table>

*Note: Upon power-on reset, the A/D is assigned to high-priority interrupt (vector address of 0008). We can use the ADIP bit of the IPR1 register to assign low priority to it, which will land at vector address 00018H. See Chapter 11.*
;Program 13-2

ORG 0000H
GOTO MAIN ;bypass interrupt vector table

;--on default all interrupts go to to address 00008
ORG 0008H ;interrupt vector table
BTFSS PIR1,ADIF ;Did we get here due to A/D int?
RETFIE ;No. Then return to main
GOTO AD_ISR ;Yes. Then go INTO ISR

;--the main program for initialization
ORG 00100H

MAIN CLRF TRISC ;make PORTC an output
CLRF TRISD ;make PORTD an output
BSF TRISA,0 ;make RA0 an input pin for analog input
MOVLW 0x81 ;Fosc/64, channel 0, A/D is on
MOVWF ADCON0
MOVLW 0xCE ;right justified, Fosc/64, AN0 = analog
MOVWF ADCON1
BCF PIR1,ADIF ;clear ADIF for the first round
BSF PIE1,ADIE ;enable A/D interrupt
BSF INTCON,PEIE ;enable peripheral interrupts
BSF INTCON,GIE ;enable interrupts globally

OVER CALL DELAY ;wait for Tacq (sample and hold time)
BSF ADCON0,GO ;start conversion
BRA OVER ;stay in this loop forever

;-----A/D Converter ISR

AD_ISR

ORG 200H

MOVFF ADRESL,PORTC ;give the low byte to PORTC
MOVFF ADRESH,PORTD ;give the high byte to PORTD
CALL QSEC_DELAY
BSF PIR1,ADIF ;clear ADIF interrupt flag bit
RETFIE

END

//Program 13-2C (This is the C version of Program 13-2)

#include <PIC18F458.h>

#pragma code My_HiPrio_Int=0x0008 //high-priority interrupt
void My_HiPrio_Int (void)
{
    chk_isr();
}
#pragma code //end high-priority interrupt
#pragma interrupt chk_isr //Which interrupt?
void chk_isr (void)
{
    if (PIR1bits.ADIF==1) //A/D caused interrupt?
        AD_ISR( ); //Yes. Execute INTO program
}
void main(void)
{
    TRISC=0; // make PORTC output port
    TRISD=0; // make PORTD output port
    TRISAbits.TRISA0=0; // RA0 = INPUT for analog input
    ADCON0 = 0x81; // Fosc/64, channel 0, A/D is on
    ADCON1 = 0xCE; // right justified, Fosc/64, AN0 = analog
    PIE1bits.ADIF=0; // clear A/D interrupt flag
    INTCONbits.PEIE=1; // enable peripheral interrupts
    INTCONbits.GIE=1; // enable all interrupts globally
    while(1) // keep looping until interrupt comes
    {
        DELAY(1);
        ADCON0bits.GO = 1; // start converting
    }
}

// ----------- A/D ISR
void AD_ISR(void)
{
    PORTC=ADRESL; // display low byte on PORTC
    PORTD=ADRESH; // display high byte on
    DELAY(250); // wait for one quarter of a
    // second before trying again
    PIE1bits.ADIF=0; // clear A/D interrupt flag
}

Review Questions

1. Give the main factor affecting the step size of A/D in PIC18.
2. The A/D of PIC18 is a(n) _______ -bit converter.
3. True or false. The A/D of PIC18 has pins for DOUT.
4. True or false. A/D in the PIC18 is an off-chip module.
5. Find the step size for an PIC18 ADC, if Vref = 1.024 V.
6. For problem 5, calculate the D0–D9 output if the analog input is: (a) 0.7 V, and (b) 1 V.
7. Indicate the number of available analog input channels for each of the following options in the ADCON0 register:
   (a) PCFG = 0100  (b) PCFG = 1001
8. True or false. The conversion time is equal to 12 x Tad.
9. The minimum Tad allowed is _______ μs.
10. Which bit is used to poll for the end of conversion?
SECTION 13.3: DAC INTERFACING

This section will show how to interface a DAC (digital-to-analog converter) to the PIC18. Then we demonstrate how to generate a sine wave on the scope using the DAC.

Digital-to-analog converter (DAC)

The digital-to-analog converter (DAC) is a device widely used to convert digital pulses to analog signals. In this section we discuss the basics of interfacing a DAC to the PIC18.

Recall from your digital electronics course the two methods of creating a DAC: binary weighted and R/2R ladder. The vast majority of integrated circuit DACs, including the MC1408 (DAC0808) used in this section, use the R/2R method because it can achieve a much higher degree of precision. The first criterion for judging a DAC is its resolution, which is a function of the number of binary inputs. The common ones are 8, 10, and 12 bits. The number of data bit inputs decides the resolution of the DAC because the number of analog output levels is equal to \(2^n\), where \(n\) is the number of data bit inputs. Therefore, an 8-input DAC such as the DAC0808 provides 256 discrete voltage (or current) levels of output. See Figure 13-10. Similarly, the 12-bit DAC provides 4,096 discrete voltage levels. There are also 16-bit DACs, but they are more expensive.

![Figure 13-10. DAC Block Diagram](image)

**MC1408 DAC (or DAC0808)**

In the MC1408 (DAC0808), the digital inputs are converted to current (\(I_{out}\)), and by connecting a resistor to the \(I_{out}\) pin, we convert the result to voltage. The total current provided by the \(I_{out}\) pin is a function of the binary numbers at the D0–D7 inputs of the DAC0808 and the reference current (\(I_{ref}\)), and is as follows:

\[
I_{out} = I_{ref} \left( \frac{D7}{2} + \frac{D6}{4} + \frac{D5}{8} + \frac{D4}{16} + \frac{D3}{32} + \frac{D2}{64} + \frac{D1}{128} + \frac{D0}{256} \right)
\]

where D0 is the LSB, D7 is the MSB for the inputs, and \(I_{ref}\) is the input current that must be applied to pin 14. The \(I_{ref}\) current is generally set to 2.0 mA. Figure 13-11 shows the generation of current reference (setting \(I_{ref} = 2\) mA) by using the
standard 5 V power supply. Now assuming that \( I_{\text{ref}} = 2 \) mA, if all the inputs to the DAC are high, the maximum output current is 1.99 mA (verify this for yourself).

**Converting \( I_{\text{out}} \) to voltage in DAC0808**

Ideally we connect the output pin \( I_{\text{out}} \) to a resistor, convert this current to voltage, and monitor the output on the scope. In real life, however, this can cause inaccuracy because the input resistance of the load where it is connected will also affect the output voltage. For this reason, the \( I_{\text{ref}} \) current output is isolated by connecting it to an op-amp such as the 741 with \( R_f = 5 \) kOhms for the feedback resistor. Assuming that \( R = 5 \) kOhms, by changing the binary input, the output voltage changes as shown in Example 13-7.

---

### Example 13-7

Assuming that \( R = 5 \) kOhms and \( I_{\text{ref}} = 2 \) mA, calculate \( V_{\text{out}} \) for the following binary inputs:

(a) 10011001 binary (99H)  
(b) 11001000 (C8H)

**Solution:**

(a) \( I_{\text{out}} = 2 \) mA \((153/256)\) = 1.195 mA and \( V_{\text{out}} = 1.195 \) mA \(\times 5K = 5.975 \) V

(b) \( I_{\text{out}} = 2 \) mA \((200/256)\) = 1.562 mA and \( V_{\text{out}} = 1.562 \) mA \(\times 5K = 7.8125 \) V

---

### Generating a sine wave

Example 13-8 shows how to generate a stair-step ramp. To generate a sine wave, we first need a table whose values represent the magnitude of the sine of angles between 0 and 360 degrees. The values for the sine function vary from \(-1.0\) to \(+1.0\) for 0- to 360-degree angles. Therefore, the table values are integer num-
Example 13-8

In order to generate a stair-step ramp, set up the circuit in Figure 13-11 and connect the output to an oscilloscope. Then write a program to send data to the DAC to generate a stair-step ramp.

Solution:

```assembly
CLRF TRISB ; PORTB as output
CLRF PORTB ; clear PORTB
AGAIN:
INCF PORTB,F ; count from 0 to FFH send it to DAC
RCALL DELAY ; let DAC recover
BRA AGAIN
```

Numbers representing the voltage magnitude for the sine of theta. This method ensures that only integer numbers are output to the DAC by the PIC18 microcontroller. Table 13-5 shows the angles, the sine values, the voltage magnitudes, and the integer values representing the voltage magnitude for each angle (with 30-degree increments). To generate Table 13-5, we assumed a full-scale voltage of 10 V for DAC output (as designed in Figure 13-11). Full-scale output of the DAC is achieved when all the data inputs of the DAC are HIGH. Therefore, to achieve the full-scale 10 V output, we use the following equation.

\[ V_{out} = 5 \text{ V} + (5 \times \sin \theta) \]

\( V_{out} \) of DAC for various angles is calculated and shown in Table 13-5. See Example 13-9 for verification of the calculations.

Table 13-5: Angle versus Voltage Magnitude for Sine Wave

<table>
<thead>
<tr>
<th>Angle ( \theta ) (degrees)</th>
<th>Sin ( \theta )</th>
<th>( V_{out} ) (Voltage Magnitude) ( 5 \text{ V} + (5 \times \sin \theta) )</th>
<th>Values Sent to DAC (decimal) (Voltage Mag. ( \times 25.6 ))</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>5</td>
<td>128</td>
</tr>
<tr>
<td>30</td>
<td>0.5</td>
<td>7.5</td>
<td>192</td>
</tr>
<tr>
<td>60</td>
<td>0.866</td>
<td>9.33</td>
<td>238</td>
</tr>
<tr>
<td>90</td>
<td>1.0</td>
<td>10</td>
<td>255</td>
</tr>
<tr>
<td>120</td>
<td>0.866</td>
<td>9.33</td>
<td>238</td>
</tr>
<tr>
<td>150</td>
<td>0.5</td>
<td>7.5</td>
<td>192</td>
</tr>
<tr>
<td>180</td>
<td>0</td>
<td>5</td>
<td>128</td>
</tr>
<tr>
<td>210</td>
<td>-0.5</td>
<td>2.5</td>
<td>64</td>
</tr>
<tr>
<td>240</td>
<td>-0.866</td>
<td>0.669</td>
<td>17</td>
</tr>
<tr>
<td>270</td>
<td>-1.0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>300</td>
<td>-0.866</td>
<td>0.669</td>
<td>17</td>
</tr>
<tr>
<td>330</td>
<td>-0.5</td>
<td>2.5</td>
<td>64</td>
</tr>
<tr>
<td>360</td>
<td>0</td>
<td>5</td>
<td>128</td>
</tr>
</tbody>
</table>

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Example 13-9

Verify the values given for the following angles: (a) 30° (b) 60°.

Solution:

(a) \( V_{\text{out}} = 5 \, \text{V} + (5 \, \text{V} \times \sin \theta) = 5 \, \text{V} + 5 \times \sin 30° = 5 \, \text{V} + 5 \times 0.5 = 7.5 \, \text{V} \)

DAC input value = 7.5 \, \text{V} \times 25.6 = 192 \text{ (decimal)}

(b) \( V_{\text{out}} = 5 \, \text{V} + (5 \, \text{V} \times \sin \theta) = 5 \, \text{V} + 5 \times \sin 60° = 5 \, \text{V} + 5 \times 0.866 = 9.33 \, \text{V} \)

DAC input value = 9.33 \, \text{V} \times 25.6 = 238 \text{ (decimal)}

To find the value sent to the DAC for various angles, we simply multiply the \( V_{\text{out}} \) voltage by 25.60 because there are 256 steps and full-scale \( V_{\text{out}} \) is 10 volts. Therefore, 256 steps / 10 V = 25.6 steps per volt. To further clarify this, look at the following code. This program sends the values to the DAC continuously (in an infinite loop) to produce a crude sine wave. See Figure 13-12.

;Program 13-3
OVER   MOVWF  TBLPTRU
MOVWF  TBLPTRH
MOVWF  TBLPTRL
CLRF   TRISB
AGAIN  TBLRD*
MOVF   TABLAT,W
XORLW  0x0
BZ      OVER
MOVF   PORTB
INCF   TBLPTRL,F
BRA    AGAIN
ORG  0x250
TABLE:  DB 'D'128',D'192',D'238',D'255',D'238',D'192'
        DB 'D'128',D'64',D'17',D'1',D'17',D'64',D'0'
END

;to get a better looking sine wave, regenerate
;Table 13-5 for 2-degree angles
Programming DAC in C

//Program 13-3C. This is the C version of Program 13-3.
#include <p18F458.h>
rom const unsigned char WAVEVALUE[12] = {128,192,238,255,
                                        238,192,128,64,
                                        17,0,17,64};

void main()
{
    unsigned char x;
    TRISB=0;
    while(1)
    {
        for(x=0;x<12;x++)
            PORTB = WAVEVALUE[x];
    }
}

Review Questions

1. In a DAC, input is _____ (digital, analog) and output is _____ (digital, analog).
2. In an ADC, input is _____ (digital, analog) and output is _____ (digital, analog).
3. DAC0808 is a(n) _____-bit D-to-A converter.
4. (a) The output of DAC0808 is in _____ (current, voltage).
    (b) True or false. The output of DAC0808 is ideal to drive a motor.
SECTION 13.4: SENSOR INTERFACING AND SIGNAL CONDITIONING

This section will show how to interface sensors to the microcontroller. We examine some popular temperature sensors and then discuss the issue of signal conditioning. Although we concentrate on temperature sensors, the principles discussed in this section are the same for other types of sensors such as light and pressure sensors.

Temperature sensors

_Transducers_ convert physical data such as temperature, light intensity, flow, and speed to electrical signals. Depending on the transducer, the output produced is in the form of voltage, current, resistance, or capacitance. For example, temperature is converted to electrical signals using a transducer called a _thermistor_. A thermistor responds to temperature change by changing resistance, but its response is not linear, as seen in Table 13-6.

<table>
<thead>
<tr>
<th>Temperature (°C)</th>
<th>Tf (K ohms)</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>29.490</td>
</tr>
<tr>
<td>25</td>
<td>10.000</td>
</tr>
<tr>
<td>50</td>
<td>3.893</td>
</tr>
<tr>
<td>75</td>
<td>1.700</td>
</tr>
<tr>
<td>100</td>
<td>0.817</td>
</tr>
</tbody>
</table>

From William Kleitz, _Digital Electronics_

The complexity associated with writing software for such nonlinear devices has led many manufacturers to market a linear temperature sensor. Simple and widely used linear temperature sensors include the LM34 and LM35 series from National Semiconductor Corp. They are discussed next.

LM34 and LM35 temperature sensors

The sensors of the LM34 series are precision integrated-circuit temperature sensors whose output voltage is linearly proportional to the Fahrenheit temperature. See Table 13-7. The LM34 requires no external calibration because it is internally calibrated. It outputs 10 mV for each degree of Fahrenheit temperature. Table 13-7 is a selection guide for the LM34.

The LM35 series sensors are precision integrated-circuit temperature sensors whose output voltage is linearly proportional to the Celsius (centigrade) tem-
perature. The LM35 requires no external calibration because it is internally calibrated. It outputs 10 mV for each degree of centigrade temperature. Table 13-8 is the selection guide for the LM35. (For further information see http://www.national.com.)

### Signal conditioning and interfacing the LM35 to the PIC18

Signal conditioning is widely used in the world of data acquisition. The most common transducers produce an output in the form of voltage, current, charge, capacitance, and resistance. We need to convert these signals to voltage, however, in order to send input to an A-to-D converter. This conversion (modification) is commonly called signal conditioning. See Figure 13-13. Signal conditioning can be a current-to-voltage conversion or a signal amplification. For example, the thermistor changes resistance with temperature. The change of resistance must be translated into voltages in order to be of any use to an ADC. Look at the case of connecting an LM34 to an ADC of the PIC18F458. The A/D has 10-bit resolution with a maximum of 1,024 steps and the LM34 (or LM35) produces 10 mV for every degree of temperature change. Now, if we use the step size of 10 mV, the $V_{out}$ will be 10,240 mV (10.24 V) for full-scale output. This is not acceptable even though the maximum temperature sensed by the LM34 is 300 degrees F, and the highest output for the A/D we will get is 3,000 mV (3.00 V). Now, if we use the step size of 2.5 mV, the $V_{out}$ will be $1,024 \times 2.5 \text{ mV} = 2,560 \text{ mV (2.56 V)}$ for full-scale output. That means we must set $V_{ref} = 2.56 \text{ V}$. This makes the binary output number for the A/D 4 times the real temperature (10 mV/2.5 mV = 4). We can scale it by dividing it by 4 to get the real number for temperature. See Table 13-9.

Figure 13-14 shows the connection of a temperature sensor to the PIC18F458. Notice that we use the LM336-2.5 zener diode to fix the voltage across the 10K pot at 2.5 volts. The use of the LM336-2.5 should overcome any fluctuations in the power supply.
Table 13-9: Temperature vs. V_{out} for PIC18 with V_{ref} = 2.56 V
(SS = 2.5 mV)

<table>
<thead>
<tr>
<th>Temp. (F)</th>
<th>V_{in} (mV)</th>
<th>#of steps</th>
<th>Binary ( V_{out} (b9-b0) )</th>
<th>Temp. in Binary</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>00 00000000</td>
<td>00000000</td>
</tr>
<tr>
<td>1</td>
<td>10</td>
<td>4</td>
<td>00 00000100</td>
<td>00000100</td>
</tr>
<tr>
<td>2</td>
<td>20</td>
<td>8</td>
<td>00 00001000</td>
<td>00000100</td>
</tr>
<tr>
<td>3</td>
<td>30</td>
<td>12</td>
<td>00 00011100</td>
<td>00011100</td>
</tr>
<tr>
<td>10</td>
<td>100</td>
<td>20</td>
<td>00 10100000</td>
<td>01001000</td>
</tr>
<tr>
<td>20</td>
<td>200</td>
<td>80</td>
<td>00 10100000</td>
<td>01001000</td>
</tr>
<tr>
<td>30</td>
<td>300</td>
<td>120</td>
<td>00 11110000</td>
<td>01110000</td>
</tr>
<tr>
<td>40</td>
<td>400</td>
<td>160</td>
<td>00 11110000</td>
<td>01110000</td>
</tr>
<tr>
<td>50</td>
<td>500</td>
<td>200</td>
<td>00 11001000</td>
<td>01100100</td>
</tr>
<tr>
<td>60</td>
<td>600</td>
<td>240</td>
<td>01 00111000</td>
<td>01001100</td>
</tr>
<tr>
<td>70</td>
<td>700</td>
<td>300</td>
<td>01 00101000</td>
<td>01001011</td>
</tr>
<tr>
<td>80</td>
<td>800</td>
<td>320</td>
<td>01 00100000</td>
<td>01001000</td>
</tr>
<tr>
<td>90</td>
<td>900</td>
<td>360</td>
<td>01 01101000</td>
<td>01011010</td>
</tr>
<tr>
<td>100</td>
<td>1000</td>
<td>400</td>
<td>01 10010000</td>
<td>01100100</td>
</tr>
</tbody>
</table>

Example 13-10

In Table 13-9, verify the PIC output for a temperature of 70 degrees. Find values in the PIC18 A/D registers of ADRESL and ADRESH.

Solution:

The step size is 2.56/1,024 = 2.5 mV because V_{ref} = 2.56 V.

For the 70 degrees temperature we have 700 mV output because the LM34 provides 10 mV output for every degree. Now, the number of steps are 700 mV/2.5 mV = 300 in decimal. Now 300 = 0101000000 in binary and the PIC18 A/D output registers have ADRESL = 0100000 and ADRESH = 00000001.
Reading and displaying temperature

Programs 13-4 and 13-4C show code for reading and displaying temperature in both Assembly and C respectively.

The programs correspond to Figure 13-14. Regarding these two programs, the following points must be noted:

(1) The LM34 (or LM35) is connected to channel 0 (RA0 pin).
(2) The channel AN3 (RA3 pin) is connected to the Vref of 2.56 V. That makes PCFG = 0010 for the ADCON1 register.
(3) The 10-bit output of the A/D is divided by 4 to get the real temperature.

The algorithm is as follows: (a) Shift right the ADRESL 2 bits, (b) rotate the ADRESH 2 bits, and (c) OR the ADRESH with ADRESL together to get the 8-bit output for temperature.

;Program 13-4
;this program reads the sensor and displays it on PORTD
L_Byte SET 0x20 ;set a location 0x20 for L_Byte
H_Byte SET 0x21 ;set a location 0x21 for H_Byte
BIN_TEMP SET 0x22 ;set a location 0x22 for BIN_TEMP
CLRF TRISD ;make PORTD an output
BSF TRISA,0 ;make RA0 an input pin for analog volt
BSF TRISA,3 ;make RA3 an input pin for Vref volt
MOVWL 0x81
MOVWF ADCON0 ;Fosc/64, channel 0, A/D is on
MOVWL 0xC5 ;right justified, Fosc/64,
MOVWF ADCON1 ;AN0 = analog, AN3 = Vref+
OVER CALL DELAY ;wait for Tacq (sample and hold time)
BACK BTFSC ADCON0, DONE;keep polling end-of-conversion (EOC)
BSF ADCON0, GO ;start conversion
BRA BACK ;wait for end-of-conversion
MOVFF ADRESL, L_Byte ;save the low byte
MOVFF ADRESH, H_Byte ;save the high byte
CALL ALGOL_10_TO_8 ;make it an 8-bit value
MOVFF BIN_TEMP, PORTD ;display the temp on PORTD
BRA OVER ;keep repeating it

;------------
ALGOL_10_TO_8
RRNCF L_Byte, F ;rotate right twice
RRNCF L_Byte, W
ANDLW 0x3F ;mask the upper 2 bits
MOVWF L_Byte
RRNCF H_Byte, F ;rotate right through carry twice
RRNCF H_Byte, W
ANDLW 0x0C0 ;mask the lower 6 bits
IORWF L_Byte, W ;combine low and high
MOVWF BIN_TEMP
RETURN

;------------

//Program 13-4C void main(void)
{
    unsigned char L_Byte, H_Byte, Bin_Temp;
    TRISD=0; //make PORTD output port
    TRISAbits.TRISA0=1; //RA0 = INPUT for analog input
TRISAbits.TRISA2=1; //RA2 = INPUT for vref input
ADCON0 = 0x81;      //Fosc/64, channel 0, A/D is on
ADCON1 = 0xC5;      //right justified, Fosc/64,
                    //AN0 = analog, AN3 = Vref+
while(1) {
    MSDelay(1);      //give A/D channel time to sample
    ADCON0bits.GO = 1; //start converting
    while(ADCON0bits.DONE == 1);     //wait for EOC
    L_Byte=ADRESL;  //save the low byte
    H_Byte=ADRESH;  //save the high byte
    L_Byte>>=2;    //shift right
    L_Byte&=0x3F;  //mask the upper 2 bits
    H_Byte<<=6;    //shift left 6 times
    H_Byte&=0xC0;  //mask the lower 6 bits
    Bin_Temp= L_Byte|H_Byte;
    PORTD=Bin_Temp;
}

Review Questions

1. True or false. The transducer must be connected to signal conditioning circuitry before it is sent to the ADC.
2. The LM35 provides _____ mV for each degree of _____ (Fahrenheit, Celsius) temperature.
3. The LM34 provides _____ mV for each degree of _____ (Fahrenheit, Celsius) temperature.
4. Why do we set the Vref of the PIC to 2.56 V if the analog input is connected to the LM35?
5. In Question 4, what is the temperature if the ADC output is 0011 1001?

SUMMARY

This chapter showed how to interface real-world devices such as DAC chips, ADC chips, and sensors to the PIC. First, we discussed both parallel and serial ADC chips, then described how the ADC module inside the PIC18 works and explained how to program it in both Assembly and C. Next we explored the DAC chip, and showed how to interface it to the PIC. In the last section we studied sensors. We also discussed the relation between the analog world and a digital device, and described signal conditioning, an essential feature of data acquisition systems.

PROBLEMS

SECTION 13.1: ADC CHARACTERISTICS

1. True or false. Sensor output is in analog.
2. True or false. A 10-bit ADC has 10-bit digital output.
3. True or false. ADC0848 is an 8-bit ADC.
4. True or false. MAX1112 is a 10-bit ADC.
5. True or false. An ADC with 8 channels of analog input must have 8 pins, one for each analog input.
6. True or false. For a serial ADC, it takes a longer time to get the converted digital data out of the chip.
7. True or false. ADC0848 has 4 channels of analog input.
8. True or false. MAX1112 has 8 channels of analog input.
9. True or false. ADC0848 is a serial ADC.
10. True or false. MAX1112 is a parallel ADC.
11. Which of the following ADC sizes provides the best resolution?
    (a) 8-bit  (b) 10-bit  (c) 12-bit  (d) 16-bit  (e) They are all the same.
12. In Question 11, which provides the smallest step size?
13. Calculate the step size for the following ADCs, if \( V_{\text{ref}} \) is 5 V:
    (a) 8-bit  (b) 10-bit  (c) 12-bit  (d) 16-bit
14. With \( V_{\text{ref}} = 1.28 \) V, find the \( V_{\text{in}} \) for the following outputs:
    (a) D7–D0 = 11111111  (b) D7–D0 = 10011001  (c) D7–D0 = 1101100
15. In the ADC0848, what should be the Vref value if we want a step size of 5 mV?
16. With Vref+ = 2.56 V and Vref− = Gnd, find the \( V_{\text{in}} \) for the following outputs:
    (a) D7–D0 = 11111111  (b) D7–D0 = 10011001  (c) D7–D0 = 01101100

SECTION 13.2: ADC PROGRAMMING IN THE PIC18

17. True or false. The PIC18F452/458 has an on-chip A/D converter.
18. True or false. A/D of the PIC18 is an 8-bit ADC.
19. True or false. PIC18F452/458 has 8 channels of analog input.
20. True or false. The unused analog pins of the PIC18F452/458 can be used for I/O pins.
21. True or false. The A/D conversion speed in the PIC18F452/458 depends on the crystal frequency.
22. True or false. Upon power-on reset, the A/D module of the PIC18F452/458 is turned on and ready to go.
23. True or false. The A/D module of the PIC18F452/458 has an external pin for the start-conversion signal.
24. True or false. The A/D module of the PIC18F452/458 can convert only one channel at a time.
25. True or false. The A/D module of the PIC18F452/458 can have multiple external Vref+ at any given time.
26. True or false. The A/D module of the PIC18F452/458 can use the Vdd for Vref+.
27. In the A/D of PIC18 what happens to the converted analog data? How do we know that the ADC is ready to provide us the data?
28. In the A/D of PIC18 what happens to the old data if we start conversion again before we pick up the last data?
29. Assume Vref− = Gnd. For the A/D of PIC18, find the step size for each of the following \( V_{\text{ref}} \):
    (a) \( V_{\text{ref}} = 1.024 \) V  (b) \( V_{\text{ref}} = 2.048 \) V  (c) \( V_{\text{ref}} = 2.56 \) V
30. In the PIC18, what should be the Vref value if we want a step size of 2 mV?
31. In the PIC18, what should be the Vref value if we want a step size of 3 mV?
32. With a step size of 1 mV, what is the analog input voltage if all outputs are 1?
33. With $V_{\text{ref}} = 1.024$ V, find the $V_{\text{in}}$ for the following outputs:
   (a) D9–D0 = 0011111111 (b) D9–D0 = 0010011000 (c) D9–D0 = 0011010000
34. In the A/D of PIC18, what should be the Vref value if we want a step size of 4 mV?
35. With Vref+ = 2.56 V and Vref− = Gnd, find the $V_{\text{in}}$ for the following outputs:
   (a) D9–D0 = 1111111111 (b) D9–D0 = 1000000001 (c) D9–D0 = 1100110000
36. Find the conversion time for the following cases if XTAL = 8 MHz:
   (a) Fosc/2 (b) Fosc/4 (c) Fosc/8 (d) Fosc/16 (e) Fosc/32
37. Find the conversion time for the following cases if XTAL = 12 MHz:
   (a) Fosc/8 (b) Fosc/16 (c) Fosc/32 (d) Fosc/64
38. How do we start conversion in the PIC18?
39. How do we recognize the end of conversion in the PIC18?
40. The PIC18F452/458 can have a minimum of _______ channels of analog input.
41. In the PIC18F452/458, what ports are used for the analog channels?
42. Which register of the PIC18 is used to designate the number of A/D channels?
43. Which register of the PIC18 is used to select the A/D’s conversion speed?
44. Which register of the PIC18 is used to select the analog channel to be converted?
45. Find the value for the ADCON0 register if we want Fosc/8, channel 0, and ADON on.
46. Find the value for the ADCON1 register if we want Fosc/64, 3 channels of analog input, and right-justified output.
47. Find the value for the ADCON0 register if we want Fosc/2, channel 2, and ADON off.
48. Find the value for the ADCON1 register if we want Fosc/32, 2 channels of analog input with external source for Vref+, and left-justified output.
49. Give the name of the interrupt flags for the A/D of the PIC18F452/458. State to which register they belong.
50. Upon power-on reset, the A/D of the PIC18F452/458 is given (low, high) priority.

SECTION 13.3: DAC INTERFACING

51. True or false. DAC0808 is the same as DAC1408.
52. Find the number of discrete voltages provided by the $n$-bit DAC for the following:
   (a) $n = 8$ (b) $n = 10$ (c) $n = 12$
53. For DAC1408, if $I_{\text{ref}} = 2$ mA, show how to get the $I_{\text{out}}$ of 1.99 when all inputs are HIGH.
54. Find the $I_{\text{out}}$ for the following inputs. Assume $I_{\text{ref}} = 2$ mA for DAC0808.
   (a) 100111001 (b) 110011100 (c) 111011100
   (d) 00100010 (e) 00001001 (f) 10001000
55. To get a smaller step, we need a DAC with ______ (more, fewer) digital inputs.
56. To get full-scale output, what should be the inputs for DAC?

SECTION 13.4: SENSOR INTERFACING AND SIGNAL CONDITIONING

57. What does it mean when a given sensor is said to have a linear output?
58. The LM34 sensor produces ______ mV for each degree of temperature.
59. What is signal conditioning?
60. What is the purpose of the LM336 Zener diode around the pot setting the V_ref in Figure 13-14?

ANSWERS TO REVIEW QUESTIONS

SECTION 13.1: ADC CHARACTERISTICS

1. Number of steps and Vref voltage
2. 8
3. True
4. (a) 8 (b) 8
5. 1.28 V/256 = 5 mV
6. (a) 0.7 V/ 5 mV= 140 in decimal and D7–D0 = 10001100 in binary.
   (a) 1 V/ 5 mV= 200 in decimal and D7–D0 = 11001000 in binary.

SECTION 13.2: ADC PROGRAMMING IN THE PIC18

1. Vref
2. 10
3. False
4. False
5. 1 mV
6. (a) 700 mV (101011100), (b) 1000 mV (1111101000)
7. (a) 2 channels (b) 6 channels
8. True
9. 1.6
10. DONE bit of the ADCON0 register

SECTION 13.3: DAC INTERFACING

1. Digital, analog
2. Analog, digital
3. 8
4. (a) current (b) true

SECTION 13.3: SENSOR INTERFACING AND SIGNAL CONDITIONING

1. True
2. 10, Celsius
3. 10, Fahrenheit
4. Using the 8-bit part of the 10-bit ADC, it gives us 256 steps, and 2.56 V/256 = 10 mV. The LM35 produces 10 mV for each degree of temperature, which matches the ADC's step size.
5. 00111001 = 57, which indicates it is 57 degrees.
CHAPTER 14

USING FLASH AND EEPROM MEMORIES FOR DATA STORAGE

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Contrast and compare various types of semiconductor memories in terms of their capacity, organization, and access time
- Describe the relationship between the number of memory locations on a chip, the number of data pins, and the chip capacity
- Define Flash ROM memory and describe its use in PIC18-based systems
- Contrast and compare PROM, EPROM, UV-EPROM, EEPROM, Flash memory EPROM, and mask ROM memories
- Code PIC18 Assembly and C programs for writing data into PIC18 Flash memory space
- Code PIC18 Assembly and C programs for erasing the Flash memory in PIC18
- Explain how to write data to EEPROM memory of the PIC18
- Explain how to read data from EEPROM memory of the PIC18
In this chapter we discuss how to access data stored in both Flash and EEPROM memories of the PIC18F. In Section 14.1 we study semiconductor memory concepts with emphasis on different types of ROM. In Section 14.2, the writing of data into PIC18F Flash memory is discussed. The accessing of EEPROM in the PIC18 is explored in Section 14.3.

SECTION 14.1: SEMICONDUCTOR MEMORY

In this section we discuss various types of semiconductor memories and their characteristics such as capacity, organization, and access time. In the design of all microprocessor-based systems, semiconductor memories are used as primary storage for code and data. Semiconductor memories are connected directly to the CPU and are the memory that the CPU first asks for information (code and data). For this reason, semiconductor memories are sometimes referred to as primary memory. The most widely used semiconductor memories are ROM and RAM. Before we discuss different types of RAM and ROM, we discuss some important terminology common to all semiconductor memories, such as capacity, organization, and speed.

Memory capacity

The number of bits that a semiconductor memory chip can store is called chip capacity. It can be in units of Kbits (kilobits), Mbits (megabits), and so on. This must be distinguished from the storage capacity of computer systems. While the memory capacity of a memory IC chip is always given in bits, the storage capacity of a computer system is given in bytes. For example, an article in a technical journal may state that the 128M chip has become popular. In that case, it is understood, although it is not mentioned, that 128M means 128 megabits because the article is referring to an IC memory chip. However, if an advertisement states that a computer comes with 128M memory, it is understood that 128M means 128 megabytes because it is referring to a computer system.

Memory organization

Memory chips are organized into a number of locations within the IC. Each location can hold 1 bit, 4 bits, 8 bits, or even 16 bits, depending on how it is designed internally. The number of bits that each location within the memory chip can hold is always equal to the number of data pins on the chip. How many locations exist inside a memory chip? That depends on the number of address pins. The number of locations within a memory IC always equals 2 to the power of the number of address pins. Therefore, the total number of bits that a memory chip can store is equal to the number of locations times the number of data bits per location. To summarize:

1. A memory chip contains \(2^x\) locations, where \(x\) is the number of address pins.
2. Each location contains \(y\) bits, where \(y\) is the number of data pins on the chip.
3. The entire chip will contain \(2^x \times y\) bits, where \(x\) is the number of address pins and \(y\) is the number of data pins on the chip.
Speed

One of the most important characteristics of a memory chip is the speed at which its data can be accessed. To access the data, the address is presented to the address pins, the READ pin is activated, and, after a certain amount of time has elapsed, the data shows up at the data pins. The shorter this elapsed time, the better, and consequently, the more expensive the memory chip. The speed of the memory chip is commonly referred to as its access time. The access time of memory chips varies from a few nanoseconds to hundreds of nanoseconds, depending on the IC technology used in the design and fabrication process.

The three important memory characteristics of capacity, organization, and access time will be explored extensively in this chapter. Table 14-1 serves as a reference for the calculation of memory characteristics. Examples 14-1 and 14-2 demonstrate these concepts.

Table 14-1: Powers of 2

<table>
<thead>
<tr>
<th>x</th>
<th>(2^x)</th>
</tr>
</thead>
<tbody>
<tr>
<td>10</td>
<td>1K</td>
</tr>
<tr>
<td>11</td>
<td>2K</td>
</tr>
<tr>
<td>12</td>
<td>4K</td>
</tr>
<tr>
<td>13</td>
<td>8K</td>
</tr>
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<td>16K</td>
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<td>17</td>
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</tr>
<tr>
<td>18</td>
<td>256K</td>
</tr>
<tr>
<td>19</td>
<td>512K</td>
</tr>
<tr>
<td>20</td>
<td>1M</td>
</tr>
<tr>
<td>21</td>
<td>2M</td>
</tr>
<tr>
<td>22</td>
<td>4M</td>
</tr>
<tr>
<td>23</td>
<td>8M</td>
</tr>
<tr>
<td>24</td>
<td>16M</td>
</tr>
<tr>
<td>25</td>
<td>32M</td>
</tr>
<tr>
<td>26</td>
<td>64M</td>
</tr>
<tr>
<td>27</td>
<td>128M</td>
</tr>
</tbody>
</table>

Example 14-1

A given memory chip has 12 address pins and 4 data pins. Find:
(a) the organization, and (b) the capacity.

Solution:

(a) This memory chip has 4,096 locations \(2^{12} = 4,096\), and each location can hold 4 bits of data. This gives an organization of 4,096 x 4, often represented as 4Kx4.
(b) The capacity is equal to 16K bits because there is a total of 4K locations and each location can hold 4 bits of data.

Example 14-2

A 512K memory chip has 8 pins for data. Find:
(a) the organization, and (b) the number of address pins for this memory chip.

Solution:

(a) A memory chip with 8 data pins means that each location within the chip can hold 8 bits of data. To find the number of locations within this memory chip, divide the capacity by the number of data pins. 512K/8 = 64K; therefore, the organization for this memory chip is 64Kx8.
(b) The chip has 16 address lines because \(2^{16} = 64K\).
ROM (read-only memory)

ROM is a type of memory that does not lose its contents when the power is turned off. For this reason, ROM is also called nonvolatile memory. There are different types of read-only memory, such as PROM, EPROM, EEPROM, flash EPROM, and mask ROM. Each is explained below.

PROM (programmable ROM) and OTP

PROM refers to the kind of ROM that the user can burn information into. In other words, PROM is a user-programmable memory. For every bit of the PROM, there exists a fuse. PROM is programmed by blowing the fuses. If the information burned into PROM is wrong, that PROM must be discarded because its internal fuses are permanently blown. For this reason, PROM is also referred to as OTP (one-time programmable). Programming ROM, also called burning ROM, requires special equipment called a ROM burner or ROM programmer.

EPROM (erasable programmable ROM) and UV-EPROM

EPROM was invented to allow changes in the contents of PROM after it is burned. In EPROM, one can program the memory chip and erase it thousands of times. This is especially necessary during development of the prototype of a microprocessor-based project. A widely used EPROM is called UV-EPROM, where UV stands for ultraviolet. The only problem with UV-EPROM is that erasing its contents can take up to 20 minutes. All UV-EPROM chips have a window through which the programmer can shine ultraviolet (UV) radiation to erase its contents. For this reason, EPROM is also referred to as UV-erasable EPROM or simply UV-EPROM. Figure 14-1 shows the pins for a UV-EPROM chip.

To program a UV-EPROM chip, the following steps must be taken:

1. Its contents must be erased. To erase a chip, remove it from its socket on the system board and place it in EPROM erase equipment to expose it to UV radiation for 15–20 minutes.
2. Program the chip. To program a UV-EPROM chip, place it in the ROM burner (programmer). To burn code or data into EPROM, the ROM burner uses 12.5 volts or higher, depending on the EPROM type. This voltage is referred to as $V_{pp}$ in the UV-EPROM data sheet.
3. Place the chip back into its socket on the system board.

As can be seen from the above steps, not only is there an EPROM programmer (burner), but there is also separate EPROM erase equipment. The main problem, and indeed the major disadvantage, of UV-EPROM is that it cannot be erased and programmed while it is in the system board. To find a solution to this problem, EEPROM was invented.

Notice the patterns of the IC numbers in Table 14-2. For example, part number 27128-25 refers to UV-EPROM that has a capacity of 128K bits and an access time of 250 nanoseconds. The capacity of the memory chip is indicated in the part number and the access time is given with a zero dropped. In part numbers, C refers to CMOS technology. Notice that 27XX refers to UV-EPROM chips.
Table 14-2: Some UV-EPROM Chips

<table>
<thead>
<tr>
<th>Part #</th>
<th>Capacity</th>
<th>Org.</th>
<th>Access</th>
<th>Pins</th>
<th>V_{PP}</th>
</tr>
</thead>
<tbody>
<tr>
<td>2716</td>
<td>16K</td>
<td>2Kx8</td>
<td>450 ns</td>
<td>24</td>
<td>25 V</td>
</tr>
<tr>
<td>2732</td>
<td>32K</td>
<td>4Kx8</td>
<td>450 ns</td>
<td>24</td>
<td>25 V</td>
</tr>
<tr>
<td>2732A-20</td>
<td>32K</td>
<td>4Kx8</td>
<td>200 ns</td>
<td>24</td>
<td>21 V</td>
</tr>
<tr>
<td>27C32-1</td>
<td>32K</td>
<td>4Kx8</td>
<td>450 ns</td>
<td>24</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>2764-20</td>
<td>64K</td>
<td>8Kx8</td>
<td>200 ns</td>
<td>28</td>
<td>21 V</td>
</tr>
<tr>
<td>2764A-20</td>
<td>64K</td>
<td>8Kx8</td>
<td>120 ns</td>
<td>28</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27C64-12</td>
<td>128K</td>
<td>16Kx8</td>
<td>250 ns</td>
<td>28</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27128-25</td>
<td>256K</td>
<td>32Kx8</td>
<td>150 ns</td>
<td>28</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27C256-15</td>
<td>512K</td>
<td>64Kx8</td>
<td>250 ns</td>
<td>28</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27C512-15</td>
<td>1,024K</td>
<td>128Kx8</td>
<td>150 ns</td>
<td>32</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27C020-15</td>
<td>2,048K</td>
<td>256Kx8</td>
<td>150 ns</td>
<td>32</td>
<td>12.5 V CMOS</td>
</tr>
<tr>
<td>27C040-15</td>
<td>4,096K</td>
<td>512Kx8</td>
<td>150 ns</td>
<td>32</td>
<td>12.5 V CMOS</td>
</tr>
</tbody>
</table>

Example 14-3

For ROM chip 27128, find the number of data and address pins.

Solution:

The 27128 has a capacity of 128K bits. It has 16Kx8 organization (all ROMs have 8 data pins), which indicates that there are 8 pins for data and 14 pins for address \(2^{14} = 16K\).

Figure 14-1. Pin Configurations for 27xx ROM Family
EEPROM (electrically erasable programmable ROM)

EEPROM has several advantages over EPROM, such as the fact that its method of erasure is electrical and therefore instant, as opposed to the 20-minute erasure time required for UV-EPROM. In addition, in EEPROM one can select the byte to be erased, in contrast to UV-EPROM, in which the entire contents of ROM are erased. The main advantage of EEPROM is that one can program and erase its contents while it is still in the system board. It does not require physical removal of the memory chip from its socket. In other words, unlike UV-EPROM, EEPROM does not require an external erasure and programming device. To utilize EEPROM fully, the designer must incorporate the circuitry to program the EEPROM into the system board. In general, the cost per bit for EEPROM is much higher than for UV-EPROM. In Section 14.3 we show how to access the PIC18 on-chip EEPROM.

Table 14-3: Some EEPROM and Flash Chips

<table>
<thead>
<tr>
<th>EEPROMs</th>
<th>Part No.</th>
<th>Capacity</th>
<th>Org.</th>
<th>Speed</th>
<th>Pins</th>
<th>V&lt;sub&gt;PP&lt;/sub&gt;</th>
</tr>
</thead>
<tbody>
<tr>
<td>2816A-25</td>
<td>16K</td>
<td>2Kx8</td>
<td>250 ns</td>
<td>24</td>
<td>5 V</td>
<td></td>
</tr>
<tr>
<td>2864A</td>
<td>64K</td>
<td>8Kx8</td>
<td>250 ns</td>
<td>28</td>
<td>5 V</td>
<td></td>
</tr>
<tr>
<td>28C64A-25</td>
<td>64K</td>
<td>8Kx8</td>
<td>250 ns</td>
<td>28</td>
<td>5 V</td>
<td>CMOS</td>
</tr>
<tr>
<td>28C256-15</td>
<td>256K</td>
<td>32Kx8</td>
<td>150 ns</td>
<td>28</td>
<td>5 V</td>
<td></td>
</tr>
<tr>
<td>28C256-25</td>
<td>256K</td>
<td>32Kx8</td>
<td>250 ns</td>
<td>28</td>
<td>5 V</td>
<td>CMOS</td>
</tr>
<tr>
<td>Flash</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Part No.</td>
<td>Capacity</td>
<td>Org.</td>
<td>Speed</td>
<td>Pins</td>
<td>V&lt;sub&gt;PP&lt;/sub&gt;</td>
<td></td>
</tr>
<tr>
<td>28F256-20</td>
<td>256K</td>
<td>32Kx8</td>
<td>200 ns</td>
<td>32</td>
<td>12 V</td>
<td>CMOS</td>
</tr>
<tr>
<td>28F010-15</td>
<td>1,024K</td>
<td>128Kx8</td>
<td>150 ns</td>
<td>32</td>
<td>12 V</td>
<td>CMOS</td>
</tr>
<tr>
<td>28F020-15</td>
<td>2,048K</td>
<td>256Kx8</td>
<td>150 ns</td>
<td>32</td>
<td>12 V</td>
<td>CMOS</td>
</tr>
</tbody>
</table>

Flash memory EPROM

Since the early 1990s, flash EPROM has become a popular user-programmable memory chip, and for good reasons. First, the erasure of the entire contents takes less than a second, or one might say in a flash, hence its name, flash memory. In addition, the erasure method is electrical, and for this reason it is sometimes referred to as flash EEPROM. To avoid confusion, it is commonly called flash memory. The major difference between EEPROM and flash memory is that when flash memory's contents are erased (or written to), the entire device is erased, in contrast to EEPROM, where one can erase a desired section or byte. In recent decades, Flash memory contents are divided into blocks and the erasure (or write) is done block by block. Unlike EEPROM, Flash memory has no byte erase option. Because Flash memory can be programmed while it is in its socket on the system board, it has replaced the UV-EPROM for the storage of BIOS ROM of the PC. Nowadays, Flash memory is widely used for mass storage devices such as PDAs, cell phones, USB memory sticks, and MP3 players. Some computer sci-
entists believe that Flash memory will replace the hard disk as a mass storage medium. This would increase the performance of the computer tremendously, because Flash memory is semiconductor memory with access time in the range of 100 ns compared with disk access time in the range of tens of milliseconds. For this to happen, flash memory's program/erase cycles must become infinite, just like hard disks. Program/erase cycle refers to the number of times that a chip can be erased and programmed (written to) before it becomes unreliable. At this time, the program/erase cycle is in the 100,000s for Flash and EEPROM, in the 1,000s for UV-EPROM, and infinite for RAM and hard disks.

Mask ROM

Mask ROM refers to a kind of ROM in which the contents are programmed by the IC manufacturer. In other words, it is not a user-programmable ROM. The term mask is used in IC fabrication. Because the burning process is costly, mask ROM is used when the needed volume is high (hundreds of thousands) and it is absolutely certain that the contents will not change. It is common practice to use UV-EPROM or flash for the development phase of a project; and only after the code/data have been finalized is the mask version of the product ordered. The main advantage of mask ROM is its cost, because it is significantly cheaper than other kinds of ROM; but if an error is found in the data/code, the entire batch must be thrown away. Many manufacturers of 8051 microcontrollers support the mask ROM version of the 8051. Note that all ROM memories have 8 bits for data pins; therefore, the organization is x8.

RAM (random access memory)

RAM memory is called volatile memory because cutting off the power to the IC results in the loss of data. Sometimes RAM is also referred to as RAWM (read and write memory), in contrast to ROM, which cannot be written to. There are three types of RAM: static RAM (SRAM), NV-RAM (nonvolatile RAM), and dynamic RAM (DRAM). Each is explained separately.

SRAM (static RAM)

Storage cells in static RAM memory are made of flip-flops and therefore do not require refreshing in order to keep their data. This is in contrast to DRAM, discussed below. The problem with the use of flip-flops for storage cells is that each cell requires at least 6 transistors to build, and the cell holds only 1 bit of data. In recent years, the cells have been made of 4 transistors, which still is too many. The use of 4-transistor cells plus the use of CMOS technology has given birth to a high-capacity SRAM, but its capacity is far below that of DRAM. Table 14-4 shows some examples of SRAM. Figure 14-2 shows the pin

| A7 | 1 | 24 | Vcc |
| A6 | 2 | 23 | A8  |
| A5 | 3 | 22 | A9  |
| A4 | 4 | 21 | WE  |
| A3 | 5 | 20 | OE  |
| A2 | 6 | 19 | A10 |
| A1 | 7 | 18 | CS  |
| A0 | 8 | 17 | I/O 8 |
| I/O 1 | 9 | 16 | I/O 7 |
| I/O 2 | 10 | 15 | I/O 6 |
| I/O 3 | 11 | 14 | I/O 5 |
| GND | 12 | 13 | I/O 4 |

Figure 14-2. 2Kx8 SRAM Pins
diagram for an SRAM chip. In Figure 14-2, notice that WE is write enable, and OE is output enable, for read and write signals, respectively.

Table 14-4: Some SRAM and NV-RAM Chips

<table>
<thead>
<tr>
<th>SRAM</th>
<th>Part No.</th>
<th>Capacity</th>
<th>Org.</th>
<th>Speed</th>
<th>Pins</th>
<th>V_{PP}</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>6116P-1</td>
<td>16K</td>
<td>2Kx8</td>
<td>100 ns</td>
<td>24</td>
<td>CMOS</td>
</tr>
<tr>
<td></td>
<td>6116P-2</td>
<td>16K</td>
<td>2Kx8</td>
<td>120 ns</td>
<td>24</td>
<td>CMOS</td>
</tr>
<tr>
<td></td>
<td>6116P-3</td>
<td>16K</td>
<td>2Kx8</td>
<td>150 ns</td>
<td>24</td>
<td>CMOS</td>
</tr>
<tr>
<td></td>
<td>6116LP-1</td>
<td>16K</td>
<td>2Kx8</td>
<td>100 ns</td>
<td>24</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>6116LP-2</td>
<td>16K</td>
<td>2Kx8</td>
<td>120 ns</td>
<td>24</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>6116LP-3</td>
<td>16K</td>
<td>2Kx8</td>
<td>150 ns</td>
<td>24</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>6264P-10</td>
<td>64K</td>
<td>8Kx8</td>
<td>100 ns</td>
<td>28</td>
<td>CMOS</td>
</tr>
<tr>
<td></td>
<td>6264LP-70</td>
<td>64K</td>
<td>8Kx8</td>
<td>70 ns</td>
<td>28</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>6264LP-12</td>
<td>64K</td>
<td>8Kx8</td>
<td>120 ns</td>
<td>28</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>62256LP-10</td>
<td>256K</td>
<td>32Kx8</td>
<td>100 ns</td>
<td>28</td>
<td>Low-power CMOS</td>
</tr>
<tr>
<td></td>
<td>62256LP-12</td>
<td>256K</td>
<td>32Kx8</td>
<td>120 ns</td>
<td>28</td>
<td>Low-power CMOS</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>NV-RAM from Dallas Semiconductor</th>
<th>Part No.</th>
<th>Capacity</th>
<th>Org.</th>
<th>Speed</th>
<th>Pins</th>
<th>V_{PP}</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>DS1220Y-150</td>
<td>16K</td>
<td>2Kx8</td>
<td>150 ns</td>
<td>24</td>
<td></td>
</tr>
<tr>
<td></td>
<td>DS1225AB-150</td>
<td>64K</td>
<td>8Kx8</td>
<td>150 ns</td>
<td>28</td>
<td></td>
</tr>
<tr>
<td></td>
<td>DS1230Y-85</td>
<td>256K</td>
<td>32Kx8</td>
<td>85 ns</td>
<td>28</td>
<td></td>
</tr>
</tbody>
</table>

**NV-RAM (nonvolatile RAM)**

Whereas SRAM is volatile, there is a new type of nonvolatile RAM called NV-RAM. Like other RAMs, it allows the CPU to read and write to it, but when the power is turned off the contents are not lost. NV-RAM combines the best of RAM and ROM: the read and write ability of RAM, plus the nonvolatility of ROM. To retain its contents, every NV-RAM chip internally is made of the following components:

1. It uses extremely power-efficient (very low power consumption) SRAM cells built out of CMOS.
2. It uses an internal lithium battery as a backup energy source.
3. It uses an intelligent control circuitry. The main job of this control circuitry is to monitor the V_{CC} pin constantly to detect loss of the external power supply. If the power to the V_{CC} pin falls below out-of-tolerance conditions, the control circuitry switches automatically to its internal power source, the lithium battery. The internal lithium power source is used to retain the NV-RAM contents only when the external power source is off.

It must be emphasized that all three of the components above are incorporated into a single IC chip, and for this reason nonvolatile RAM is a very expen-
sive type of RAM as far as cost per bit is concerned. Offsetting the cost, however, is the fact that it can retain its contents up to ten years after the power has been turned off and allows one to read and write in exactly the same way as SRAM. See Table 14-4 for NV-RAM parts made by Dallas Semiconductor.

**DRAM (dynamic RAM)**

Since the early days of the computer, the need for huge, inexpensive read/write memory has been a major preoccupation of computer designers. In 1970, Intel Corporation introduced the first dynamic RAM (random access memory). Its density (capacity) was 1,024 bits and it used a capacitor to store each bit. Using a capacitor to store data cuts down the number of transistors needed to build the cell; however, the capacitor requires constant refreshing due to leakage. This is in contrast to SRAM (static RAM), whose individual cells are made of flip-flops. Because each bit in SRAM uses a single flip-flop, and each flip-flop requires 6 transistors, SRAM has much larger memory cells and consequently lower density. The use of capacitors as storage cells in DRAM results in much smaller net memory cell size.

The advantages and disadvantages of DRAM memory can be summarized as follows. The major advantages are high density (capacity), cheaper cost per bit, and lower power consumption per bit. The disadvantage is that it must be refreshed periodically because the capacitor cell loses its charge; furthermore, while DRAM is being refreshed, the data cannot be accessed. This is in contrast to SRAM's flip-flops, which retain data as long as the power is on, do not need to be refreshed, and whose contents can be accessed at any time. Since 1970, the capacity of DRAM has exploded. After the 1K-bit (1,024-bit) chip came the 4K-bit in 1973, and then the 16K chip in 1976. The 1980s saw the introduction of 64K, 256K, and finally 1M and 4M memory chips. The 1990s saw 16M, 64M, 256M, and the beginning of 1G-bit DRAM chips. In the 2000s, 2G-bit chips are standard, and as the fabrication process gets smaller, larger memory chips will be rolling off the manufacturing line. Keep in mind that when talking about IC memory chips, the capacity is always assumed to be in bits. Therefore, a 1M chip means a 1-megabit chip and a 256K chip means a 256-kilobit memory chip. When talking about the memory of a computer system, however, it is always assumed to be in bytes.

**Packaging issue in DRAM**

In DRAM there is a problem in packing a large number of cells into a single chip with the normal number of pins assigned to addresses. For example, a 64K chip (64Kx1) must have 16 address lines and 1 data line, requiring 16 pins to send in the address if the conventional method is used. This is in addition to $V_{CC}$ power, ground, and read/write control pins. Using the conventional method of data access, the large number of pins defeats the purpose of high density and small packaging, so dearly cherished by IC designers. Therefore, to reduce the number of pins needed for addresses, multiplexing/demultiplexing is used. The method used is to split the address in half and send in each half of the address through the same pins, thereby requiring fewer address pins. See Table 14-5. Internally, the DRAM structure is divided into a square of rows and columns. The first half of the address is called the row and the second half is called the column. For example, in the case
of DRAM of 64Kx1 organization, the first half of the address is sent in through the 8 pins A0–A7, and by activating RAS (row address strobe), the internal latches inside DRAM grab the first half of the address. After that, the second half of the address is sent in through the same pins, and by activating CAS (column address strobe), the internal latches inside DRAM latch the second half of the address. This results in using 8 pins for addresses plus RAS and CAS, for a total of 10 pins, instead of the 16 pins that would be required without multiplexing. To access a bit of data from DRAM, both row and column addresses must be provided. For this concept to work, there must be a 2-by-1 multiplexer outside the DRAM circuitry and a demultiplexer inside every DRAM chip. Due to the complexities associated with DRAM interfacing (RAS, CAS, the need for multiplexer and refreshing circuitry), there are DRAM controllers designed to make DRAM interfacing much easier. However, many small microcontroller-based projects that do not require much RAM (usually less than 64K bytes) use SRAM of types EEPROM and NV-RAM, instead of DRAM.

### Table 14-5: Some Widely Used DRAMs

<table>
<thead>
<tr>
<th>Part No.</th>
<th>Speed</th>
<th>Capacity</th>
<th>Org.</th>
<th>Pins</th>
</tr>
</thead>
<tbody>
<tr>
<td>4164-15</td>
<td>150 ns</td>
<td>64K</td>
<td>64Kx1</td>
<td>16</td>
</tr>
<tr>
<td>41464-8</td>
<td>80 ns</td>
<td>256K</td>
<td>64Kx4</td>
<td>18</td>
</tr>
<tr>
<td>41256-15</td>
<td>150 ns</td>
<td>256K</td>
<td>256Kx1</td>
<td>16</td>
</tr>
<tr>
<td>41256-6</td>
<td>60 ns</td>
<td>256K</td>
<td>256Kx1</td>
<td>16</td>
</tr>
<tr>
<td>414256-10</td>
<td>100 ns</td>
<td>1M</td>
<td>256Kx1</td>
<td>20</td>
</tr>
<tr>
<td>511000P-8</td>
<td>80 ns</td>
<td>1M</td>
<td>1Mx1</td>
<td>18</td>
</tr>
<tr>
<td>514100-7</td>
<td>70 ns</td>
<td>4M</td>
<td>4Mx1</td>
<td>20</td>
</tr>
</tbody>
</table>

### DRAM organization

In the discussion of ROM, we noted that all of them have 8 pins for data. This is not the case for DRAM memory chips, which can have x1, x4, x8, or x16 organizations. See Example 14-4.

### Example 14-4

Discuss the number of pins set aside for addresses in each of the following memory chips: (a) 16Kx4 DRAM (b) 16Kx4 SRAM

**Solution:**

Because \(2^{14} = 16K\):

(a) For DRAM we have 7 pins (A0–A6) for the address pins and 2 pins for RAS and CAS.

(b) For SRAM we have 14 pins for address and no pins for RAS and CAS because they are associated only with DRAM. In both cases we have 4 pins for the data bus.
In memory chips, the data pins are also called I/O. In some DRAMs there are separate $D_{in}$ and $D_{out}$ pins. Figure 14-3 shows a 256Kx1 DRAM chip with pins A0–A8 for address, RAS and CAS, WE (write enable), and data in and data out, as well as power and ground.

**Review Questions**

1. The speed of semiconductor memory is in the range of
   (a) microseconds (b) milliseconds
   (c) nanoseconds (d) picoseconds

2. Find the organization and chip capacity for each ROM with the indicated number of address and data pins:
   (a) 14 address, 8 data (b) 16 address, 8 data (c) 12 address, 8 data

3. Find the organization and chip capacity for each RAM with the indicated number of address and data pins:
   (a) 11 address, 1 data SRAM (b) 13 address, 4 data SRAM
   (c) 17 address, 8 data SRAM (d) 8 address, 4 data DRAM
   (e) 9 address, 1 data DRAM (f) 9 address, 4 data DRAM

4. Find the capacity and number of pins set aside for address and data for memory chips with the following organizations:
   (a) 16Kx4 SRAM (b) 32Kx8 EPROM (c) 1Mx1 DRAM
   (d) 256Kx4 SRAM (e) 64Kx8 EEPROM (f) 1Mx4 DRAM

5. Which of the following is (are) volatile memory?
   (a) EEPROM (b) SRAM (c) DRAM (d) NV-RAM

**SECTION 14.2: ERASING AND WRITING TO FLASH IN THE PIC18F**

The PIC18F comes with three types of memory (a) SRAM, (b) Flash, and (c) EEPROM. The SRAM is for general purpose usage including function registers, as we have seen throughout the book. The EEPROM is used for storing data only. While the Flash memory is used primarily to store program (code), we can also use it for storing fixed data such as look-up tables as we have seen throughout the book. In Chapter 6 we discussed how to use the TBLRD instruction to read the fixed data stored in program Flash. In this section, we discuss how to write to Flash memory. In the next section, we discuss how to access the EEPROM memory in the PIC18.

There are two ways to store (write) information (code or data) to the Flash memory or erase its content: (a) using an external Flash programmer (burner) such as PICSTART, and (b) using instructions such as TBLWR. In this section, we show how to use the TBLWR (table write) instruction to write to Flash memory. We will also show how to erase the contents of Flash memory. Due to similarities between the TBLRD and TBLWR instructions, it is very helpful to understand the material in Section 6.3 of Chapter 6, where we showed how to use the TBLRD instruction to read data stored in the Flash ROM.
Using TBLWR to write data to Flash

There are some major similarities between the TBLRD and TBLWR instructions. From Chapter 6, recall that in using the TBLRD instruction, we use the TBLPTR register as pointer to the data in Flash and the TABLAT register as a temporary place to store the data fetched from Flash. In the same way, the TBLWR instruction writes data held in the TABLAT register to the Flash ROM location whose address is pointed to by the TBLPTR register. In terms of autoincrement/autodecrement, the TBLRD and TBLWR instructions are exactly the same. See Table 14-6.

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBLWT*</td>
<td>Table Write</td>
<td>After write, TBLPTR stays the same</td>
</tr>
<tr>
<td>TBLWT*+</td>
<td>Table Write with post-inc</td>
<td>Write and increment TBLPTR</td>
</tr>
<tr>
<td>TBLWT*-</td>
<td>Table Write with post-dec</td>
<td>Write and decrement TBLPTR</td>
</tr>
<tr>
<td>TBLWT++</td>
<td>Table Write with pre-inc</td>
<td>Increment TBLPTR and then write</td>
</tr>
</tbody>
</table>

There is a major difference between the TBLRD and TBLWR instructions. While the TBLRD reads individual bytes from Flash, the TBLWR writes a block of 8 bytes to Flash. The TBLRD instruction reads one byte at a time from the Flash into the TABLAT registers, which means we must save the contents of the TABLAT before the next read wipes it out. The TBLWR instruction uses what is called short write and long write to write to Flash. In the short write, we use the TBLWR instruction to write a block of 8 bytes of data into 8 TABLAT registers one byte at a time. After the short write is done, we use the long write to actually store (write, or one might say burn) the entire block of 8 bytes into the Flash. The long write is done with the help of a register called EECON1, shown in Figure 14-4. Notice that the EECON1 register is used for both Flash and EEPROM memory, as we will see in the next section. Also contrast the difference between the Flash and EEPTROM memory. In Flash memory, the write or erase process is done on a block of data, while in EEPROM we can write or erase one byte at a time, which means it is byte-accessible memory. Readings for both Flash and EEPROM memories are in byte sizes. The block size for Flash memory varies among the Flash memories depending on their size and intended application. The block size for write/erase in the PIC184580 is 8 bytes, while in other Flash memories it is 64 or 256 bytes. The breaking of the PIC18F Flash into blocks of 8 bytes means the memory addresses must be on the 8-byte boundaries. This means that the lower three bits of the address A21–A0 of the Flash ROM location must be all zeros. See Figure 14-5.
<table>
<thead>
<tr>
<th>EEPGD</th>
<th>CFGS</th>
<th>FREE</th>
<th>WRERR</th>
<th>WERN</th>
<th>WR#</th>
<th>RD#</th>
</tr>
</thead>
</table>

**EEPGD** Flash Program or Data EEPROM Memory select bit  
1 = Access Program Flash memory  
0 = Access Data EEPROM memory

**CFGs** Flash Program/Data EE or Configuration Select bit  
1 = Access Configuration Registers  
0 = Access Program Flash or Data EEPROM memory

**FREE** Flash Row Erase Enable bit  
1 = Erase the Program Flash memory row addresses by TBLPTR on the next  
WR command (this bit is cleared when the Erase operation is completed)  
0 = Perform write only

**WERR** Write Error Flag bit  
1 = A write operation is prematurely terminated  
0 = The write operation is completed

**WREN** Write Enable bit  
1 = Allows write cycle  
0 = Inhibits write to the EEPROM or Flash memory

**WR#** Write Control bit. This is an an active-LOW signal used for both Flash and  
EEPROM. We can only make it HIGH by software and the PIC will make it  
LOW automatically when the write cycle is completed  
1 = initiates the write cycle to Flash or EEPROM (also used for initiating  
Erase / write cycle).  
0 = Write cycle is completed

**RD#** Read Control bit. This is an an active-LOW signal used by EEPROM only. We  
can only make it HIGH by software and the PIC will make it LOW  
automatically when the read cycle is completed.  
1 = Initiates the read cycle to EEPROM  
0 = Does not initiate an EEPROM read

---

Figure 14-4. EECO1 (EEPROM Control Register, also used for Flash)
Although there are 8 TABLAT registers for the short write, they are not accessible individually. These 8 TABLAT registers are internal and are used solely for the purpose of the short write using the TBLWRT instruction. Compare Figures 14-6 and 14-7 to contrast reading and writing to the Flash memory.

**Steps in writing to Flash memory**

Assuming that an area of Flash memory is erased, we can use the following steps to write a block of 8 bytes of data to the Flash memory:

1. Load the TBLPTR registers with the address of the first byte being written.
2. Using the TBLWR instruction, write 8 bytes of data to the TABLAT registers one after another. This is the end of the short write.
3. Set the EECON1 register for the write operation by setting (a) EEPOD = 1, (b) CFGS = 0, and (c) WREN = 1.
4. Disable all interrupts globally with "BCF INTCON, GIE".
5. Write 55H to the EECON2 dummy register. This is the start of a long write.
6. Write AAH to the EECON2 dummy register.
7. Set WR# to 1 with the instruction "BSF EECON1, WE". With WE = 1, the write cycle begins.
8. It will take about 2 ms to finish writing the 8 bytes to Flash. During this write cycle, the CPU is stalled and will not allow fetching any opcode. Upon completion of the write cycle, the WE# bit will go back low automatically to indicate that the write cycle is finished. This step concludes the end of long-write cycle.
9. Reenable the interrupts globally with "BSF INTCON, GIE".

Notice from step 4 that we must disable the interrupts to prevent any interruption of the write cycle (long write). If writing to Flash is interrupted by the reset pin (MCLR) or the WDT (watch dog timer), the WERR (write error) bit of the
EECON1 will go HIGH to indicate that. The good thing is the EEPGD bit of the EECON1 remains HIGH, allowing us to fix the error by rewriting the data to Flash. The EECON2 register does not exist physically and cannot be accessed. It is used exclusively for the purpose of writing/erasing the Flash/EEPROM memory. Program 14-1 shows how to write 8 bytes of data to Flash locations starting at address 400H. After writing the bytes, we read and display them on PORTB one byte at a time to verify the write operation.

The C language version of Program 14-1 is given at the end of this section.
Program 14-1 (a) writes the message “GOOD BYE” to Flash memory starting at location 400H, and (b) reads the data from Flash and places it in PORTB one byte at a time.

; Program 14-1
COUNT EQU 0x20
MOVLW 0x00
MOVWF TBLPTRL ;load the low byte of address
MOVLW 0x04
MOVWF TBLPTRH ;load the high byte of address
; start a short write
MOVLW A'G' ;load the 'G' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write increment address
MOVLW A'O' ;load the 'O' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write increment address
MOVLW A'O' ;load the 'O' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'D' ;load the 'D' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A' ' ;load the space into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'B' ;load the 'B' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'Y' ;load the 'Y' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'E' ;load the 'E' byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
; start the long write cycle (write to Flash itself)
MOVLW 0x00
MOVWF TBLPTRL ;load the low byte of address
MOVLW 0x04
MOVWF TBLPTRH ;load the high byte of address
BSF EECON1,EEPGD ;point to Flash memory
BCF EECON1,CFGS ;access Flash program
BSF EECON1,WREN ;enable write
BCF INTCON,GIE ;disable all interrupts
MOVLW 55H ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0AAH ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1, WR ;now write it to Flash
NOP
BSF INTCON, GIE ;enable all interrupts
BCF EECON1, WREN ;disable write to memory

;read them back one byte at a time and examine the
;bytes on PORTB
MOVLW 0x00
MOVWF TBLPTRL; reload the low byte of address
MOVLW 0x04
MOVWF TBLPTRH; reload the high byte of address
CLRF TRISB ; PORTB an output port
MOVLW 0x8 ; counter = 8
MOVWF COUNT
OVER TBLRD*+ ; read the byte to TABLAT and increment
MOVFF TABLAT, PORTB ; send it to PORTB
CALL DELAY ; wait enough to see the byte
DECFSZ COUNT, F ; decrement counter
DECF COUNT, F
BRA OVER ; continue for all the bytes

If the size of the block is not 8 bytes, then the rest of the unused block will
be untouched. See Program 14-2.

Program 14-2 (a) writes the message “HELLO” to Flash memory starting
at location 450H, and (b) reads the data from Flash and places it in PORTB one
byte at a time.

Program 14-2
COUNT EQU 0x20
MOVLW 0x50
MOVWF TBLPTRL ; load the low byte of address
MOVLW 0x04
MOVWF TBLPTRH; load the high byte of address
; start a short write
MOVLW A'H' ; load the 'H' byte into WREG
MOVWF TABLAT ; move it to TABLATch reg
TBLWT*+ ; perform short write and increment
MOVLW A'E' ; load the 'E' byte into WREG
MOVWF TABLAT ; move it to TABLATch reg
TBLWT*+ ; perform short write and increment
MOVLW A'L' ; load the 'L' byte into WREG
MOVWF TABLAT ; move it to TABLATch reg
TBLWT*+ ; perform short write
MOVLW A'O' ; load the 'O' byte into WREG
MOVWF TABLAT ; move it to TABLATch reg

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;TBLWT*+ ;perform short write
;start the long write cycle (write to Flash itself)
  MOVLW 0x50
  MOVWF TBLPTRL ;load the low byte of address
  MOVLW 0x04
  MOVWF TBLPTRH ;load the high byte of address
  BSF EECON1,EEPGD ;point to Flash memory
  BCF EECON1,CFGS ;access Flash program
  BSF EECON1,WREN ;enable write
  BCF INTCON,GIE ;disable all interrupts
  MOVLW 55H ;wreg = 55h
  MOVWF EECON2 ;write to dummy reg
  MOVLW 0AAH ;wreg = aah
  MOVWF EECON2 ;write to dummy reg
  BSF EECON1,WR ;now write it to Flash
  NOP ;wait
  BSF INTCON,GIE ;enable all interrupts
  BCF EECON1,WREN ;disable write to memory
;read them back one byte at a time and examine the
;bytes on PORTB
  MOVLW 0x50
  MOVWF TBLPTRL ;reload the low byte of address
  MOVLW 0x04
  MOVWF TBLPTRH;reload the high byte of address
  CLRF TRISB ;PORTB an output port
  MOVLW 0x05 ;counter = 5
  MOVWF COUNT
OVER TBLRD*+ ;read the byte and incremment
  MOVF TABLAT,PORTB ;send it to PORTB
  CALL DELAY ;wait enough to see the byte
  DECF COUNT,F ;dec counter
  DECFZ COUNT,F
  BRA OVER ;continue for all the bytes

Program 14-3 (a) transfers a block of data from the code space of the PIC18 chip into RAM, (b) then writes the same data from RAM to Flash, and (c) reads the data from new Flash locations and sends it to the serial port of the PIC18 one byte at a time.

;Program 14-3
COUNT EQU 0x0B
BUFRAM EQU 0x20
  MOVLW D'8' ;number of bytes to retrieve
  MOVWF COUNT
  MOVLW high (BUFRAM) ;point to buffer
  MOVWF FSROH
MOVLW low (BUFRAM)
MOVF FSR0L
MOVLW upper (CODE_DATA) ;load TBLPTR
MOVF TBLPTRU
MOVLW high (CODE_DATA)
MOVF TBLPTRH
MOVLW low (CODE_DATA)
MOVF TBLPTRL
;retrieve the data from program memory
READ_BLOCK
TBLRD*+ ;read into TABLAT, and increment
MOVF TABLAT, W ;get data
MOVF POSTINCO ;store data
DECFSZ COUNT ;done?
BRA READ_BLOCK ;repeat
MOVLW upper (NEW_DATA) ;load TBLPTR
MOVF TBLPTRU
MOVLW high (NEW_DATA)
MOVF TBLPTRH
MOVLW low (NEW_DATA)
MOVF TBLPTRL
MOVLW high (BUFRAM) ;point to buffer
MOVF FSR0H
MOVLW low (BUFRAM)
MOVF FSR0L
MOVLW 8 ;number of bytes in RAM
MOVF COUNT
;move the data back to program memory
WRITE_BACK
MOVF POSTINCO, W ;get a byte from RAM
MOVF TABLAT ;store the byte in table latch
TBLWT*+ ;perform a short write
DECFSZ COUNT ;loop until buffers are full
BRA WRITE_BACK
MOVLW upper (NEW_DATA) ;load TBLPTR
MOVF TBLPTRU
MOVLW high (NEW_DATA)
MOVF TBLPTRH
MOVLW low (NEW_DATA)
MOVF TBLPTRL
BSF EECON1, EEPGD ;point to Flash program memory
BCF EECON1, CFGS ;access Flash program memory
BSF EECON1, WREN ;enable write to memory
BCF INTCON, GIE ;disable interrupts
MOVLW 55h ;write 55h
MOVF EECON2
MOVLW 0AAh ;write 0AAh

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MOVWF EECON2 ; start program (CPU stall)
BSF EECON1, WR
NOP
BSF INTCON, GIE ; re-enable interrupts
BCF EECON1, WREN ; disable write to memory
; read them back one byte at a time and send serially
BSF TRISD, 7 ; PORTD.7 as in input
MOVLW 0x20 ; enable transmit and low baud rate
MOVWF TXSTA ; write to reg
BCF PIR1, TXIF
MOVLW D'15' ; 9600 bps (Fosc/(64*Speed)-1)
MOVWF SPBRG ; write to reg
BCF TRISC, TX ; make TX pin of PORTC an output
BSF RCSTA, SPEN ; enable the entire serial port
MOVLW 8 ; number of bytes in RAM
MOVWF COUNT
MOVLW upper (NEW_DATA) ; load TBLPTR
MOVWF TBLPTRU
MOVLW high (NEW_DATA)
MOVWF TBLPTRH
MOVLW low (NEW_DATA)
MOVWF TBLPTRL
CLRF TRISB ; PORTB an output port
MOVLW 0x8 ; counter = 8
MOVWF COUNT
LN TBLRD*+
; read the character
MOVF TABLAT, W
R1 BTFSS PIR1, TXIF ; wait until the last bit is gone
BRA R1 ; stay in loop
MOVF TXREG ; load the value to be transmitted
DECFSZ COUNT ; loop until buffers are full
BRA LN ; repeat

Steps in erasing Flash memory

Although we can use external Flash programmers to erase the Flash memory contents, the PIC18 allows us to write a program to erase the Flash memory. The erase process works on block-size, not byte-size data. The minimum block size for the erase is 64 bytes. That means the lowest 6 bits of addresses are all zeros, making them 64-byte block boundaries. We can use the following steps to erase a single 64-byte block of Flash memory:

1. Load the TBLPTR registers with the address of the block being erased.
2. Set the EECON1 register for the erase operation by setting
   (a) EEPGD = 1, (b) CFGS = 0, (c) WREBN = 1, and (d) FREE = 1.
3. Disable all interrupts globally using “BCF INTCON, GIE”.
4. Write 55H to the EECON2 dummy register.
5. Write AAH to the EECON2 dummy register.
6. Set WR# to 1 with the instruction “BSF EECON1, WE”. With WE = 1, the erase cycle begins.
7. It will take about 2 ms to finish erasing the block of 64 bytes. During this erase cycle, the CPU is installed and will not allow fetching of any opcode. Upon completion of the erase cycle, the WE# bit will go back HIGH automatically to indicate the erase cycle is finished.
8. Reenable the interrupts globally using “BCF INTCON, GIE”.

Program 14-4 shows how to erase the 64-byte block.

;Program 14-4: This program erases the Flash memory starting at location 0x500.
ORG 0
MOVLW upper(MYDATA)
MOVWF TBLPTRLH ;load the upper address
MOVLW high(MYDATA)
MOVWF TBLPTRL ;load the high byte of address
MOVLW low(MYDATA)
MOVWF TBLPTRL ;load the low byte of address
BSF EECON1,EEPGD ;point to Flash memory
BCF EECON1,CFGS ;access Flash program
BSF EECON1,WREN ;enable write
BSF EECON1, FREE ;enable row erase operation
BCF INTCON,GIE ;disable all interrupts
MOVLW 55H ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0AAH ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1,WR ;now write it to Flash
NOP ;wait
BSF INTCON,WR ;enable all interrupts
BCF EECON1,WREN ;disable write to memory
HERE BRA HERE
ORG 500H
MYDATA data "ABCDEFGH"
END

Examine Program 14-5. It combines erasing, writing, and reading of the Flash memory.

;Program 14-5: This program erases the message of “GOOD BYE” from Flash addresses 0x1200 and replaces it with “HELLO”.
MOVLW upper(MYDATA)
MOVWF TBLPTRU ;load the upper address
MOVLW high(MYDATA)
MOVWF TBLPTRH ;load the high byte of address
MOVWF TBLPTRL ;load the low byte of address
BSF EECON1,EEPGD ;point to Flash memory
BCF EECON1,CFGS ;access Flash program
BSF EECON1,WREN ;enable write
BSF EECON1, FREE ;enable row erase operation
BCF INTCON,GIE ;disable all interrupts
MOVLW 55H ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0AAH ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1,WR ;now write it to Flash
NOP ;wait
BSF INTCON,GIE ;enable all interrupts
BCF EECON1,WREN ;disable write to memory
MOVLW upper(MYDATA)
MOVWF TBLPTRU ;load the upper address
MOVLW high(MYDATA)
MOVWF TBLPTRH ;load the high byte of address
MOVLW low(MYDATA)
MOVWF TBLPTRL ;load the low byte of address
;start a short write
MOVLW A'H' ;load the byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write and increment
MOVLW A'E' ;load the byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write and increment
MOVLW A'L' ;load the byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'L' ;load the byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
MOVLW A'O' ;load the byte into WREG
MOVWF TABLAT ;move it to TABLATch reg
TBLWT*+ ;perform short write
;start the long write cycle (write to Flash itself)
BSF EECON1,EEPGD ;point to Flash memory
BCF EECON1,CFGS ;
BSF EECON1,WREN ;enable write
BCF INTCON,GIE ;disable all interrupts
MOVLW 55H ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0AAH ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF      EECON1,WR  ;now write it to Flash
NOP      ;wait
BSF      INTCON,GIE ;enable all interrupts
BCF      EECON1,WREN  ;disable write to memory
;read them back one byte at a time and examine the
;bytes on PORTB
MOVLW    upper(MYDATA)
MOVWF    TBLPTRU  ;load the upper address
MOVLW    high(MYDATA)
MOVWF    TBLPTRH ;load the high byte of address
MOVLW    low(MYDATA)
MOVWF    TBLPTRL ;load the low byte of address
CLRF     TRISB   ;PORTB an output port
MOVLW    0x05    ;counter = 5
MOVWF    COUNT
OVER      TBLRD++ ;read byte to TABLAT and point to next
MOVFF    TABLAT,PORTB ;send it to PORTB
CALL      DELAY ;wait enough to see byte on PORTB
DECF      COUNT,F ;decrement counter
BNZ       OVER  ;continue for all the bytes
ORG       1200H
MYDATA  data "GOOD BYE"
END

Erasing and writing to Flash memory in C

Programs 14-6C through 14-8C are the C versions of earlier programs.

/*Program 14-6C: This C program (a) writes the mes-
 sage "GOOD BYE" to Flash memory starting at location
400H, (b) reads the data from Flash and places it in
PORTB one byte at a time. */

#include <pl8Cxxx.h>
void Delay(unsigned int itime);

void main()
{
    unsigned char x;  
    //write to program memory
    TBLPTR  = (short long)0x0400;  //load TBLPTR
    TABLAT='G';                //load in TABLAT
    _asm TBLWTPOSTINC _endasm  //short write
    TABLAT='O';                //load in TABLAT
    _asm TBLWTPOSTINC _endasm  //short write
    TABLAT='O';                //load in TABLAT
    _asm TBLWTPOSTINC _endasm  //short write
    TABLAT='O';                //load in TABLAT
    _asm TBLWTPOSTINC _endasm  //short write

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TABLAT='D'; //load in TABLAT
_asm TBLWTPPOSTINC _endasm //short write
TABLAT=' '; //load in TABLAT
_asm TBLWTPPOSTINC _endasm //short write
TABLAT='B'; //load in TABLAT
_asm TBLWTPPOSTINC _endasm //short write
TABLAT='Y'; //load in TABLAT
_asm TBLWTPPOSTINC _endasm //short write
TABLAT='E'; //load in TABLAT
_asm TBLWTPPOSTINC _endasm //short write

//long write
TBLPTR = (short long)0x0400; //reload TBLPTR
EECON1bits.EEPGD=1;
EECON1bits.CFGS=0;
EECON1bits.WREN=1;
INTCONbits.GIE=0;
EECON2='0x55;
EECON2='0xAA;
EECON1bits.WR=1;
_asm NOP _endasm
INTCONbits.GIE=1;
EECON1bits.WREN=0;

//read from program memory send to PORTB
TBLPTR = (short long)0x0400; //reload TBLPTR
for(x=0;x<8;x++){
    _asm TBLRDPOSTINC _endasm
    PORTB=TABLAT;
    Delay(250);
}

//Program 14-7C: This C program erases the Flash
//memory starting at location 0x500.

#include <pl8Cxxx.h>

#pragma romdata const_table = 0x500
const rom char my_const_array[10] = "GOOD BYE";
#pragma romdata
void main()
{

    //erase program memory
    TBLPTR = (short long)0x0500; //load TBLPTR
    EECON1bits.EEPGD=1;


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EECON1bits.CFGS=0;
EECON1bits.WREN=1;
EECON1bits.FREE=1;
INTCONbits.GIE=0;
EECON2=0x55;
EECON2=0xAA;
EECON1bits.WR=1;
_asm NOP _endasm
INTCONbits.GIE=1;
EECON1bits.WREN=0;
}

//Program 14-8C: This C program erases the message of
//"GOOD BYE" from Flash addresses 0x1200 and replaces
//it with "HELLO".

#include <pl8Cxxx.h>

void Delay(unsigned int itime);

#pragma romdata const_table = 0x1200
const rom char my_const_array[10] = "GOOD BYE";
#pragma romdata

void main()
{
    unsigned char x;

    //erase program memory
    TBLPTR = (short long)0x1200;  //load TBLPTR
    EECON1bits.EEPGD=1;
    EECON1bits.CFGS=0;
    EECON1bits.WREN=1;
    EECON1bits.FREE=1;
    INTCONbits.GIE=0;
    EECON2=0x55;
    EECON2=0xAA;
    EECON1bits.WR=1;
    _asm NOP _endasm
    INTCONbits.GIE=1;
    EECON1bits.WREN=0;

    TBLPTR = (short long)0x1200;  //load TBLPTR
    TABLAT='H';                //load in TABLAT
    _asm TBLWTPPOSTINC _endasm  //short write
    TABLAT='E';                //load in TABLAT
    _asm TBLWTPPOSTINC _endasm  //short write
    TABLAT='L';                //load in TABLAT

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_asm TBLWTPOSTINC _endasm //short write
TABLAT='L'; //load in TABLAT
.asm TBLWTPOSTINC _endasm //short write
TABLAT='O'; //load in TABLAT
.asm TBLWTPOSTINC _endasm //short write

//long write
TBLPTR = (short long)0x1200; //reload TBLPTR
EECON1bits.EEPGD=1;
EECON1bits.CFGS=0;
EECON1bits.WREN=1;
INTCONbits.GIE=0;
EECON2=0x55;
EECON2=0xAA;
EECON1bits.WR=1;
.asm NOP _endasm
INTCONbits.GIE=1;
EECON1bits.WREN=0;

//read from program memory send to PORTB
TBLPTR = (short long)0x1200; //reload TBLPTR
for(x=0;x<8;x++){
  _asm TBLRDPOSTINC _endasm
  PORTB=TABLAT;
  Delay(250);
}

Review Questions

1. True or false. The PIC18F Flash memory can be used for both program code and data.
2. True or false. The PIC18F SRAM memory can be used for both program code and data.
3. True or false. In the PIC18F, writing to Flash is not allowed.
4. True or false. Reading from Flash memory is in byte size, while writing to it is in block size.
5. True or false. During the long write, the CPU keeps fetching and executing the instructions.
6. What is the size of the block for writing to Flash memory in the PIC18F458?
7. What is the size of the block for erasing the Flash memory in the PIC18F458?
SECTION 14.3: READING AND WRITING TO DATA EEPROM IN THE PIC18

The vast majority of the members of the PIC18 family come with some EEPROM memory. The amount varies from 256 bytes to a few K depending on the family member. For example, the PIC18F4520 has 256 bytes of EEPROM while PIC18F4585 has only 1,024 bytes. Table 14-7 shows some of the family members and their EEPROM space. While the Flash memory in PIC18F can be used for storing both code and data, the EEPROM space is used exclusively for storing data. Of the three memory spaces that PIC18 has, the SRAM and EEPROM are used for data only while the Flash is used mainly for program and sometimes for fixed data storage. See Figure 14-8.

Table 14-7: EEPROM Size for Some PIC18 Chips

<table>
<thead>
<tr>
<th>Part No.</th>
<th>On-chip Flash</th>
<th>On-chip RAM</th>
<th>On-chip EEPROM</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F1220</td>
<td>4 KB</td>
<td>256 B</td>
<td>256 B</td>
</tr>
<tr>
<td>PIC18F1230</td>
<td>4 KB</td>
<td>256 B</td>
<td>128 B</td>
</tr>
<tr>
<td>PIC18F2410</td>
<td>16 KB</td>
<td>768 B</td>
<td>0 B</td>
</tr>
<tr>
<td>PIC18F4520</td>
<td>32 KB</td>
<td>1,536 B</td>
<td>256 B</td>
</tr>
<tr>
<td>PIC18F4580</td>
<td>32 KB</td>
<td>1,536 B</td>
<td>256 B</td>
</tr>
<tr>
<td>PIC18F4585</td>
<td>48 KB</td>
<td>3,328 B</td>
<td>1,024 B</td>
</tr>
</tbody>
</table>

Note: On-chip RAM does not include the SFR space.

Figure 14-8. Contrasting PIC18F Memories
Writing data to EEPROM

There are four registers associated with the EEPROM. They are as follows:
(a) EEADR: An 8-bit register, used as pointer to EEPROM location.
(b) EEDATA: An 8-bit register, holds data to be written to EEPROM.
(c) EECON1: See Figure 14-4. Used by both EEPROM and Flash.
(d) EECON2: The dummy register. Used by both EEPROM and Flash.

Notice that the EEADR (EE address) register is only 8 bits wide in the PIC18F452/458. The 8-bit address gives us a total space of 256 bytes, which should cover the size of the EEPROM in these chips. In microcontroller chips such as PIC18F4585, which have 1,024 bytes of EEPROM, we have the low-byte and high-byte addresses for the EEADR and they are called EEADRL and EEADRH.

Steps in writing to EEPROM

To write a byte of data to a location in the EEPROM memory, we go through the following steps:

1. Load the EEADR registers with the address of the EEPROM location we want to write the data byte to.
2. Load the EEDATA registers with the data byte we want to write to EEPROM.
3. Set the EECON1 register for the EEPROM write by making
   (a) EEPGD = 0, (b) CFGS = 0, and (c) WREN = 1.
4. Disable all interrupts globally using “BCF INTCON, GIE”.
5. Write 55H to the EECON2 dummy register.
6. Write AAH to the EECON2 dummy register.
7. Set WR# to 1 with the instruction “BSF EECON1, WE”.
   With WE = 1, the write cycle begins.
8. Upon completion of the write cycle, the WE# bit will be cleared automatically to indicate that the write cycle is finished.
9. Re-enable the interrupts globally using “BCF INTCON, GIE”.
10. The WREN bit should be cleared to prevent an accidental write to the EEPROM by some runaway program.

In the above steps notice the last one. It is important to make WREN = 0, because the PIC18 will not do that automatically. The following program writes a single ASCII letter of ‘H’ to EEPROM address 10H.

```
MOVLW 0x10 ;starts at location 10H of EEPROM
MOVF  EEADRD ;load the EEPROM address
MOVLW A'H' ;load the byte into WREG
MOVF  EEDATA ;move it to EEDATA reg
BCF EECON1, EEPGD ;point to EEPROM memory
BCF EECON1, CFGS ;
BSF EECON1, WREN ;enable write
BCF INTCON, GIE ;disable all interrupts
MOVLW 0x55 ;wreg = 55h
```
MOVWF EECON2 ;write to dummy reg
MOVLW 0xA0 ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1, WR ;now write it to Flash
BSF INTCON, GIE ;enable all interrupts
BCF EECON1, WREN ;disable write to memory

Steps in reading from EEPROM

Reading a byte from the EEPROM memory is simple and straightforward as shown in the following steps:

1. Load the EEARD register with the address of the EEPROM location we want to read from.
2. Set the EECON1 register for the EEPROM read by making (a) EEPGD = 0, (b) CFGS = 0, and (c) RD = 1.
3. Within the next instruction cycle, the PIC18 will automatically fetch the data from the EEPROM location and place it in the EEDATA register. The only thing we have to do is to move data from the EEDTAT register to a safe place before we do another read. The following shows how to read a byte from EEPROM and place it in PORTB:

MOVLW 0x10 ;read location 10H of EEPROM
MOVWF EEAR; load the EEPROM address
BCF EECON1, EEPGD ;point to EEPROM memory
BCF EECON1, CFGS ;
BSF EECON1, RD ;enable read
NOP ;data is fetched from EEPROM to EEDATA reg
MOVFF EEDATA, PORTB ;place the data in PORTB

Program 14-9 (a) writes the message “HELLO” to EEPROM memory starting at location 0, and (b) reads the data back from EEPROM and places it in PORTB one byte at a time.

;Program 14-9: Writing to EEPROM
MOVWLW 0x00 ;starts at location 00H of EEPROM
MOVWF EEADR ;load the EEPROM address
MOVLW A'H' ;load the byte into WREG
MOVWF EEDATA ;move it to EEDATA reg
CALL EE_WRT
INCF EEADR, F ;point to next location
MOVLW A'E' ;load the byte into WREG
MOVWF EEDATA ;move it to EEDATA reg
CALL EE_WRT
INCF EEADR, F ;point to next location
MOVLW A'L' ;load the byte into WREG
MOVWF EEDATA ;move it to EEDATA reg
CALL EE_WRT

CHAPTER 14: USING FLASH AND EEPROM MEMORIES
INCF EEADR,F ;point to next location
MOVLW A'L' ;load the byte into WREG
MOVWF EEDATA ;move it to EEDATA reg
CALL EE_WRT
INCF EEADR,F ;point to next location
MOVLW A'O' ;load the byte into WREG
MOVWF EEDATA ;move it to EEDATA reg
CALL EE_WRT
INCF EEADR,F ;point to next location

;read EEPROM one byte at a time and send it to PORTB
MOVLW 0x0 ;starts at location 0H of EEPROM
MOVWF EEADR ;load the EEPROM address
BCF EECON1,EEPGD ;point to EEPROM memory
BCF EECON1,CFGS ;
MOVLW 0x05 ;count = 5
MOVWF COUNT
CLRF TRISB ;make PORTB output port
OVER BSF EECON1,RD ;enable read
NOP
MOVF EEDATA,PORTB ;read the data to PORTB
CALL DELAY ;wait
INCF EEADR,F ;point to next location
DECF COUNT,F ;decrement counter
BNZ OVER ;keep repeating
HERE BRA HERE

EE_WRT

BCF EECON1,EEPGD ;point to EEPROM memory
BCF EECON1,CFGS ;
BSF EECON1,WREN ;enable write
BCF INTCON,GIE ;disable all interrupts
MOVLW 0x55 ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0xAA ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1,WR ;now write it to Flash
BSF INTCON,GIE ;enable all interrupts

EE_WAIT BTFSS PIR2,EEIF
BRA EE_WAIT
BCF PIR2,EEIF
RETURN

558
Program 14-10 (a) moves a block of data from the code space of the PIC18 chip into EEPROM, and (b) then reads the same data from EEPROM and sends it to the serial port of the PIC18, one byte at a time.

```c
#include pl8f458.inc

COUNT EQU 0x0B
BUFRAM EQU 0x20

MOVLW D'8' ;number of bytes to retrieve
MOVWF COUNT
MOVLW 0H ;starts at location 0H of EEPROM
MOVWF EEDATA ;load the EEPROM address
MOVLW upper (CODE_DATA) ;load TBLPTR
MOVWF TBLPTRU
MOVLF high (CODE_DATA)
MOVWF TBLPTRH
MOVLF low (CODE_DATA)
MOVWF TBLPTRL
;retrieve the data from program memory
READ_BLOCK
    TBLRD*+ ;read into TABLAT, and increment
    MOVF TABLAT, W ;get data
    MOVWF EEDATA ;load data
    CALL EE_WRT ;save data
    INCF EEADR, F ;point to next location
    DECFSZ COUNT ;done?
    BRA READ_BLOCK ;repeat
;read them back one byte at a time and send to serial
;port
    BSF TRISD, 7 ;PORTD.7 as in input
    MOVLW 0x20 ;enable transmit and low baud rate
    MOVWF TXSTA ;write to reg
    BCF PIR1, TXIF
    MOVLW D'15' ;9600 bps (Fosc / (64 * Speed) - 1)
    MOVWF SPBRG ;write to reg
    BCF TRISC, TX ;make TX pin of PORTC an output
    BSF RCSTA, SPEN ;enable the entire serial port
    MOVLW 8 ;number of bytes in RAM
    MOVWF COUNT
    MOVLW 0x0 ;start at location 0H of EEPROM
    MOVWF EEADR ;load the EEPROM address
    CLR TRISB ;make PORTB an output port
    MOVLW 0x8 ;counter = 8
    MOVWF COUNT
```

CHAPTER 14: USING FLASH AND EEPROM MEMORIES
LN CALL EE_RD ;read the character
CALL SENDCOM ;send character to serial port
INCF EEADR,F
DECF COUNT ;loop until buffers are full
BRA LN ;repeat
HERE BRA HERE
;----------
SENDCOM
S1 BTFSS PIR1, TXIF ;wait until the last bit is gone
BRA S1 ;stay in loop
MOVWF TXREG ;load the value to be transferred
RETURN ;return to caller
;----------
 EE_WRT
BCF EECON1,EEPGD ;point to EEPROM memory
BCF EECON1,CFGS ;
BSF EECON1,WREN ;enable write
BCF INTCON,GIE ;disable all interrupts
MOVLW 0x55 ;wreg = 55h
MOVWF EECON2 ;write to dummy reg
MOVLW 0xAA ;wreg = aah
MOVWF EECON2 ;write to dummy reg
BSF EECON1,WR ;now write it to Flash
BSF INTCON,GIE ;enable all interrupts
EE_WAIT BTFSS PIR2,EEIF
BRA EE_WAIT
BCF PIR2,EEIF
RETURN
;----------
 EE_RD
BCF EECON1, EE PGD ;point to DATA memory
BCF EECON1, CFGS
BSF EECON1, RD ;EEPROM read
MOVF EEDATA, W ;W = EEDATA
RETURN
;----------
ORG 0x0300
CODE_DATA
DATA "MOVE ME"
END
Accessing the EEPROM in C

Program 14-11C shows how to write and read the EEPROM memory in C language. This is the C version of an earlier program.

Program 14-11C (a) writes the message "YES" to EEPROM memory, and (b) then reads the same data from EEPROM and sends it to PORTB one byte at a time.

//Program 14-11C
#include <pl8F458.h>
void EE_WRT(void);
unsigned char EE_READ(void);
void Delay(unsigned int itime);
void main()
{
    unsigned char x;
    TRISB=0;       //make PORTB output
    //write to EEPROM
    EEADR=0x00;    //EEPROM location
    EEDATA='Y';    //write this char to it
    EE_WRT();
    EEADR=0x10;
    EEDATA='E';
    EE_WRT();
    EEADR=0x20;
    EEDATA='S';
    EE_WRT();
    EECON1bits.WREN=0; //disable write
    //read from EEPROM and place it on PORTB
    EECON1bits.RD=1;  //enable read
    EEADR =0x00;     //EEPROM location
    x=EE_READ();     //read data from EEPROM
    PORTB=x;        //place it on PORTB
    Delay(250);
    EEADR =0x10;    //EEPROM location
    x=EE_READ();    //place it on PORTB
    PORTB=x;
    Delay(250);
    EEADR =0x20;    //EEPROM location
    x=EE_READ();    //place it on PORTB
    PORTB=x;
    while(1);
}
void EE_WRT()
{
    EECON1bits.EEPGD=0; //point to EEPROM
    EECON1bits.CFGS=0;
    EECON1bits.WREN=1;  //enable write
    INTCONbits.GIE=0;  //disable interrupts
    EECON2=0x55;
    EECON2=0xAA;
}
Program 14-12C (a) transfers a block of data from Flash to RAM, (b) writes the block to EEPROM memory, and (b) then reads the same data from EEPROM and sends it to the serial port one byte at a time.

```c
#include <pl8f458.h>
void EE_WRT(void);
unsigned char EE_READ(void);
void SerTx(unsigned char);

void main()
{
    char *RomPointer="MOVE ME";
    char RamString[7];
    unsigned char x, ch, k =sizeof(RomPointer);
    TXSTA=0x20;  //choose low baud rate, 8-bit
    SPBRG=15;    //9600 baud rate, XTAL = 10 MHz
    TXSTAbits.TXEN=1;
    RCSTAbits.SPEN=1;

    //move the string to RAM
    for(x=0;x<7;x++)
    {
        RamString[x]=RomPointer[x];
    }

    //move the string to EEPROM
    for(x=0;x<7;x++)
    {
        EEADR=x;
        EEDATA=RamString[x];
        EE_WRT();
    }
    EECON1bits.WREN=0; //disable write

    //read from EEPROM and send serially
    for(x=0;x<7;x++)
    {
```
void EE_WRT()
{
    EECON1bits.EEPGD=0;  //point to EEPROM
    EECON1bits.CFGS=0;
    EECON1bits.WREN=1;   //enable write
    INTCONbits.GIE=0;   //disable interrupts
    EECON2=0x55;
    EECON2=0xAA;
    EECON1bits.WR=1;
    INTCONbits.GIE=1;
    while(!PIR2bits.EEIF);
    PIR2bits.EEIF=0;
}

unsigned char EE_READ()
{
    EECON1bits.EEPGD=0;  //point to EEPROM
    EECON1bits.CFGS=0;
    EECON1bits.RD=1;
    return(EEDATA);
}

void SerTx(unsigned char c)
{
    while(PIR1bits.TXIF==0);  //wait until transmitted
    TXREG=c;                  //place character in buffer
}

Review Questions

1. True or false. The PIC18 EEPROM memory is used for both program code and data.
2. True or false. The PIC18F4580 has 1,024 bytes of EEPROM memory.
3. True or false. In the PIC18, EEPROM contents are lost when power is cut off to the chip.
4. True or false. In the PIC18, EEPROM memory is read and write memory.
5. True or false. Every PIC18F chip comes with 1 KB of EEPROM.
6. What is the advantage of the EEPROM over Flash?
SUMMARY

This chapter described memory interfacing with 8031/51-based systems. We began with an overview of semiconductor memories. Types of memories were compared in terms of their capacity, organization, and access time.

ROM (read-only memory) is nonvolatile memory typically used to store programs. The relative advantages of various types of ROM were described in this chapter, including PROM, EPROM, UV-EPROM, EEPROM, flash memory EPROM, and mask ROM.

RAM (random-access memory) is typically used to store data or programs. The relative advantages of its various types, including SRAM, NV-RAM, checksum byte RAM, and DRAM, were discussed.

The Flash memory space of the PIC18 was discussed, and programs were written in both Assembly and C to access it. Finally, the EEPROM memory of the PIC18 chip was explored and we showed how to access it in both Assembly and C.

PROBLEMS

SECTION 14.1: SEMICONDUCTOR MEMORY

1. What is the difference in capacity between a 4M memory chip and 4M of computer memory?
2. True or false. The more address pins, the more memory locations are inside the chip. (Assume that the number of data pins is fixed.)
3. True or false. The more data pins, the more each location inside the chip will hold.
4. True or false. The more data pins, the higher the capacity of the memory chip.
5. True or false. The more data pins and address pins, the greater the capacity of the memory chip.
6. The speed of a memory chip is referred to as its ________________.
7. True or false. The price of memory chips varies according to capacity and speed.
8. The main advantage of EEPROM over UV-EPROM is ________________.
9. True or false. SRAM has a larger cell size than DRAM.
10. Which of the following, EPROM, DRAM, or SRAM, must be refreshed periodically?
11. Which memory is used for PC cache?
12. Which of the following, SRAM, UV-EPROM, NV-RAM, or DRAM, is volatile memory?
13. RAS and CAS are associated with which memory?
   (a) EPROM  (b) SRAM  (c) DRAM  (d) all of the above
14. Which memory needs an external multiplexer?
   (a) EPROM  (b) SRAM  (c) DRAM  (d) all of the above
15. Find the organization and capacity of memory chips with the following pins:
   (a) EEPROM A0–A14, D0–D7  (b) UV-EPROM A0–A12, D0–D7
16. Find the capacity, address, and data pins for the following memory organizations:

(a) 16Kx8 ROM
(b) 32Kx8 ROM
(c) 64Kx8 SRAM
(d) 256Kx8 EEPROM
(e) 64Kx8 ROM
(f) 64Kx4 DRAM
(g) 1Mx8 SRAM
(h) 4Mx4 DRAM
(i) 64Kx8 NV-RAM

SECTION 14.2: ERASING AND WRITING TO FLASH IN THE PIC18F

17. True or false. The Flash memory in PIC18F is used primarily for the program code.
18. True or false. The Flash memory in PIC18F can be also used for storing fixed data.
19. True or false. The maximum memory space for program memory in PIC18F is 2M bytes.
20. True or false. Reading data from Flash memory can be done one byte at a time.
21. True or false. Writing data to Flash memory can be done one byte at a time.
22. True or false. Writing data to Flash memory must be done in blocks of 64 bytes.
23. True or false. Erasing of Flash memory can be done one byte at a time.
24. True or false. The use of the EECON2 register in writing/erasing of Flash memory must be optional.
25. What registers are used in reading the fixed data stored in Flash memory?
26. What registers are used in writing fixed data to Flash memory?
27. What registers are used in erasing the Flash memory?
28. What is the difference between the WREN and WR bits in the EECON1 register?
29. What registers are used by the TBLRD instruction?
30. What registers are used by the TBLWRT instruction?
31. Explain the difference between the short write and the long write in the PIC18.
32. During which write is the fetching of the opcode suspended by the CPU?
33. What is the size of the block of data for writing to Flash memory in PIC184580?
34. What is the size of the block of data for erasing the Flash memory in PIC184580?
35. Indicate all the addresses that have an 8-byte boundary:
   (a) 510H    (b) 512H    (c) 514H    (d) 516H    (e) 518H
   (f) 51AH    (g) 51CH    (h) 51EH
36. Indicate all the addresses that have a 64-byte boundary:
   (a) 500H    (b) 520H    (c) 540H    (d) 560H    (e) 580H
   (f) 5A0H    (g) 5C0H
37. Give the boundary addresses for the address range of 2000–2020H that can be
used in writing to Flash.
38. Give the boundary addresses for the address range of 2000–2100H that can be used in erasing of Flash memory.
39. Write a program to erase a section of Flash and then write the message “Hello World” to it.
40. For Problem 39, write a program to verify the write operation by reading it and sending it to the serial port one byte at a time.

SECTION 14.3: READING AND WRITING TO DATA EEPROM IN THE PIC18F

41. True or false. The EEPROM memory in the PIC18F is used primarily for the program code.
42. True or false. The EEPROM memory in the PIC18F is used for data only.
43. True or false. Every PIC18F member has at least 256 bytes of EEPROM memory.
44. True or false. Reading data from EEPROM memory can be done one byte at a time.
45. True or false. Writing data to EEPROM memory can be done one byte at a time.
46. True or false. Writing data to EEPROM memory must be done in blocks of 64 bytes.
47. True or false. Erasing of data in EEPROM memory can be done one byte at a time.
48. True or false. The use of the EECON2 register in reading and writing of EEPROM memory is optional.
49. True or false. The EECON2 register is used by both the Flash and EEPROM memory write operation.
50. What registers are used in reading data from EEPROM memory?
51. What registers are used in writing data to EEPROM memory?
52. Give the major differences between Flash and EEPROM in the PIC18.
53. What is the size of the block of data for writing to EEPROM memory in the PIC18?
54. Which bits of the EECON1 are used by the read operation of the EEPROM?
55. Why do we disable the interrupts during the write cycle of Flash/EEPROM memory?
56. Why don’t we disable the interrupts during the read cycle of Flash/EEPROM memory?
57. Write a program to write the message “Hello World” to EEPROM.
58. For Problem 57, write a program to verify the write operation by reading it and sending it to the serial port one byte at a time.

ANSWERS TO REVIEW QUESTIONS

SECTION 14.1: SEMICONDUCTOR MEMORY

1. c
2. (a) 16Kx8, 128K bits (b) 64Kx8, 512K (c) 4Kx8, 32K
### Section 14.2: Erasing and Writing to Flash in the PIC18F

1. True  
2. False  
3. False  
4. True  
5. False  
6. 8 bytes  
7. 64 bytes  

### Section 14.3: Reading and Writing to Data EEPROM in the PIC18

1. False  
2. False  
3. False  
4. True  
5. False  
6. In EEPROM we can write a single byte of data, while in Flash, we must write a block of data.
CHAPTER 15

CCP AND ECCP PROGRAMMING

OBJECTIVES

Upon completion of this chapter, you will be able to:

»» Understand the compare and capture features of the PIC18
»» Examine the use of timers in CCP and ECCP modules
»» Explain how the compare feature of CCP and ECCP modules works
»» Explain how the capture feature of CCP and ECCP modules works
»» Code programs for compare and capture features in Assembly and C
»» Explain how the PWM (pulse width modulation) works in both CCP and ECCP
»» Code programs to create PWM in Assembly and C
This chapter discusses the capture/compare/pulse width modulation (CCP) features of the PIC18. In Section 15.1, we show the difference between standard and enhanced CCP modules. In Section 15.2, we describe the compare feature while Section 15.3 deals with the capture feature of the PIC18. The pulse width modulation (PWM) of the PIC18 is shown in Section 15.4. An overview of ECCP is given in Section 15.5. In all these sections we use both Assembly and C language programs to show these important features of the PIC18.

SECTION 15.1: STANDARD AND ENHANCED CCP MODULES

Depending on the family member, the PIC18 has anywhere between 0 and 5 CCP modules inside it. The multiple CCP modules are designated as CCP1, CCP2, CCP3, and so on (CCPx). In recent years, the PWM feature of the CCP has been enhanced greatly for better DC motor control, producing what is called enhanced CCP (ECCP). Therefore, a given family member can have two standard CCP modules and one or more ECCP modules, all on a single chip. See Table 15-1. The ECCP modules are discussed in Chapter 17.

### CCP and timers

To program these CCP modules, we must understand PIC18 timers. Review timers in Chapter 9 before you embark on this chapter. Depending on the CCP feature used, the timer usage is different. The allocation of the timers among the CCP features is shown in Table 15-2.

#### Table 15-1: PIC18 CCP and ECCP Modules

<table>
<thead>
<tr>
<th>Chip</th>
<th># of CCP</th>
<th># of ECCP</th>
</tr>
</thead>
<tbody>
<tr>
<td>PIC18F2220</td>
<td>2</td>
<td>0</td>
</tr>
<tr>
<td>PIC18F4220</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PIC18F452/4520</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PIC18F458/4580</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>PIC18F65J10</td>
<td>2</td>
<td>3</td>
</tr>
</tbody>
</table>

#### Table 15-2: PIC18 Usage of Timers

<table>
<thead>
<tr>
<th>CCP mode</th>
<th>Timer</th>
</tr>
</thead>
<tbody>
<tr>
<td>Capture</td>
<td>Timer1 or Timer3</td>
</tr>
<tr>
<td>Compare</td>
<td>Timer1 or Timer3</td>
</tr>
<tr>
<td>PWM</td>
<td>Timer2</td>
</tr>
</tbody>
</table>

*Note: The T3CON register is used to choose the timer for the compare and capture modes.*

### The CCP registers

Each CCP module has three registers associated with it. They are as follows:

(a) CCPxCON is an 8-bit control register. We select one of the compare, capture, and PWM modes using this register. See Figure 15-1.
(b) and (c) CCPRxL and CCPxH form the low byte and the high byte of the 16-bit register. This 16-bit register can be used either as a 16-bit compare register, or a 16-bit capture register, or an 8-bit duty cycle register by the PWM, but not all at the same time. See Figures 15-1 and 5-2. The CCP1CON register selects the mode of operation.

<table>
<thead>
<tr>
<th>DC1B1</th>
<th>DC1B2</th>
<th>CCP1M3</th>
<th>CCP1M2</th>
<th>CCP1M1</th>
<th>CCP1M0</th>
</tr>
</thead>
<tbody>
<tr>
<td>DC1B1 Duty Cycle Bit 1. Used only in PWM mode.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Bit 1 of the 10-bit duty cycle register used in PWM</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DC1B0 Duty Cycle Bit 0. Used only in PWM mode.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>The least-significant bit (bit 0) of the 10-bit duty cycle register. Used in PWM.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>The CCPxL register is used as bit 2 to bit 9 of the 10-bit duty cycle register.</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

CCP1M3–CC1M0 CCP1 Mode Select

| 0000 | CCP1 is off |
| 0001 | Reserved |
| 0010 | Compare mode. Toggle CCP1 output pin on match. (CCP1IF bit is set.) |
| 0011 | Reserved |
| 0100 | Capture mode, every falling edge |
| 0101 | Capture mode, every rising edge |
| 0110 | Capture mode, every 4th rising edge |
| 0111 | Capture mode, every 16th rising edge |
| 1000 | Compare mode. Initialize CCP1 pin LOW, on compare match force CCP1 pin HIGH. (CCP1IF is set.) |
| 1001 | Compare mode. Initialize CCP1 pin HIGH, on compare match force CCP1 pin LOW. (CCP1IF is set.) |
| 1010 | Compare mode. Generate software interrupt on compare match. (CCP1IF bit is set, CCP1 pin is unaffected.) |
| 1011 | Compare mode. Trigger special event. (CCP1IF bit is set, and Timer 1 or Timer 3 is reset to zero.) |
| 11xx | PWM mode |

Figure 15-1. CCP1 Control Register. (This register selects one of the operation modes of Capture, Compare, or PWM.)

<table>
<thead>
<tr>
<th>CCPR1H</th>
<th>CCPR1L</th>
</tr>
</thead>
<tbody>
<tr>
<td>D15</td>
<td>D14</td>
</tr>
<tr>
<td>D13</td>
<td>D12</td>
</tr>
<tr>
<td>D11</td>
<td>D10</td>
</tr>
<tr>
<td>D9</td>
<td>D8</td>
</tr>
<tr>
<td>D7</td>
<td>D6</td>
</tr>
<tr>
<td>D5</td>
<td>D4</td>
</tr>
<tr>
<td>D3</td>
<td>D2</td>
</tr>
<tr>
<td>D1</td>
<td>D0</td>
</tr>
</tbody>
</table>

Figure 15-2. CCP High and Low Registers

CHAPTER 15: CCP AND ECCP PROGRAMMING 571
CCP pins

Each CCP module has a single pin assigned to it. That means that a PIC18 family member with two standard CCP modules (e.g., PIC18F65J10) has two pins, one assigned to each of the CCPs. See Figure 15-3. In the case of the enhanced CCP (ECCP), although it has a single pin, we can program up to four pins to be used by the PWM feature of the ECCP, as we will see at the end of this chapter and in Chapter 17.

![Diagram of CCP pins]

Figure 15-3. Standard CCP Pins in PIC18F458/4580 (452/4520)

Review Questions

1. True or false. The PIC18 chip can have multiple CCP modules inside a single chip.
2. True or false. The CCP1 register is a 16-bit register.
3. True or false. A single pin is associated with each of the standard CCP modules.
4. Give the pin number used for the standard CCP1 in the PIC18F452/458 (or PIC18F4520/4580) chip.
SECTION 15.2: COMPARE MODE PROGRAMMING

The Compare mode of the CCP module is selected using the select bits in the CCPxCON register. The Compare mode can cause an event outside the microcontroller. This event can be simply turning on a device connected to the CCP pin, or the start of an ADC conversion. This event is caused when the content of the Timer1 (or Timer3) register is equal to the 16-bit CCPR1H:CCPR1L register. To use the compare mode of the CCP, we must load both the 16-bit (CCPR1H:CCPR1L) and the Timer1 (or Timer3) register with some initial values. As Timer1 (or Timer3) counts up, its value is constantly compared with the CCPR1H:CCPR1L register and when a match occurs, the CCP1 pin can perform one of the following actions:

(a) Drive high the CCP1 pin
(b) Drive low the CCP1 pin
(c) Toggle the CCP1 pin
(d) Remain unaffected
(e) Trigger a special event with a hardware interrupt and clear the timer

We use the CCP1CON register to select one of the above actions. See Example 15-1. Note that upon match, the CCP1IF will also go HIGH. See Figure 15-5. Notice that for the above options of (a), (b), and (c) to work, the CCP pin must be configured as an output pin. From Figure 15-4 we use the T3CON register to select Timer1 or Timer3 for the Compare mode. In PIC18F452/458 (or their newer version, PIC18F4520/4580) chips with both CCP1 and ECCP1 modules on the chip, we can assign Timer1 to CCP1 and Timer3 to ECCP1, therefore making them work independently of each other. Also note that only option (e), the special event trigger, will cause Timer1 (or Timer3) to clear, while in other cases we must clear the timer.

Example 15-1

Using Figures 15-1 and 15-4, find the following:
(a) The CCP1CON register value for Compare mode if we want to toggle the CCP1 pin upon match
(b) The T3CON register value if we want to to use Timer3 for the Compare mode of CCP1 with no prescaler

Solution:

(a) From Figure 15-1 we have 00000010 (binary) or 0x20 for the CCP1CON register.
(b) From Figure 15-4 we have 01000010 (binary) or 0x42 for the T3CON register.

There are many applications for the compare feature. One application can be to count the number of people going through a door and closing the door when it reaches a certain number.
RD16 | T3CCP2 | T3CKPS1 | T3CKPS0 | T3CCP1 | T3SYNC | TMR3CS | TMR3ON
 RD16 |       D7       | 16-bit read/write enable bit  
 1 = Timer3 16-bit is accessible in one 16-bit operation.  
 0 = Timer3 16-bit is accessible in two 8-bit operations.  

T3CCP2:T3CCP1 | D6 D3 assigns Timer3 or Timer1 to CCP1 and CCP2 modules  
CCP1 | ECCP1 (or CCP2)  
0 0 = Timer1  
0 1 = Timer1  
1 x = Timer3  
    Timer3 (clock source for compare/capture)  
    Timer3 (clock source for compare/capture)  
    Timer3 (clock source for compare/capture)  

T3CKPS1:T3CKPS0 | D5 D4 Timer3 Input Clock Prescaler Selector  
0 0 = 1:1  
0 1 = 1:2  
1 0 = 1:4  
1 1 = 1:8  
    Prescale value  
    Prescale value  
    Prescale value  
    Prescale value  

T1SYNC | D2 Timer3 External Clock Input Synchronization Control bit  
Used only when TMR3CS = 1 and clock comes from an  
external source. If TMR3CS = 0, this bit is not used.  
1 = Do not synchronize external clock input.  
0 = Synchronize external clock input.  

TMR3CS | D1 Timer 3 Clock Source Select bit  
1 = External clock from pin RC0 (T1OSI or T1CKI)  
0 = Internal clock (Fosc/4)  

TMR3ON | D0 Timer3 ON and OFF Control bit  
1 = Enable (start) Timer3  
0 = Stop Timer3  

Figure 15-4. T3CON (Timer 3 Control) Register

CCP1IF | CCP1IF Interrupt Flag bit  
Compare Mode  
0 = Timer1 (or Timer3) match did not occur.  
1 = Timer1 (or Timer3) match occurred (must be cleared by software).  
Capture Mode  
0 = Timer1 (or Timer3) register capture did not occur.  
1 = Timer1 (or Timer3) register capture occurred (must be cleared by software).  

Figure 15-5. PIR1 (Peripheral Interrupt flag register 1) Contains the  
CCP1IF Flag
Steps for programming the Compare mode

The following steps are taken in programming the Compare mode:

1. Initialize the CCP1CON register for the compare option.
2. Initialize the T3CON register for Timer1 (or Timer3).
3. Initialize the CCPR1H:CCPR1L registers.
4. Make the CCP1 pin an output pin.
5. Initialize Timer1 (or Timer3) register values.
6. Start Timer1 (or Timer3).
7. Monitor the CCP1IF flag (or use an interrupt).

Program 15-1 shows an example of the Compare mode. It uses Timer3 as a counter and counts the number of pulses fed to Timer3. The pulses could be the number of people going into an elevator. When the count reaches 10, it toggles the LED connected to the CCP1 pin.

For Program 15-1 assume that a 1-Hz pulse is connected to the Timer3 pin and an LED is connected to the CCP1 pin. Timer3 is being used as a counter. Using the Compare mode, this Assembly language program will toggle the LED every 10 pulses.
; Program 15-1
MOVLW 0x02
MOVWF CCP1CON; Compare mode, toggle upon match
MOVLW 0x42
MOVWF T3CON; Timer3 for Compare, 1:1 prescaler
BCF TRISC, CCP1; CCP1 pin as output
BSF TRISC, T1CKI; T3CLK pin as input pin
MOVLW D'10';
MOVWF CCP1L; CCPR1L = 10
MOVLW 0x0; CCPR1H = 0
MOVF CCP1H
OVER CLRF TMR3H; clear TMR3H
CLRF TMR3L; clear TMR3L
BCF PIR1, CCP1IF; clear CCP1IF
BSF T3CON, TMR3ON; start Timer3
B1 BTFSS PIR1, CCP1IF
BRA B1
;------- CCP toggle CCP pin upon match
B2 BCF T3CON, TMR3ON; stop Timer3
GOTO OVER; keep doing it

// Program 15-1C is a C version of Program 15-1
CCP1CON=0x02; // Compare mode, toggle upon match
T3CON=0x42; // Timer3 for Compare, 1:1 prescaler
TRISCbits.TRISC2=0; // CCP1 pin an output
TRISCbits.TRISC0=1; // T3CLK pin an input
CCPR1L=10; // load CCPR1L
CCPR1H=0; // load CCPR1H
while(1)
{
    TMR3H=0;
    TMR3L=0;
    PIR1bits.CCP1IF=0; // clear CCP1IF flag
    T3CONbits.TMR3ON=1; // turn on Timer3
    while(PIR1bits.CCP1IF==0); // wait for CCP1IF
    // CCP toggles CCP pin upon match
    T3CONbits.TMR3ON=0; // stop Timer3
}

Examine Program 15-2: For this program we assume that the
PIC18452/458 has Fosc = 10 MHz. It programs the CCP1 module in Compare
mode to create a square wave with a period of 40 ms on the CCP1 pin continuously. The square wave has a 50% duty cycle, which means it is high 50% of each period. This is an example of how Timer1 is used in compare mode. See Figure
15-8. Because the timer uses the Fosc/4, we have 1/2.5 MHz = 0.4 μs for the clock. A 40 ms period gives us 20 ms for high and low portions of the square wave. Now 20 ms / 0.4 μs = 50,000 or C350 in hex. This is the value we load into CCPR1H:CCPR1L for the Compare mode.
<table>
<thead>
<tr>
<th>RD16</th>
<th>RD15--</th>
<th>T1CKPS0</th>
<th>T1CKPS1</th>
<th>TIOSCEN</th>
<th>T1SYNC</th>
<th>TMR1CS</th>
<th>TMR1ON</th>
</tr>
</thead>
</table>

RD16

- **D7**: 16-bit read/write enable bit
  - 1 = Timer1 16-bit is accessible in one 16-bit operation.
  - 0 = Timer1 16-bit is accessible in two 8-bit operations.

- **D6**: Not used

**T1CKPS2:T1CKPS0**

- **D5 D4**: Timer1 prescaler selector
  - 0 0 = 1:1 Prescale value
  - 0 1 = 1:2 Prescale value
  - 1 0 = 1:4 Prescale value
  - 1 1 = 1:8 Prescale value

**TIOSCEN**

- **D3**: Timer1 oscillator enable bit
  - 1 = Timer1 oscillator is enabled
  - 0 = Timer1 oscillator is shut off

**T1SYNC**

- **D2**: Timer1 synchronization (used only when TMR1CS = 1 for counter mode to synchronize external clock input)
  - If TMR1CS = 0, this bit is not used.

**TMR1CS**

- **D1**: Timer1 clock source select bit
  - 1 = External clock from pin RC0/T1CKI
  - 0 = Internal clock (Fosc/4 from XTAL)

**TMR1ON**

- **D0**: Timer1 ON and OFF control bit
  - 1 = Enable (start) Timer 1
  - 0 = Stop Timer 1

**Figure 15-8. T1CON (Timer 1 Control) Register**

Program 15-2 creates a square wave with a 40 ms period and 50% duty cycle on CCP1 pin using the Compare mode.

;Program 15-2
MOVLW 0x02
MOVWF CCP1CON ;Compare mode, toggle upon match
MOVLW 0x0
MOVWF T3CON ;use Timer1 for Compare mode
MOVLF 0x00
MOVWF T1CON ;Timer1, internal CLK, 1:1 prescale
BCF TRISC,CCP1 ;CCP1 pin as output
MOVLW 0xC3
MOVWF CCPR1H ;CCPR1H = 0xC3
MOVLW 0x50

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MOVWF CCP1L ;CCPR1L = 0x50
OVER CLRF TMR1H ;clear TMR1H
CLRF TMR1L ;clear TMR1L
BCF PIR1,CCP1IF ;clear CCP1IF
BSF T1CON,TMR1ON ;start Timer1
B1 BTFSS PIR1,CCP1IF
BRA B1 ;CCP toggles CCP1 pin upon match
BCF T1CON,TMR1ON ;stop Timer1
GOTO OVER ;keep doing it

//Program 15-2C is the C version of Program 15-2.
CCP1CON=0x02; //Compare mode, toggle upon match
T3CON=0x0; //Timer1 for Compare, 1:1 prescaler
T1CON=0x0; //Timer1 internal clk, 1:1 prescaler
TRISCbits.TRISC2=0; //make CCP1 pin an output
TRISCbits.TRISC0=1; //make T1CLK pin an input
CCPR1H=0xC3; //load CCPR1H
CCPR1L=0x50; //load CCPR1L
while(1)
{
    TMR1H=0; //clear Timer1
    TMR1L=0;
    PIR1bits.CCP1IF=0; //clear CCP1IF flag
    T1CONbits.TMR1ON=1; //turn on Timer1
    while(PIR1bits.CCP1IF==0); //wait for CCP1IF
    //CCP toggles CCP1 pin upon match
    T1CONbits.TMR1ON=0; //stop Timer1
}

Review Questions
1. True or false. We can use any timers we want for the Compare mode.
2. True or false. There is a single pin associated with the Compare mode.
3. True or false. In Compare mode, the CCP pin must be configured as an input pin.
4. Which register is used to choose the timer for the Compare mode?
SECTION 15.3: CAPTURE MODE PROGRAMMING

We select Capture mode with the bit selection in the CCP1CON register. In Capture mode, an event at the CCP pin will cause the contents of the Timer1 (or Timer3) register to be loaded into the 16-bit CCPR1H:CCPR1L register. That means, for the Capture mode to work, the CCP pin must be configured as an input pin. The event that causes the contents of Timer1 (or Timer 3) to be captured into the CCPR1H:CCPR1L register can be a High-to-Low (falling-edge) pulse or Low-to-High (rising-edge) pulse. As far as the edge-triggering pulse is concerned, we have the following four options to choose from:

(a) every falling-edge pulse
(b) every rising-edge pulse
(c) every fourth rising-edge pulse
(d) every 16th rising-edge pulse

One of the above options can be chosen by selection bits in the CCP1CON register. See Example 15-2. Notice that for any of the above options to work, the CCP pin must be configured as an input pin.

Example 15-2

Using Figures 15-1 and 15-4, find the following:
(a) The CCP1CON register value for Capture mode if we want to capture on the rising edge of every pulse
(b) The T3CON register value if we want to use Timer3 for Capture mode of the CCP1 with no prescaler

Solution:

(a) From Figure 15-1, we have 00000101 (binary) or 0x05 for the CCP1CON register.
(b) From Figure 15-4, we have 01000010 (binary) or 0x42 for the T3CON register.

One application of Capture mode is measuring the frequency of an incoming pulse. See Program 15-3.

Steps for programming Capture mode

The following steps are used in programming Capture mode for measuring the period of a pulse:

1. Initialize the CCP1CON register for capture.
2. Make the CCP1 pin an input pin.
3. Initialize the T3CON register to select Timer1 or Timer3.
4. Read the Timer1 (or Timer3) register value on the first rising edge and save it.
5. Read the Timer1 (or Timer3) register value on the second rising edge and save it.
6. Subtract the value in step 4 from the value in step 5.
Measuring the period of a pulse

Program 15-3 shows an example of capture mode. See Figure 15-9. It measures the period of the pulse fed to the CCP pin. The measurement is in terms of the number of clock cycles, Tsclk \((1/(\text{Fosc}/4))\). See Figure 15-10.

For Program 15-3 assume a pulse is being fed to the CCP1 pin. Using Capture mode, this Assembly language program measures the period of the pulse and puts the results on PORTB and PORTD. The measurement is in terms of the Fosc/4 clock period. See Figure 15-11.
; Program 15-3

MOVLW 0x05
MOVWF CCP1CON ; Capture mode rising edge
MOVLW 0x0
MOVWF T3CON ; Timer 1 for Capture
MOVLW 0x0
MOVWF T1CON ; Timer 1, internal CLK, 1:1 prescaler
CLRF TRISB ; make PORTB output port
CLRF TRISD ; make PORTD output port
BSF TRISC, CCP1 ; make CCP1 pin an input
MOVLW 0x0
MOVWF T1CON
MOVLW 0x0
MOVWF CCPR1H ; CCPR1H = 0
MOVLW 0x0
MOVWF CCPR1L ; CCPR1L = 0

OVER
CLRF TMR1H ; clear TMR1H
CLRF TMR1L ; clear TMR1L
BCF PIR1, CCP1IF ; clear CCP1IF

RE_1
BTFSS PIR1, CCP1IF
BRA RE_1 ; stay here for 1st rising edge
BSF T1CON, TMR1ON ; start Timer 1
BCF PIR1, CCP1IF ; clear CCP1IF for next edge

RE_2
BTFSS PIR1, CCP1IF
BRA RE_2 ; stay here for 2nd rising edge
BCF T1CON, TMR1ON ; stop Timer 1
MOVFF TMR1L, PORTB ; put low byte on PORTB
MOVFF TMR1H, PORTD ; put high byte on PORTD
GOTO OVER ; keep doing it

// Program 15-3C is the C version of Program 15-3.
CCP1CON=0x05; // Capture mode on every rising edge
T3CON=0x0; // Timer 1 for capture
T1CON=0x0; // Timer 1 internal clk, 1:1 prescaler
TRISB=0; // make PORTB output port
TRISD=0; // make PORTD output port
TRISCbits.TRISC2=1; // make CCP1 pin an input
CCPR1L=0; // CCPR1L = 0
CCPR1H=0; // CCPR1H = 0

while(1)
{
TMR1H=0; // clear Timer 1
TMR1L=0;
PIR1bits.CCP1IF=0; // clear CCP1IF flag
while(PIR1bits.CCP1IF==0); // wait for 1st rising edge
T1CONbits.TMR1ON=1; // start Timer 1
PIR1bits.CCP1IF=0; // clear CCP1IF for next edge
while(PIR1bits.CCP1IF==0); // wait for 2nd rising edge
T1CONbits.TMR1ON=0; // stop Timer 1
PORTB=CCPR1L;
PORTD=CCPR1H; // display the clock count
}

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One problem in measuring the period in the above program is the rate of error due to overhead associated with the program. One way to reduce the effect of the overhead is to use every fourth or every sixteenth rising edge.

**Measuring pulse width**

One of the most widely used applications of Capture mode is measuring the pulse width. A large number of devices measure things such as distance, temperature, and so on, in which the quantity is provided in terms of the pulse width instead of traditional voltage or current. In these devices, the output is provided in pulse-width-modulated (PWM) form. In a device with PWM output, the output has a fixed frequency and the variable duty cycle provides the quantity we are measuring. For example, the MAX6666/6667 temperature sensors from Maxim Corp. "convert the ambient temperature into a ratiometric PWM output with temperature information contained in the duty cycle of the output square wave." According to their data sheets the output is a square wave with a nominal frequency of 35 Hz at +25°C. The output format is decoded as follows:

Temperature (°C) = 235 - (400 × t1) / t2  

(Equation 15-1)

where t1 is fixed with a typical value of 10 ms and t2 is modulated by the temperature. In the above formula, T = t1 + t2 where T is the period of the pulse, t1 is the high portion of the pulse, and t2 is the low portion, as shown in Figure 15-12. With t1 = 10 and t2 = 20 ms, we get temperature = 235 - (400 × 10 ms) / 20 ms = 235 - 200 = 35°C. Program 15-4 shows how to measure the duty cycle using Capture mode.

Figure 15-12. Duty Cycle for MAX6666/76 Temperature Sensor (Maxim Corp.)
; Program 15-4
FLAG EQU 0x10 ; flag register for steps in detection
DISP EQU 0x0 ; flag for capture complete
RF EQU 0x1 ; flag for rising or falling edge
ORG 0x0000
GOTO MAIN
ORG 0x0008
BTFSC PIR1, CCP1IF ; Is it CCP1?
GOTO CCP_ISR ; service CCP1
RETFIE ; else return
MAIN MOVLW 0x05
MOVWF CCP1CON ; Capture mode rising edge
MOVFL 0x0
MOVWF T3CON ; Timer1 for Capture
MOVFL 0x0
MOVWF T1CON ; Timer1, internal CLK, 1:1 prescale
CLRF TRISB ; make PORTB output port
CLRF TRISD ; make PORTD output port
BSF TRISC, CCP1 ; make CCP1 pin an input
CLRF CCPR1H ; CCPR1H = 0
CLRF CCPR1L ; CCPR1L = 0
BSF PIE1, CCP1IE ; enable CCP1 interrupt
BSF INTCON, PEIE ; enable peripheral interrupt
BSF INTCON, GIE ; enable all interrupts
OVER CLRF TMR1H ; clear TMR1H
CLRF TMR1L ; clear TMR1L
WAIT BTFSS FLAG, DISP ; Is capture complete?
BRA WAIT ; else wait
BCF FLAG, DISP ; clear flag for next capture
MOVFL 0x03
SUBF TMR1L, F ; subtract the overhead
MOVFF TMR1L, PORTB ; put low byte on PORTB
MOVFF TMR1H, PORTD ; put high byte on PORTD
GOTO OVER ; keep doing it
CCP_ISR BTFSS FLAG, RF ; Is it rising edge?
GOTO RISE_ISR ; service rising edge
GOTO FALL_ISR ; else service falling edge
RISE_ISR BSF T1CON, TMR1ON ; start Timer1
BSF FLAG, RF ; ready for falling edge
BCF CCP1CON, 0 ; detect falling edge
BCF P1R1, CCP1IF ; clear interrupt
RETFIE ; return and wait for falling edge
FALL_ISR BCF T1CON, TMR1ON ; stop Timer1
BSF FLAG, DISP ; capture complete
BCF FLAG, RF ; ready for rising edge
BSF CCP1CON, 0 ; detect rising edge
BCF P1R1, CCP1IF ; clear interrupt
RETFIE ; return capture complete
END
//Program 15-4C
#include "pl8f458.h"

void CCP1_ISR(void);
void rising(void);
void falling(void);
unsigned char disp = 0;
unsigned char rf = 0;

#pragma interrupt chk_isr
void chk_isr (void)
{
    if (PIR1bits.CCP1IF==1)
        CCP1_ISR();
}

#pragma code My_HiPrio_Int=0x0008
void My_HiPrio_Int (void)
{
    _asm
    GOTO chk_isr
    _endasm
}

#pragma code
void main()
{
    CCP1CON=0x05;    //Capture mode rising edge
    T3CON=0x0;       //Timer1 for Capture
    T1CON=0x0;       //Timer1, internal CLK, 1:1 prescale
    TRISB=0x0;       //make PORTB output port
    TRISD=0x0;       //make PORTD output port
    TRISCbits.TRISC2=1; //make CCP1 pin an input
    CCPR1H=0x0;      //CCPR1H = 0
    CCPR1L=0x0;      //CCPR1L = 0
    PIE1bits.CCP1IE=1; //enable CCP1 interrupt
    INTCONbits.PEIE=1; //enable peripheral interrupt
    INTCONbits.GIE=1; //enable all interrupts
    while(1)
    {
        TMR1H=0x0;     //clear TMR1H
        TMR1L=0x0;     //clear TMR1L
        while(disp==0); //Is data ready to display?
        disp=0;       //clear the flag
        TMR1L-=15;     //subtract the overhead
        PORTB=TMR1L;   //put low byte on PORTB
        PORTD=TMR1H;   //put high byte on PORTD
    }
}
void CCP1_ISR()
{
    if(rf==0) rising();
    else falling();
}

void rising()
{
    TICONbits.TMR1ON=1; //start Timer1
    rf=1;
    CCP1ICONbits.CCP1M0=0; //ready for falling edge
    PIR1bits.CCP1IF=0; //detect falling edge
}

void falling()
{
    TICONbits.TMR1ON=0; //stop Timer1
    disp=1; //capture complete
    rf=0; //ready for rising edge
    CCP1ICONbits.CCP1M0=1; //detect rising edge
    PIR1bits.CCP1IF=0; //clear interrupt
}

Notice that in the company’s web site for data sheets the output for a given device is identified as analog (voltage or current) or PWM.

Review Questions
1. True or false. In Capture mode, the CCP pin must be configured as an input pin.
2. True or false. We can use only Timers 1 and 3 for Capture mode.
3. True or false. The timer’s register values are transferred to CCPR1H:CCPR1L every time the CPU is reset.
4. True or false. After the timer’s register values are transferred to CCPR1H:CCPR1L, the timer’s registers are cleared.
5. Which register is used to choose the timer for Capture mode?
SECTION 15.4: PWM PROGRAMMING

Another feature of CCP is pulse width modulation (PWM). The PWM feature allows us to create pulses with variable widths. Although we can program timers to create PWM, the CCP module makes the programming of PWM much easier and less tedious. PWM is widely used in industrial controls such as DC motor controls, as we will see in Chapter 17. Indeed the PWM is so widely used that Microchip has enhanced the PWM capabilities of the newer generation of the PIC18 family members and has designated them as ECCP (enhanced CCP). We will study ECCP in the next section. The main difference between ECCP and standard CCP is the PWM capability. In creating pulses with variable widths for the PWM, two factors are important: The period of the pulse and its duty cycle. The duty cycle (DC) is the portion of the pulse that stays HIGH relative to the entire period. Very often the DC is stated in the form of percentages. For example, a pulse with a 4 ms period that stays HIGH for 1 ms has DC of 25% (1 ms / 4 ms = 25%), as shown in Figure 15-13.

![Figure 15-13. Period and Duty Cycle](image)

The period of PWM

The CCP module uses Timer2 and its associated register, PR2, for the PWM time-base, which means that the frequency of the PWM is a fraction of the Fosc, the crystal frequency. It uses the PR2 register to set the PWM period as follows:

\[ T_{\text{pwm}} = \left[ \frac{(\text{PR2}) + 1}{4} \times N \times T_{\text{osc}} \right] \]  

(Equ. 15-2)

where Tosc is the inverse of 1/Fosc, the crystal frequency, Tpwm is the desired PWM period, and N is the prescaler of 1, 4, or 16 set by the Timer2 control register (T2CON). Therefore, we can get the value for the PR2 register as follows:

\[ \text{PR2} = \left[ \frac{\text{Fosc}}{(F_{\text{pwm}} \times 4 \times N)} \right] - 1 \]  

(Equ. 15-3)

From Equation 15-2, we can conclude that the maximum value for Tpwm can be achieved when N = 16 and PR2 = 255. Therefore, we have:

\[ T_{\text{pwm}} = \left[ (255) + 1 \right] \times 4 \times 16 \times T_{\text{osc}} = 16,382 \times T_{\text{osc}} \]

which means that the minimum allowed Fpwm = Fosc/16,382.
Examine Examples 15-3 to 15-5 to see the calculation of the PWM period.

### Example 15-3

Find the PR2 value and the prescaler needed to get the following PWM frequencies. Assume XTAL = 20 MHz.
(a) 1.22 kHz, (b) 4.88 kHz, (c) 78.125 kHz

**Solution:**

(a) PR2 value \( = \left[\frac{20 \text{ MHz}}{4 \times 1.22 \text{ kHz}}\right] - 1 \) = 4,097, which is larger than 255, the maximum value allowed for the PR2. Now choosing the prescaler of 16 we get PR2 value \( = \left[\frac{20 \text{ MHz}}{4 \times 1.22 \text{ kHz} \times 16}\right] - 1 \) = 255

(b) PR2 value \( = \left[\frac{20 \text{ MHz}}{4 \times 4.88 \text{ kHz}}\right] - 1 \) = 1,023, which is larger than 255, the maximum value allowed for the PR2. Now choosing the prescaler of 4 we get PR2 value \( = \left[\frac{20 \text{ MHz}}{4 \times 4.88 \text{ kHz} \times 4}\right] - 1 \) = 255

(c) PR2 value \( = \left[\frac{20 \text{ MHz}}{4 \times 78.125 \text{ kHz}}\right] - 1 \) = 63

### Example 15-4

Find the PR2 value for the following PWM frequencies. Assume XTAL = 10 MHz and prescaler = 1.
(a) 10 kHz, (b) 25 kHz

**Solution:**

(b) PR2 value \( = \left[\frac{10 \text{ MHz}}{4 \times 10 \text{ kHz} \times 1}\right] - 1 \) = 250 - 1 = 249
(c) PR2 value \( = \left[\frac{10 \text{ MHz}}{4 \times 25 \text{ kHz} \times 1}\right] - 1 \) = 100 - 1 = 99

### Example 15-5

Find the minimum and maximum Fpwm frequency allowed for XTAL = 10 MHz. State the PR2 and prescaler values for the minimum and maximum Fpwm.

**Solution:**

We get the minimum Fpwm by making PR2 = 255 and prescaler = 16, which gives us \( \frac{10 \text{ MHz}}{4 \times 16 \times 256} \) = 610 Hz.

We get the maximum Fpwm by making PR2 = 1 and prescaler = 1, which gives us \( \frac{10 \text{ MHz}}{4 \times 1 \times 1} \) = 2.5 MHz.
The duty cycle of PWM

As stated earlier, the duty cycle of PWM is the portion of the pulse that stays HIGH relative to the entire period. To set the duty cycle, the CCP module uses the 10-bit register of DC1B9:DC1B0. The 10-bit register of DC1B9:DC1B0 is formed from 8 bits of CCPR1L and 2 bits from the CCP1CON register, where CCPR1L is the upper 8 bits and DC1B2:DC1B1 of the CCP1CON are the lower 2 bits of the 10-bit register. In reality, CCPRL1 is the main register for the duty cycle and the lower 2 bits of DC1B2:DC1B1 are for the decimal point portion of the duty cycle and are set as follows:

<table>
<thead>
<tr>
<th>DC1B2</th>
<th>DC1B1</th>
<th>Decimal points</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>0.25</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>0.5</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>0.75</td>
</tr>
</tbody>
</table>

It must be noted that the value for the duty cycle register of the CCPR1L is always some percentage of the PR2 register. For example, if PR2 = 50, and we need a 20% duty cycle, then CCPRL1 = 10 because 20% \times 50 = 10. In this case, DC1DB2:DC1B1 = 00. Now assume that we want a 25% duty cycle for the same PR2. Because 50 \times 25% = 12.5, we make CCPRL1 = 12 and DC1B2:DC1B1 = 10 to take care of the 0.5 part. See Example 15-6 for further clarification.

**Example 15-6**

Find the values of registers PR2, CCP1RL, and DC1B2:DC1B1 for the following PWM frequencies if we want a 75% duty cycle. Assume XTAL = 10 MHz.
(a) 1 kHz  b) 2.5 kHz

**Solution:**
(a) Using the \( \text{PR2} = \text{Fosc} / (4 \times \text{Fpwm} \times N) \) equation, we must set \( N = 16 \) for prescale. Therefore, we have
\[
\text{PR2} = \left( \frac{10 \text{ MHz}}{4 \times 1 \text{ kHz} \times 16} \right) - 1 = 156 - 1 = 155 \text{ and because } 155 \times 75\% = 116.25 \text{ we have CCPRL1 = 116 and DC1B2:DC1B1 = 01 for the 0.25 portion.}
\]
(b) Using the \( \text{PR2} = \text{Fosc} / (4 \times \text{Fpwm} \times N) \) equation, we can set \( N = 4 \) for prescale. Therefore, we have
\[
\text{PR2} = \left( \frac{10 \text{ MHz}}{4 \times 2.5 \text{ kHz} \times 4} \right) - 1 = 250 - 1 = 249 \text{ and because } 249 \times 75\% = 186.75 \text{ we have CCPRL1 = 186 and DC1B2:DC1B1 = 11 for the 0.75 portion.}
\]
The following steps are taken to program the PWM feature of the CCP module:

1. Set the PWM period by writing to the PR2 register.
2. Set the PWM duty cycle by writing to CCPR1L for the higher 8 bits.
3. Set the CCP pin as an output.
4. Using the T2CON register, set the prescale value. See Figure 15-14.
5. Clear the TMR2 register.
6. Configure the CCP1ICON register for PWM and set DC1B2:DC1B1 bits for the decimal portion of the duty cycle.
7. Start Timer2.

Examine Programs 15-5 and 15-5C to see how the PWM feature is programmed. These programs use the TMR2IF flag. See Figure 15-15.
Using data from Example 15-6, Program 15-5 will create a 2.5 kHz PWM frequency with a 75% duty cycle on the CCP1 pin.

;Program 15-5
CLRF CCP1CON ;clear CCP1CON reg
MOVlw D'249'    
MOVwf PR2
MOVlw D'186'    ;75% duty cycle
MOVWF CCPRL
BCF TRISC,CCP1 ;make PWM pin an output
MOVlw 0x01     ;Timer2, 4 prescale, no postscaler
MOVWF T2CON
MOVlw 0x0C     ;PWM mode, 11 for DC1B1:B0
MOVWF CCP1CON
CLRF TMR2      ;clear Timer2
BSF T2CON,TMR2ON ;turn on Timer2
AGAIN BCF PIR1,TMR2IF ;clear Timer2 flag
OVER BTFSS PIR1,TMR2IF ;wait for end of period
BRA OVER
GOTO AGAIN    ;continue

//Program 15-5C is the C version of Program 15-5.
CCP1CON=0;       //clear CCP1CON reg
PR2=249;
CCPR1L=186;      //75% duty cycle
TRISCCbits.TRISC2=0;  //make PWM pin an output
T2CON=0x01;      //Timer2, 4 prescale, no postscaler
CCP1CON=0x3C;   //PWM mode, 11 for DC1B1:B0
TMR2=0;          //clear Timer2
T2CONbits.TMR2ON=1; //turn on Timer2
while(1)
{
PIR1bits.TMR2IF=0; //clear Timer2 flag
while(PIR1bits.TMR2IF==0); //wait for end of period
}

The role of CCPR1H in the process of creating the duty cycle must be noted. A copy of the duty cycle value in register CCPR1L is given to CCPR1H as soon as we start Timer2. Timer2 goes through the following stages in creating the PWM:

(a) The CCPR1L is loaded into CCPR1H and the CCP1 pin goes HIGH to start the beginning of the period.
(b) As TMR2 counts up, the TMR2 value is compared with both the CCPR1H and PR2 registers.
(c) When the TMR2 and CCPR1H (which is the same as CCPR1L) values are equal, the CCP pin is forced low. That ends the duty cycle portion of the period.
(d) The TMR2 keeps counting up until its value matches the PR2. At that point, the CCP pin goes high, indicating the end of one period and the beginning
Notice that because the CCPR1L is a fraction of PR2, Timer2 matches CCPR1L first before it matches PR2, unless we have a 100% duty cycle. In that case, Timer2 matches both CCPR1L and PR2 at same time because they have equal values for the 100% duty cycle.

**Duty cycle and Fosc**

The PIC18 datasheet gives the relation between the Fosc and duty cycle period as follows:

\[
T_{dutycycle} = (DC1B9:DC1B0 \text{ value}) \times T_{osc} \times N
\]

(Equ. 15-4)

where \(T_{osc} = 1 / F_{osc}\) and \(N\) is the prescaler of 1, 4, or 16 set by the Timer2 control register. To get the value for the DC1B9:DC1B0 register, we can rearrange the above equation as follows:

\[
DC1B9:DC1B0 = [(F_{osc} / (F_{dutycycle} \times N))]
\]

(Equ. 15-5)

To calculate the maximum resolution (number of bits) that can be used for the PWM, the PIC manual gives the following equation:

Maximum PWM Resolution (bits) = \(\log(F_{osc} / F_{PWM}) / \log(2)\) bits.

Notice that the maximum resolution is 10 bits.
Review Questions

1. True or false. Every standard CCP module has only one PWM pin.
2. How many standard CCP modules do we have in the PIC18F458/4580?
3. True or false. For CCP1, we must use PR2 to set the PWM period.
4. True or false. For CCP1, we must use CCPR1L to set the PWM duty cycle.
5. Which pin of the PIC18F458/4580 is used for PWM?
6. True or false. The duty cycle is always a fraction of the period, unless we want a 100% duty cycle.

SECTION 15.5: ECCP PROGRAMMING

A large number of the PIC18F family members come with ECCP (enhanced CCP) in addition to the standard CCP. While the standard CCP modules are called CCP1, CCP2, and so on, the ECCP modules are designated as ECCP1, ECCP2, and so on. Just like standard CCP, the ECCP has its own pins and registers. The PIC18F452/458 chip uses pin RD4 (PORTD.4) for the ECCP1 pin, while pin RC2 (PORTC.2) is used by the standard CCP1. See Figure 15-18. Figure 15-19 shows the ECCP1 control register.

![ECCP Pins for PWM in PIC18F458/4580 (452/4520)](image-url)
EPWM1M1:EPWM1M0 PWM output pin configuration. It allows the use of a single pin for the capture/compare mode, or four pins for the PWM.

In compare/capture mode, only pin P1A (RD4) is used. In that case, there is no selection for these two bits.

In PWM mode, the options for these two bits are as follows:

<table>
<thead>
<tr>
<th>Code</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>P1A is used for a modulated output. P1B, P1C, and P1D are used as I/O.</td>
</tr>
<tr>
<td>01</td>
<td>Full-Bridge output forward. P1D modulated, P1A active. P1B and P1C inactive.</td>
</tr>
<tr>
<td>10</td>
<td>Half-Bridge output. P1A and P1D modulated with deadband control, P1C and P1D used as I/O.</td>
</tr>
<tr>
<td>01</td>
<td>Full-Bridge output reverse. P1B modulated, P1C active. P1A and P1D inactive.</td>
</tr>
</tbody>
</table>

EDC1B10:EDC1B1 PWM Duty Cycle least-significant bits. Used in PWM only.

The least-significant bits (Bit 1 and Bit 0) of the 10-bit duty cycle register are used in PWM. The ECCPR1L register is used as Bit 2 to Bit 9 of the 10-bit duty cycle register.

ECCP1M3-ECC1M0 ECCP1 Mode Select

<table>
<thead>
<tr>
<th>Code</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>ECCP1 is off</td>
</tr>
<tr>
<td>0001</td>
<td>Reserved</td>
</tr>
<tr>
<td>0010</td>
<td>Compare Mode. Toggle ECCP1 output pin on match. (ECCP1IF bit is set.)</td>
</tr>
<tr>
<td>0011</td>
<td>Reserved</td>
</tr>
<tr>
<td>0100</td>
<td>Capture Mode, every falling edge</td>
</tr>
<tr>
<td>0101</td>
<td>Capture Mode, every rising edge</td>
</tr>
<tr>
<td>0110</td>
<td>Capture Mode, every 4th rising edge</td>
</tr>
<tr>
<td>0111</td>
<td>Capture Mode, every 16th rising edge</td>
</tr>
<tr>
<td>1000</td>
<td>Compare Mode. Initialize ECCP1 pin LOW, on compare match, force CCP1 pin HIGH. (ECCP1IF is set.)</td>
</tr>
<tr>
<td>1001</td>
<td>Compare Mode. Initialize CCP1 pin HIGH, on compare match, force CCP1 pin LOW. (ECCP1IF is set.)</td>
</tr>
<tr>
<td>1010</td>
<td>Compare Mode. Generate software interrupt on compare match. (ECCP1IF bit is set, ECCP1 pin is unaffected.)</td>
</tr>
<tr>
<td>1011</td>
<td>Compare Mode. Trigger special event (ECCP1IF bit is set, and Timer1 or Timer3 is reset to zero.)</td>
</tr>
<tr>
<td>1100</td>
<td>PWM Mode; P1A, P1C active-HIGH; P1B and P1D active-HIGH</td>
</tr>
<tr>
<td>1101</td>
<td>PWM Mode; P1A, P1C active-HIGH; P1B and P1D active-LOW</td>
</tr>
<tr>
<td>1110</td>
<td>PWM Mode; P1A, P1C active-LOW; P1B and P1D active-HIGH</td>
</tr>
<tr>
<td>1111</td>
<td>PWM Mode; P1A, P1C active-LOW; P1B and P1D active-LOW</td>
</tr>
</tbody>
</table>

Figure 15-19. ECCP1 Control Register. (This register selects one of the operation modes of Capture, Compare, or PWM of ECCP1.)
The ECCP1 also has the registers of ECCPR1L, ECCPR1H, and ECCP-CON1. Register PIR2 has the ECCP1IF flag. See Figures 15-20 and 15-21. Just like the standard CCP, it uses Timer1, Timer2, and Timer3 to program the features of compare-capture and PWM. See Table 15-3.

![ECCPR1H ECCPR1L](image)

**Figure 15-20. ECCP High and Low Registers**

<table>
<thead>
<tr>
<th>ECCP1IF</th>
<th>ECCP1IF Interrupt Flag bit.</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Compare Mode</td>
</tr>
<tr>
<td></td>
<td>0 = Timer1 (or Timer3) match did not occur</td>
</tr>
<tr>
<td></td>
<td>1 = Timer1 (or Timer3) match occurred (must be cleared by software)</td>
</tr>
<tr>
<td></td>
<td>Capture Mode</td>
</tr>
<tr>
<td></td>
<td>0 = Timer1 (or Timer3) register capture did not occur</td>
</tr>
<tr>
<td></td>
<td>1 = Timer1 (or Timer3) register capture occurred (must be cleared by software)</td>
</tr>
</tbody>
</table>

**Figure 15-21. PIR2 (Peripheral Interrupt Flag Register 2) Contains the ECCP1IF Flag**

<table>
<thead>
<tr>
<th>ECCP mode</th>
<th>Timer</th>
</tr>
</thead>
<tbody>
<tr>
<td>Capture</td>
<td>Timer1 or Timer3</td>
</tr>
<tr>
<td>Compare</td>
<td>Timer1 or Timer3</td>
</tr>
<tr>
<td>PWM</td>
<td>Timer2</td>
</tr>
</tbody>
</table>

**Table 15-3: PIC18 Use of Timers for ECCP**

**Steps for programming the Compare mode in ECCP**

Programming the ECCP1 in compare mode is identical to the standard CCP, except we use the ECCP registers. The following steps are taken in programming the Compare mode for ECCP1:

1. Initialize the ECCPICON register for the compare option.
2. Initialize the T3CON register for Timer1 (or Timer3).
3. Initialize the ECCPR1H:ECCPR1L registers.
4. Make the ECCP1 pin an output pin.
5. Initialize the Timer1 (or Timer3) register values.
6. Start Timer1 (or Timer3).
7. Monitor the ECCP1IF flag (or use an interrupt).
Program 15-6 shows an example of the Compare mode. It uses Timer3 as a counter and counts the number of pulses fed to Timer3. When the count reaches 20, it toggles the LED connected to the ECCP1 pin.

For Program 15-6 assume that a 1-Hz pulse is connected to the Timer3 pin and an LED is connected to the CCP1 pin. Timer3 is being used as a counter. Using the Compare mode, this Assembly language program will toggle the LED every 20 pulses.

;Program 15-6
MOVLW 0x02
MOVWF ECCP1CON ;Compare mode, toggle upon match
MOVLW 0x42
MOVWF T3CON ;Timer3 for Compare, 1:1 prescaler
BCF TRISD,ECCP1 ;ECCP pin as output
BSF TRISC,T1CKI ;T3CLK pin as input pin
MOVLW D'20'
MOVWF ECCPR1L ;ECCPR1L = 20
MOVLW 0x0 ;ECCPR1H = 0
MOVWF ECCPR1H
OVER CLRF TMR3H ;clear TMR3H
CLRF TMR3L ;clear TMR3L
BCF PIR2,ECCP1IF ;clear ECCP1IF
BSF T3CON,TMR3ON ;start Timer3
B1 BTFSS PIR2,ECCP1IF
BRA B1
;-------- CCP toggle CCP pin upon match
B2 BCF T3CON,TMR3ON ;stop Timer3
GOTO OVER ;keep doing it

//Program 15-6C is the C version of Program 15-6.
ECCP1CON=0x02; //Compare mode, toggle upon match
T3CON=0x42; //Timer3 for Compare, 1:1 prescaler
TRISDbits.TRISD4=0; //make ECCP1 pin an output
TRISCbits.TRISC0=1; //make T3CLK pin an input
ECCPR1L=20; //load ECCPR1L
ECCPR1H=0; //load ECCPR1H
while(1)
{
    TMR3H=0;
    TMR3L=0;
    PIR2bits.ECCP1IF=0; //clear ECCP1IF flag
    T3CONbits.TMR3ON=1; //turn on Timer3
    while(PIR2bits.ECCP1IF==0); //wait for CCP1IF
    //ECCP toggles ECCP pin upon match
    T3CONbits.TMR3ON=0; //stop Timer3
}
Steps for programming the Capture mode in ECCP

Programming the ECCP1 in capture mode is identical to the standard CCP, except that we use the ECCP registers. The following steps are taken in programming the Capture mode of ECCP1 for measuring the period of a pulse:

1. Initialize the ECCP1CON register for the Capture option.
2. Make the ECCP1 pin an input pin.
3. Initialize the T3CON register to select Timer1 or Timer3.
4. Read the Timer1 (or Timer3) register value on the first rising edge and save it.
5. Read the Timer1 (or Timer3) register value on the second rising edge and save it.
6. Subtract the value in step 4 from the value in step 3.

For Program 15-7 assume that a pulse is being fed to the ECCP1 pin. Using the Capture mode, this Assembly language program measures the period of the pulse and puts it on PORTB and PORTC. The measure is in terms of the Fosc/4 clock period.

;Program 15-7
MOVlw 0x05
MOVWF ECCP1CON ;Capture mode on rising edge
MOVlw 0x0
MOVWF T3CON ;Timer1 for capture
MOVlw 0x0
MOVWF T1CON ;Timer1, internal clk, 1:1 prescale
CLRF TRISB ;make PORTB output port
CLRF TRISC ;make PORTC output port
BSF TRISD,ECCP1 ;make ECCP1 pin an input
MOVlw 0x0
MOVWF CCPR1H ;ECCPR1H = 0
MOVwf CCPR1L ;ECCPR1L = 0
OVER CLRF TMR1H ;clear TMR1H
CLRF TMR1L ;clear TMR1L
BCF PIR2,ECCP1IF ;clear ECCP1IF
RE_1 BFSS PIR2,ECCP1IF
BRA RE_1 ;stay here for 1st rising edge
BSF T1CON,TMR1ON ;start Timer1
BCF PIR2,ECCP1IF ;clear ECCP1IF for next
RE_2 BFSS PIR2,ECCP1IF
BRA RE_2 ;stay here for 2nd rising edge
BCF T1CON,TMR1ON ;stop Timer1
MOVFF TMR1L,PORTC ;put low byte on PORTC
MOVFF TMR1H,PORTD ;put high byte on PORTD
GOTO OVER ;keep doing it

//Program 15-7C is the C version of Program 15-7.
ECCP1CON=0x05;  //Capture mode on every rising edge

596
T3CON=0x0; //Timer1 for capture
T1CON=0x0;  //Timer1, internal clk, 1:1 prescaler
TRISC=0;    //make PORTB output port
TRISD=0;    //make PORTD output port
TRISDbits.TRISD4=1; //make ECCP1 pin an input
ECCPR1L=0;  //ECCPR1L = 0
ECCPR1H=0;  //ECCPR1H = 0

//Timer1 for capture
while(1)
{
    TMR1H=0;    //clear Timer1
    TMR1L=0;
    PIR2bits.ECCP1IF=0; //clear ECCP1IF flag
    while(PIR2bits.ECCP1IF==0); //wait for 1st rising edge
    T1CONbits.TMR1ON=1; //start Timer1
    PIR2bits.ECCP1IF=0; //clear ECCP1IF for next edge
    while(PIR2bits.ECCP1IF==0); //wait for 2nd rising edge
    T1CONbits.TMR1ON=0; //stop Timer1
    PORTC=CCPR1L;
    PORTD=CCPR1H;    //display the clock count
}

PWM features of ECCP

The main difference between the ECCP and standard CCP module is the PWM capability. The standard CCP allows only a single pin for PWM output. This is not enough for implementation of the H-Bridge used widely in DC motor control. As we will see in Chapter 17, we need four pins to drive the H-Bridge for DC motor control. The ECCP allows the use of four pins for the implementation of Full-Bridge or two pins for the Half-Bridge. The four pins used by the ECCP are shown in Table 15-4. In terms of the duty cycle calculation, ECCP1 is the same as CCP1. It uses the PR2 for the duty cycle.

Table 15-4: PIC18 Usage of Pins for ECCP1

<table>
<thead>
<tr>
<th>ECCP mode</th>
<th>RD4</th>
<th>RD5</th>
<th>RD6</th>
<th>RD7</th>
</tr>
</thead>
<tbody>
<tr>
<td>Compare/Capture</td>
<td>ECCP1</td>
<td>I/O</td>
<td>I/O</td>
<td>I/O</td>
</tr>
<tr>
<td>Dual Output PWM</td>
<td>P1A</td>
<td>P1B</td>
<td>I/O</td>
<td>I/O</td>
</tr>
<tr>
<td>Quad Output PWM</td>
<td>P1A</td>
<td>P1B</td>
<td>P1C</td>
<td>P1D</td>
</tr>
</tbody>
</table>

Note: I/O means they are used for input/output purpose or other functions associated with the pins.

Steps in programming PWM of ECCP

The following steps are taken to program the PWM feature of the ECCP module:

1. Set the PWM period by writing to the PR2 register.
2. Set the PWM duty cycle by writing to ECCPR1L for the higher 8 bits.
3. Set the ECCP pins as output.
4. Using the T2CON register, set the prescale value.
5. Clear the TMR2 register.
6. Configure the ECCP1CON register for PWM and set the EDC1B2:EDC1B1 bits for the decimal portion of the duty cycle.
7. Start Timer2.

Notice that in programming the compare/capture features, we can assign Timer1 to standard CCP1 and Timer3 to ECCP1 (or vice versa). For the PWM, however, there is only one register for setting the duty cycle. As a result, if we program the PWM feature for both CCP1 and ECCP1, then they will have the same period because there is only one PR2 to set the period. In Chapter 17 we will show how to use ECCP for DC motor control using all four pins in H-Bridge implementations.

Review Questions
1. True or false. Every ECCP module can use only one pin for PWM.
2. How many ECCP modules does the PIC18F458/4580 have?
3. True or false. For ECCP1, we must use PR2 to set the PWM period.
4. True or false. For ECCP1, we must use CCPR1L to set the PWM duty cycle.
5. Which pins of the PIC18F458/4580 are used for PWM?

SUMMARY

This chapter began by describing the CCP features of the PIC18 family. We discussed both the standard CCP and enhanced CCP (ECCP) modules and described each of the compare, capture, and PWM features. We showed how to use Timer1 or Timer3 as the time basis for the compare and capture modes. We also showed how PWM uses Timer2 to create the pulse width modulation.

PROBLEMS

SECTION 15.1: STANDARD AND ENHANCED CCP MODULES

1. True or false. Every member of the PIC18 family has an on-chip CCP module.
2. True or false. The PIC18F452/458 has only one standard CCP.
3. True or false. The PIC18F452/458 has only one ECCP module.
4. True or false. Each CCP module has a 16-bit register accessible as CCPRL and CCPRH.
5. True or false. Each CCP module has a single pin.
6. Give the number of standard and enhanced CCP (ECCP) modules in the PIC18F4520/4580.
7. Give the pin used for standard CCP in the 40-pin DIP package of the PIC18F458/4580.
SECTION 15.2: COMPARE MODE PROGRAMMING

8. True or false. We use register CCP1CON to choose the Compare mode.
9. True or false. We can use Timer0 and Timer2 for Compare mode.
10. True or false. To use Compare mode, we must make the CCP pin an output pin.
11. Which timers can be used for the Compare mode?
12. Assuming that we are using Timer1 for the Compare mode, indicate when the CCP pin is driven HIGH.
13. Which register holds the CCP flag bit?
14. Find the value for the CCP1CON register in compare mode if we want to drive HIGH the CCP pin upon match.
15. Find the value for the CCP1CON register in compare mode if we want to drive LOW the CCP pin upon match.
16. Find the value for the CCP1CON register in compare mode if we want to toggle the CCP pin upon match.
17. Rewrite Program 15-1 (or 15-1C) for Timer1.
18. Rewrite Program 15-1 (or 15-1C) for the count of 1000.
20. Rewrite Program 15-2 (or 15-2C) to create a square wave with a frequency of 100 Hz.

SECTION 15.3: CAPTURE MODE PROGRAMMING

21. True or false. We use the CCP1CON register to choose the Capture mode.
22. True or false. We can use Timer0 and Timer2 for Capture mode.
23. True or false. To use Capture mode, we must make the CCP pin an output pin.
24. Which timers can be used for the capture mode?
25. Find the value for the CCP1CON register in capture mode if we want to capture on the falling edge.
26. Find the value for the CCP1CON register in capture mode if we want to capture every fourth rising edge.
27. Find the value for the T3CON register if we want to use Timer1 for capture mode.
28. Rewrite Program 15-3 (or 15-3C) for Timer3.

SECTION 15.4: PWM PROGRAMMING

29. True or false. We use the CCP1CON register to choose the PWM mode.
30. True or false. We can use Timer0 and Timer1 for the PWM mode.
31. True or false. To use PWM mode, we must make the CCP pin an output pin.
32. Which timer can be used for PWM mode for the standard CCP1?
33. Find the value for the CCP1CON register for PWM mode.
34. Of the CCPR1L and CCPR1H registers, which one is used to set the duty cycle?
35. Which register holds the DC1B2:DC1B1 bits?
36. What is the role of the DC1B2:DC1B1 bits in creating the duty cycle?
37. What is the value for the DC1B2:DC1B1 bits if we want 0.75 for the decimal
38. In programming the PWM, the value loaded into the CCPRL1 is always a _____________ (fraction, multiple) of the PR2 value.
39. Find the values of registers PR2, CCP1RL, and DCB1B2:DC1B1 bits for the PWM frequency of 2 kHz with 25% duty cycle. Assume XTAL = 10 MHz.
40. Find the values of registers PR2, CCP1RL, and DCB1B2:DC1B1 bits for the PWM frequency of 1.8 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.
41. Find the values of registers PR2, CCP1RL, and DCB1B2:DC1B1 bits or the PWM frequency of 1.5 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.
42. Find the values of registers PR2, CCP1RL, and DCB1B2:DC1B1 bits for the PWM frequency of 1.2 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.

SECTION 15.5: ECCP PROGRAMMING

43. True or false. We use ECCP1CON to choose the PWM mode.
44. True or false. We can use Timer1 or Timer3 for the PWM mode in ECCP.
45. True or false. To use capture mode, we must make the ECCP pin an output pin.
46. Which timer can be used for the PWM mode for ECCP1?
47. Which register holds the ECCP1IF flag bit?
48. Find the value for the ECCP1CON register in compare mode if we want to drive HIGH the ECCP pin upon match.
49. In the PIC18F452/458, give the pin used for ECCP for compare/capture mode.
50. Which timers can be used for the compare mode in ECCP?
51. Which pins are used for PWM in ECCP1?
52. Find the value for the ECCP1CON register in compare mode if we want to drive HIGH the ECCP pin upon match.
53. Find the value for the ECCP1CON register in PWM mode if we want to have H-Bridge where P1A and P1C are active high and the rest are active low.
54. Of the ECCPR1L and ECCPR1H registers, which one is used to set the duty cycle?
55. Which register holds the EDC1B2:EDC1B1 bits?
56. What is role of the EDC1B2:EDC1B1 bits in creating duty cycle?
57. What is value for the EDC1B2:EDC1B1 bits if we want 0.5 for the decimal points part of the duty cycle?
58. In programming the PWM, the value loaded into ECCPRL1 is always a _____________ (fraction, multiple) of the PR2 value.
59. Find the values of registers PR2, ECCP1RL, and EDCB1B2:EDC1B1 bits for the PWM frequency of 2 kHz with 25% duty cycle. Assume XTAL = 10 MHz.
60. Find the values of registers PR2, ECCP1RL, and EDCB1B2:DC1B1 bits for the PWM frequency of 1.8 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.
61. Find the values of registers PR2, ECCP1RL, and EDCB1B2:DC1B1 bits for the PWM frequency of 1.5 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.
62. Find the values of registers PR2, ECCP1RL, and EDCB1B2:DC1B1 bits for the PWM frequency of 1.2 kHz with duty cycle of 25%. Assume XTAL = 10 MHz.
ANSWERS TO REVIEW QUESTIONS

SECTION 15.1: STANDARD AND ENHANCED CCP MODULES
1. True
2. True
3. True
4. RC2 (PORTC.2)

SECTION 15.2: COMPARE MODE PROGRAMMING
1. False
2. True
3. True
4. T3CON

SECTION 15.3: CAPTURE MODE PROGRAMMING
1. True
2. True
3. False
4. False
5. T3CON

SECTION 15.4: PWM PROGRAMMING
1. True
2. One
3. True
4. True
5. RC2
6. True

SECTION 15.5: ECCP PROGRAMMING
1. False. Up to four pins.
2. One
3. True
4. False
5. RD4–RD7
CHAPTER 16

SPI PROTOCOL AND DS1306 RTC INTERFACING

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Understand the Serial Peripheral Interfacing (SPI) protocol
- Explain how the SPI read and write operations work
- Examine the SPI pins SDO, SDI, CE, and SCLK
- Code programs in Assembly and C for SPI
- Explain how the real-time clock (RTC) chip works
- Explain the function of the DS1306 RTC pins
- Explain the function of the DS1306 RTC registers
- Understand the interfacing of the DS1306 RTC to the PIC18
- Code programs to display time and date in Assembly and C
- Explore and program the alarm and interrupt features of the RTC
This chapter discusses the SPI bus and shows the interfacing and programming of the DS1306 real-time clock (RTC), an SPI chip. In Section 16.1, we describe SPI bus connection and protocol. In Section 16.2, we describe the DS1306 RTC's pin functions and show its interfacing and programming with the PIC18. The C programming of DS1306 is shown in Section 16.3. The alarm feature of the DS1306 is discussed in Section 16.4.

SECTION 16.1: SPI BUS PROTOCOL

The SPI (serial peripheral interface) is a bus interface connection incorporated into many devices such as ADC, DAC, and EEPROM. In this section we examine the pins of the SPI bus and show how the read and write operations in the SPI work.

SPI bus

The SPI bus was originally started by Motorola Corp. (now Freescale), but in recent years has become a widely used standard adapted by many semiconductor chip companies. SPI devices use only 2 pins for data transfer, called SDI (Din) and SDO (Dout), instead of the 8 or more pins used in traditional buses. This reduction of data pins reduces the package size and power consumption drastically, making them ideal for many applications in which space is a major concern. The SPI bus has the SCLK (shift clock) pin to synchronize the data transfer between two chips. The last pin of the SPI bus is CE (chip enable), which is used to initiate and terminate the data transfer. These four pins, SDI, SDO, SCLK, and CE, make the SPI a 4-wire interface. See Figure 16-1. There is also a widely used standard called a 3-wire interface bus. In a 3-wire interface bus, we have SCLK and CE, and only a single pin for data transfer. The SPI 4-wire bus can become a 3-wire interface when the SDI and SDO data pins are tied together. However, there are some major differences between the SPI and 3-wire devices in the data transfer protocol. For that reason, a device must support the 3-wire protocol internally in order to be used as a 3-wire device. Many devices such as the DS1306 RTC (real-time clock) support both SPI and 3-wire protocols.

![Figure 16-1. SPI Bus vs. Traditional Parallel Bus Connection to Microcontroller](image-url)
SPI read and write protocol

In connecting a device with an SPI bus to a microcontroller, we use the microcontroller as the master while the SPI device acts as a slave. This means that the microcontroller generates the SCLK, which is fed to the SCLK pin of the SPI device. The SPI protocol uses SCLK to synchronize the transfer of information one bit at a time, where the most-significant bit (MSB) goes in first. During the transfer, the CE must stay HIGH. The information (address and data) is transferred between the microcontroller and the SPI device in groups of 8 bits, where the address byte is followed immediately by the data byte. To distinguish between the read and write, the D7 bit of the address byte is always 1 for write, while for the read, the D7 bit is LOW, as we will see next.

Steps for writing data to an SPI device

In accessing SPI devices, we have two modes of operation: single-byte and multibyte. We will explain each one separately.

Single-byte write

The following steps are used to send (write) data in single-byte mode for SPI devices, as shown in Figure 16-2:

1. Make CE = 1 to begin writing.
2. The 8-bit address is shifted in one bit at a time, with each edge of SCLK. Notice that A7 = 1 for the write operation, and the A7 bit goes in first.
3. After all 8 bits of the address are sent in, the SPI device expects to receive the data belonging to that address location immediately.
4. The 8-bit data is shifted in one bit at a time, with each edge of the SCLK.
5. Make CE = 0 to indicate the end of the write cycle.

Figure 16-2. SPI Single-Byte Write Timing (Notice A7 = 1)

Multibyte burst write

Burst mode writing is an effective means of loading consecutive locations. In burst mode, we provide the address of the first location, followed by the data for that location. From then on, while CE = 1, consecutive bytes are written to consecutive memory locations. In this mode, the SPI device internally increments the
address location as long as CE is HIGH. The following steps are used to send (write) multiple bytes of data in burst mode for SPI devices as shown in Figure 16-3:

1. Make CE = 1 to begin writing.
2. The 8-bit address of the first location is provided and shifted in one bit at a time, with each edge of SCLK. Notice that A7 = 1 for the write operation and the A7 bit goes in first.
3. The 8-bit data for the first location is provided and shifted in one bit at a time, with each edge of the SCLK. From then on, we simply provide consecutive bytes of data to be placed in consecutive memory locations. In the process, CE must stay high to indicate that this is a burst mode multibyte write operation.
4. Make CE = 0 to end writing.

![Figure 16-3. SPI Burst (MultiByte) Mode Writing](image)

**Steps for reading data from an SPI device**

In reading SPI devices, we also have two modes of operation: single-byte and multibyte. We will explain each one separately.

**Single-byte read**

The following steps are used to get (read) data in single-byte mode from SPI devices as shown in Figure 16-4:

1. Make CE = 1 to begin writing.
2. The 8-bit address is shifted in one bit at a time, with each edge of SCLK. Notice that A7 = 0 for the read operation, and the A7 bit goes in first.
3. After all 8 bits of the address are sent in, the SPI device sends out data belonging to that location.
4. The 8-bit data is shifted out one bit at a time, with each edge of the SCLK.
5. Make CE = 0 to indicate the end of the read cycle.

![Figure 16-4. SPI Single-Byte Read Timing (Notice A7 = 0)](image)
**Multibyte burst read**

Burst mode reading is an effective means of bringing out the contents of consecutive locations. In burst mode, we provide the address of the first location only. From then on, while CE = 1, consecutive bytes are brought out from consecutive memory locations. In this mode, the SPI device internally increments the address location as long as CE is HIGH. The following steps are used to get (read) multiple bytes of data in burst mode for SPI devices, as shown in Figure 16-5:

1. Make CE = 1 to begin reading.
2. The 8-bit address of the first location is provided and shifted in one bit at a time, with each edge of SCLK. Notice that A7 = 0 for the read operation, and the A7 bit goes in first.
3. The 8-bit data for the first location is shifted out one bit at a time, with each edge of the SCLK. From then on, we simply keep getting consecutive bytes of data belonging to consecutive memory locations. In the process, CE must stay HIGH to indicate that this is a burst mode multibyte read operation.
4. Make CE = 0 to end reading.

![Figure 16-5. SPI Burst (MultiByte) Mode Reading](image-url)

**Review Questions**

1. True or false. The SPI protocol writes and reads information in 8-bit chunks.
2. True or false. In SPI, the address is immediately followed by the data.
3. True or false. In an SPI write cycle, bit A7 of the address is LOW.
4. True or false. In an SPI write, the LSB goes in first.
5. State the difference between the single-byte and burst modes in terms of the CE signal.
SECTION 16.2: DS1306 RTC INTERFACING AND PROGRAMMING

The real-time clock (RTC) is a widely used device that provides accurate time and date information for many applications. Many systems such as the x86 IBM PC come with such a chip on the motherboard. The RTC chip in the IBM PC provides the time components of hour, minute, and second, in addition to the date/calendar components of year, month, and day. Many RTC chips use an internal battery, which keeps the time and date even when the power is off. Although some microcontrollers, such as the DS5000T, come with the RTC already embedded into the chip, we have to interface the vast majority of them to an external RTC chip. One of the most widely used RTC chips is the DS12887 from Dallas Semiconductor/Maxim Corp. This chip is found in the vast majority of x86 PCs. The original IBM PC/AT used the MC14618B RTC from Motorola. The DS12887 is the replacement for that chip. It uses an internal lithium battery to keep operating for over 10 years in the absence of external power. The DS12887 is a parallel RTC with 8 pins for the data bus. In this chapter, we interface and program the DS1306 RTC, which has an SPI bus. According to the DS1306 data sheet from Maxim, it keeps track of “seconds, minutes, hours, day of week, date, month, and year with leap-year compensation valid up to year 2099.” The DS1306 RTC provides the above information in BCD format only. It supports both 12-hour and 24-hour clock modes, with AM and PM in the 12-hour mode. It does not support the Daylight Savings Time option. The DS1306 has a total of 128 bytes of nonvolatile RAM. It uses 28 bytes of RAM for clock/calendar and control registers, and the other 96 bytes of RAM are for general-purpose data storage. Next, we describe the pins of the DS1306. See Figure 16-6.

$V_{CC2}$

Pin 1 provides an external back-up supply voltage to the chip. This pin is connected to an external rechargeable power source. This option is called trickle charge. If this pin is not used, it must be grounded.

$V_{bat}$

Pin 2 can be connected to an external +3 V lithium battery, thereby providing the power source to the chip externally as back-up supply voltage. We must connect this pin to ground if it is not used.

$V_{CC1}$

Pin 16 is used as the primary external voltage supply to the chip. This primary external voltage source is generally set to +5 V. When $V_{CC1}$ falls below the $V_{bat}$ voltage level, the DS1306 switches to $V_{bat}$ and the external lithium battery provides power to the RTC. According to the DS1306 data sheet “upon power-up, the device switches from $V_{bat}$ to $V_{CC1}$ when $V_{CC1}$ is greater than $V_{bat}+0.2$ Volts.” Because we can connect the standard 3 V lithium battery to the $V_{bat}$ pin, the $V_{CC1}$ voltage level must remain above 3.2 V in order for the $V_{CC1}$ to remain as the primary voltage source to the chip. This nonvolatile capability of the RTC prevents
any loss of data. See Figure 16-7.

**GND**

Pin 8 is the ground.

**SDI (Serial Din)**

The SDI pin provides the path to bring data into the chip, one bit at a time.

**SDO (Serial Dout)**

The SDO pin provides the path to bring data out of the chip, one bit at a time.

**32KHz**

This is an output pin providing a 32.768 kHz frequency. This frequency is always present at the pin.

**X1–X2**

These are input pins that allow the DS1306 connection to an external crystal oscillator to provide the clock source to the chip. We must use the standard 32.768 kHz quartz crystal. The accuracy of the clock depends on the quality of this crystal oscillator. See Figure 16-8. Heat can cause a drift on the oscillator. To avoid this, we use the DS32KHZ chip, which automatically adjusts for temperature variations. Note that when using the DS32KHZ or similar clock generators, we only need to connect X1 because the X2 loopback is not required.

**SCLK (serial clock)**

An input pin is used for the serial clock to synchronize the data transfer between the DS1306 and the microcontroller.

**1-Hz**

An output pin provides a 1-Hz square wave frequency. The DS1306 creates the 1-Hz square wave automatically. To get this 1-Hz frequency to show up on the pin, however, we must enable the associated bit in the DS1306 control register.

**CE**

Chip enable is an input pin and an active-HIGH signal. During the read and write cycle time, CE must be high.

**INT0#**

Interrupt request is an output pin and an active-LOW signal. To use INT0, the interrupt-enable bit in the RTC control register must be set HIGH. The interrupt feature of the DS1306 is discussed in Section 16.4.

**INT1**

Interrupt request is an output pin and an active-HIGH signal. To use INT1, the interrupt-enable bit in the RTC control register must be set HIGH. The inter-
rupt feature of the DS1306 is discussed in Section 16.4.

**SERMODE (serial mode selection)**

Pin 9 is an input pin. If it is HIGH, then the SPI mode is selected. If it is connected to ground, the 3-wire mode is used. In our application, the SERMODE pin is connected to the \( V_{CC} \) pin because we program the 1306 chip using the SPI protocol.

**\( V_{CC} \)**

Pin 14 is the interface logic power-supply input. This pin allows interfacing of the DS1306 with systems with 3 V logic in mixed supply systems. See the DS1306 data sheet if you are using a power source other than 5 V in your system.

![Simplified Block Diagram of DS1306 (Maxim/Dallas Semiconductor)](image)

**Figure 16-8. Simplified Block Diagram of DS1306 (Maxim/Dallas Semiconductor)**

**Importance of the WP bit in the Control register**

As shown in Table 16-1, the Control register has an address of 8FH for write and 0FH for read. The most important bit in the Control register is the WP bit. The WP bit is undefined upon reset. In order to write to any of the registers of the DS1306, we must clear the WP bit first. See Figure 16-9. Upon powering up the DS1306, we have to clear the WP bit at least once. This means that after initializing the DS1306 we can write protect all the registers by making WP = 1.

![WP Bit of DS1306 Control Register (write location address is 8FH)](image)

**Figure 16-9. WP Bit of DS1306 Control Register (write location address is 8FH)**
Address map of the DS1306

The DS1306 has a total of 128 bytes of RAM space with addresses 00–7FH. The first fifteen locations, 00–0E, are set aside for RTC values of time, date, and alarm data. The next three bytes are used for the control and status registers. They are located at addresses 0F–11 in hex. The next 14 bytes from addresses 12H to 1FH are reserved and cannot be used. That leaves 96 bytes, from addresses 20H to 7FH, available for general purpose data storage. That means the entire 128 bytes of RAM are accessible directly for read or write except the addresses 12–1FH. Table 16-1 shows the address map of the DS1306. In this section we study the time and date. The alarm is examined in Section 16.4.

Table 16-1: Registers of the DS1306 (modified from datasheet)

<table>
<thead>
<tr>
<th>HEX ADDRESS</th>
<th>D7</th>
<th>D6</th>
<th>D5</th>
<th>D4</th>
<th>D3</th>
<th>D2</th>
<th>D1</th>
<th>D0</th>
<th>RANGE in HEX</th>
</tr>
</thead>
<tbody>
<tr>
<td>READ WRITE</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>00-59</td>
</tr>
<tr>
<td>0x00</td>
<td>0x80</td>
<td>0</td>
<td></td>
<td></td>
<td>10 SEC</td>
<td></td>
<td></td>
<td>SEC</td>
<td>00-59</td>
</tr>
<tr>
<td>0x01</td>
<td>0x81</td>
<td>0</td>
<td></td>
<td></td>
<td>10 MIN</td>
<td></td>
<td></td>
<td>MIN</td>
<td>00-59</td>
</tr>
<tr>
<td>0x02</td>
<td>0x82</td>
<td>0</td>
<td></td>
<td>24 / 20 HR</td>
<td>10 HR / 12</td>
<td></td>
<td></td>
<td>P/A</td>
<td>01-12 P/A</td>
</tr>
<tr>
<td>0x03</td>
<td>0x83</td>
<td>0</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>0</td>
<td>01-07</td>
</tr>
<tr>
<td>0x04</td>
<td>0x84</td>
<td>0</td>
<td></td>
<td></td>
<td>10 DATE</td>
<td></td>
<td></td>
<td>DATE</td>
<td>01-31</td>
</tr>
<tr>
<td>0x05</td>
<td>0x85</td>
<td>0</td>
<td></td>
<td></td>
<td>10 MONTH</td>
<td></td>
<td></td>
<td>MONTH</td>
<td>01-12</td>
</tr>
<tr>
<td>0x06</td>
<td>0x86</td>
<td>0</td>
<td></td>
<td></td>
<td>10 YEAR</td>
<td></td>
<td></td>
<td>YEAR</td>
<td>00-99</td>
</tr>
<tr>
<td>0x07</td>
<td>0x87</td>
<td>M</td>
<td></td>
<td></td>
<td>10 SEC ALARM0</td>
<td></td>
<td></td>
<td>SEC ALARM0</td>
<td>00-59</td>
</tr>
<tr>
<td>0x08</td>
<td>0x88</td>
<td>M</td>
<td></td>
<td></td>
<td>10 MIN ALARM0</td>
<td></td>
<td></td>
<td>MIN ALARM0</td>
<td>00-59</td>
</tr>
<tr>
<td>0x09</td>
<td>0x89</td>
<td>M</td>
<td>24 / 20 HR</td>
<td>10 HR / 12</td>
<td></td>
<td></td>
<td>P/A</td>
<td>00-23</td>
<td></td>
</tr>
<tr>
<td>0x0A</td>
<td>0x8A</td>
<td>M</td>
<td>0</td>
<td></td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>DAY ALARM0</td>
<td>01-07</td>
</tr>
<tr>
<td>0x0B</td>
<td>0x8B</td>
<td>M</td>
<td>10 SEC ALARM1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>SEC ALARM1</td>
<td>00-59</td>
</tr>
<tr>
<td>0x0C</td>
<td>0x8C</td>
<td>M</td>
<td>10 MIN ALARM1</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>MIN ALARM1</td>
<td>00-59</td>
</tr>
<tr>
<td>0x0D</td>
<td>0x8D</td>
<td>M</td>
<td>24 / 20 HR</td>
<td>10 HR / 12</td>
<td></td>
<td></td>
<td>P/A</td>
<td>00-23</td>
<td></td>
</tr>
<tr>
<td>0x0E</td>
<td>0x8E</td>
<td>M</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>DAY ALARM1</td>
<td>01-07</td>
</tr>
<tr>
<td>0x0F</td>
<td>0x8F</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>CONTROL REGISTER</td>
<td></td>
</tr>
<tr>
<td>0x10</td>
<td>0x90</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>STATUS REGISTER</td>
<td></td>
</tr>
<tr>
<td>0x11</td>
<td>0x91</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td>TRICKLE CHARGER REGISTER</td>
<td>RESERVE</td>
</tr>
<tr>
<td>0x12-0x1F</td>
<td>0x92-0x9F</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Time and date address locations and modes

The byte addresses 0–6 are set aside for the time and date, as shown in Table 16-2. Table 16-2 is extracted from Table 16-1. It shows a summary of the address locations in read/write modes with data ranges for each location. The DS1306 provides data in BCD format only. Notice the data range for the hour mode. We can select 12-hour or 24-hour mode with bit 6 of hour location 02. When D6 = 1, the 12-hour mode is selected, and D6 = 0 provides us the 24-hour mode. In the 12-hour mode, we decide the AM and PM with the bit 5. If D5 = 0, the AM is selected and D5 = 1 is for the PM. Example 16-1 shows how to get the range of the data acceptable for the hour location.
Table 16-2: DS1306 Address Locations for Time and Date (extracted from Table 16-1)

<table>
<thead>
<tr>
<th>Hex Address Read</th>
<th>Location Write</th>
<th>Function</th>
<th>Data Range BCD</th>
<th>Range in hex</th>
</tr>
</thead>
<tbody>
<tr>
<td>00</td>
<td>80</td>
<td>Seconds</td>
<td>00–59</td>
<td>00–59</td>
</tr>
<tr>
<td>01</td>
<td>81</td>
<td>Minutes</td>
<td>00–59</td>
<td>00–59</td>
</tr>
<tr>
<td>02</td>
<td>82</td>
<td>Hours, 12-Hour Mode</td>
<td>01–12</td>
<td>41–52 AM</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Hours, 24-Hour Mode</td>
<td>00–23</td>
<td>00–23</td>
</tr>
<tr>
<td>03</td>
<td>83</td>
<td>Day of the Week, Sun = 1</td>
<td>01–07</td>
<td>01–07</td>
</tr>
<tr>
<td>04</td>
<td>84</td>
<td>Day of the Month</td>
<td>01–31</td>
<td>01–31</td>
</tr>
<tr>
<td>05</td>
<td>85</td>
<td>Month</td>
<td>01–12</td>
<td>01–12</td>
</tr>
<tr>
<td>06</td>
<td>86</td>
<td>Year</td>
<td>00–99</td>
<td>00–99</td>
</tr>
</tbody>
</table>

Example 16-1

Using Table 16-1, verify the hour location values in Table 16-2.

Solution:

(a) For 24-hour mode, we have D6 = 0. Therefore, the range goes from 0000 0000 to 0010 0011, which is 00–23 in BCD.

(b) For 12-hour mode, we have D6 = 1 and D5 = 0 for AM. Therefore, the range goes from 0100 0001 to 0101 0010, which is 41–52 in BCD.

(c) For 12-hour mode, we have D6 = 1 and D5 = 1 for PMn. Therefore, the range goes from 0110 0001 to 0111 0010, which is 61–72 in BCD.

PIC18 interfacing to DS1306 using MSSP module

The DS1306 supports both SPI and 3-wire modes. In DS1306, we select the SPI mode by connecting the SERMODE pin to Vcc. If SERMODE = Gnd, then the 3-wire protocol is used. In this section, we use SPI mode only. The MSSP (Master Synchronous Serial Port) module inside the PIC18 supports SPI bus protocol. Three registers are associated with SPI of the MSSP module. They are SSPBUF, SSPCON1, and SSPSTAT. To transfer a byte of data, we place it in SSPBUF. The SSPBUF register also holds the byte received via the SPI bus. Figures 16-10 and 16-11 show the other two major registers of the PIC18 for SPI interfacing. We use SSPCON1 to select the SPI mode operation of the PIC18. Notice that the SSPEN bit in the SSPCON1 register must be set to HIGH to allow the use of the PIC18 pins for SPI data bus protocol. We must also choose the SPI Master mode using the SSPM3:SSPM0 bits of SSPCON1. In our application, we will use Fosc/64 speed for best performance in data transfer between the PIC18 and the DS1306 RTC.

After the selection of SSPCON1, we must also select the proper bits for timing in the SSPSTAT register, as shown in Figure 16-11. In our application, we send data to an SPI device on the rising edge, and receive data from the SPI device in the middle of the SCLK clock pulse.

Because we are using the SPI feature of the PIC18 to communicate with
our SPI device, we must use the designated pins for the SPI signals. They are RC2 (CE), RC3 (SCLK), RC4 (SDI), and RC5 (SDO), as shown in Figure 16-12.

<table>
<thead>
<tr>
<th></th>
<th>SSPEN</th>
<th>SSPM3</th>
<th>SSPM2</th>
<th>SSPM1</th>
<th>SSPM0</th>
</tr>
</thead>
<tbody>
<tr>
<td>SSPEN</td>
<td>D5</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>Synchronous Serial Port Enable bit</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>1 = Enables serial port and configures SCK, SDO, and SDI as serial port pins</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td></td>
<td>0 = Disables serial port and configures these pins as I/O ports</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>SSPM3:SSPM0</th>
<th>D3–D0</th>
<th>SPI Mode Selection bits</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>0010</td>
<td>SPI Master, clock = Fosc/64</td>
</tr>
<tr>
<td></td>
<td>0001</td>
<td>SPI Master, clock = Fosc/16</td>
</tr>
<tr>
<td></td>
<td>0000</td>
<td>SPI Master, clock = Fosc/4</td>
</tr>
</tbody>
</table>

The rest of the bits are unused in our implementation of SPI.
We use SPI in master mode.

![Figure 16-10. SSPCON1 - SSP Control Register 1](image)
*Note:* Portion shown is used for SPI.

<table>
<thead>
<tr>
<th>SMP</th>
<th>CKE</th>
<th>D7</th>
<th>Sample bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>1</td>
<td>Input data sampled at end of data output time</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0</td>
<td>Input data sampled at middle of data output time</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>CKE</th>
<th>D6</th>
<th>SPI Clock Edge Select bit</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1</td>
<td>Transmit occurs on transition from active to idle clock state.</td>
</tr>
<tr>
<td></td>
<td>0</td>
<td>Transmit occurs on transition from idle to active clock state.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>BF</th>
<th>D0</th>
<th>Buffer Full Status bit. Used for receive only.</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td></td>
<td>Receive complete, SSPBUF is full.</td>
</tr>
<tr>
<td>0</td>
<td></td>
<td>Receive not complete, SSPBUF is empty.</td>
</tr>
</tbody>
</table>

The rest of the bits are used for I²C module.

![Figure 16-11. SSPSTAT - SSP Status Register](image)
*Note:* Portion shown is used for SPI.
Figure 16-12. DS1306 Connection to PIC18

*Note:* For more accuracy, we use the DS32KHZ chip in place of a crystal.

### Setting the time in Assembly

Program 16-1 initializes the clock at 16:58:55 using the 24-hour clock mode. It uses the single-byte operation for writing into the control register of the DS1306 and multibyte burst mode for writing seconds, minutes, and hours. Regarding the SPI subroutine in Program 16-1, we must note the following points:

1. In order for the PIC18 to transfer a byte of data using SPI protocol, it must be placed in SSPBUF.
2. After writing to SSPBUF, we must monitor the BF flag bit of the SSPSTAT register to ensure the entire byte has been transferred.
3. SSPBUF is also used as the destination for incoming data from an SPI device. This happens as data is being sent. The BF flag indicates that the entire byte has been received.

```assembly
;Program 16-1: Setting the Time
MOVLW 0x00
MOVWF SSPSTAT ;read at middle, send on active edge
MOVLW 0x22
MOVWF SSPCON1 ;enable master SPI, Fosc / 64
CLRF TRISC ;make PORTC output
BSF TRISC, SDI ;except SDI
;-- send control byte to DS1306 in single-byte mode
BSF PORTC, RC2 ;make CE = 1 for single-byte
CALL SDELAY
MOVLW 0x8F
MOVLW 0x00 ;DS1306 control register address
CALL SPI
```

614
CALL SPI
BCF PORTC,RC2
; make CE = 0 to end write (single-byte)
CALL SDELAY
;-- send the data to DS1306 in burst mode
BSF PORTC,RC2 ; make CE = 1 (start multibyte write)
MOVLW 0x80 ; seconds register address
CALL SPI ; send address
MOVLW 0x55 ; 55 seconds
CALL SPI ; send seconds
MOVLW 0x58 ; 58 minutes
CALL SPI ; send minutes
MOVLW 0x16 ; 24-hour clock at 16 hours
CALL SPI ; send hour
BCF PORTC,RC2 ; make CE = 0 (end multibyte write)
;-- SPI write/read subroutine
SPI MOVWF SSPBUF ; load SSPBUF for transfer
WAIT BTFSS SSPSTAT,BF ; wait for all bits
BRA WAIT
MOVF SSPBUF,W ; get the received byte
RETURN ; return with byte in WREG
END

Setting the date in Assembly

Program 16-2 shows how to set the date to October 19th, 2004.

; Program 16-2: Setting the Date
MOVLW 0x00
MOVWF SSPSTAT ; read at middle, send on active edge
MOVLW 0x22
MOVF SSPCON1 ; master SPI enable, Fosc / 64
CLR TRISC ; make PORTC output
BSF TRISC,SDI ; except SDI
BSF PORTC,RC2 ; enable the RTC
MOVLW 0x8F ; DS1306 control register address
CALL SPI
MOVLW 0x00 ; clear WP bit for write
CALL SPI
BCF PORTC,RC2 ; turn off RTC
;-- send the date to DS1306
BSF PORTC,RC2 ; enable the RTC
MOVLW 0x84 ; date register address
CALL SPI ; send address
MOVLW 0x19 ; 19th of the month
CALL SPI ; send date
MOVLW 0x10 ; October
CALL SPI ; send month
MOVLW 0x04 ; 2004
CALL SPI ; send year
BCF PORTC,RC2 ; disable RTC
;-- SPI write/read subroutine
RTCs setting, reading, and displaying time and date

Program 16-3 is the complete Assembly code for setting, reading, and displaying the time and date. The times and dates are sent to the IBM PC screen via the serial port after they are converted from packed BCD to ASCII.

```assembly
; Program 16-3
#include p18f458.inc
D1uL EQU D'2' ; 1 microsecond delay byte
DR1uL EQU 0x0D ; register for 1 microsecond delay
DAY EQU 10H ; for day of the week
MON EQU 11H ; for day of the week
DAT EQU 12H ; for day of the month
YR EQU 13H ; for year
HR EQU 14H ; for hour
MIN EQU 15H ; for minutes
SEC EQU 16H ; for seconds
CNT EQU 20H ; for counter
TMP EQU 21H ; for conversions
MOVLW 0x00
MOVWF SSPSTAT ; read at middle, send on active edge
MOVLW 0x22
MOVWF SSPCON1 ; master SPI enable, Fosc / 64
CLRF TRISC ; make PORTC output
BSF TRISC, SDI ; except SDI
BSF TRISC, RX ; and RX
; -- enable USART communication
MOVLW B'00100000' ; enable transmit and low baud
MOVWF TXSTA ; write to reg
MOVLW D'15' ; 9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ; write to reg
BCF TRISC, TX ; make TX pin of PORTC an output pin
BSF RCSTA, SPEN ; enable the serial port
; -- start a new line for USART communications
MOVLW 0x0A ; form feed
CALL TRANS
MOVLW 0x0D ; new line
CALL TRANS
; -- send control byte to DS1306
BSF PORTC, RC2 ; enable the RTC
CALL SDELAY
MOVLW 0x8F ; control register address
CALL SPI
MOVLW 0x00 ; clear WP bit for write
```
CALL SPI
BCF PORTC,RC2 ;disable RTC
CALL SDELAY

;-- send the time followed by date
BSF PORTC,RC2 ;enable the RTC
MOVLW 0x80 ;seconds register address for write
CALL SPI ;send address
MOVLW 0x55 ;55 seconds
CALL SPI ;send seconds
MOVLW 0x58 ;58 minutes
CALL SPI ;send minutes
MOVLW 0x16 ;24-hour clock at 16 hours
CALL SPI ;send hour
MOVLW 0x3 ;Tuesday
CALL SPI ;send day of the week
MOVLW 0x19 ;19th of the month
CALL SPI ;send day of the month
MOVLW 0x10 ;October
CALL SPI ;send month
MOVLW 0x04 ;2004
CALL SPI ;send year
BCF PORTC,RC2 ;disable RTC
CALL SDELAY

;-- get the time and date from DS1306
RDA BSF PORTC,RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x00 ;seconds register address for read
CALL SPI ;send address to DS1306
CALL SPI ;start getting time/date
MOVWF SEC ;save the seconds
CALL SPI ;get the minutes
MOVWF MIN ;save the minutes
CALL SPI ;get the hour
MOVWF HR ;save the hour
CALL SPI ;get the day
MOVWF DAY ;save the day
CALL SPI ;get the date
MOVWF DAT ;save the date
CALL SPI ;get the month
MOVWF MON ;save the month
CALL SPI ;get the year
MOVWF YR ;save the year
BCF PORTC,RC2 ;disable RTC

;-- convert packed BCD to ASCII and display
LFSR FSR0,0x11 ;address of fileReg for time/date
MOVLW D'6'; ;6 bytes of data to display
MOVWF CNT ;set up the counter
SND MOVFF INDF0,TMP ;get the data for high nibble
MOVlw 0xF0 ;clear low nibble
ANDWF TMP,F ;keep in TMP register
SWAPF TMP,F ;switch high and low nibbles
MOVLW 0x30 ;convert to ASCII
IORWF TMP,W ;put in WREG
CALL TRANS ;display the data
MOVFF POSTINC0,TMP ;get the data and point to next
MOVLW 0x0F ;clear high nibble
ANDWF TMP,F ;keep in TMP register
MOVLW 0x30 ;convert to ASCII
IORWF TMP,W ;put in WREG
CALL TRANS ;display the data
MOVLW ':'
CALL TRANS
DECFSZ CNT ;Is it the last one?
BRA SND ;no
MOVLW 0x0D ;line feed
CALL TRANS
BRA RDA ;keep reading time/date and display them

;-- SPI write/read subroutine
SPI MOVWF SSPBUF ;load SSPBUF for transfer
WAIT BTFSS SSPSTAT,BF ;wait for all bits
BRA WAIT
MOVF SSPBUF,W ;get the received byte
RETURN ;return with byte in WREG

;----serial data transfer subroutine
TRANS BTFSS PIR1, TXIF ;wait until the last bit is gone
BRA TRANS ;stay in loop
MOVWF TXREG ;load the value to be transmitted
RETURN ;return to caller

;----short delay
SDELAY: MOVLW DluL ;low byte of delay
MOVF DRluL ;store in register
DS1 DECF DRluL,F ;stay until DRluL becomes 0
BNZ DS1
RETURN
END

Review Questions

1. True or false. All of the RAM contents of the DS1306 are nonvolatile.
2. How many bytes of RAM in the DS1306 are set aside for the clock and date?
3. How many bytes of RAM in the DS1306 are set aside for general-purpose applications?
4. True or false. The DS1306 has a single pin for Din.
5. Which pin of the DS1306 is used for Clock in SPI connection?
6. True or false. To use the DS1306 in SPI mode, we make SERMODE = GND.
SECTION 16.3: DS1306 RTC PROGRAMMING IN C

In this section, we program the DS1306 in PIC18 C language. Before you embark on this section, make sure you understand the basic concepts of the DS1306 chip covered in the first section.

Setting the time and date in C

Program 16-4C shows how to set the time and date for the DS1306 configuration in Figure 16-12.

//Program 16-4C : Setting time and date
#include <pl8f458.h>
unsigned char SPI(unsigned char);
void SDELAY(int ms);
void main()
{
    SSPSTAT = 0; //read at middle, send on active edge
    SSPCON1 = 0x22; //master SPI enable, Fosc / 64
    TRISC = 0; //make PORTC output
    TRISCbits.TRISC4 = 1; //except SDI
    TRISCbits.TRISC7 = 1; //and RX
    PORTCbits.RC2 = 1; //enable the RTC
    SDELAY(1);
    SPI(0x8F); //control register address
    SPI(0x00); //clear WP bit for write
    PORTCbits.RC2 = 0; //end of single-byte write
    SDELAY(1);
    PORTCbits.RC2 = 1; //begin multibyte write
    SPI(0x80); //seconds register address
    SPI(0x55); //55 seconds
    SPI(0x58); //58 minutes
    SPI(0x16); //24-hour clock at 16 hours
    SPI(0x3); //Tuesday
    SPI(0x19); //19th of the month
    SPI(0x10); //October
    SPI(0x04); //2004
    PORTCbits.RC2 = 0; //end multibyte write
    SDELAY(1);
}

//-- SPI Write/Read subroutine
unsigned char SPI(unsigned char myByte)
{
    SSPBUF = myByte; //load SSPBUF for transfer
    while(!SSPSTATbits.BF); //wait for all bits
    return SSPBUF; //return with received byte
}
Reading and displaying the time and date in C

Program 16-5C shows how to read the time, convert it to ASCII, and send it to the PC screen via the serial port.

```c
#include <p18f458.h>

unsigned char SPI(unsigned char);
void TRANS(unsigned char);
void BCDtoASCIIandSEND(unsigned char);
void SDELAY(int ms);

void main()
{
    unsigned char data[7]; //holds date and time
    unsigned char tmp; //for BCD to ASCII conversion
    int i;
    SSPSTAT = 0; //read at middle, send on active edge
    SSPCON1 = 0x22; //master SPI enable, Fosc / 64
    TRISC = 0; //make PORTC output
    TRISCbits.TRISC4 = 1; //except SDI
    TRISCbits.TRISC7 = 1; //and RX
    TXSTA = 0x20; //enable transmit and low baud
    SPBRG = 15; //9600 bps (Fosc / (64 * Speed) - 1)
    RCSTAbits.SPEN = 1; //enable the serial port
    TRANS(0x0A); //form feed
    TRANS(0x0D); //new line

    //-- get the time and date from RTC and save them
    while(1)
    {
        PORTCbits_RC2 = 1; //begin multibyte read
        SDELAY(1);
        SPI(0x00); //seconds register address
        for(i=0;i<7;i++)
        {
            data[i] = SPI(0x00); //get time/date and save
        }
        PORTCbits_RC2 = 0; //end of multibyte read

        //-- convert time/date and display MM:DD:YY:HH:MM:SS
        BCDtoASCIIandSEND(data[5]); //the month
        BCDtoASCIIandSEND(data[4]); //the date
        BCDtoASCIIandSEND(data[6]); //the year
        BCDtoASCIIandSEND(data[2]); //the hour
        BCDtoASCIIandSEND(data[1]); //the minute
        BCDtoASCIIandSEND(data[0]); //the second
        TRANS(0x0D); //new line
    }
}
```
//-- SPI Write/Read

unsigned char SPI(unsigned char myByte)
{
    SSPBUF = myByte;
    while(!SSPSTATbits.BF);
    return SSPBUF;
}

void TRANS(unsigned char myChar) //serial data transfer
{
    while(!PIRlbits.TXIF);
    TXREG = myChar; //load the value to be transmitted
}

void BCDtoASCIIandSEND(unsigned char myValue)
{
    unsigned char tmp = myValue;
    tmp = tmp & 0xF0;  //mask lower nibble
    tmp = tmp >> 4;   //swap it
    tmp = tmp | 0x30;  //make it ASCII
    TRANS(tmp);       //display
    tmp = myValue;    //for other digit
    tmp = tmp & 0x0F;  //mask upper nibble
    tmp = tmp | 0x30;  //make it ASCII
    TRANS(tmp);       //display
    TRANS(':');       //display separator
}

void SDELAY(int ms)
{
    for(i=0;i<ms;i++)
        for(j=0;j<135;j++);
}

Review Questions

1. True or false. All of the RAM contents of the DS1306 are volatile.
2. What locations of RAM in the DS1306 are set aside for the clock and date?
3. What locations of RAM in the DS1306 are set aside for general-purpose applications?
4. True or false. The DS1306 has a single pin for Dout.
5. True or false. CE is an output pin.
6. True or false. To use the DS1306 in SPI mode, we make SERMODE = VCC.
SECTION 16.4: ALARM AND INTERRUPT FEATURES OF THE DS1306

In this section, we program the alarm and interrupt features of the DS1306 chip using Assembly and C languages. These powerful features of the DS1306 can be very useful in many real-world applications. In the DS1306 there are two alarms, called Alarm0 and Alarm1, each with their own hardware interrupts. There is also a 1-Hz square wave output pin, which we discuss next. These features are accessed with the Control register shown in Figure 16-13.

<table>
<thead>
<tr>
<th>0</th>
<th>WP</th>
<th>0</th>
<th>0</th>
<th>0</th>
<th>1-Hz</th>
<th>AIE1</th>
<th>AIE0</th>
</tr>
</thead>
</table>

**WP (Write Protect)** If the WP bit is set high, the DS1306 prevents any write operation to its registers. We must make WP = 0 before we can write to any of the registers. Upon power-up, the WP bit is undefined. Therefore, we must make WP = 0 before we can write to any of the registers.

**1-Hz (1-Hz output enable)** If this bit is set HIGH, it allows the 1-Hz frequency to come out of the 1-Hz pin of the DS1306. By making it LOW, we get High-Z on the 1-Hz pin. Notice that the 1-Hz frequency is automatically generated by the DS1306, but it will not show up at the 1-Hz pin unless we set this bit to HIGH.

**AIE0** Alarm interrupt0 enable. If AIE 0= 1, the INT0 pin will be asserted LOW when all three bytes of the real time (hh:mm:ss) are the same as the alarm bytes of hh:mm:ss. Also, if AIE = 1, the cases of once-per-second, once-per-minute, and once-per-hour will assert LOW the INT0 pin.

**AIE1** Alarm interrupt1 enable. If AIE1 = 1, the INT1 pin will be asserted HIGH when all three bytes of the real time (hh:mm:ss) are the same as the alarm bytes of hh:mm:ss. Also, if AIE = 1, the cases of once-per-second, once-per-minute, and once-per-hour will assert HIGH the INT1 pin.

**Figure 16-13. DS1306 Control Register (Write location address is 8FH)**

**Programming the 1-Hz feature**

The 1-Hz pin of the DS1306 provides us a square wave output of 1-Hz frequency. Internally, the DS1306 generates the 1-Hz square wave automatically but it is blocked. We must enable the 1-Hz bit in the Control register to let it show up on the 1-Hz pin. This is shown below. Because we are writing to a single location, burst mode is not used.

```
MOVWF SSPSTAT ;middle read, active edge send
MOVWF 0x22
MOVLW 0x00
MOVWF SSPCON1 ;master SPI enable, Fosc / 64
CLRF TRISC ;make PORTC output
BSF TRISC,SDI ;except SDI
;-- send control byte to enable write first (Figure 16-13)
BSF PORTC,RC2 ;enable the RTC
```
CALL SDELAY
MOVLW 0x8F ;Control register address
CALL SPI
MOVLW 0x0 ;clear WP bit for write
CALL SPI
BCF PORTC,RC2 ;disable RTC
CALL SDELAY
--;– send control byte to enable 1 Hz signal after WP = 0
BSF PORTC,RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x8F ;Control register address
CALL SPI
MOVLW 0x04 ;enable 1 Hz signal in Control register
CALL SPI
BCF PORTC,RC2 ;disable RTC
CALL SDELAY

Alarm0, Alarm1, and interrupt

There are two time-of-day alarms in the DS1306 chip. They are referred to as Alarm0 and Alarm1. We can access Alarm0 by writing to its registers located at addresses 87H through 8AH, as shown in Table 16-3. Alarm1 is accessed by writing to its registers located at addresses 8BH through 8EH, as shown in Table 16-3. During each clock update, the RTC compares the clock registers and alarm registers. When the values stored in the timekeeping registers of 0, 1, and 2 match the values stored in the alarm registers, the corresponding alarm flag bit (IRQF0 or IRQF1) in the status register will go HIGH. See Figure 16-14. Because polling the IRQxF is too time-consuming, we can enable the AIEx bit in the Control register, and make it a hardware interrupt coming out of the INT0 and INT1 pins.

<table>
<thead>
<tr>
<th>0</th>
<th>0</th>
<th>0</th>
<th>0</th>
<th>0</th>
<th>0</th>
<th>0</th>
<th>IRQF1</th>
<th>IRQF0</th>
</tr>
</thead>
</table>

**IRQF0 (Interrupt 0 Request Flag)** The IRQF0 bit will go HIGH when all three bytes of the current real time (hh:mm:ss) are the same as the Alarm0 bytes of hh:mm:ss. Also, the cases of once-per-second, once-per-minute, and once-per-hour will assert HIGH the IRQ0 bit. We can use polling to see the status of IRQF0. However, in the Control register, if we make AIE0 = 1, IRQF0 will assert LOW the INT0 pin, making it a hardware interrupt. Any read or write of the Alarm0 registers will clear IRQF0.

**IRQF1 (Interrupt 0 Request Flag)** The IRQF1 bit will go HIGH when all three bytes of the current real time (hh:mm:ss) are the same as the Alarm1 bytes of hh:mm:ss. Also, the cases of once-per-second, once-per-minute, and once-per-hour will assert HIGH the IRQ1 bit. We can use polling to see the status of IRQF1. However, in the Control register, if we make AIE1 = 1, IRQF1 will assert HIGH the INT1 pin, making it a hardware interrupt. Any read or write of the Alarm1 registers will clear IRQF1.

Figure 16-14. Status Register (Read location address is 10H)
Alarm and IRQ output pins

The alarm interrupts of INT0 and INT1 can be programmed to occur at rates of (a) once per week, (b) once per day, (c) once per hour, (d) once per minute, and (f) once per second. Next, we look at each of these.

Once-per-day alarm

Table 16-3 shows the address locations belonging to the alarm seconds, alarm minutes, alarm hours, and alarm days. Notice the D7 bits of these locations. An alarm is generated every day when D7 of the day alarm location is set to HIGH. Therefore, to program the alarm for once-per-day, we must (a) write the desired time for the alarm into the hour, minute, and second of Alarm locations, and (b) set HIGH D7 of the alarm day. See Table 16-4. As the clock keeps the time, when all three bytes of hour, minute, and second for the real-time clock match the values in the alarm hour, minute, and second, the IRQxF flag bit in the Status register of the DS1306 will go high. We can poll the IRQxF bit in the Status register, which is a waste of microcontroller resources, or allow the hardware INTx pin to be activated upon matching the alarm time with the real time. It must be noted that in order
to use the hardware INTx pin of the DS1306 for an alarm, the interrupt-enable bit for alarm in control register (AIEx) must be set HIGH. We will examine the process shortly.

**Once-per-hour alarm**

To program the alarm for once per hour, we must set HIGH D7 of both the day alarm and hour alarm registers. See Table 16-4.

**Once-per-minute alarm**

To program the alarm for once per minute, we must set HIGH D7 of all three, day alarm, hour alarm, and minute alarm locations. See Table 16-4.

**Once-per-second alarm**

To program the alarm for once per second, we must set HIGH D7 of all four locations of alarm day, alarm hour, alarm minute, and alarm second. See Table 16-4.

**Once-per-week alarm**

To program the alarm for once per week, we must clear D7 of all four locations of alarm day, alarm hour, alarm minute, and alarm second. See Table 16-4.

### Table 16-4: DS1306 Time-of-day Alarm Mask Bits

<table>
<thead>
<tr>
<th>Seconds</th>
<th>Minutes</th>
<th>Hours</th>
<th>Days</th>
<th>Function</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>Alarm once per second</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
<td>1</td>
<td>Alarm when seconds match (once-per-minute)</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>Alarm when minutes and seconds match (once-per-hour)</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>Alarm when hours, minutes, and seconds match (once-per-day)</td>
</tr>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>Alarm when day, hours, minutes, and seconds match (once-per-week)</td>
</tr>
</tbody>
</table>

**Example 16-2**

Using Table 16-4, find the values we must place in the Alarm1 register if we want to have an alarm activated at 16:05:07, and from then on once-per-minute at 7 seconds past the minute.

**Solution:**

Because we use 24-hour clock, we have D6 = 0 for the HR register. Therefore, we have 1001 0110 for 16 in BCD. This means that we must put value 96H into register location 8D of the DS1306. Notice that D7 is 1, according to Table 16-4.

For the MIN register, we have 1000 0101 for 05 in BCD. This means that we must put value 85H into register location 8C of the DS1306. Notice that D7 is 1, according to Table 16-4.

For the SEC register we have 0000 0111 for 07 in BCD. This means that we must put value 07H into register location 8B of the DS1306. Notice that D7 is 0 according to Table 16-4.

For once-per-minute to work, we must make sure that D7 of Alarm1 day is also set to 1. See Table 16-4.
Using INT0 of DS1306 to activate the PIC18 interrupt

We can connect the INT0 bit of the DS1306 to the external interrupt pin of the PIC18 (INT0). See Figure 16-15. This allows us to perform a task once per day, once per minute, and so on. Example 16-2 shows the values needed for the Alarm0 registers. Program 16-6 uses the Alarm0 interrupt (INT0) to send the message “YES” to the serial port once per minute, at exactly 8 seconds past the minute.

![Figure 16-15. DS1306 Connection to PIC18 with Hardware INT0](image)

Program 16-6

```assembly
; Program 16-6
D1uL EQU D'2' ; 1 microsecond delay byte
DR1uL EQU 0x0D ; register for 1 microsecond delay
ORG 0x00
BRA MAIN ; bypass INT vector table
ORG 0x08
BTFSC INTCON,INT0IF ; Was it INT0?
BRA INT0_ISR ; yes, go to INT0 ISR
RETFIE
ORG 0x28
"

; -- initialize SPI, INT0, and USART
MAIN CLRF TRISC ; make PORTC output
BSF TRISC,SDI ; except SDI
BSF TRISC,RX ; and RX
BSF TRISB,INT0 ; make RB0 input for interrupt
MOVlw 0x00
MOVWF SSPSTAT ; middle read, active edge send
MOVlw 0x22
MOVWF SSPCON1 ; master SPI enable, Fosc / 64
BCF INTCON2,INTEDGO ; make INTO negative-edge triggered
BSF INTCON,INT0IE ; enable INTO
MOVlw B'00100000'; enable transmit and choose low baud
MOVWF TXSTA ; write to reg
MOVlw D'15'; ; 9600 bps (Fosc / (64 * Speed) - 1)
MOVWF SPBRG ; write to reg
```

626
BCF TRISC, TX ;make TX pin of PORTC an output pin
BSF RCSTA, SPEN ;enable the serial port
BSF INTCON, GIE ;enable interrupts globally

;-- send control byte to enable write
BSF PORTC, RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x8F ;control register
CALL SPI
MOVLW 0x0 ;clear WP bit for write
CALL SPI
BCF PORTC, RC2 ;disable RTC
CALL SDELAY

;-- send the data
BSF PORTC, RC2 ;enable for multibyte write
MOVLW 0x87 ;Alarm0 address
CALL SPI ;send address
MOVLW 0x08 ;alarm at 8 seconds
CALL SPI ;send second
MOVLW 0x80 ;once-per-minute
CALL SPI ;send minute
MOVLW 0x80 ;once-per-minute
CALL SPI ;send hour
MOVLW 0x80 ;once-per-minute
CALL SPI ;send day
BCF PORTC, RC2 ;end of multibyte write
CALL SDELAY

;-- send control byte to enable INTO
BSF PORTC, RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x8F ;control register of DS1306
CALL SPI
MOVLW 0x01 ;enable INTO pin of DS1306
CALL SPI
BCF PORTC, RC2 ;disable RTC
CALL SDELAY
LOOP BRA LOOP ;wait for interrupt

;-- service Alarm0
INT0_ISR
BSF PORTC, RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x8F ;control register
CALL SPI
MOVLW 0x04 ;1 Hz on, Alarm0 off
CALL SPI
BCF PORTC, RC2 ;disable RTC
CALL SDELAY

;-- send Alarm0 seconds to reset alarm
BSF PORTC, RC2 ;enable the RTC
CALL SDELAY
MOVLW 0x87 ;Alarm0 seconds register
CALL SPI
MOV LW 0x08 ;at 8 seconds
CALL SPI
BCF PORTC,RC2 ;disable RTC
CALL SDELAY

;-- begin displaying
MOV LW upper (MESSAGE)
MOVWF TBLPTRU
MOV LW high (MESSAGE)
MOVWF TBLPTRH
MOV LW low (MESSAGE)
MOVWF TBLPTRL

NEXT TBLRD*+

MOVF TABLAT,W
IORLW 0x0
BZ OVER
CALL TRANS
BRA NEXT

;-- send control byte to enable INTO
OVER BSF PORTC,RC2 ;enable the RTC
CALL SDELAY
MOV LW 0x8F ;control register
CALL SPI
MOV LW 0x01 ;1 Hz off, Alarm0 on
CALL SPI
BCF PORTC,RC2 ;disable RTC
CALL SDELAY
BCF INTCON,INT0IF
RETFIE

;-- SPI subroutine

;-- serial data transfer subroutine
;-- delay for SPI communications
RETURN;SEE PREVIOUS PROGRAMS FOR ABOVE SUBROUTINES

;--message to be displayed upon interrupt
MESSAGE: DB 0x0A,0x0D,"Yes",0

END

The following is the C version of the above program.

//Program 16-6C
#include <pl8f458.h>
//INSERT FUNCTION PROTOTYPES
#pragma interrupt chk_isr //used for high priority interrupt only
void chk_isr (void)
{
    if (INTCONbits.INT0IF==1) //INT0 caused interrupt?
        INTO_ISR( ); //Yes. Execute INTO program
}
#pragma code My_HiPrio_Int=0x0008 //high-priority interrupt
void My_HiPrio_Int (void)
{
#pragma code

```c
void main(void)
{
    //-- initialize SPI, INTO, and USART
    TRISC=0x90; //make PORTC output, except SDI and RX
    TRISBbits.TRISB0=1; //make RB0 input for interrupt
    SSPSTAT=0x00; //middle read, active edge send
    SSPCON1=0x22; //master SPI enable, Fosc / 64
    INTCON2bits.INTEDGO=0; //make INTO negative edge
    //triggered
    INTO=1; //enable INTO
    TXSTA=0x20; //enable transmit and choose low baud
    SPBRG=15; //9600 bps (Fosc / (64 * Speed) - 1)
    RCSTAbits.SPEN=1; //enable the serial port
    INTCONbits.GIE=1; //enable interrupts globally

    //-- send control byte to enable write
    PORTCbits.RC2=1; //enable the RTC
    MSDelay(1);
    SPI(0x8F); //control register address
    PortCbits.RC2=0; //disable RTC
    MSDelay(1);

    //-- send the data
    PORTCbits.RC2=1; //enable the RTC
    MSDelay(1);
    SPI(0x87); //AlarmO address
    SPI(0x08); //alarm at 8 seconds
    SPI(0x80); //once-per-minute
    SPI(0x80); //once-per-minute
    SPI(0x80); //once-per-minute
    PORTCbits.RC2=0; //disable RTC
    MSDelay(1);

    //-- send control byte to enable INTO
    PORTCbits.RC2=1; //enable the RTC
    MSDelay(1);
    SPI(0x8F); //control register
    SPI(0x01); //enable INTO
    PORTCbits.RC2=0; //disable RTC
    MSDelay(1);
    while(1); //wait for interrupt
}
```

void INTO_ISR()
{
    unsigned char mess[]={0x0D,0x0A,'Y','E','S',0};
    unsigned char i;
    PORTCbits.RC2=1; //enable the RTC
}
```

CHAPTER 16: SPI PROTOCOL AND DS1306 RTC INTERFACING
MSDelay(1);
SPI(0x8F); //control register
SPI(0x04); //1 Hz on, Alarm0 off
PORTCbits.RC2=0; //disable RTC
MSDelay(1);
//-- send Alarm0 seconds to reset alarm
PORTCbits.RC2=1; //enable the RTC
MSDelay(1);
SPI(0x87); //Alarm0 seconds register
SPI(0x08); //at 8 seconds
PORTCbits.RC2=0; //disable RTC
MSDelay(1);
//-- begin sending the data
for(i=0; mess[i]!=0; i++)
    TRANS(mess[i]);
//-- send control byte to enable INT0
PORTCbits.RC2=1; //enable the RTC
MSDelay(1);
SPI(0x8F); //control register
SPI(0x01); //1 Hz off bits. Alarm0 on
PORTCbits.RC2=0; //turn off RTC
INTCONbits.INT0IF=0;
}
//--SEE PREVIOUS EXAMPLES FOR SUBROUTINES

In the last program, we send a message to the serial port to indicate that the alarm has occurred. We can use the 32 kHz output to sound an actual alarm. Because 32 kHz is too high a frequency for human ears, however, we can use multiple D flip flops to bring down the frequency. See Figure 16-16. The modification of Program 16-6 for Figure 16-16 is left to the reader.

Figure 16-16. DS1306 Connection to PIC18 with Buzzer Control

Review Questions

1. Which bit of the Control register belongs to the 1-Hz pin?
2. True or false. The INT0 pin is an input for the DS1306.
3. True or false. The INT0 pin is active-LOW.
4. Which bit of the Control register belongs to the Alarm1 interrupt?
5. Give the address locations for Alarm1.
SUMMARY

This chapter began by describing the SPI bus connection and protocol. We also discussed the function of each pin of the DS1306 RTC chip. The DS1306 can be used to provide a real-time clock and dates for many applications. Various features of the RTC were explained, and numerous programming examples were given.

PROBLEMS

SECTION 16.1: SPI BUS PROTOCOL

1. True or false. The SPI bus needs an external clock.
2. True or false. The SPI CE is active-LOW.
3. True or false. The SPI bus has a single Din pin.
4. True or false. The SPI bus has multiple Dout pins.
5. True or false. When the SPI device is used as a slave, the SCLK is an input pin.
6. True or false. In SPI devices, data is transferred in 8-bit chunks.
7. True or false. In SPI devices, each bit of information (data, address) is transmitted with a single clock pulse.
8. True or false. In SPI devices, the 8-bit data is followed by an 8-bit address.
9. In term of data pins, what is the difference between the SPI and 3-wire connections?
10. How does the SPI protocol distinguish between the read and write cycles?

SECTION 16.2: DS1306 RTC INTERFACING AND PROGRAMMING

11. The DS1306 DIP package is a(n) _______ -pin package.
12. Which pin is assigned as primary Vcc?
13. In the DS1306, how many pins are designated as address/data pins?
14. True or false. The DS1306 needs an external crystal oscillator.
15. True or false. The DS1306’s crystal oscillator and heat affect the time-keeping accuracy.
16. In DS1306, what is the maximum year that it can provide?
17. Describe the functions of pins SDI, SDO, and SCLK.
18 CE is an _______ (input, output) pin.
19. The CE pin is normally _______ (LOW, HIGH) and needs a _______ (LOW, HIGH) signal to be activated.
20. Who keeps the contents of the DS1306 time and date registers if power to the primary Vcc pin is cut off?
21. Vbat pin stands for __________ and is an _______ (input, output) pin.
22. For the DS1306 chip, pin Vcc2 is connected to _______ (Vcc, GND).
23. SERMODE is an _______ (input, output) pin and it is connected to _______ for SPI mode.
24. Vcc1 is an _______ (input, output) pin and it is connected to _______ voltage.
25. 1-Hz is an _______ (input, output).
26. INTO is an _______ (input, output) pin.
27. 32KHz is an _______ (input, output) pin.
28. INT1 is an _______ (input, output) pin.
29. DS1306 has a total of _____ bytes of locations. Give the addresses for read and write operations.
30. What are the contents of the DS1306 time and date registers if power to the Vcc pin is lost?
31. What are the contents of the general-purpose RAM locations if power to the Vcc1 is lost?
32. When does the DS1306 switch to a battery energy source?
33. What are the addresses assigned to the real-time clock (time) registers?
34. What are the addresses assigned to the calendar?
35. Which register is used to set the AM/PM mode? Give the bit location of that register.
36. Which register is used to set the 24-hour mode? Give the bit location of that register.
37. At what memory location does the DS1306 store the year 2007?
38. What is the address of the last location of RAM for the DS1306?
39. True or false. The DS1306 provides data in BCD format only.
40. Write a program to get the year data in BCD and send it to ports PORTB and PORTD.
41. Write a program to get the hour and minute data in BCD and send it to ports PORTB and PORTD.
42. Write a program to set the time to 9:15:05 PM.
43. Write a program to set the time to 22:47:19.
44. Write a program to set the date to May 14, 2009.
45. What are the roles of Vbat and Vcc2?

SECTION 16.3: DS1306 RTC PROGRAMMING IN C

46. Write a C program to display the time in AM/PM mode.
47. Write a C program to get the year data in BCD and send it to ports PORTB and PORTD.
48. Write a C program to get the hour and minute data and send it to ports PORTB and PORTD.
49. Write a C program to set the time to 9:15:05 PM.
50. Write a C program to set the time to 22:47:19.
51. Write a C program to set the date to May 14, 2009.
52. In Question 51, how does the RTC keep track of the century?

SECTION 16.4: ALARM AND INTERRUPT FEATURES OF THE DS1306

53. INTO is an _______ (input, output) pin and active-_______(LOW, HIGH).
54. 1-Hz is an _______ (input, output) pin.
55. Give the bit location of the Control register belonging to the alarm interrupt. Show how to enable it.
56. Give the bit location of the Control register belonging to the 1-Hz pin. Show how to enable it.
57. Give the bit location of the Status register belonging to the Alarm0 interrupt.
58. Give the bit location of the Status register belonging to the Alarm1 interrupt.
59. True or False. For the 32KHz output pin, the frequency is set and cannot be changed.
60. Give sources of interrupts that can activate the INT1 pin.
61. Why do we want to direct the AIE0 (Alarm0 flag) to an IRQ pin?
62. What is the difference between the IRQF0 and AIE0 bits?
63. What is the difference between the IRQF1 and AIE1 bits?
64. How do we allow the square wave to come out of the 1-Hz pin?
65. Which register is used to set the once-per-second Alarm1?
66. Explain how the IRQ1F pin is activated due to the once-per-minute alarm option.

ANSWERS TO REVIEW QUESTIONS

SECTION 16.1: SPI BUS PROTOCOL

1. True
2. True
3. False
4. False
5. In single-byte mode, after each byte, the CE pin must go LOW before the next cycle. In burst mode, the CE pin stays HIGH for the duration of the burst (multibyte) transfer.

SECTION 16.2: DS1306 RTC INTERFACING AND PROGRAMMING

1. True. Only if Vbat is connected to an external battery.
2. 7
3. 96
4. True
5. Pin 11 is SCLK.
6. False. SERMODE = Vcc

SECTION 16.3: DS1306 RTC PROGRAMMING IN C

1. True
2. 0–6
3. 20–7FH
4. True
5. False
6. False

SECTION 16.4: ALARM AND INTERRUPT FEATURES OF THE DS1306

1. Bit 2
2. False
3. True
4. Bit 1
5. Byte addresses of 0B–0E (in hex) for read and 8B–8E (in hex) for write
CHAPTER 17

MOTOR CONTROL: RELAY, PWM, DC, AND STEPPER MOTORS

OBJECTIVES

Upon completion of this chapter, you will be able to:

- Describe the basic operation of a relay
- Interface the PIC18 with a relay
- Describe the basic operation of an optoisolator
- Interface the PIC18 with an optoisolator
- Describe the basic operation of a stepper motor
- Interface the PIC18 with a stepper motor
- Code PIC18 programs to control and operate a stepper motor
- Define stepper motor operation in terms of step angle, steps per revolution, tooth pitch, rotation speed, and RPM
- Describe the basic operation of a DC motor
- Interface the PIC18 with a DC motor
- Code PIC18 programs to control and operate a DC motor
- Describe how PWM is used to control motor speed
- Code CCP programs to control and operate a DC motor
- Code ECCP programs to control and operate a DC motor
This chapter discusses motor control and shows PIC18 interfacing with relays, optoisolators, stepper motors, and DC motors. In Section 17.1, the basics of relays and optoisolators are described. Then we show their interfacing with the PIC18. In Section 17.2, stepper motor interfacing with the PIC18 is shown. The characteristics of DC motors are discussed in Section 17.3, along with their interfacing to the PIC18. We will also discuss the topic of PWM (pulse width modulation). In Section 17.4, the CCP feature of PIC18 is used to control DC motors, while the ECCP usage in motor control is shown in Section 17.5. We use both Assembly and C in our programming examples.

SECTION 17.1: RELAYS AND OPTOISOLATORS

This section begins with an overview of the basic operations of electromechanical relays, solid-state relays, reed switches, and optoisolators. Then we describe how to interface them to the PIC18. We use both Assembly and C language programs to demonstrate their control.

Electromechanical relays

A relay is an electrically controllable switch widely used in industrial controls, automobiles, and appliances. It allows the isolation of two separate sections of a system with two different voltage sources. For example, a +5 V system can be isolated from a 120 V system by placing a relay between them. One such relay is called an electromechanical (or electromagnetic) relay (EMR) as shown in Figure 17-1. The EMRs have three components: the coil, spring, and contacts. In Figure 17-1, a digital +5 V on the left side can control a 12 V motor on the right side without any physical contact between them. When current flows through the coil, a magnetic field is created around the coil (the coil is energized), which causes the armature to be attracted to the coil. The armature's contact acts like a switch and closes or opens the circuit. When the coil is not energized, a spring pulls the armature to its normal state of open or closed. In the block diagram for electromechanical relays (EMR) we do not show the spring, but it does exist internally. There are all types of relays for all kinds of applications. In choosing a relay the following characteristics need to be considered:

1. The contacts can be normally open (NO) or normally closed (NC). In the NC type, the contacts are closed when the coil is not energized. In the NO, the contacts are open when the coil is unenergized.
2. There can one or more contacts. For example, we can have SPST (single pole, single throw), SPDT (single pole, double throw), and DPDT (double pole, double throw) relays.
3. The voltage and current needed to energize the coil. The voltage can vary from a few volts to 50 volts, while the current can be from a few mA to 20 mA. The relay has a minimum voltage, below which the coil will not be energized. This minimum voltage is called the "pull-in" voltage. In the datasheet for relays we might not see current, but rather coil resistance. The V/R will give you the pull-in current. For example, if the coil voltage is 5 V, and the coil resistance is 500 ohms, we need a minimum of 10 mA (5 V/500 ohms = 10 mA) pull-in current.
4. The maximum DC/AC voltage and current that can be handled by the contacts. This is in the range of a few volts to hundreds of volts, while the current can be from a few amps to 40 A or more, depending on the relay. Notice the difference between this voltage/current specification and the voltage/current needed for energizing the coil. The fact that one can use such a small amount of voltage/current on one side to handle a large amount of voltage/current on the other side is what makes relays so widely used in industrial controls. Examine Table 17-1 for some relay characteristics.

Table 17-1: Selected DIP Relay Characteristics (www.Jameco.com)

<table>
<thead>
<tr>
<th>Part No.</th>
<th>Contact Form</th>
<th>Coil Volts</th>
<th>Coil Ohms</th>
<th>Contact Volts-Current</th>
</tr>
</thead>
<tbody>
<tr>
<td>106462CP</td>
<td>SPST-NO</td>
<td>5 VDC</td>
<td>500</td>
<td>100 VDC-0.5 A</td>
</tr>
<tr>
<td>138430CP</td>
<td>SPST-NO</td>
<td>5 VDC</td>
<td>500</td>
<td>100 VDC-0.5 A</td>
</tr>
<tr>
<td>106471CP</td>
<td>SPST-NO</td>
<td>12 VDC</td>
<td>1000</td>
<td>100 VDC-0.5 A</td>
</tr>
<tr>
<td>138448CP</td>
<td>SPST-NO</td>
<td>12 VDC</td>
<td>1000</td>
<td>100 VDC-0.5 A</td>
</tr>
<tr>
<td>129875CP</td>
<td>DPDT</td>
<td>5 VDC</td>
<td>62.5</td>
<td>30 VDC-1 A</td>
</tr>
</tbody>
</table>

Figure 17-1. Relay Diagrams
Driving a relay

Digital systems and microcontroller pins lack sufficient current to drive the relay. While the relay's coil needs around 10 mA to be energized, the microcontroller's pin can provide a maximum of 1–2 mA current. For this reason, we place a driver, such as the ULN2803, or a power transistor between the microcontroller and the relay as shown in Figure 17-2.

![Figure 17-2. PIC18 Connection to Relay](image)

Program 17-1 turns the lamp on and off shown in Figure 17-2 by energizing and de-energizing the relay every few ms.

```assembly
; Program 17-1
R3 SET 0x20 ; set aside location 0x20 for R3
R4 SET 0x21 ; loc. 0x21 for R4
ORG 0H
BCF TRISB, 0 ; PORTB.0 as output
OVER BSF PORTB, 0 ; turn on the lamp
CALL DELAY
BCF PORTB, 0 ; turn off the lamp
CALL DELAY
BRA OVER
DELAY MOVLW 0xFF
MOVWF R4
D1 MOVLW 0xFF
MOVWF R3
D2 NOP
NOP
DECF R3, F
BNZ D2
DECF R4, F
BNZ D1
RETURN
```
Solid-state relay

Another widely used relay is the solid-state relay. See Table 17-2. In this relay, there is no coil, spring, or mechanical contact switch. The entire relay is made out of semiconductor materials. Because no mechanical parts are involved in solid-state relays, their switching response time is much faster than that of electromechanical relays. Another advantage of the solid-state relay is its greater life expectancy. The life cycle for the electromechanical relay can vary from a few hundred thousand to a few million operations. Wear and tear on the contact points can cause the relay to malfunction after a while. Solid-state relays, however, have no such limitations. Extremely low input current and small packaging make solid-state relays ideal for microprocessor and logic control switching. They are widely used in controlling pumps, solenoids, alarms, and other power applications. Some solid-state relays have a phase control option, which is ideal for motor-speed control and light-dimming applications. Figure 17-3 shows control of a fan using a solid-state relay (SSR).

Table 17-2: Selected Solid-State Relay Characteristics (www.Jameco.com)

<table>
<thead>
<tr>
<th>Part No.</th>
<th>Contact Style</th>
<th>Control Volts</th>
<th>Contact Volts</th>
<th>Contact Current</th>
</tr>
</thead>
<tbody>
<tr>
<td>143058CP</td>
<td>SPST</td>
<td>4–32 VDC</td>
<td>240 VAC</td>
<td>3 A</td>
</tr>
<tr>
<td>139053CP</td>
<td>SPST</td>
<td>3–32 VDC</td>
<td>240 VAC</td>
<td>25 A</td>
</tr>
<tr>
<td>162341CP</td>
<td>SPST</td>
<td>3–32 VDC</td>
<td>240 VAC</td>
<td>10 A</td>
</tr>
<tr>
<td>172591CP</td>
<td>SPST</td>
<td>3–32 VDC</td>
<td>60 VDC</td>
<td>2 A</td>
</tr>
<tr>
<td>175222CP</td>
<td>SPST</td>
<td>3–32 VDC</td>
<td>60 VDC</td>
<td>4 A</td>
</tr>
<tr>
<td>176647CP</td>
<td>SPST</td>
<td>3–32 VDC</td>
<td>120 VDC</td>
<td>5 A</td>
</tr>
</tbody>
</table>

Figure 17-3. PIC18 Connection to a Solid-State Relay
Reed switch

Another popular switch is the reed switch. When the reed switch is placed in a magnetic field, the contact is closed. When the magnetic field is removed, the contact is forced open by its spring. See Figure 17-4. The reed switch is ideal for moist and marine environments where it can be submerged in fuel or water. Reed switches are also widely used in dirty and dusty atmospheres because they are tightly sealed.

![Figure 17-4. Reed Switch and Magnet Combination](image)

Optoisolator

In some applications we use an optoisolator (also called optocoupler) to isolate two parts of a system. An example is driving a motor. Motors can produce what is called back EMF, a high-voltage spike produced by a sudden change of current as indicated in the \( V = L \frac{di}{dt} \) formula. In situations such as printed circuit board design, we can reduce the effect of this unwanted voltage spike (called ground bounce) by using decoupling capacitors (see Appendix C). In systems that have inductors (coil winding), such as motors, a decoupling capacitor or a diode will not do the job. In such cases we use optoisolators. An optoisolator has an LED (light-emitting diode) transmitter and a photosensor receiver, separated from each other by a gap. When current flows through the diode, it transmits a signal light across the gap and the receiver produces the same signal with the same phase but a different current and amplitude. See Figure 17-5. Optoisolators are also widely used in communication equipment such as modems. This device allows a computer to be connected to a telephone line without risk of damage from power surges. The gap between the transmitter and receiver of optoisolators prevents the electrical current surge from reaching the system.
Interfacing an optoisolator

The optoisolator comes in a small IC package with four or more pins. There are also packages that contain more than one optoisolator. When placing an optoisolator between two circuits, we must use two separate voltage sources, one for each side, as shown in Figure 17-6. Unlike relays, no drivers need to be placed between the microcontroller/digital output and the optoisolators.
Review Questions

1. Give one application where would you use a relay.
2. Why do we place a driver between the microcontroller and the relay?
3. What is an NC relay?
4. Why are relays that use coils called electromechanical relays?
5. What is the advantage of a solid-state relay over EMR?
6. What is the advantage of an optoisolator over an EM relay?

SECTION 17.2: STEPPER MOTOR INTERFACING

This section begins with an overview of the basic operation of stepper motors. Then we describe how to interface a stepper motor to the PIC18. Finally, we use Assembly language programs to demonstrate control of the angle and direction of stepper motor rotation.

Stepper motors

A stepper motor is a widely used device that translates electrical pulses into mechanical movement. In applications such as disk drives, dot matrix printers, and robotics, the stepper motor is used for position control. Stepper motors commonly have a permanent magnet rotor (also called the shaft) surrounded by a stator (see Figure 17-7). There are also steppers called variable reluctance stepper motors that do not have a permanent magnet rotor. The most common stepper motors have four stator windings that are paired with a center-tapped common as shown in Figure 17-8. This type of stepper motor is commonly referred to as a four-phase or unipolar stepper motor. The center tap allows a change of current direction in

Figure 17-7. Rotor Alignment
each of two coils when a winding is grounded, thereby resulting in a polarity change of the stator. Notice that while a conventional motor shaft runs freely, the stepper motor shaft moves in a fixed repeatable increment, which allows one to move it to a precise position. This repeatable fixed movement is possible as a result of basic magnetic theory where poles of the same polarity repel and opposite poles attract. The direction of the rotation is dictated by the stator poles. The stator poles are determined by the current sent through the wire coils. As the direction of the current is changed, the polarity is also changed causing the reverse motion of the rotor. The stepper motor discussed here has a total of six leads: four leads representing the four stator windings and two commons for the center-tapped leads. As the sequence of power is applied to each stator winding, the rotor will rotate. There are several widely used sequences, each of which has a different degree of precision. Table 17-3 shows a two-phase, four-step stepping sequence.

Note that although we can start with any of the sequences in Table 17-3, once we start we must continue in the proper order. For example, if we start with step 3 (0110), we must continue in the sequence of steps 4, 1, 2, etc.

Table 17-3: Normal Four-Step Sequence

<table>
<thead>
<tr>
<th>Clockwise</th>
<th>Step #</th>
<th>Winding A</th>
<th>Winding B</th>
<th>Winding C</th>
<th>Winding D</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>3</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td></td>
</tr>
<tr>
<td>Counter-clockwise</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Step angle

How much movement is associated with a single step? This depends on the internal construction of the motor, in particular the number of teeth on the stator and the rotor. The step angle is the minimum degree of rotation associated with a single step. Various motors have different step angles. Table 17-4 shows some step angles for various motors. In Table 17-4, notice the term steps per revolution. This is the total number of steps needed to rotate one complete rotation or 360 degrees (e.g., 180 steps × 2 degrees = 360).

It must be noted that perhaps contrary to one’s initial impression, a stepper motor does not need more terminal leads for the stator to achieve smaller steps. All the stepper motors

![Figure 17-8. Stator Winding Configuration](image_url)
discussed in this section have four leads for the stator winding and two COM wires for the center tap. Although some manufacturers set aside only one lead for the common signal instead of two, they always have four leads for the stators. See Example 17-1. Next we discuss some associated terminology in order to understand the stepper motor further.

### Example 17-1

Describe the PIC18 connection to the stepper motor of Figure 17-9 and code a program to rotate it continuously.

**Solution:**

The following steps show the PIC18 connection to the stepper motor and its programming:

1. Use an ohmmeter to measure the resistance of the leads. This should identify which COM leads are connected to which winding leads.
2. The common wire(s) are connected to the positive side of the motor's power supply. In many motors, +5 V is sufficient.
3. The four leads of the stator winding are controlled by four bits of the PIC18 port (RB0–RB3). Because the PIC18 lacks sufficient current to drive the stepper motor windings, we must use a driver such as the ULN2003 to energize the stator. Instead of the ULN2003, we could have used transistors as drivers, as shown in Figure 17-11. However, notice that if transistors are used as drivers, we must also use diodes to take care of inductive current generated when the coil is turned off. One reason that using the ULN2003 is preferable to the use of transistors as drivers is that the ULN2003 has an internal diode to take care of back EMF.

```plaintext
<table>
<thead>
<tr>
<th>MyReg</th>
<th>SET 0x30</th>
<th>; loc 30H for MyReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>R2</td>
<td>SET 0x20</td>
<td>; loc 20H for R2 Reg</td>
</tr>
<tr>
<td></td>
<td>CLRQ TRISB</td>
<td>; Port B as output</td>
</tr>
<tr>
<td></td>
<td>MOVFLW 0x66</td>
<td>; load step sequence</td>
</tr>
<tr>
<td></td>
<td>MOVFF MyReg</td>
<td>; issue sequence to motor</td>
</tr>
<tr>
<td></td>
<td>RRNCF MyReg,F</td>
<td>; rotate right clockwise</td>
</tr>
<tr>
<td></td>
<td>CALL DELAY</td>
<td>; wait</td>
</tr>
<tr>
<td></td>
<td>BRA BACK</td>
<td>; keep going</td>
</tr>
<tr>
<td></td>
<td>MOVFLW 0xFF</td>
<td></td>
</tr>
<tr>
<td></td>
<td>MOVWF R2</td>
<td></td>
</tr>
<tr>
<td></td>
<td>NOP</td>
<td></td>
</tr>
<tr>
<td></td>
<td>DECF R2,F</td>
<td></td>
</tr>
<tr>
<td></td>
<td>BNZ D1</td>
<td></td>
</tr>
<tr>
<td></td>
<td>RETURN</td>
<td></td>
</tr>
<tr>
<td></td>
<td>END</td>
<td></td>
</tr>
</tbody>
</table>
```

Change the value of DELAY to set the speed of rotation. We can use the single-bit instructions BSF and BCF instead of RRNCF to create the sequences.
Steps per second and rpm relation

The relation between rpm (revolutions per minute), steps per revolution, and steps per second is as follows.

$$Steps \ per \ second = \frac{rpm \times Steps \ per \ revolution}{60}$$

The 4-step sequence and number of teeth on rotor

The switching sequence shown earlier in Table 17-3 is called the 4-step switching sequence because after four steps the same two windings will be "ON". How much movement is associated with these four steps? After completing every four steps, the rotor moves only one tooth pitch. Therefore, in a stepper motor with 200 steps per revolution, the rotor has 50 teeth because $4 \times 50 = 200$ steps are needed to complete one revolution. This leads to the conclusion that the minimum step angle is always a function of the number of teeth on the rotor. In other words, the smaller the step angle, the more teeth the rotor passes. See Example 17-2.

Example 17-2

Give the number of times the four-step sequence in Table 17-3 must be applied to a stepper motor to make an 80-degree move if the motor has a 2-degree step angle.

Solution:
A motor with a 2-degree step angle has the following characteristics:
Step angle: 2 degrees    Steps per revolution: 180
Number of rotor teeth: 45    Movement per 4-step sequence: 8 degrees
To move the rotor 80 degrees, we need to send 10 consecutive 4-step sequences, because $10 \times 4 \times 2 \ degrees = 80 \ degrees$. 

Figure 17-9. PIC18 Connection to Stepper Motor
Looking at Example 17-2, one might wonder what happens if we want to move 45 degrees, because the steps are 2 degrees each. To allow for finer resolutions, all stepper motors allow what is called an 8-step switching sequence. The 8-step sequence is also called half-stepping, because in the 8-step sequence each step is half of the normal step angle. For example, a motor with a 2-degree step angle can be used as a 1-degree step angle if the sequence of Table 17-5 is applied.

Table 17-5: Half-Step 8-Step Sequence

<table>
<thead>
<tr>
<th>Step #</th>
<th>Winding A</th>
<th>Winding B</th>
<th>Winding C</th>
<th>Winding D</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>2</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>4</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>5</td>
<td>0</td>
<td>1</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>6</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>7</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>8</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
</tbody>
</table>

Motor speed

The motor speed, measured in steps per second (steps/s), is a function of the switching rate. Notice in Example 17-1 that by changing the length of the time delay loop, we can achieve various rotation speeds.

Holding torque

The following is a definition of holding torque: “With the motor shaft at standstill or zero rpm condition, the amount of torque, from an external source, required to break away the shaft from its holding position. This is measured with rated voltage and current applied to the motor.” The unit of torque is ounce-inch (or kg-cm).

Wave drive 4-step sequence

In addition to the 8-step and the 4-step sequences discussed earlier, there is another sequence called the wave drive 4-step sequence. It is shown in Table 17-6. Notice that the 8-step sequence of Table 17-5 is simply the combination of the wave drive 4-step and normal 4-step normal sequences shown in Tables 17-6 and 17-3, respectively. Experimenting with the wave drive 4-step sequence is left to the reader.

Table 17-6: Wave Drive 4-Step Sequence

<table>
<thead>
<tr>
<th>Step #</th>
<th>Winding A</th>
<th>Winding B</th>
<th>Winding C</th>
<th>Winding D</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>2</td>
<td>0</td>
<td>1</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>0</td>
<td>0</td>
<td>1</td>
<td>0</td>
</tr>
<tr>
<td>4</td>
<td>0</td>
<td>0</td>
<td>0</td>
<td>1</td>
</tr>
</tbody>
</table>
Table 17-7: Selected Stepper Motor Characteristics (www.Jameco.com)

<table>
<thead>
<tr>
<th>Part No.</th>
<th>Step Angle</th>
<th>Drive System</th>
<th>Volts</th>
<th>Phase Resistance</th>
<th>Current</th>
</tr>
</thead>
<tbody>
<tr>
<td>151861CP</td>
<td>7.5</td>
<td>unipolar</td>
<td>5 V</td>
<td>9 ohms</td>
<td>550 mA</td>
</tr>
<tr>
<td>171601CP</td>
<td>3.6</td>
<td>unipolar</td>
<td>7 V</td>
<td>20 ohms</td>
<td>350 mA</td>
</tr>
<tr>
<td>164056CP</td>
<td>7.5</td>
<td>bipolar</td>
<td>5 V</td>
<td>6 ohms</td>
<td>800 mA</td>
</tr>
</tbody>
</table>

Unipolar versus bipolar stepper motor interface

There are three common types of stepper motor interfacing: universal, unipolar, and bipolar. They can be identified by the number of connections to the motor. A universal stepper motor has eight, while the unipolar has six and the bipolar has four. The universal stepper motor can be configured for all three modes, while the unipolar can be either unipolar or bipolar. Obviously the bipolar cannot be configured for universal nor unipolar mode. Table 17-7 shows selected stepper motor characteristics. Figure 17-10 shows the basic internal connections of all three type of configurations.

Unipolar stepper motors can be controlled using the basic interfacing shown in Figure 17-11, whereas the bipolar stepper requires H-Bridge circuitry. Bipolar stepper motors require a higher operational current than the unipolar; the advantage of this is a higher holding torque.

Using transistors as drivers

Figure 17-11 shows an interface to a unipolar stepper motor using transistors. Diodes are used to reduce the back EMF spike created when the coils are energized and de-energized, similar to the electromechanical relays discussed earlier. TIP transistors can be used to supply higher current to the motor. Table 17-8 lists the common industrial Darlington transistors. These transistors can accommodate higher voltages and currents.
Figure 17-11. Using Transistors for Stepper Motor Driver

Table 17-8: Darlington Transistor Listing

<table>
<thead>
<tr>
<th>NPN</th>
<th>PNP</th>
<th>Vceo (volts)</th>
<th>Ic (amps)</th>
<th>hfe (common)</th>
</tr>
</thead>
<tbody>
<tr>
<td>TIP110</td>
<td>TIP115</td>
<td>60</td>
<td>2</td>
<td>1000</td>
</tr>
<tr>
<td>TIP111</td>
<td>TIP116</td>
<td>80</td>
<td>2</td>
<td>1000</td>
</tr>
<tr>
<td>TIP112</td>
<td>TIP117</td>
<td>100</td>
<td>2</td>
<td>1000</td>
</tr>
<tr>
<td>TIP120</td>
<td>TIP125</td>
<td>60</td>
<td>5</td>
<td>1000</td>
</tr>
<tr>
<td>TIP121</td>
<td>TIP126</td>
<td>80</td>
<td>5</td>
<td>1000</td>
</tr>
<tr>
<td>TIP122</td>
<td>TIP127</td>
<td>100</td>
<td>5</td>
<td>1000</td>
</tr>
<tr>
<td>TIP140</td>
<td>TIP145</td>
<td>60</td>
<td>10</td>
<td>1000</td>
</tr>
<tr>
<td>TIP141</td>
<td>TIP146</td>
<td>80</td>
<td>10</td>
<td>1000</td>
</tr>
<tr>
<td>TIP142</td>
<td>TIP147</td>
<td>100</td>
<td>10</td>
<td>1000</td>
</tr>
</tbody>
</table>
Controlling stepper motor via optoisolator

In the first section of this chapter we examined the optoisolator and its use. Optoisolators are widely used to isolate the stepper motor’s EMF voltage and keep it from damaging the digital/microcontroller system. This is shown in Figure 17-12. See Examples 17-3 and 17-4.

![Diagram of controlling stepper motor via optoisolator](image)

**Figure 17-12. Controlling Stepper Motor via Optoisolator**

**Example 17-3**

A switch is connected to pin RD7 (PORTD.7). Write a program to monitor the status of SW and perform the following:
(a) If SW = 0, the stepper motor moves clockwise.
(b) If SW = 1, the stepper motor moves counterclockwise.

**Solution:**

- **MyReg**
  - `SET 0x30 ;loc 30H for MyReg`
  - `BSF TRISD, RD7 ;RD7 as input pin`
  - `CLRF TRISB ;Port B as output`
  - `MOVLW 0x66 ;load step sequence`
  - `MOVWF MyReg`

- **BACK**
  - `BTFSS PORTD, RD7 ;check the SW`
  - `BRA OVER ;It is high. Make it clockwise`
  - `MOVFF MyReg, PORTB ;issue sequence to motor`
  - `RRNCF MyReg, F ;rotate right clockwise`
  - `CALL DELAY ;wait`
  - `BRA BACK ;keep going`

- **OVER**
  - `MOVFF MyReg, PORTB ;issue sequence to motor`
  - `RLNCF MyReg, F ;rotate left clockwise`
  - `CALL DELAY ;wait`
  - `BRA BACK ;keep going`
Stepper motor control with PIC18 C

The PIC18 C version of the stepper motor control is given below. In this program we could have used \(<<\) (shift left) and \(>>\) (shift right) as was shown in Chapter 7.

```c
#include <pl8f458.h>
void main()
{
    TRISB=0x0; //PORTB as output
    while(1)
    {
        PORTB = 0x66;
        MSDelay(100);
        PORTB = 0xCC;
        MSDelay(100);
        PORTB = 0x99;
        MSDelay(100);
        PORTB = 0x33;
        MSDelay(100);
    }
}
```

Example 17-4

A switch is connected to pin RD7. Write a C program to monitor the status of SW and perform the following:

(a) If SW = 0, the stepper motor moves clockwise.
(b) If SW = 1, the stepper motor moves counterclockwise.

Solution:

```c
#include <pl8f458.h>
#define SW PORTDbits.RD7
void MSDelay(int ms);
void main()
{
    TRISD=0x80; //RD7 as input pin
    TRISB=0x0; //PORTB as output
    while(1)
    {
        if(SW == 0)
        {
            PORTB = 0x66;
            MSDelay(100);
            PORTB = 0xCC;
            MSDelay(100);
            PORTB = 0x99;
            MSDelay(100);
            PORTB = 0x33;
            MSDelay(100);
        }
        else
        {
            PORTB = 0x66;
            MSDelay(100);
            PORTB = 0x33;
        }
    }
}
```
Example 17-4 Cont.

```c
MSDelay(100);
PORTB = 0x99;
MSDelay(100);
PORTB = 0xCC;
MSDelay(100);
}

void MSDelay(unsigned int value)
{
    unsigned int x, y;
    for(x=0; x<1275; x++)
        for(y=0; y<value; y++);
}
```

Review Questions

1. Give the 4-step sequence of a stepper motor if we start with 0110.
2. A stepper motor with a step angle of 5 degrees has ____ steps per revolution.
3. Why do we put a driver between the microcontroller and the stepper motor?

SECTION 17.3: DC MOTOR INTERFACING AND PWM

This section begins with an overview of the basic operation of DC motors. Then we describe how to interface a DC motor to the PIC18. Finally, we use Assembly and C language programs to demonstrate the concept of pulse width modulation (PWM) and show how to control the speed and direction of a DC motor.

DC motors

A direct current (DC) motor is another widely used device that translates electrical pulses into mechanical movement. In the DC motor we have only + and − leads. Connecting them to a DC voltage source moves the motor in one direction. By reversing the polarity, the DC motor will move in the opposite direction. One can easily experiment with the DC motor. For example, small fans used in many motherboards to cool the CPU are run by DC motors. By connecting their leads to the + and − voltage source, the DC motor moves. While a stepper motor moves in steps of 1 to 15 degrees, the DC motor moves continuously. In a stepper motor, if we know the starting position we can easily count the number of steps the motor has moved and calculate the final position of the motor. This is not possible in a DC motor. The maximum speed of a DC motor is indicated in rpm and is given in the data sheet. The DC motor has two rpms: no-load and loaded. The manufacturer’s data sheet gives the no-load rpm. The no-load rpm can be from a few thousand to tens of thousands. The rpm is reduced when moving a load and it decreases as the load is increased. For example, a drill turning a screw has a much lower rpm speed than when it is in the no-load situation. DC motors also have voltage and current ratings. The nominal voltage is the voltage for that motor under normal conditions, and can be from 1 to 150 V, depending on the motor. As we increase the voltage, the rpm goes up. The current rating is the current consump-
tion when the nominal voltage is applied with no load, and can be from 25 mA to a few amps. As the load increases, the rpm is decreased, unless the current or voltage provided to the motor is increased, which in turn increases the torque. With a fixed voltage, as the load increases, the current (power) consumption of a DC motor is increased. If we overload the motor it will stall, and that can damage the motor due to the heat generated by high current consumption.

**Unidirectional control**

Figure 17-13 shows the DC motor rotation for clockwise (CW) and counterclockwise (CCW) rotations. See Table 17-9 for selected DC motors.

<table>
<thead>
<tr>
<th>Part No.</th>
<th>Nominal Volts</th>
<th>Volt Range</th>
<th>Current</th>
<th>RPM</th>
<th>Torque</th>
</tr>
</thead>
<tbody>
<tr>
<td>154915CP</td>
<td>3 V</td>
<td>1.5–3 V</td>
<td>0.070 A</td>
<td>5,200</td>
<td>4.0 g-cm</td>
</tr>
<tr>
<td>154923CP</td>
<td>3 V</td>
<td>1.5–3 V</td>
<td>0.240 A</td>
<td>16,000</td>
<td>8.3 g-cm</td>
</tr>
<tr>
<td>177498CP</td>
<td>4.5 V</td>
<td>3–14 V</td>
<td>0.150 A</td>
<td>10,300</td>
<td>33.3 g-cm</td>
</tr>
<tr>
<td>181411CP</td>
<td>5 V</td>
<td>3–14 V</td>
<td>0.470 A</td>
<td>10,000</td>
<td>18.8 g-cm</td>
</tr>
</tbody>
</table>

**Figure 17-13. DC Motor Rotation (Permanent Magnet Field)**

**Bidirectional control**

With the help of relays or some specially designed chips we can change the direction of the DC motor rotation. Figures 17-14 through 17-17 show the basic concepts of H-Bridge control of DC motors.

Figure 17-14 shows the connection of an H-Bridge using simple switches. All the switches are open, which does not allow the motor to turn.
Figure 17-14. H-Bridge Motor Configuration

Figure 17-15 shows the switch configuration for turning the motor in one direction. When switches 1 and 4 are closed, current is allowed to pass through the motor.

Figure 17-15. H-Bridge Motor Clockwise Configuration

Figure 17-16 shows the switch configuration for turning the motor in the opposite direction from the configuration of Figure 17-15. When switches 2 and 3 are closed, current is allowed to pass through the motor.
Figure 17-16. H-Bridge Motor Counterclockwise Configuration

Figure 17-17 shows an invalid configuration. Current flows directly to ground, creating a short circuit. The same effect occurs when switches 1 and 3 are closed or switches 2 and 4 are closed.

Table 17-10 shows some of the logic configurations for the H-Bridge design.

H-Bridge control can be created using relays, transistors, or a single IC solution such as the L293. When using relays and transistors, you must ensure that invalid configurations do not occur.
Table 17-10: Some H-Bridge Logic Configurations for Figure 17-14

<table>
<thead>
<tr>
<th>Motor Operation</th>
<th>SW1</th>
<th>SW2</th>
<th>SW3</th>
<th>SW4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Off</td>
<td>Open</td>
<td>Open</td>
<td>Open</td>
<td>Open</td>
</tr>
<tr>
<td>Clockwise</td>
<td>Closed</td>
<td>Open</td>
<td>Open</td>
<td>Open</td>
</tr>
<tr>
<td>Counterclockwise</td>
<td>Open</td>
<td>Closed</td>
<td>Closed</td>
<td>Open</td>
</tr>
<tr>
<td>Invalid</td>
<td>Closed</td>
<td>Closed</td>
<td>Closed</td>
<td>Closed</td>
</tr>
</tbody>
</table>

Although we do not show the relay control of an H-Bridge, Example 17-5 shows a simple program to operate a basic H-Bridge.

Example 17-5

A switch is connected to pin RD7 (PORTD.7). Using a simulator, write a program to simulate the H-Bridge in Table 17-10. We must perform the following:
(a) If DIR = 0, the DC motor moves clockwise.
(b) If DIR = 1, the DC motor moves counterclockwise.

Solution:

```
BCF TRISB,0          ;PORTB.0 as output for switch 1
BCF TRISB,1          ;   1 "    switch 2
BCF TRISB,2          ;   2 "    switch 3
BCF TRISB,3          ;   3 "    switch 4
BSF TRISD,7          ;make PORTD.7 an input DIR

MONITOR:
  BTFSS PORTD,7
  BRA CLOCKWISE
  BSF PORTB,0          ;switch 1
  BCF PORTB,1          ;switch 2
  BCF PORTB,2          ;switch 3
  BSF PORTB,3          ;switch 4
  BRA MONITOR

CLOCKWISE:
  BCF PORTB,0          ;switch 1
  BSF PORTB,1          ;switch 2
  BSF PORTB,2          ;switch 3
  BCF PORTB,3          ;switch 4
  BRA MONITOR
```

View the results on your simulator. This example is for simulation only and should not be used on a connected system.


Figure 17-18 shows the connection of the L293 to an PIC18. Be aware that the L293 will generate heat during operation. For sustained operation of the motor, use a heat sink. Example 17-6 shows control of the L293.
Figure 17-18. Bidirectional Motor Control Using an L293 Chip

Example 17-6

Figure 17-18 shows the connection of an L293. Add a switch to pin RD7 (PORTD.7). Write a program to monitor the status of SW and perform the following:

(a) If SW = 0, the DC motor moves clockwise.
(b) If SW = 1, the DC motor moves counterclockwise.

Solution:

```
BCF TRISB,0 ;enable the chip
BCF TRISB,1
BCF TRISB,2
BSF TRISD,7
BSF PORTB,0
CHK BTFSS PORTD,7 ;turn the motor counterclockwise
BRA CWISE
BCF PORTB,1
BSF PORTB,2
BRA CHK
CWISE BSF PORTB,1
BCF PORTB,2 ;turn motor clockwise
BRA CHK
```
Pulse width modulation (PWM)

The speed of the motor depends on three factors: (a) load, (b) voltage, and (c) current. For a given fixed load we can maintain a steady speed by using a method called pulse width modulation (PWM). By changing (modulating) the width of the pulse applied to the DC motor we can increase or decrease the amount of power provided to the motor, thereby increasing or decreasing the motor speed. Notice that, although the voltage has a fixed amplitude, it has a variable duty cycle. That means the wider the pulse, the higher the speed. PWM is so widely used in DC motor control that some microcontrollers come with the PWM circuitry embedded in the chip. In such microcontrollers all we have to do is load the proper registers with the values of the high and low portions of the desired pulse, and the rest is taken care of by the microcontroller. This allows the microcontroller to do other things. For microcontrollers without PWM circuitry, we must create the various duty cycle pulses using software, which prevents the microcontroller from doing other things. The ability to control the speed of the DC motor using PWM is one reason that DC motors are preferable over AC motors. AC motor speed is dictated by the AC frequency of the voltage applied to the motor and the frequency is generally fixed. As a result, we cannot control the speed of the AC motor when the load is increased. As was shown earlier, we can also change the DC motor’s direction and torque. See Figure 17-19 for PWM comparisons.

<table>
<thead>
<tr>
<th>Power</th>
<th>DC</th>
</tr>
</thead>
<tbody>
<tr>
<td>¼ POWER</td>
<td>25%</td>
</tr>
<tr>
<td>½ POWER</td>
<td>50%</td>
</tr>
<tr>
<td>¾ POWER</td>
<td>75%</td>
</tr>
<tr>
<td>FULL</td>
<td>100%</td>
</tr>
</tbody>
</table>

Figure 17-19. Pulse Width Modulation Comparison

DC motor control with optoisolator

As we discussed in the first section of this chapter, the optoisolator is indispensable in many motor control applications. Figures 17-20 and 17-21 show the connections to a simple DC motor using a bipolar and a MOSFET transistor. Notice that the PIC18 is protected from EMI created by motor brushes by using an optoisolator and a separate power supply.

Figures 17-20 and 17-21 show optoisolators for control of single directional motor control, and the same principle should be used for most motor applications. Separating the power supplies of the motor and logic will reduce the possibility of damage to the control circuitry.
Figure 17-20 shows the connection of a bipolar transistor to a motor. Protection of the control circuit is provided by the optoisolator. The motor and PIC18 use separate power supplies. The separation of power supplies also allows the use of high-voltage motors. Notice that we use a decoupling capacitor across the motor; this helps reduce the EMI created by the motor. The motor is switched on by clearing bit P1.0.

![Figure 17-20. DC Motor Connection using a Darlington Transistor](image)

Figure 17-21 shows the connection of a MOSFET transistor. The optoisolator protects the PIC18 from EMI. The zener diode is required for the transistor to reduce gate voltage below the rated maximum value. See Example 17-7.

![Figure 17-21. DC Motor Connection using a MOSFET Transistor](image)
Example 17-7

Refer to the figure in this example. Write a program to monitor the status of the switch and perform the following:

(a) If PORTD.7 = 1, the DC motor moves with 25% duty cycle pulse.
(b) If PORTD.7 = 0, the DC motor moves with 50% duty cycle pulse.

Solution:

```
BCF TRISB,RB0 ; PORTB.0 as output
BSF TRISD,RD7 ; PORTD.7 as input
BCF PORTB,RB0 ; turn off motor

CHK
BTFSS PORTD,RD7
BRA PWM_50
BSF PORTB,RB0 ; high portion of pulse
CALL DELAY
BCF PORTB,RB0 ; low portion of pulse
CALL DELAY
CALL DELAY
CALL DELAY
BRA CHK

PWM_50
BSF PORTB,RB0 ; high portion of pulse
CALL DELAY
CALL DELAY
BCF PORTB,RB0 ; low portion of pulse
CALL DELAY
CALL DELAY
BRA CHK
```

CHAPTER 17: MOTOR CONTROL: RELAY, PWM, DC, AND STEPPER MOTORS 659
DC motor control and PWM using C

Examples 17-8 through 17-10 show the PIC18 C version of the earlier programs controlling the DC motor.

Example 17-8

Refer to Figure 17-18 for connection of the motor. A switch is connected to pin RD7. Write a C program to monitor the status of SW and perform the following:

(a) If SW = 0, the DC motor moves clockwise.
(b) If SW = 1, the DC motor moves counterclockwise.

Solution:

```c
#include <pl8f458.h>
#define SW PORTDbits.RD7
#define ENABLE PORTBbits.RB0
#define MTR_1 PORTBbits.RB1
#define MTR_2 PORTBbits.RB2

void main()
{
    TRISD=0x80;  // make RD7 input pin
    TRISB=0x0;   // make PORTB output
    SW = 1;
    ENABLE = 0;
    MTR_1 = 0;
    MTR_2 = 0;

    while(1)
    {
        ENABLE = 1;
        if(SW == 1)
        {
            MTR_1 = 1;
            MTR_2 = 0;
        }
        else
        {
            MTR_1 = 0;
            MTR_2 = 1;
        }
    }
}
```
Example 17-9

Refer to the figure in this example. Write a C program to monitor the status of SW and perform the following:

(a) If SW = 0, the DC motor moves with 50% duty cycle pulse.
(b) If SW = 1, the DC motor moves with 25% duty cycle pulse.

Solution:

```c
#include <p18f458.h>
#define SW PORTDbits.RD7
#define MTR PORTBbits.RB1
void MSDelay(unsigned int value);
void main()
{
  TRISD=0x80; // make RD7 input pin
  TRISB=0xFD; // make RB1 output pin
  while(1)
  {
    if(SW == 1)
    {
      MTR = 1;
      MSDelay(25);
      MTR = 0;
      MSDelay(75);
    }
    else
    {
      MTR = 1;
      MSDelay(50);
      MTR = 0;
      MSDelay(50);
    }
  }
  void MSDelay(unsigned int value)
  {
    unsigned char x, y;
    for(x=0; x<1275; x++)
      for(y=0; y<value; y++);
  }
```

CHAPTER 17: MOTOR CONTROL: RELAY, PWM, DC, AND STEPPER MOTORS 661
Example 17-10

Refer to Figure 17-20 for connection to the motor. Two switches are connected to pins RD0 and RD1. Write a C program to monitor the status of both switches and perform the following:

<table>
<thead>
<tr>
<th>SW2 (RD1)</th>
<th>SW1 (RD0)</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>DC motor moves slowly (25% duty cycle).</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>DC motor moves moderately (50% duty cycle).</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>DC motor moves fast (75% duty cycle).</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>DC motor moves very fast (100% duty cycle).</td>
</tr>
</tbody>
</table>

Solution:

```c
#include <p18f458.h>
#define MTR PORTBbits.RB1
void MSDelay(unsigned int value);

void main()
{
    unsigned int duty;
    TRISB = 0xFD;
    TRISD = 0xFF;
    while(1)
    {
        duty = PORTD&0x03;
        duty++;
        duty *= 25;
        MTR = 1;
        MSDelay(duty);
        MTR = 0;
        MSDelay(100-duty);
    }
}
```

Review Questions

1. True or false. The permanent magnet field DC motor has only two leads for + and − voltages.
2. True or false. Just like a stepper motor, one can control the exact angle of a DC motor’s move.
3. Why do we put a driver between the microcontroller and the DC motor?
4. How do we change a DC motor’s rotation direction?
5. What is stall in a DC motor?
6. True or false. PWM allows the control of a DC motor with the same phase, but different amplitude pulses.
7. The RPM rating given for the DC motor is for _________ (no-load, loaded).
SECTION 17.4: PWM MOTOR CONTROL WITH CCP

We examined the CCP (Compare Capture Pulse-Width-Modulation) part of the PIC452/458 in Chapter 15. One of the features of the CCP is the pulse width modulation (PWM) as we saw in Section 15.4 of Chapter 15. In this section we use the PWM feature of the CCP to control DC motors. Review the programming of the PWM in Section 15.4 before embarking on this section.

DC motor control with CCP

Recall from Section 15.4 that the PWM part of the CCP is programmed by using the PR2 and Timer2 registers. Program 17-2 is the rewrite of Example 17-7 using the PWM feature of the CCP1. Notice that Program 17-2 is the modified version of Program 15-5 in Chapter 15. Program 17-2C is the C version of Program 17-2. In Program 17-2 (and 17-2C), an input switch is being monitored. If the switch is low, the PIC18 creates a 50% duty cycle PWM using the CCP1 module. If the switch is high, a 25% duty cycle PWM is created. Recall from Chapter 15 that we must use PR2 and Timer2 registers for creating PWM pulses.

![DC Motor Control Using CCP1 Pin](image)

Program 17-2

```assembly
BCF TRISC, CCP1 ;make PWM output pin
BSF TRISD, RD7  ;make RD7 input pin
MOVLW 0x3C      ;PWM MODE, 11 for DC1B1:B0
MOVF CCP1CON
MOVLW D'100'    ;set period to 100 * Fosc/4
MOVF PR2
MOVLW 0x01      ;Timer2, 4 prescale, no postscaler
MOVF T2CON
AGAIN BTFSS PORTD, RD7 ;Is the switch high?
BRA T2DUTY ;no, then 50%
MOVLW D'25'     ;25% duty cycle
BRA LOAD
T2DUTY MOVLW D'50'; 50% duty cycle
```

Figure 17-22: DC Motor Control Using CCP1 Pin
BRA LOAD
LOAD MOVWF CCPR1L ;load duty cycle
CLRF TMR2 ;clear Timer2
BSF T2CON,TMR2ON ;turn on Timer2
BCF PIR1,TMR2IF ;clear Timer2 flag
OVER BTFSS PIR1,TMR2IF ;wait for end of period
BRA OVER
GOTO AGAIN ;continue

The following is the C version of the above program.

//Program 17-2C
#include <p18f458.h>
void main()
{
    TRISC = 0xFB; //make CCP1 output pin
    TRISD = 0x80; //make RD7 input pin
    CCP1CON = 0x3C; //PWM MODE, 11 for DC1B1:B0
    PR2=100; //set period to 100 * 16/Fosc
    T2CON=0x01; //4 prescaler, no postscaler
    while(1)
    {
        if(PORTDbits.RD7==1)
        {
            CCPR1L = 25; //25% duty cycle
        }
        else
        {
            CCPR1L = 50; //50% duty cycle
            TMR2=0x0; //clear Timer2
            PIR1bits.TMR2IF=0; //clear Timer2 flag
            T2CONbits.TMR2ON=1; //start Timer2
            while(PIR1bits.TMR2IF==0); //wait for end of period
        }
    }
}

Review Questions
1. True or false. For standard CCP1, we use the RC2 pin for PWM.
2. True or false. For standard CCP1, the CCP1 pin must be configured as output.
3. In standard CCP1, we use _________________ to set the period for PWM.
4. In standard CCP1, we use _________________ to set the duty cycle for PWM.
5. True or false. In standard CCP1, we must use Timer1 for PWM.
SECTION 17.5: DC MOTOR CONTROL WITH ECCP

The PIC18F452/458 (or 4520/4580) comes with one standard CCP and one enhanced CCP (ECCP). Indeed, in recent years the CCP module has been de-emphasized while the ECCP is becoming more prominent in the PIC18 family. The reason is that ECCP allows the implementation of the H-Bridge for bidirectional control of the DC motor in addition to the capture/compare mode present in the standard CCP. In this section, we use the ECCP feature of the PIC18 to control the DC motor. Before embarking on this section, the basic concept of ECCP programming in Chapter 15 needs to be reviewed.

Bidirectional DC motor control with ECCP

ECCP allows the implementation of the H-Bridge for bidirectional movement of the DC motor because it uses 4 pins instead of a single pin as is used in standard CCP. As we saw in Section 17.3 of this chapter, the bidirectional DC movement needs some kind of H-Bridge circuitry. The ECCP module of the PIC18 implements the entire H-Bridge circuitry internally. It uses RD7–RD4 (PORTD.7–PORTD.4) for this purpose as shown in Figures 17-23 through 17-26.

| MCLR/VPP | 1 | 40 | RB7/PGD |
| RA0/AN0/CVREF | 2 | 39 | RB6/PGC |
| RA1/AN1 | 3 | 38 | RB5/PGM |
| RA2/AN2/VREF- | 4 | 37 | RB4 |
| RA3/AN3/VREF+ | 5 | 36 | RB3/CANRX |
| RA4/T0CKI | 6 | 35 | RB2/CANTX/INT2 |
| RA5/AN4/SS/LVDIN | 7 | 34 | RB1/INT1 |
| RE0/AN5/RD | 8 | 33 | RB0/INT0 |
| RE1/AN6/WR/C1OUT | 9 | 32 | VDD |
| RE2/AN7/CS/C2OUT | 10 | 31 | VSS |
| VDD | 11 | 30 | RD7/PSP7/P1D |
| VSS | 12 | 29 | RD6/PSP6/P1C |
| OSC1/CLKI | 13 | 28 | RD5/PSP5/P1B |
| OSC2/CLK0/RA6 | 14 | 27 | RD4/PSP4/ECCP/P1A |
| RC0/T1OSO/T1CKI | 15 | 26 | RC7/RX/DT |
| RC1/T1OSI | 16 | 25 | RC6/TX/CK |
| RC2/CCP1 | 17 | 24 | RC5/SDO |
| RC3/SCK/SCL | 18 | 23 | RC4/SDI/SDA |
| RD0/PSP0/C1IN+ | 19 | 22 | RD3/PSP3/C2IN- |
| RD1/PSP1/C1IN- | 20 | 21 | RD2/PSP2/C2IN+ |

Figure 17-23. ECCP Pins for PWM in PIC18F458/4580 (452/4520)
Figure 17-24. Forward Current Flow Using ECCP (from Microchip)

Figure 17-25. Reverse Current Flow Using ECCP (from Microchip)
EPWM1M1:EPWM1M0 PWM output pin configuration. It allows the use of a single pin for the capture/compare mode, or four pins for the PWM. In compare/capture mode, only pin PA (RD4) is used. In that case, there is no selection for these two bits. In the PWM mode the options for these two bits are as follows:

- **00**: P1A is used as a modulated output. P1B, P1C, and P1D are used as I/O.
- **01**: Full-Bridge output forward. P1D modulated, P1A active. P1B and P1C inactive.
- **10**: Half-Bridge output. P1A and P1D modulated with deadband control, P1C and P1D used as I/O.
- **11**: Full-Bridge output reverse. P1B modulated, P1C active. P1A and P1D inactive.

EDC1B0:EDC1B1 PWM Duty Cycle least-significant bits. Used in PWM only. The least-significant bits (Bit 1 and Bit 0) of the 10-bit duty cycle register are used in PWM. The ECCPR1L register is used as Bit 2 to Bit 9 of the 10-bit duty cycle register.

ECCP1M3–ECC1M0 ECCP1 Mode Select

- **0000**: ECCP1 is off
- **0001**: Reserved
- **0010**: Compare Mode. Toggle ECCP1 output pin on match. (ECCP1IF bit is set.)
- **0011**: Reserved
- **0100**: Capture mode, every falling edge
- **0101**: Capture mode, every rising edge
- **0110**: Capture mode, every 4th rising edge
- **0111**: Capture mode, every 16th rising edge
- **1000**: Compare mode. Initialize ECCP1 pin low; on compare match, force CCP1 pin HIGH. (ECCP1IF is set.)
- **1001**: Compare mode. Initialize CCP1 pin HIGH; on compare match, force CCP1 pin LOW. (ECCP1IF is set.)
- **1010**: Compare mode. Generate software interrupt on compare match. (ECCP1IF bit is set, ECCP1 pin is unaffected.)
- **1011**: Compare mode. Trigger special event. (ECCP1IF bit is set, and Timer1 or Timer3 is reset to zero.)
- **1100**: PWM Mode; P1A, P1C active-HIGH; P1B and P1D active-HIGH
- **1101**: PWM Mode; P1A, P1C active-HIGH; P1B and P1D active-LOW
- **1110**: PWM Mode; P1A, P1C active-LOW; P1B and P1D active-HIGH
- **1111**: PWM Mode; P1A, P1C active-LOW; P1B and P1D active-LOW

Figure 17-26. ECCP1 Control Register. (This register selects one of the operation modes of Capture, Compare, or PWM of ECCP1)
Program 17-3 shows Full-Bridge implementation of the PWM for ECCP module. For the implementation of Half-Bridge and other applications of PWM using the ECCP module, see the PIC18 manual.

Program 17-3

;Program 17-3
CLRF TRISD ;make PORTD output
MOVLW D'100'
MOVWF PR2 ;period = 100 * 16/Fosc
MOVLW D'50'
MOVWF ECCPR1L ;duty = 50%
MOVWF ECCP1CON ;reverse full-bridge PWM
MOVFW 0x24
MOVLW T2CON ;4 postscaler, turn on Timer2
AGAIN CLRF TMR2 ;start pulse
BCF PIR1,TMR2IF ;clear flag
WAIT BTFSS PIR1,TMR2IF ;wait for period
BRA WAIT
BRA AGAIN ;do it again

The following is the C version of the above program.

//Program 17-3C
#include <p18f458.h>

void main()
{
    TRISD=0; //make PORTD output
    PR2=100; //period = 100 * 16/Fosc
    ECCPR1L=50; //duty = 50%
    ECCP1CON=0xCF; //reverse full-bridge PWM
    T2CON=0x24; //4 postscaler, turn on Timer2

    while(1)
    {
        TMR2=0; //start pulse
        PIR1bits.TMR2IF=0; //clear flag
        while(PIR1bits.TMR2IF==0); //wait for period
    }
}

Review Questions

1. True or false. For ECCP1, we use the RD3–RD0 pins for Full-Bridge.
2. True or false. For ECCP1, the P1A to P1D pins must be configured as output.
3. In ECCP1, we use _______________ to set the period for PWM.
4. In ECCP1, we use _______________ to set the duty cycle for PWM.
5. True or false. In ECCP1, we must use Timer2 for PWM.
SUMMARY

This chapter continued showing how to interface the PIC18 with real-world devices. Devices covered in this chapter were the relay, optoisolator, stepper motor, and DC motor.

First, the basic operation of relays and optoisolators was defined, along with key terms used in describing and controlling their operations. Then the PIC18 was interfaced with a stepper motor. The stepper motor was then controlled via an optoisolator using PIC18 Assembly and C programming languages.

The PIC18 was interfaced with DC motors. A typical DC motor will take electronic pulses and convert them to mechanical motion. This chapter showed how to interface the PIC18 with a DC motor. Then, simple Assembly and C programs were written to show the concept of PWM.

Control systems that require motors must be evaluated for the type of motor needed. For example, you would not want to use a stepper in a high-velocity application or a DC motor for a low-speed, high-torque situation. The stepper motor is ideal in an open-loop positional system and a DC motor is better for a high-speed conveyer belt application. DC motors can be modified to operate in a closed-loop system by adding a shaft encoder, then using a microcontroller to monitor the exact position and velocity of the motor. In the last two sections, we showed how to use CCP and ECCP features of PIC18 to control DC motors.

PROBLEMS

SECTION 17.1: RELAYS AND OPTOISOLATORS

1. True or false. The minimum voltage needed to energize a relay is the same for all relays.
2. True or false. The minimum current needed to energize a relay depends on the coil resistance.
3. Give the advantages of a solid-state relay over an EM relay.
4. True or false. In relays, the energizing voltage is the same as the contact voltage.
5. Find the current needed to energize a relay if the coil resistance is 1,200 ohms and the coil voltage is 5 V.
6. Give two applications for an optoisolator.
7. Give the advantages of an optoisolator over an EM relay.
8. Of the EM relay and solid-state relay, which has the problem of back EMF?
9. True or false. The greater the coil resistance, the worse the back EMF voltage.
10. True or false. We should use the same voltage sources for both the coil voltage and contact voltage.

SECTION 17.2: STEPPER MOTOR INTERFACING

11. If a motor takes 90 steps to make one complete revolution, what is the step angle for this motor?
12. Calculate the number of steps per revolution for a step angle of 7.5 degrees.
13. Finish the normal four-step sequence clockwise if the first step is 0011 (binary).
14. Finish the normal four-step sequence clockwise if the first step is 1100 (binary).
15. Finish the normal four-step sequence counterclockwise if the first step is 1001 (binary).
16. Finish the normal four-step sequence counterclockwise if the first step is 0110 (binary).
17. What is the purpose of the ULN2003 placed between the PIC18 and the stepper motor? Can we use that for 3A motors?
18. Which of the following cannot be a sequence in the normal four-step sequence for a stepper motor?
   (a) CCH  (b) DDH  (c) 99H  (d) 33H
19. What is the effect of a time delay between issuing each step?
20. In Question 19, how can we make a stepper motor go faster?

SECTION 17.3: DC MOTOR INTERFACING AND PWM

21. Which motor is best for moving a wheel exactly 90 degrees?
22. True or false. Current dissipation of a DC motor is proportional to the load.
23. True or false. The rpm of a DC motor is the same for no-load and loaded.
24. The rpm given in data sheets is for __________ (no-load, loaded).
25. What is the advantage of DC motors over AC motors?
26. What is the advantage of stepper motors over DC motors?
27. True or false. Higher load on a DC motor slows it down if the current and voltage supplied to the motor are fixed.
28. What is PWM, and how is it used in DC motor control?
29. A DC motor is moving a load. How do we keep the rpm constant?
30. What is the advantage of placing an optoisolator between the motor and the microcontroller?

ANSWERS TO REVIEW QUESTIONS

SECTION 17.1: RELAYS AND OPTOISOLATORS

1. With a relay we can use a 5 V digital system to control 12 V–120 V devices such as horns and appliances.
2. Because microcontroller/digital outputs lack sufficient current to energize the relay, we need a driver.
3. When the coil is not energized, the contact is closed.
4. When current flows through the coil, a magnetic field is created around the coil, which causes the armature to be attracted to the coil.
5. It is faster and needs less current to get energized.
6. It is smaller and can be connected to the microcontroller directly without a driver.

SECTION 17.2: STEPPER MOTOR INTERFACING

1. 0110, 0011, 1001, 1100 for clockwise; and 0110, 1100, 1001, 0011 for counterclockwise
2. 72
3. Because the microcontroller pins do not provide sufficient current to drive the stepper motor
SECTION 17.3: DC MOTOR INTERFACING AND PWM

1. True
2. False
3. Because microcontroller/digital outputs lack sufficient current to drive the DC motor, we need a driver.
4. By reversing the polarity of voltages connected to the leads
5. The DC motor is stalled if the load is beyond what it can handle.
6. False
7. No-load

SECTION 17.4: PWM MOTOR CONTROL WITH CCP

1. True
2. True
3. PR2
4. CCPR1L
5. False

SECTION 17.5: DC MOTOR CONTROL WITH ECCP

1. False
2. True
3. PR2
4. CCPR1L
5. True
APPENDIX A

PIC18 INSTRUCTIONS: FORMAT AND DESCRIPTION

OVERVIEW

In the first section of this appendix, we describe the instruction format of the PIC18. Special emphasis is placed on the instructions using both WREG and file registers. This section includes a list of machine cycles (clock counts) for each of the PIC18 instructions.

In the second section of this appendix, we describe each instruction of the PIC18. In many cases, a simple programming example is given to clarify the instruction.
This Appendix deals mainly with PIC18 instructions. In Section A.1, we describe the instruction formats and categories. In Section A.2, we describe each instruction of PIC18 with some examples.

SECTION A.1: PIC18 INSTRUCTION FORMATS AND CATEGORIES

As shown in Figure A-1, the PIC18 instructions fall into five categories:

1. Bit-oriented instructions
2. Instructions using a literal value
3. Byte-oriented instructions
4. Table read and write instructions
5. Control instructions using branch and call

In this section, we describe the format and syntax with special emphasis placed on byte-oriented instructions. For some of the instructions, the reader needs to review the concepts of access bank and bank registers in Chapter 6 (Section 6.3).

**Bit-oriented instructions**

The bit-oriented instructions perform operations on a specific bit of a file register. After the operation, the result is placed back in the same file register. For example, the “BCF f,b,a” instruction clears a specific bit of fileReg. See Table A-1. In these types of instructions, the b is the specific bit of the fileReg, which can be 0 to 7, representing the D0 to D7 bits of the register. The fileReg location can be in the bank register called access bank (if a = 0) or a location within other bank registers (if a = 1). Notice that if a = 0, the assembler assumes the access bank automatically.

**Table A-1: Bit-Oriented Instructions** (from Microchip datasheet)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
</tr>
</thead>
<tbody>
<tr>
<td>BCF f, b, a</td>
<td>Bit Clear f</td>
<td>1</td>
</tr>
<tr>
<td>BSF f, b, a</td>
<td>Bit Set f</td>
<td>1</td>
</tr>
<tr>
<td>BTFSC f, b, a</td>
<td>Bit Test f, Skip if Clear</td>
<td>1 (2 or 3)</td>
</tr>
<tr>
<td>BTFSS f, b, a</td>
<td>Bit Test f, Skip if Set</td>
<td>1 (2 or 3)</td>
</tr>
<tr>
<td>BTG f, d, a</td>
<td>Bit Toggle f</td>
<td>1</td>
</tr>
</tbody>
</table>

Look at the examples that follow for clarification of bit-oriented instructions:

674
Byte-oriented File Register operations

```
| 15 10 9 8 7 0 |
| OPCODE | d | a | f (FILE #) |
| ADDWF MYREG, W, B |
```

- d = 0 for result destination to be WREG Register
- d = 1 for result destination to be File Register (f)
- a = 0 to force Access Bank
- a = 1 for BSR to select bank
- f = 8-bit File Register address

Byte to Byte move operations (2-word)

```
| 15 12 11 0 |
| OPCODE | f (Source FILE #) |
| MOVFF MYREG1, MYREG2 |
```

- f = 12-bit File Register address

Bit-oriented File Register operations

```
| 15 12 11 9 8 7 0 |
| OPCODE | b (BIT #) | a | f (FILE #) |
| BSF MYREG, bit, B |
```

- b = 3-bit position of bit in File Register (f)
- a = 0 to force Access Bank
- a = 1 for BSR to select bank
- f = 8-bit File Register address

Literal operations

```
| 15 8 7 0 |
| OPCODE | k (literal) |
| MOVWL 0x7F |
```

- k = 8-bit immediate value

Control operations

CALL, GOTO, and Branch operations

```
| 15 8 7 0 |
| OPCODE | n<7:0> (literal) |
| GOTO label |
```

```
| 15 12 11 0 |
| 1111 | n<19:8> (literal) |
```

- n = 20-bit immediate value

Figure A-1. General Formatting of PIC18 Instructions (From MicroChip)
BCF PORTB,5 ;clear bit D5 of PORTB
BCF TRISB,4 ;clear bit D4 of TRISC reg
BTG PORTC,7 ;toggle bit D7 of PORTC
BTG PORTD,0 ;toggle bit D0 of PORTD
BSF STATUS,C ;set carry flag to one

The following example uses the fileReg in the access bank:

MyReg SET 0x30 ;set aside loc 30H for MyReg
MOVLW 0x0 ;WREG = 0
MOVWF MyReg ;MyReg = 0
BTG MYReg,7 ;toggle bit D7 of MyReg
BTG MYReg,5 ;toggle bit D5 of MyReg

The following example uses the fileReg in the access bank:

MyReg SET 0x50 ;set aside loc. 50H for MyReg
MOVLW 0x0 ;WREG = 0
MOVWF MyReg ;MyReg = 0
BTG MYReg,2 ;toggle bit D2 of MyReg
BTG MYReg,4 ;toggle bit D4 of MyReg

As we discuss in Chapter 6, when using a bank other than the access bank, we must load the BSR (bank select register) with the desired bank number, which can go from 1 to F (in hex), depending on the family member. We do that by using the MOVLB instruction. Look at the following examples.

The example below uses a location in Bank 2 (RAM locations 200–2FFH).

YReg SET 0x30 ;set aside loc 30H for YReg
MOVLB 0x2 ;use Bank 2 (address loc 230H)
MOVLW 0x0 ;WREG = 0
MOVWF YReg ;YReg = 0
BTG YReg,7,1 ;toggle bit D7 of YReg in bank 2
BTG YReg,5,1 ;toggle bit D5 of YReg in bank 2

The example below uses a location in Bank 4 (RAM locations 400–4FFH).

ZReg SET 0x10 ;set aside loc 10H for ZReg
MOVLB 0x4 ;use Bank 4 (address loc 410H)
MOVLW 0x0 ;WREG = 0
MOVWF ZReg ;ZReg = 0
BSF ZReg,6,1 ;set HIGH bit D6 of ZReg in bank 4
BSF ZReg,1,1 ;set HIGH bit D1 of ZReg in bank 4
Notice that all the bit-oriented instructions start with letter B (bit). The branch instructions also start with letter B, like “BZ target” for branch if zero, but they are not bit-oriented.
<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDLW k</td>
<td>Add literal and WREG</td>
<td>1</td>
</tr>
<tr>
<td>ANDLW k</td>
<td>AND literal with WREG</td>
<td>1</td>
</tr>
<tr>
<td>IORLW k</td>
<td>Inclusive OR literal with WREG</td>
<td>1</td>
</tr>
<tr>
<td>LFSR f, k</td>
<td>Move literal (12-bit) 2nd word to FSRx 1st word</td>
<td>2</td>
</tr>
<tr>
<td>MOVLB k</td>
<td>Move literal to BSR&lt;3:0&gt;</td>
<td>1</td>
</tr>
<tr>
<td>MOV LW k</td>
<td>Move literal to WREG</td>
<td>1</td>
</tr>
<tr>
<td>MULLW k</td>
<td>Multiply literal with WREG</td>
<td>1</td>
</tr>
<tr>
<td>RET LW k</td>
<td>Return with literal in WREG</td>
<td>2</td>
</tr>
<tr>
<td>SUBLW k</td>
<td>Subtract WREG from literal</td>
<td>1</td>
</tr>
<tr>
<td>XORLW k</td>
<td>Exclusive OR literal with WREG</td>
<td>1</td>
</tr>
</tbody>
</table>

**Instructions using literal values**

In this type of instruction, an operation is performed on the WREG register and a fixed value called k. See Table A-2. Because WREG is only 8-bit, the k value cannot be greater than 8-bit. Therefore, the k value is between 0–255 (00–FF in hex). After the operation, the result is placed back in WREG. Look at the following examples for clarification:

```
MOVLW 0x45 ;WREG = 45H
ADDLW 0x24 ;WREG = 45H + 24H = 69H

MOVLW 0x35 ;WREG = 35H
ANDLW 0x0F ;WREG = 35H ANDed with 0FH = 05H

MOVLW 0x55 ;WREG = 55H
XORLW 0xAA ;WREG = 55H EX-ORed with AAH = FFH
```

**Byte-oriented instructions**

There are two groups of instructions in this category. In the first group, the operation is performed on the file register and the result is placed back in the file register. The instruction “CLRF f,a” is an example in this group. See Table A-3. In the second group, the operation involves both fileReg and WREG. As a result, we have the options of placing the result in fileReg or in WREG. As an example in this group, examine the “ADDWF f,d,a” instruction. The destination for the result can be WREG (if d = 0) or fileReg (if d = 1). For the fileReg location, it can be in the access bank (if a = 0) or in other bank registers (if a = 1). Also notice that if a = 0, the assembler assumes that automatically.
Table A-3: Byte-Oriented Instructions (from Microchip datasheet)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>BYTE-ORIENTED FILE REGISTER OPERATIONS</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDWF f, d, a</td>
<td>Add WREG and f</td>
<td>1</td>
</tr>
<tr>
<td>ADDWFC f, d, a</td>
<td>Add WREG and Carry bit to f</td>
<td>1</td>
</tr>
<tr>
<td>ANDWF f, d, a</td>
<td>Add WREG with f</td>
<td>1</td>
</tr>
<tr>
<td>CLRF f, a</td>
<td>Clear f</td>
<td>1</td>
</tr>
<tr>
<td>COMF f, d, a</td>
<td>Complement f</td>
<td>1</td>
</tr>
<tr>
<td>CPFSEQ f, a</td>
<td>Compare f with WREG, skip =</td>
<td>1</td>
</tr>
<tr>
<td>CPFSGT f, a</td>
<td>Compare f with WREG, skip &gt;</td>
<td>1</td>
</tr>
<tr>
<td>CPFSLT f, a</td>
<td>Compare f with WREG, skip &lt;</td>
<td>1</td>
</tr>
<tr>
<td>DECF f, d, a</td>
<td>Decrement f</td>
<td>1</td>
</tr>
<tr>
<td>DECFSZ f, d, a</td>
<td>Decrement f, Skip if 0</td>
<td>1</td>
</tr>
<tr>
<td>DCFSNZ f, d, a</td>
<td>Decrement f, Skip if Not 0</td>
<td>1</td>
</tr>
<tr>
<td>INCf f, d, a</td>
<td>Increment f</td>
<td>1</td>
</tr>
<tr>
<td>INCFSZ f, d, a</td>
<td>Increment f, Skip if 0</td>
<td>1</td>
</tr>
<tr>
<td>INFSNZ f, d, a</td>
<td>Increment f, Skip if Not 0</td>
<td>1</td>
</tr>
<tr>
<td>IORWF f, d, a</td>
<td>Inclusive OR WREG with f</td>
<td>1</td>
</tr>
<tr>
<td>MOVf f, d, a</td>
<td>Move f</td>
<td>1</td>
</tr>
<tr>
<td>MOVFF f_s, f_d</td>
<td>Move f_s (source) to 1st word f_d (destination) 2nd word</td>
<td>2</td>
</tr>
<tr>
<td>MOVWF f, a</td>
<td>Move WREG to f</td>
<td>1</td>
</tr>
<tr>
<td>MULWF f, a</td>
<td>Multiply WREG with f</td>
<td>1</td>
</tr>
<tr>
<td>NEGF f, a</td>
<td>Negate f</td>
<td>1</td>
</tr>
<tr>
<td>RLCF f, d, a</td>
<td>Rotate Left f through Carry</td>
<td>1</td>
</tr>
<tr>
<td>RLNCf f, d, a</td>
<td>Rotate Left f (No Carry)</td>
<td>1</td>
</tr>
<tr>
<td>RRCF f, d, a</td>
<td>Rotate Right f through Carry</td>
<td>1</td>
</tr>
<tr>
<td>RRNCf f, d, a</td>
<td>Rotate Right f (No Carry)</td>
<td>1</td>
</tr>
<tr>
<td>SETf f, a</td>
<td>Set f</td>
<td>1</td>
</tr>
<tr>
<td>SUBFWB f, d, a</td>
<td>Subtract f from WREG with borrow</td>
<td>1</td>
</tr>
<tr>
<td>SUBWF f, d, a</td>
<td>Subtract WREG from f</td>
<td>1</td>
</tr>
<tr>
<td>SUBWFB f, d, a</td>
<td>Subtract WREG from f with borrow</td>
<td>1</td>
</tr>
<tr>
<td>SWAPF f, d, a</td>
<td>Swap nibbles in f</td>
<td>1</td>
</tr>
<tr>
<td>TSTFSZ f, a</td>
<td>Test f, Skip if 0</td>
<td>1</td>
</tr>
<tr>
<td>XORWF f, d, a</td>
<td>Exclusive OR WREG with f</td>
<td>1</td>
</tr>
</tbody>
</table>
Look at the following examples.

When $d = 0$ and $a = 0$:

MyReg SET 0x20 ;loc 20H for MyReg
MOVLIW 0x45 ;WREG = 45H
MOVWF MyReg ;MyReg = 45H
MOVLIW 0x23 ;WREG = 23H
ADDWF MyReg ;WREG = 68H (45H + 23H = 68H)

In the above example, the last instruction could have been coded as “ADDWF MyReg,0,0”.

When $d = 1$ and $a = 0$:

MyReg SET 0x20 ;loc 20H for MyReg
MOVLIW 0x45 ;WREG = 45H
MOVWF MyReg ;MyReg = 45H
MOVLIW 0x23 ;WREG = 23H
ADDWF MyReg,F ;MyReg = 68H (45H + 23H = 68H)

In the above example, the last instruction could have been coded as “ADDWF MyReg,F,0” or “ADDWF MyReg,1,0”. As far as the MPLAB is concerned, they mean the same thing. Notice that the use of letter F in “ADDWF MyReg,F” is being used in place of 1.

To use banks other than the access bank, we must load the BSR register first. The following example uses a location in Bank 2 (RAM location 200–2FFH).

When $d = 0$ and $a = 1$:

MyReg SET 0x30 ;set aside location 30H for MyReg
MOVLB 0x20 ;use Bank 2 (address loc 230H)
MOVLIW 0x45 ;WREG = 45H
MOVWF MyReg,1 ;MyReg = 45H (loc 230H)
MOVLIW 0x23 ;WREG = 23H
ADDWF MyReg,1 ;WREG = 68H (add loc 230H to W)

When $d = 1$ and $a = 1$:

MyReg SET 0x20 ;loc 20H for MyReg
MOVLB 0x4 ;use bank 4
MOVLIW 0x45 ;WREG = 45H
MOVWF MyReg ;MyReg = 45H (loc 420H)
MOVLIW 0x23 ;WREG = 23H
ADDWF MyReg,F,1 ;MyReg = 68H (loc 420H)
Register-indirect addressing mode uses FSRx as a pointer to RAM location. We have three registers, FSR0, FSR1, and FSR2, that can be used for pointers.

Examples:

ADDWF POSTINC0 ;add to W data pointed to by FSR0, ;also increment FSR0
ADDWF POSTINC1 ;add to W data pointed to by FSR1 ;also increment FSR1

See Example 6-6 in Chapter 6.

Table processing instructions

The table processing instructions allow us to read fixed data located in the program ROM of the PIC18. See Table A-4. They also allow us to write into the program ROM if it is Flash memory. Chapter 14 discusses the TBLRD and TBLWRT instructions in detail. It also shows how to use table read and write to access the EEPROM.

Table A-4: Table Processing Instructions (from Microchip datasheet)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
</tr>
</thead>
<tbody>
<tr>
<td>DATA ←→ PROGRAM MEMORY OPERATIONS</td>
<td></td>
<td></td>
</tr>
<tr>
<td>TBLRD*</td>
<td>Table Read</td>
<td>2</td>
</tr>
<tr>
<td>TBLRD*+</td>
<td>Table Read with post-increment</td>
<td>2</td>
</tr>
<tr>
<td>TBLRD*-</td>
<td>Table Read with post-decrement</td>
<td>2</td>
</tr>
<tr>
<td>TBLRD+*</td>
<td>Table Read with pre-increment</td>
<td>2</td>
</tr>
<tr>
<td>TBLWT*</td>
<td>Table Write</td>
<td>2</td>
</tr>
<tr>
<td>TBLWT*+</td>
<td>Table Write with post-increment</td>
<td>2</td>
</tr>
<tr>
<td>TBLWT*-</td>
<td>Table Write with post-decrement</td>
<td>2</td>
</tr>
<tr>
<td>TBLWT+*</td>
<td>Table Write with pre-increment</td>
<td>2</td>
</tr>
</tbody>
</table>

Control instructions

The control instructions such as branch and call deal mainly with flow control. See Table A-5. We must pay special attention to the target address of the control instructions. The target address for some of the branch instructions such as BZ (branch if zero) cannot be farther than 128 bytes away from the current instruction. The CALL instruction allows us to call a subroutine located anywhere in the 2M ROM space of the PIC18. See the individual instructions in the next section for further discussion on this issue.
Table A-5: Control Instructions (from Microchip datasheet)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>CONTROL OPERATIONS</strong></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BC n</td>
<td>Branch if Carry</td>
<td>1</td>
</tr>
<tr>
<td>BN n</td>
<td>Branch if Negative</td>
<td>1</td>
</tr>
<tr>
<td>BNC n</td>
<td>Branch if Not Carry</td>
<td>1</td>
</tr>
<tr>
<td>BNN n</td>
<td>Branch if Not Negative</td>
<td>1</td>
</tr>
<tr>
<td>BNOV n</td>
<td>Branch if Not Overflow</td>
<td>1</td>
</tr>
<tr>
<td>BNZ n</td>
<td>Branch if Not Zero</td>
<td>1</td>
</tr>
<tr>
<td>BOV n</td>
<td>Branch if Overflow</td>
<td>1</td>
</tr>
<tr>
<td>BRA n</td>
<td>Branch Unconditionally</td>
<td>2</td>
</tr>
<tr>
<td>BZ n</td>
<td>Branch if Zero</td>
<td>1</td>
</tr>
<tr>
<td>CALL n, s</td>
<td>Call subroutine</td>
<td>2</td>
</tr>
<tr>
<td></td>
<td>1st word</td>
<td></td>
</tr>
<tr>
<td></td>
<td>2nd word</td>
<td></td>
</tr>
<tr>
<td>CLRWD T</td>
<td>Clear Watchdog Timer</td>
<td>1</td>
</tr>
<tr>
<td>DAW —</td>
<td>Decimal Adjust WREG</td>
<td>1</td>
</tr>
<tr>
<td>GOTO n</td>
<td>Go to address</td>
<td>2</td>
</tr>
<tr>
<td></td>
<td>1st word</td>
<td></td>
</tr>
<tr>
<td></td>
<td>2nd word</td>
<td></td>
</tr>
<tr>
<td>NOP —</td>
<td>No Operation</td>
<td>1</td>
</tr>
<tr>
<td>NOP —</td>
<td>No Operation</td>
<td>1</td>
</tr>
<tr>
<td>POP —</td>
<td>Pop top of return stack (TOS)</td>
<td>1</td>
</tr>
<tr>
<td>PUSH —</td>
<td>Push top of return stack (TOS)</td>
<td>1</td>
</tr>
<tr>
<td>RCALL n</td>
<td>Relative Call</td>
<td>2</td>
</tr>
<tr>
<td>RESET</td>
<td>Software device RESET</td>
<td>1</td>
</tr>
<tr>
<td>RETFIE s</td>
<td>Return from interrupt enable</td>
<td>2</td>
</tr>
<tr>
<td>RETLW k</td>
<td>Return with literal in WREG</td>
<td>2</td>
</tr>
<tr>
<td>RETURN s</td>
<td>Return from Subroutine</td>
<td>2</td>
</tr>
<tr>
<td>SLEEP —</td>
<td>Go into standby mode</td>
<td>1</td>
</tr>
</tbody>
</table>
SECTION A.2: THE PIC18 INSTRUCTION SET

In this section we provide a brief description of each instruction with some examples.

**ADDLW K**

**Add Literal to WREG**

Function: ADD literal value of k to WREG  
Syntax: ADDLW k

This adds the literal value of k to the WREG register, and places the result back into WREG. Because register WREG is one byte in size, the operand k must also be one byte.

The ADD instruction is used for both signed and unsigned numbers. Each one is discussed separately. See Chapter 5 for discussion of signed numbers.

**Unsigned addition**

In the addition of unsigned numbers, the status of C, DC, Z, N, and OV may change. The most important of these flags is C. It becomes 1 when there is a carry from D7 out in 8-bit (D0–D7) operations.

Example:

```
MOVLW 0x45
ADDLW 0x4F
```

; WREG = 94H (45H + 4FH = 94H)  
; C = 0

Example:

```
MOVLW 0xFE
ADDLW 0x75
```

; WREG = FE + 75 = 73H  
; C = 1

Example:

```
MOVLW 0x25
ADDLW 0x42
```

; WREG = 67H (25H + 42H = 67H)  
; C = 0

Notice that in all the above examples we ignored the status of the OV flag. Although ADD instructions do affect OV, it is in the context of signed numbers that the OV flag has any significance. This is discussed next.

**Signed addition and negative numbers**

In the addition of signed numbers, special attention should be given to the overflow flag (OV) because this indicates if there is an error in the result of the addition. There are two rules for setting OV in signed number operation. The overflow flag is set to 1:

1. If there is a carry from D6 to D7 and no carry from D7 out.
2. If there is a carry from D7 out and no carry from D6 to D7.

Notice that if there is a carry both from D7 out and from D6 to D7, OV = 0.
Example:

```asm
MOVLW +D'8'; W = 0000 1000
ADDLW +D'4'; W = 0000 1100 OV = 0,
; C = 0, N = 0
```

Notice that N = D7 = 0 because the result is positive, and OV = 0 because there is neither a carry from D6 to D7 nor any carry beyond D7. Because OV = 0, the result is correct [(+8) + (+4) = (+12)].

Example:

```asm
MOVLW +D'66'; W = 0100 0010
ADDLW +D'69'; W = 1000 0101 = -121
ADDWF ; (INCORRECT) C = 0, N = D7 = 1, OV = 1
```

In the above example, the correct result is +135 [(+66) + (+69) = (+135)], but the result was -121. OV = 1 is an indication of this error. Notice that N = 1 because the result is negative; OV = 1 because there is a carry from D6 to D7 and C = 0.

Example:

```asm
MOVLW -D'12'; W = 1111 0100
ADDLW +D'18'; W = W + (+0001 0010)
; W = 0000 0110 (+6) correct
; N = 0, OV = 0, and C = 1
```

Notice above that the result is correct (OV = 0), because there is a carry from D6 to D7 and a carry from D7 out.

Example:

```asm
MOVLW -D'30'; W = 1110 0010
ADDLW +D'14'; W = W + 0000 1110
; W = 1111 0000 (-16, CORRECT)
; N = D7 = 1, OV = 0, C = 0
```

OV = 0 because there is no carry from D7 out nor any carry from D6 to D7.

Example:

```asm
MOVLW -D'126'; W = 1000 0010
ADDLW -D'127'; W = W + 1000 0001
; W = 0000 0011 (+3, INCORRECT)
; D7 = N = 0, OV = 1
```

C = 1 because there is a carry from D7 out but no carry from D6 to D7.

From the above discussion we conclude that while Carry is important in any addition, OV is extremely important in signed number addition because it is used to indicate whether or not the result is valid. As we will see in instruction "DAW", the DC flag is used in the addition of BCD numbers.
ADDWF  Add WREG and f

Function: ADD WREG and fileReg
Syntax: ADDWF f,d,a

This adds the fileReg value to the WREG register, and places the result in WREG (if d = 0) or fileReg (if d = 1).
The ADDWF instruction is used for both signed and unsigned numbers.
(See ADDLW instruction.)

Example:

MyReg SET 0x20 ;loc 20H for MyReg
MOVlw 0x45 ;WREG = 45H
MOVWF MyReg ;MyReg = 45H
MOVlw 0x4F ;WREG = 4FH
ADDWF MyReg ;WREG = 94H (45H + 4FH = 94H) ;C = 0

We can place the result in fileReg, as shown in the following example:

MyReg SET 0x20 ;loc 20H for MyReg
MOVlw 0x45 ;WREG = 45H
MOVWF MyReg ;MyReg = 45H
MOVlw 0x4F ;WREG = 4FH
ADDWF MyReg,F ;MyReg = 94H
; (45H + 4FH = 94H), C = 0

For cases of a = 0 and a = 1, see Section A.1 in this chapter.

ADDWFC  Add WREG and Carry flag to fileReg

Function: ADD WREG and Carry bit to fileReg
Syntax: ADDWFC f,d,a

This will add WREG and the C flag to fileReg (Destination = WREG + fileReg + C). If C = 1 prior to this instruction, 1 is also added to destination. If C = 0 prior to the instruction, source is added to destination plus 0. This instruction is used in multibyte additions. In the addition of 25F2H to 3189H, for example, we use the ADDWFC instruction as shown below.

Example when d = 0:
Assume we have the following data in RAM locations 0x10 and 0x11
0x10 = (F2)
0x11 = (25)

Reg_L SET 0x10 ;loc 0x10 for Reg_L
Reg_H SET 0x11 ;loc 0x11 for Reg_H
BCF STATUS,C ;make carry = 0
MOVlw 89H ;WREG = 89H
ADDWFC Reg_L,1 ;Reg_L = 89H + F2H + 0 = 7BH
; and C = 1
MOVLW 0x31 ; WREG = 31H
ADDFC Reg_2,1 ; Reg_H = 31H + 25H + 1 = 57H

Therefore the result is:
25F2H
+3189H
577BH

**ANDLW**  
**AND Literal byte with WREG**

Function: Logical AND literal value k with WREG  
Syntax: ANDLW k

This performs a logical AND on the WREG and the Literal byte operand, bit by bit, storing the result in the WREG.

Example:

MOVLW 0x39 ; W = 39H
ANDLW 0x09 ; W = 39H ANDed with 09

39H 0011 1001
09H 0000 1001
09H 0000 1001

Example:

MOVLW 32H ; W = 32H
ANDLW 50H ; AND W with 50H 0101 0000
; (W = 10H) 10H 0001 0000

**ANDWF**  
**AND WREG with fileReg**

Function: Logical AND for byte variables  
Syntax: ANDWF f,d,a

This performs a logical AND on the fileReg value and the WREG register, bit by bit, and places the result in WREG (if d = 0) or fileReg (if d = 1).

Example:

MyReg SET 0x40; set MyReg loc at 0x40
MOVLW 0x39 ; W = 39H
MOVWF MyReg ; MyReg = 39H
MOVLW 0x09
ANDWF MyReg ; 39H ANDed with 09 (W = 09)

39H 0011 1001
09H 0000 1001
09H 0000 1001
Example:

```
MyReg  SET 0x40; set MyReg loc at 0x40
MOVLW 0x32  ; W = 32H
MOVWF MyReg ; MyReg = 32H
MOVLW 0x0F  ; WREG = 0FH
ANDLW MyReg ; 32H ANDed with 0FH (W = 02)

32H 0011 0010
0FH 0000 1111
02H 0000 0010
```

We can place the result in fileReg as shown in the examples below:

```
MyReg  SET 0x40; set MyReg loc at 0x40
MOVLW 0x32  ; W = 32H
MOVWF MyReg ; MyReg = 32H
MOVLW 0x50  ; WREG = 50H
ANDLW MyReg, F ; MyReg = 09, WREG = 50H
```

The instructions below clear (mask) certain bits of the output ports, assuming the ports are configured as output ports:

```
MOVLW 0xFE
ANDWF PORTB, F ; mask PORTB.0 (D0 of Port B)
MOVLW 0x7F
ANDWF PORTC, F ; mask PORTC.7 (D7 of Port C)
MOVLW 0xF7
ANDWF PORTD, F ; mask PORTD.3 (D3 of Port D)
```

### Branch Condition

**Function:** Conditional Branch (jump)

In this type of Branch (jump), control is transferred to a target address if certain conditions are met. The following is list of branch instructions dealing with the flags:

<table>
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<tr>
<th>Branch Instruction</th>
<th>Function</th>
<th>Target Address Constraints</th>
</tr>
</thead>
<tbody>
<tr>
<td>BC</td>
<td>Branch if carry</td>
<td>jump if C = 1</td>
</tr>
<tr>
<td>BNC</td>
<td>Branch if no carry</td>
<td>jump if C = 0</td>
</tr>
<tr>
<td>BZ</td>
<td>Branch if zero</td>
<td>jump if Z = 1</td>
</tr>
<tr>
<td>BNZ</td>
<td>Branch if no zero</td>
<td>jump if Z = 0</td>
</tr>
<tr>
<td>BN</td>
<td>Branch if negative</td>
<td>jump if N = 1</td>
</tr>
<tr>
<td>BNN</td>
<td>Branch if no negative</td>
<td>jump if N = 0</td>
</tr>
<tr>
<td>BOV</td>
<td>Branch if overflow</td>
<td>jump if OV = 1</td>
</tr>
<tr>
<td>BNOV</td>
<td>Branch if no overflow</td>
<td>jump if OV = 0</td>
</tr>
</tbody>
</table>

Notice that all “Branch condition” instructions are short jumps, meaning that the target address cannot be more than -128 bytes backward or +127 bytes forward of the PC of the instruction following the jump. In other words, the target address cannot be more than -128 to +127 bytes away from the current PC. What
happens if a programmer needs to use a "Branch condition" to go to a target address beyond the -128 to +127 range? The solution is to use the "Branch condition" along with the unconditional GOTO instruction, as shown below.

```
ORG 0x100
MOVLW 0x87      ;WREG = 87H
ADDLW 0x95      ;C = 1 after addition
BNC NEXT       ;branch if C = 0
GOTO OVER      ;target more than 128 bytes away

NEXT: ...
...
...

ORG 0x5000

OVER: MOVWF PORTD
```

**BC**

**Branch if C = 1**

Function: Branch if Carry flag bit = 1
Syntax: BC target_address

This instruction branches if C = 1.

Example:

```
MOVLW 0x0     ;WREG = 0
BACK ADDLW 0x1 ;add 1 to WREG
BC EXIT      ;exit if C = 1
BRA BACK     ;keep doing it
EXIT .......
.....
```

Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.

**BCF**

**Bit Clear fileReg**

Function: Clear bit of a fileReg
Syntax: BCF f,b,a

This instruction clears a single bit of a given file register. The bit can be the directly addressable bit of a port, register, or RAM location. Here are some examples of its format:

```
BCF STATUS,C ;C = 0
BCF PORTB,5 ;CLEAR PORTB.5 (PORTB.5 = 0)
BCF PORTC,7 ;CLEAR PORTC.7 (PORTC.7 = 0)
BCF MyReg,1 ;CLEAR D1 OF File Register MyFile
```
<table>
<thead>
<tr>
<th>BN</th>
<th>Branch if N = 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Jump if Negative flag bit = 1</td>
</tr>
<tr>
<td>Syntax:</td>
<td>BN target_address</td>
</tr>
</tbody>
</table>

This instruction branches if $N = 1$. It is used in signed number addition. See ADDLW instruction. Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.

<table>
<thead>
<tr>
<th>BNC</th>
<th>Branch if no Carry</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Branch if Carry flag is 0</td>
</tr>
<tr>
<td>Syntax:</td>
<td>BNC target_address</td>
</tr>
</tbody>
</table>

This instruction examines the C flag, and if it is zero it will jump (branch) to the target address.

Example: Find the total sum of the bytes F6H, 98H, and 8AH. Save the carries in register C_Reg.

```assembly
C_Reg SET 0x20 ; set aside loc 0x20 for carries
    MOVLW 0x0 ; W = 0
    MOVWF C_Reg ; C_Reg = 0
    ADDLW 0xF6
    BNC OVER1
    INCF C_Reg,F

OVER1:    ADDLW 0x98
    BNC OVER2
    INCF C_Reg,F

OVER2:    ADDWF 0x8A
    BNC OVER3
    INCF C_Reg

OVER3:    
```

Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this.

<table>
<thead>
<tr>
<th>BNN</th>
<th>Branch if Not Negative</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Branch if Negative flag bit = 0</td>
</tr>
<tr>
<td>Syntax:</td>
<td>BNN target_address</td>
</tr>
</tbody>
</table>

This instruction branches if $N = 0$. It is used in signed number addition. See ADDLW instruction. Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.
BNOV  Branch if No Overflow

Function: Jump if overflow flag bit = 0
Syntax: BNOV target_address

This instruction branches if OV = 0. It is used in signed number addition. See ADDLW instruction. Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.

BNZ  Branch if No Zero

Function: Jump if Zero flag is 0
Syntax: BNZ target_address

This instruction branches if Z = 0.

Example:

    CLRF TRISB ;PORTB as output
    CLRF PORTB ;clear PORTB
    OVER INCF PORTB,F ;INC PORTB
    BNZ OVER ;do it until it becomes zero

Example: Add value 7 to WREG five times.

    COUNTER SET 0x20 ;loc 20H for COUNTER
    MOVLW 0x5 ;WREG = 5
    MOVWF COUNTER ;COUNTER = 05
    MOVLW 0x0 ;WREG = 0
    OVER ADDLW 0x7 ;add 7 to WREG
    DECF COUNTER,F ;decrement counter
    BNZ OVER ;do it until counter is zero

Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.

BOV  Branch if Overflow

Function: Jump if Overflow flag = 1
Syntax: BOV target_address

This instruction jumps if OV = 1. It is used in signed number addition. See ADDLW instruction. Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this issue.

APPENDIX A: PIC18 INSTRUCTIONS: FORMAT AND DESCRIPTION
BRA  Branch unconditional

Function: Branch unconditionally
Syntax: BRA target_address

BRA stands for "Branch." It transfers program execution to the target address unconditionally. The target address for this instruction must be within 1K of program memory. This is a 2-byte instruction. The first 5 bits is the opcode and the rest is the signed number displacement, which is added to the PC (program counter) of the instruction following the BRA to get the target address. Therefore, in this branch, the target address must be within −1024 to +1023 bytes of the PC (program counter) of the instruction after the BRA because the 11-bit address can take values of +1024 to −1023. This address is often referred to as a relative address because the target address is −1024 to +1023 bytes relative to the program counter (PC).

BSF  Bit Set fileReg

Function: Set bit
Syntax: BSF f, b, a

This sets HIGH the indicated bit of a file register. The bit can be any directly addressable bit of a port, register, or RAM location.

Examples:

```
BSF PORTB, 3 ; make PORTB.3 = 1
BSF PORTC, 6 ; make PORTC.6 = 1
BSF MyReg, 2 ; make bit D2 of MyReg = 1
BSF STATUS, C ; set Carry Flag C = 1
```

BTFSC  Bit Test fileReg, Skip if Clear

Function: Skip the next instruction if bit is 0
Syntax: BTFSC f, b, a

This instruction is used to test a given bit and skip the next instruction if the bit is low. The given bit can be any of the bit-addressable bits of RAM, ports, or registers of the PIC18.

Example: Monitor the PORTB.5 bit continuously and, when it becomes low, put 55H in WREG.

```
BSF TRISB, 5 ; make PORTB.5 an input bit
BTFSC PORTB, 5 ; skip if PORTB.5 = 0
HERE
BRA HERE
MOVLW 0x55 ; because PORTB.5 = 0,
; put 55H in WREG
```
Example: See if WREG has an even number. If so, make it odd.

```
BTFSC WREG,0 ;skip if it is odd
BRA NEXT
ADDLW 0x1 ;it is even, make it odd
```

**BTFSS**  
**Bit Test fileReg, Skip if Set**

Function: Skip the next instruction if bit is 1  
Syntax: `BTFSS f, b, a`

This instruction is used to test a given bit and skip the next instruction if the bit is HIGH. The given bit can be any of the bit-addressable bits of RAM, ports, or registers of the PIC18.

Example: Monitor the PORTB.5 bit continuously and when it becomes HIGH, put 55H in WREG.

```
BSF TRISB,5 ;make PORTB.5 an input bit
HERE  BTFSS PORTB,5 ;skip if PORTB.5 = 1
BRA HERE
MOVLW 55H ;because PORTB.5 = 0 WREG = 55H
```

Example: See if WREG has an odd number. If so, make it even.

```
BTFSS WREG,0 ;skip if it is even
BRA NEXT
ADDLW 0x01 ;it is even, make it odd
```

**BTG**  
**Bit Toggle fileReg**

Function: Toggle (Complement) bit  
Syntax: `BTG f, b, a`

This instruction complements a single bit. The bit can be any bit-addressable location in the PIC18.

Example: Toggle PORTB.7 a total of 150 times.

```
MOVLW 'D'150 ;WREG = 150
MOVWF COUNTER ;COUNTER = 150
BCF TRISB,7 ;make PORTB.7 an output
```

**APPENDIX A: PIC18 INSTRUCTIONS: FORMAT AND DESCRIPTION**

691
OVER BTG PORTB.7 ;toggle PORTB.7
DECF COUNTER,F ;decrement and put it in
        ;COUNTER
BNZ OVER ;do it 150 times

**BZ**  
**Branch if Zero**

Function: Branch if \(Z = 1\)
Syntax: BZ target_address

Example: Keep checking PORTB for value 99H.
SETF TRISB ;port B as input
BACK MOVFW PORTB ;get PORTB into WREG
SUBLW 0x99 ;subtract 99H from it
BZ EXIT ;if 0x99, exit
BRA BACK ;keep checking

... EXIT:

Example: Toggle PORTB 150 times.
MyReg SET 0x40 ;loc 40H for MyReg
SETF TRISB ;port B as output
MOVLW D'150' ;WREG = 150
MOVWF MyReg
BACK COMF PORTB ;toggle PORTB
DECF MyReg,F ;decrement MyReg
BZ EXIT ;if MyReg = 0, exit
BRA BACK ;keep toggling

... EXIT:

Notice that this is a 2-byte instruction; therefore, the target address cannot be more than -128 to +127 bytes away from the program counter. See Branch Condition for further discussion on this.

**CALL**

Function: Transfers control to a subroutine
Syntax: CALL k,s ;s is used for fast context switching

The Call instruction is a 4-byte instruction. The first 12 bits are used for the opcode and the rest (20 bits) are set aside for the address. A 20-bit address allows us to reach the target address anywhere in the 2M ROM space of the PIC18. If calling a subroutine, the PC register (which has the address of the instruction after the CALL) is pushed onto the stack and the stack pointer (SP) is incremented by 1. Then the program counter is loaded with the new address and control is transferred to the subroutine. At the end of the procedure, when RETURN is executed, PC is popped off the stack, which returns control to the instruction after the CALL.

Notice that CALL is a 4-byte instruction, in which 12 bits are the opcode, and the other 20 bits are the 20-bit address of an even address location. Because
all the PIC18 instructions are 2 bytes in size, the lowest address bit, A0, is automatically set to zero to make sure that the CALL instruction will not land at the middle of the targeted instruction. The 20-bit address of the CALL provides the A20–A1 part of the address and with the A0 = 0, we have the 21-bit address needed to go anywhere in the 2M address space of the PIC18.

We have two options for the “CALL k,s” instruction. They are s = 0, and s = 1. When s = 0, it is simply calling a subroutine. With s = 1, we are calling a subroutine and we are also asking the CPU to save the three major registers of WREG, STATUS, and BSR in internal buffers (shadow registers) for the purpose of context-switching. This fast context-switching can be used only in the main subroutine because the depth of the shadow registers is only one. That means no nested call with the s = 1. Look at the following case:

```
ORG 0x0
MAIN
    ....
    ....
    ....
    CALL M_SUB,1 ;call and save the registers
    MOVLW 0x55 ;address of this instruction is saved on stack
    ....
```

```
;-----------------------------
ORG 0x2000
M_SUB
    ....
    ....
    CALL Y_SUB ;we cannot use CALL Y_SUB,1
    MOVLW 0xAA ;address of this instruction is saved on stack
    ....
    ....
    RETURN,1 ;return to caller and restore the registers
    ;notice the s = 1 for RETURN
```

```
;-----------------------------
ORG 0x3000
Y_SUB
    ....
    ....
    RETURN
```

As shown in RETURN instruction, we also have two options for the RETURN: s = 0 and s = 1. If we use s = 1 for the CALL, we must also use s = 1 for the RETURN. Notice that “CALL Target” with no number after it is interpreted as s = 0 by the assembler. Likewise, the “RETURN” with no number after it is interpreted as s = 0 by the assembler.
**CLRF**

<table>
<thead>
<tr>
<th>Function:</th>
<th>Clear fileReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>CLRF f, a</td>
</tr>
</tbody>
</table>

This instruction clears the entire byte in the fileReg. All bits of the register are cleared to 0.

Example:

```plaintext
MyReg SET 0x20 ;loc 20H for MyReg
CLRF MyReg ;clear MyReg
CLRF TRISB ;clear TRISB (make PORTB output)
CLRF PORTB ;clear PORTB
CLRF TMR01L ;TMR0L = 0
```

Notice that in this instruction the result can be placed in fileReg only and there is no option for the WREG to be used as the destination.

**CLRWDT**

Function: Clear Watchdog Timer

Syntax: CLRWDT

This instruction clears the Watchdog Timer.

**COMF**

<table>
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<tr>
<th>Function:</th>
<th>Complement the fileReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>COMF f, d, a</td>
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</table>

This complements the contents of a given fileReg. The result is the 1's complement of the register; that is, 0s become 1s and 1s become 0s. The result can be placed in WREG (if \(d = 0\)) or fileReg (if \(d = 1\)).

Example:

```plaintext
MOVLW 0x0 ;WREG = 0
MOVWF TRISB ;Make PORTB an output port
MOVLW 0x55 ;WREG = 01010101
MOVWF PORTB
AGAIN COMF PORTB,F ;complement (toggle) PORTB
CALL DELAY
BRA AGAIN ;continuously (notice WREG = 55H)
```

Example:

```plaintext
MyReg SET 0x40;set MyReg loc at 0x40
MOVLW 0x39 ;W = 39H
MOVWF MyReg ;MyReg = 39H
COMPF MyReg,F ;MyReg = C6H and WREG = 39H
```

Where 39H (0011 1001 bin) becomes C6H (1100 0110).
Example:

```
MyReg SET 0x40 ; set MyReg loc at 0x40
MOVLW 0x55 ; W = 55H
MOVWF MyReg ; MyReg = 55H
COMPF MyReg, F ; MyReg AAH, WREG = 55H
```

where 55H (0101 0101) becomes AAH (1010 1010).

Example: Toggle PORTB 150 times.

```
COUNTER SET 0x40 ; loc 40H for COUNTER
SETF TRISB ; port B as output
MOVLW D'150' ; WREG = 150
MOVWF COUNTER ; COUNTER = 150
MOVLW 0x55 ; WREG = 55H
MOVWF PORTB
BACK COMF PORTB, F ; toggle PORTB
DECF COUNTER, F ; decrement COUNTER
BNZ BACK ; toggle until counter becomes 0
```

We can place the result in WREG as shown in the examples below:

```
MyReg SET 0x40 ; set MyReg loc at 0x40
MOVLW 0x39 ; W = 39H
MOVWF MyReg ; MyReg = 39H
COMPF MyReg ; MyReg = 39H and WREG = C6H
```

Example:

```
MyReg SET 0x40 ; set MyReg loc at 0x40
MOVLW 0x55 ; W = 55H
MOVWF MyReg ; MyReg = 55H
COMPF MyReg ; WREG = AA and MyReg 55H SETF
```

**CPFSEQ** Compare FileReg with WREG and skip if equal (F = W)

**Function:** Compare fileReg and WREG and skip if they are equal

**Syntax:** CPFSEQ f, a

The magnitudes of the fileReg byte and WREG byte are compared. If they are equal, it skips the next instruction.

Example: Keep monitoring PORTB indefinitely for the value of 99H. Get out only when PORTB has the value 99H.

```
SETF TRISB ; PORTB an input port
MOVLW 0x99 ; WREG = 99h
BACK CPFSEQ PORTB ; skip if PORTB has 0x99
BRA BACK ; keep monitoring
```
Notice that CPFSEQ skips only when fileReg and WREG have equal values.

**CPFSGT**  \(\text{Compare FileReg with WREG and skip if greater (F > W)}\)

Function: Compare fileReg and WREG and skip if fileReg > WREG.
Syntax: \text{CPFSGT } f, a

The magnitudes of the fileReg byte and WREG byte are compared. If fileReg is larger than the WREG, it skips the next instruction.

Example: Keep monitoring PORTB indefinitely for the value of 99H. Get out only when PORTB has a value greater than 99H.

```
SETF TRISB ; PORTB an input port
MOVLW 0x99 ; WREG = 99H
BACK: CPFSGT PORTB ; skip if PORTB > 99H
BRA BACK ; keep monitoring
```

Notice that CPFSGT skips only if FileReg is greater than WREG.

**CPFSLT**  \(\text{Compare FileReg with WREG and skip if less than (F < W)}\)

Function: Compare fileReg and WREG and skip if fileReg < WREG.
Syntax: \text{CPFSLT } f, a

The magnitudes of the fileReg byte and WREG byte are compared. If fileReg is less than the WREG, it skips the next instruction.

Example: Keep monitoring PORTB indefinitely for the value of 99H. Get out only when PORTB has a value less than 99H.

```
SETF TRISB ; PORTB an input port
MOVLW 0x99 ; WREG = 99H
BACK: CPFSEQ PORTB ; skip if PORTB < 99H
BRA BACK ; keep monitoring
```

Notice that CPFSLT skips only if FileReg < WREG.

**DAW**

Function: Decimal-adjust WREG after addition
Syntax: \text{DAW}

This instruction is used after addition of BCD numbers to convert the result back to BCD. The data is adjusted in the following two possible cases:

1. It adds 6 to the lower 4 bits of WREG if it is greater than 9 or if DC = 1.
2. It also adds 6 to the upper 4 bits of WREG if it is greater than 9 or if C = 1.
Example:
```
MOVLW 0x47 ; WREG = 0100 0111
ADDLW 0x38 ; WREG = 47H + 38H = 7FH, invalid BCD
DAW ; WREG = 1000 0101 = 85H, valid BCD
```

47H + 38H
7FH (invalid BCD)
+ 6H (after DAW)
85H (valid BCD)

In the above example, because the lower nibble was greater than 9, DAW added 6 to WREG. If the lower nibble is less than 9 but DC = 1, it also adds 6 to the lower nibble. See the following example:
```
MOVLW 0x29 ; WREG = 0010 1001
ADDLW 0x18 ; WREG = 0100 0001 INCORRECT
DAW ; WREG = 0100 0111 = 47H VALID BCD
```

29H + 18H
41H (incorrect result in BCD)
+ 6H
47H correct result in BCD

The same thing can happen for the upper nibble. See the following example:
```
MOVLW 0x52 ; WREG = 0101 0010
ADDLW 0x91 ; WREG = 1110 0011 INVALID BCD
DAW ; WREG = 0100 0011 AND C = 1
```

52H + 91H
E3H (invalid BCD)
+ 6
143H valid BCD

Similarly, if the upper nibble is less than 9 and C = 1, it must be corrected. See the following example:
```
MOVLW 0x94 ; W = 1001 0100
ADDLW 0x91 ; W = 0010 0101 INCORRECT
DAW ; W = 1000 0101, VALID BCD
```

; FOR 85, C = 1
\[ \begin{align*}
94H &+ 94H & (\text{incorrect BCD}) \\
125H &+ 6 & (\text{after DAW, adding to upper nibble}) \\
185 & & \\
\end{align*} \]

It is possible that 6 is added to both the high and low nibbles. See the following example:

\begin{align*}
\text{MOVLW} & \ 0x54 \ ; \text{WREG} = 0101 \ 0100 \\
\text{ADDLW} & \ 0x87 \ ; \text{WREG} = 1101 \ 1011 \ \text{INVALID BCD} \\
\text{DAW} & \ ; \text{WREG} = 0100 \ 0001, \ C = 1 \ (\text{BCD} \ 141) \\
54H &+ 87H \\
\text{DBH} & \ (\text{invalid result in BCD}) \\
66H &+ 6H \\
141H & \ \text{valid BCD} \\
\end{align*}

**DECF**

**Decrement fileReg**

Function: Decrement fileReg  
Syntax: DECF f, d, a

This instruction subtracts 1 from the byte operand in fileReg. The result can be placed in WREG (if d = 0) or fileReg (if d = 1).

Example:

\begin{verbatim}
MyReg SET 0x40 ; set aside loc 40H for MyReg
MOVLW 0x99 ; WREG = 99H
MOVWF MyReg ; MyReg = 99H
DECF MyReg, F ; MyReg = 98H, WREG 99H
DECF MyReg, F ; MyReg = 97H, WREG 99H
DECF MyReg, F ; MyReg = 96H, WREG 99H
\end{verbatim}

Example: Toggle PORTB 250 times.

\begin{verbatim}
COUNTER SET 0x40 ; loc 40H for COUNTER
SETF TRISB ; PORTB as output
MOVLW D'250' ; WREG = 250
MOVWF COUNTER ; COUNTER = 250
MOVLW 0x55 ; WREG = 55H
MOVWF PORTB
BACK COMF PORTB, F ; toggle PORTB
DECF COUNTER, F ; decrement COUNTER
BNZ BACK ; toggle until counter becomes 0
\end{verbatim}
We can place the result in WREG as shown in the examples below:

MyReg SET 0x40 ; set aside loc for MyReg
MOVLW 0x99 ; WREG = 99H
MOVWF MyReg ; MyReg = 99H
DECF MyReg ; WREG = 98H, MyReg = 99H
DECF MyReg ; WREG = 97H, MyReg = 99H
DECF MyReg ; WREG = 96H, MyReg = 99H

Example:
MyReg SET 0x50 ; set MyReg loc at 0x50
MOVLW 0x99 ; W = 39H
MOVWF MyReg ; MyReg = 39H
DECF MyReg ; WREG = 38H and MyReg = 39H
DECF MyReg ; WREG = 37H and MyReg = 39H
DECF MyReg ; WREG = 36H and MyReg = 39H
DECF MyReg ; WREG = 35H and MyReg = 39H

**DECFSZ**

Decrement fileReg and Skip if zero

Function: Decrement fileReg and skip if fileReg has zero in it
Syntax: DECFSZ f, d, a

This instruction subtracts 1 from the byte operand of fileReg. If the result is zero, then it skips execution of the next instruction.

Example: Toggle PORTB 250 times.

COUNT SET 0x40 ; loc 40H for COUNT
CLRF TRISB ; PORTB an output
MOVLW D'250' ; WREG = 250
MOVWF COUNT ; COUNT = 250
MOVLW 0x55 ; WREG = 55H
MOVWF PORTB
BACK COMF PORTB,F ; toggle PORTB
DECFSZ COUNT,F ; decrement COUNT and
; skip if zero
BRA BACK ; toggle until counter becomes 0

**DECFSNZ**

Decrement fileReg and skip if not zero

Function: Decrement fileReg and skip if fileReg has other than zero
Syntax: DECFSNZ f, d, a

This instruction subtracts 1 from the byte operand of fileReg. If the result is not zero, then it skips execution of the next instruction.
Example: Toggle PORTB 250 times continuously.

```
COUNT SET 0x40 ;loc 40H for COUNT
CLRF TRISB ;PORTB an output
OVER MOVLW D'250' ;WREG = 250
MOVF COUNT ;COUNT = 250
MOVLW 0x55 ;WREG = 55H
MOVF PORTB
BACK COMF PORTB,F ;toggle PORTB
DECFSNZ COUNT,F ;decrement COUNT and
BRA OVER ;start over
BRA BACK ;toggle until counter becomes 0
```

GOTO  

<table>
<thead>
<tr>
<th>Unconditional Branch</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Function:</strong></td>
</tr>
<tr>
<td><strong>Syntax:</strong></td>
</tr>
</tbody>
</table>

In the PIC18 there are two unconditional branches (jumps): GOTO (long jump) and BRA (short jump). Each is described next.

1. **GOTO** (long jump): This is a 4-byte instruction. The first 12 bits are the opcode, and the next 20 bits are an even address of the target location. Because all the PIC18 instructions are 2 bytes in size, the lowest address bit, A0, is automatically set to zero to make sure that the GOTO instruction will not land at the middle of the targeted instruction. The 20-bit address of the GOTO provides the A20–A1 part of the address and with A0 = 0, we have the 21-bit address needed to go anywhere in the 2M address space of the PIC18.

2. **BRA**: This is a 2-byte instruction. The first 5 bits are the opcode and the remaining 11 bits are the signed number displacement, which is added to the PC (program counter) of the instruction following the BRA to get the target address. Therefore, for the BRA instruction the target address must be within −1023 to +1024 bytes of the PC of the instruction after the BRA because a 11-bit address can take values of +1023 to −1024.

While GOTO is used to jump to any address location within the 2M code space of the PIC18, BRA is used to jump to a location within the 1K ROM space. The advantage of BRA is the fact that it takes 2 bytes of program ROM, while GOTO takes 4 bytes. BRA is widely used in chips with a small amount of program ROM and a limited number of pins.

Notice that the difference between GOTO and CALL is that the CALL instruction will return and continue execution with the instruction following the CALL, whereas GOTO will not return.
INCF

<table>
<thead>
<tr>
<th>Function:</th>
<th>Increment</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>INCF f, d, a</td>
</tr>
</tbody>
</table>

This instruction adds 1 to the byte operand in fileReg. The result can be placed in WREG (if d = 0) or fileReg (if d = 1).

Example:

```
MyReg SET 0x40 ; set aside loc 40H for MyReg
MOVLW 0x99 ; WREG = 99H
MOVWF MyReg
INCF MyReg,F ; MyReg = 9AH, WREG 99H
INCF MyReg,F ; MyReg = 9BH, WREG 99H
DEC MyReg,F ; MyReg = 9CH, WREG 99H
```

Example: Toggle PORTB 5 times.

```
COUNTER SET 0x40 ; loc 40H for COUNTER
SETF TRISB ; PORTB as output
MOVLW D'251' ; WREG = 251
MOVWF COUNTER ; COUNTER = 251
MOVLW 0x55 ; WREG = 55H
MOVWF PORTB
BACK COMF PORTB,F ; toggle PORTB
INCF COUNTER,F ; INC COUNTER
BNC BACK ; toggle until counter becomes 0
```

We can place the result in fileReg as shown in the examples below:

```
MyReg SET 0x40 ; set aside loc for MyReg
MOVLW 0x99 ; WREG = 99H
MOVWF MyReg ; MyReg = 99H
INCF MyReg ; WREG = 9AH, MyReg = 99H
INCF MyReg ; WREG = 9BH, MyReg = 99H
```

Example:

```
MyReg SET 0x40 ; set MyReg loc at 0x40
MOVLW 0x5 ; W = 05H
MOVWF MyReg ; MyReg = 05H
INCF MyReg ; WREG = 06H and MyReg = 05H
```

INCFSZ

<table>
<thead>
<tr>
<th>Function:</th>
<th>Increment</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>INCFSZ f, d, a</td>
</tr>
</tbody>
</table>

APPENDIX A: PIC18 INSTRUCTIONS: FORMAT AND DESCRIPTION
This instruction adds 1 to fileReg and if the result is zero it skips the next instruction.

Example: Toggle PORTB 156 times.

```assembly
COUNTER SET 0x40 ;loc 40H for COUNTER
SETF TRISB ;PORTB as output
MOVLW D'156' ;WREG = 156
MOVWF COUNTER ;COUNTER = 156
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB
BACK COMF PORTB,F ;toggle PORTB
INCFSNZ COUNTER,F ;INC COUNTER and skip if 0
BRA BACK ;toggle until counter becomes 0

INCFSNZ Increment fileReg and skip if not zero

Function: Increment
Syntax: INCFSNZ f, d, a

This instruction adds 1 to the register or memory location specified by the operand. If the result is not zero, it skips the next instruction.

Example: Toggle PORTB 156 times continuously.

```assembly
COUNTER SET 0x40 ;loc 40H for COUNTER
SETF TRISB ;PORTB as output
OVER MOVFW D'156' ;WREG = 156
MOVWF COUNTER ;COUNTER = 156
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB
BACK COMF PORTB,F ;toggle PORTB
INCFSNZ COUNTER,F;INC COUNTER, skip if not 0
BRA OVER ;start over
BRA BACK ;toggle until counter becomes 0

IORLW OR K value with WREG

Function: Logical-OR WREG with value k
Syntax: IORLW k

This performs a logical OR on the WREG register and k value, bit by bit, and stores the result in WREG.

<table>
<thead>
<tr>
<th>A</th>
<th>B</th>
<th>A OR B</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>1</td>
</tr>
</tbody>
</table>

Example:

```assembly
MOVLW 0x30 ;W = 30H
IORLW 0x09 ;now W = 39H
```
Example:

```assembly
MOVLW 0x32 ; W = 32H
IORLW 0x50 ; (W = 72H)
```

```
32H 0011 0010
50H 0101 0000
72H 0111 0010
```

**IORWF**  
**OR FileReg with WREG**

**Function:** Logical-OR fileReg and WREG  
**Syntax:** IORWF f, d, a

This performs a logical OR on the fileReg value and the WREG register, bit by bit, and places the result in WREG (if d = 0) or fileReg (if d = 1).

Example:

```assembly
MyReg SET 0x40; set MyReg loc at 0x40
MOVLW 0x39 ; WREG = 39H
MOVWF MyReg ; MyReg = 39H
MOVLW 0x07
IORWF MyReg ; 39H ORed with 07 (W = 3F)
```

```
39 0011 1001
07 0000 0111
3F 0011 1111
```

Example:

```assembly
MyReg SET 0x40; set MyReg loc at 0x40
MOVLW 0x05 ; WREG = 05H
MOVWF MyReg ; MyReg = 05H
MOVLW 0x30
IORWF MyReg ; 30H ORed with 05 (W = 35H)
```

```
05 0000 0101
30 0011 0000
35 0011 0101
```

We can place the result in fileReg as shown in the examples below:

```assembly
MOVLW 0x30 ; W = 30H
IORWF PORTB,F ; W and PORTB are ORed and result goes to PORTB
```
Example:
MyReg SET 0x20
MOVlw 0x54 ;WREG = 54H
MOVF MyReg
MOVlw 0x67 ;WREG = 67H
IORWF MyReg,F ;OR WREG and MyReg
;after the operation MyReg = 77H

44H 0101 0100
67H 0110 0111
77H 0111 0111 Therefore MyReg will have 77H, WREG = 54H.

**LFSR**

**Load FSR**

Function: Load into FSR registers a 12-bit value of k
Syntax: LFSR f,k ;k is between 000 and FFFH

This loads a 12-bit value into one of the FSR registers of FSR0, FSR1, or FSR2.

LFSR 0 , 0x200 ;FSR0 = 200H
LFSR 1 , 0x050 ;FSR1 = 050H
LFSR 2 , 0x160 ;FSR2 = 160H

This is widely used in register indirect addressing mode. See Chapter 6.

**MOVF (or MOVFW)**

**Move fileReg to WREG**

Function: Copy byte from fileReg to WREG
Syntax MOVF f, d, a:

This instruction is widely used for moving data from a fileReg to WREG. Look at the following examples:

CLRF TRISC ;PORTC output
SETF TRISB ;PORTB as input
MOVFW PORTB ;copy PORTB to WREG
ANDLW 0x0F ;mask the upper 4 bits
MOVF PORTC ;put it in PORTC

Example:
CLRF TRISD ;PORTD as output
SETF TRISB ;PORTB as input
MOVFW PORTB ;copy PORTB to WREG
IORW 0x30 ;OR it with 30H
MOVF PORTD ;put it in PORTD

This instruction can be used to copy the fileReg to itself in order to get the status of the N and Z flags. Look at the following example.
Example:
MyReg SET 0x20 ;set aside loc 0x20 to MyReg
MOVLW 0x54 ;W = 54H
MOVWF MyReg ;MyReg = 54H
MOVFW MyReg,F ;My Reg = 54, also N = 0 and Z = 0

<table>
<thead>
<tr>
<th>MOVFF</th>
<th>Move FileReg to FileReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Copy byte from one fileReg to another fileReg</td>
</tr>
<tr>
<td>Syntax:</td>
<td>MOVFF fs, fd</td>
</tr>
</tbody>
</table>

This copies a byte from the source location to the destination. The source and destination locations can be any of the file register locations, SFRs, or ports.

MOVFF PORTB,MyReg
MOVFF PORTC,PORTD
MOVFF RCREG,PORTC
MOVFF Reg1,REG2

Notice that this a 4-byte instruction because the source and destination address each take 12 bits of the instruction. That means the 24 bits of the instruction are used for the source and destination addresses. The 12-bit address allows data to be moved from any source location to any destination location within the 4K RAM space of the PIC18.

<table>
<thead>
<tr>
<th>MOVLB</th>
<th>Move Literal 4-bit value to lower 4-bit of the BSR</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Move 4-bit value k to lower 4 bits of the BSR registers</td>
</tr>
<tr>
<td>Syntax:</td>
<td>MOVLB k ;k is between 0 and 15 (0–F in hex)</td>
</tr>
</tbody>
</table>

We use this instruction to select a register bank other than the access bank. With this instruction we can load into the BSR (bank selector register) a 4-bit value representing one of 16 banks supported by the PIC18. That means the values between 0000 and 1111 (0–F in hex). For examples of the MOVLB instruction, see Chapter 6 and Section A.1 in this chapter.

<table>
<thead>
<tr>
<th>MOVLW</th>
<th>K</th>
<th>Move Literal to WREG</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Move 8-bit value k to WREG</td>
<td></td>
</tr>
<tr>
<td>Syntax:</td>
<td>MOVLW k ;k is between 0 and 255 (0–FF in hex)</td>
<td></td>
</tr>
</tbody>
</table>

Example:
MOVLW 0x55 ;WREG = 55H
MOVLW 0x0 ;clear WREG (WREG = 0)
MOVLW 0xC2 ;WREG = C2H
MOVLW 0x7F ;WREG = 7FH

This instruction, along with the MOVWF, is widely used to load fixed values into any port, SFR, or fileReg location. See the next instruction to see how it is used.
MOVWF

Move WREG to a fileReg

Function: Copy the WREG contents to a fileReg
Syntax: MOVWF f, a

This copies a byte from WREG to fileReg. This instruction is widely used along with the MOVCLW instruction to load any of the fileReg locations, SFRs, or PORTs with a fixed value. See the following examples:

Example: Toggle PORTB.
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB
MOVLW 0xAA ;WREG = AAH
MOVWF PORTB
BRA OVER ;keep toggling the PORTB

Example: Load RAM location 20H with value 50H.
MyReg SET 0x20 ;set aside the loc 0x20 for MyReg
MOVLW 0x50
MOVWF MyReg ;MyReg = 50H (loc 20H has 50H)

Example: Initialize the Timer0 low and high registers.
MOVLW 0x05 ;WREG = 05H
MOVWF TMR0H
MOVLW 0x30 ;WREG = 30H
MOVWF TMR0L

MULLW Multiply Literal with WREG

Function: Multiply k x WREG
Syntax: MULLW k

This multiplies an unsigned byte k by an unsigned byte in register WREG and the 16-bit result is placed in registers PRODH and PRODL, where PRODL has the lower byte and PRODH has the higher byte.

Example:
MOVLW 0x5 ;WREG = 5H
MULLW 0x07 ;PRODL = 35 = 23H, PRODH = 00

Example:
MOVLW 0x0A ;WREG = 10
MULLW 0x0F ;PRODL = 10 x 15 = 150 = 96H
;PRODH = 00

Example:
MOVLW 0x25
MULLW 0x78 ;PRODL = 58H, PRODH = 11H
;because 25H x 78H = 1158H
Example:
```
MOVLW D'100' ;WREG = 100
MULLW D'200' ;PRODL = 20H, PRODH = 4EH
       ;(100 x 200 = 20,000 = 4E20H)
```

**MULWF**

**Multiply WREG with fileReg**

**Function:** Multiply WREG x fileReg and place the result in PRODH:PROFDL registers

**Syntax:** MULWF f, a

This multiplies an unsigned byte in WREG by an unsigned byte in the fileReg register and the result is placed in PRODL and PRODH, where PRODL has the lower byte and PRODH has the higher byte.

Example:
```
MyReg  SET 0x20 ;MyReg has location of 0x20
MOVLW 0x5
MOVWF MyReg ;MyReg has 0x5
MOVLW 0x7
MOVWF MyReg ;WREG = 0x7
MULWF MyReg ;PRODL = 35 = 23H, PRODH = 00
```

Example:
```
MOVLW 0x0A
MOVWF MyReg ;MyReg = 10
MOVLW 0x0F
MOVWF MyReg ;WREG = 15
MULFW MyReg ;PRODL = 150 = 96H, PRODH = 00
```

Example:
```
MOVLW 0x25
MOVWF MyReg ;MyReg = 0x25
MOVLW 0x78
MOVWF MyReg ;WREG 78H
MULWF Myreg ;PRODL = 58H, PRODH = 11H
       ;(25H x 78H = 1158H)
```

Example:
```
MOVLW D'100'
MOVWF MyReg ;MyReg = 100
MOVLW D'200'
MOVWF MyReg ;WREG = 200
MULWF MyReg ;PRODL = 20H, PRODH = 4EH
       ;(100 x 200 = 20,000 = 4E20H)
```

**NEGF**

**Negate fileReg**

**Function:** No operation

**Syntax:** NEGF f, a

This performs 2's complement on the value stored in fileReg and places it back in fileReg.
Example:

```
MyReg    SET 0x30
MOVLW 0x98    ;WREG = 0x98
MOVWF MyReg    ;MyReg = 0x98
NEGF          ;2’s complement fileReg
```

98H 10011000
01100111 1’s complement
+ 1
01101000 Now FileReg = 68H

Example:

```
MyReg    SET 0x10
MOVLW 0x75    ;WREG = 0x75
MOVWF MyReg    ;MyReg = 0x75
NEGF          ;2’s complement fileReg
```

75H 01110101
10001010 1’s complement
+ 1
10001011 Now FileReg = 7AH

Notice that in this instruction we cannot place the result in the WREG register.

**NOP**

**No Operation**

Function: No operation
Syntax: NOP

This performs no operation and execution continues with the next instruction. It is sometimes used for timing delays to waste clock cycles. This instruction only updates the PC (program counter) to point to the next instruction following NOP. In PIC18, this is a 2-byte instruction.

**POP**

**POP Top of Stack**

Function: Pop from the stack
Syntax: POP

This takes out the top of stack (TOS) pointed to by SP (stack pointer) and discards it. It also decrements SP by 1. After the operation, the top of the stack will be the value pushed onto the stack previously.

**PUSH**

**PUSH Top of the Stack**

Function: Push the PC onto the stack
Syntax: PUSH

This copies the program counter (PC) onto the stack and increments SP by 1, which means the previous top of the stack is pushed down.
**RCALL**  
**Relative Call**

- **Function:** Transfers control to a subroutine within 1K space
- **Syntax:** RCALL target_address

There are two types of CALLs: RCALL and CALL. In RCALL, the target address is within 1K of the current PC (program counter). To reach the target address in the 2M ROM space of the PIC18, we must use CALL. In calling a subroutine, the PC register (which has the address of the instruction after the RCALL) is pushed onto the stack and the stack pointer (SP) is incremented by 1. Then the program counter is loaded with the new address and control is transferred to the subroutine. At the end of the procedure, when RETURN is executed, PC is popped off the stack, which returns control to the instruction after the RCALL.

Notice that RCALL is a 2-byte instruction, in which 5 bits are used for the opcode and the remaining 11 bits are used for the target subroutine address. An 11-bit address limits the range to $-1024$ to $+1023$. See the CALL instruction for discussion of the target address being anywhere in the 2M ROM space of the PIC18. Notice that RCALL is a 2-byte instruction while CALL is a 4-byte instruction. Also notice that the RCALL does not have the option of context saving, as CALL has.

**RESET**  
**Reset (by software)**

- **Function:** Reset by software
- **Syntax:** RESET

This instruction is used to reset the PIC18 by way of software. After execution of this instruction, all the registers and flags are forced to their reset condition. The reset condition is created by activating the hardware pin MCLR. In other words, the RESET instruction is the software version of the MCLR pin.

**RETFIE**  
**Return from Interrupt Exit**

- **Function:** Return from interrupt
- **Syntax:** RETFIE s

This is used at the end of an interrupt service routine (interrupt handler). The top of the stack is popped into the program counter and program execution continues at this new address. After popping the top of the stack into the program counter (PC), the stack pointer (SP) is decremented by 1.

Notice that while the RETURN instruction is used at the end of a subroutine associated with the CALL and RCALL instructions, RETFIE must be used for the interrupt service routines (ISRs).
RETLW  Return with Literal in WREG

Function:  The k value is placed in WREG and the top of the stack is placed in PC (program counter)
Syntax: RETLW k

After execution of this instruction, the k value is loaded into WREG and the top of the stack is popped into the program counter (PC). After popping the top of the stack into the program counter, the stack pointer (SP) is decremented by 1. This instruction is used for the implementation of a look-up table. See Section 6.3 in Chapter 6.

RETURN  Return

Function:  Return from subroutine
Syntax:  RETURN s ; where s = 0 or s = 1

This instruction is used to return from a subroutine previously entered by instructions CALL or RCALL. The top of the stack is popped into the program counter (PC) and program execution continues at this new address. After popping the top of the stack into the program counter, the stack pointer (SP) is decremented by 1. For the case of "RETURN s" where s = 1, the RETURN will also restore the context registers. See the CALL instruction for the case of s = 1. Notice that "RETURN 1" cannot be used for subroutines associated with RCALL.

RLCF  Rotate Left Through Carry the fileReg

Function:  Rotate fileReg left through carry
Syntax: RLCF f, d, a

This rotates the bits of a fileReg register left. The bits rotated out of fileReg are rotated into C, and the C bit is rotated into the opposite end of the fileReg register.

Example:
MyReg SET 0x30 ; set aside loc 30H for MyReg
BCF STATUS, C ; C = 0
MOVLW 0x99 ; WREG = 99H
MOVWF MyReg ; MyReg = 99H = 10011001
RLCF MyReg, F ; now MyReg = 00110010 and ; C = 1
RLCF MyReg, F ; now MyReg = 01100101 and ; C = 0
### RLNCF

<table>
<thead>
<tr>
<th><strong>Function:</strong></th>
<th>Rotate left not through Carry</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax:</strong></td>
<td>RLCNF f, d, a</td>
</tr>
</tbody>
</table>

This rotates the bits of a fileReg register left. The bits rotated out of fileReg are rotated back into fileReg at the opposite end.

**Example:**
```
MyReg SET 0x20 ;set aside loc 20 for MyReg
MOVLW 0x69 ;WREG = 01101001
MOVF MyReg ;MyReg = 69H = 01101001
RLNCF MyReg, F ;now MyReg = 11010010
RLNCF MyReg, F ;now MyReg = 10100101
RLNCF MyReg, F ;now MyReg = 01001011
RLNCF MyReg, F ;now MyReg = 10010110
```

Notice that after four rotations, the upper and lower nibbles are swapped.

### RRCF

<table>
<thead>
<tr>
<th><strong>Function:</strong></th>
<th>Rotate Right through Carry</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax:</strong></td>
<td>RRCF f, d, a</td>
</tr>
</tbody>
</table>

This rotates the bits of a fileReg register right. The bits rotated out of the register are rotated into C, and the C bit is rotated into the opposite end of the register.

**Example:**
```
MyReg SET 0x20 ;set aside loc 20 for MyReg
BSF STATUS, C ;C = 1
MOVLW 0x99 ;WREG = 10011001
MOVF MyReg ;MyReg = 99H = 10011001
RRCF MyReg, F ;now MyReg = 11001100, C = 1
RRCF MyReg, F ;now MyReg = 11100110, C = 0
```

### RRNCF

<table>
<thead>
<tr>
<th><strong>Function:</strong></th>
<th>Rotate Right not through Carry</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Syntax:</strong></td>
<td>RRNCF f, d, a</td>
</tr>
</tbody>
</table>

This rotates the bits of a fileReg register right. The bits rotated out of the register are rotated back into fileReg at the opposite end.
Example:
MyReg SET 0x20 ; set aside loc 20H for MyReg
MOVlw 0x66 ; WREG = 66H = 01100110
MOVwf MyReg ; MyReg = 66H = 01100110
RRncf MyReg,F ; now MyReg = 00110011
RRncf MyReg,F ; now MyReg = 10011001
RRncf MyReg,F ; now MyReg = 11001100
RRncf MyReg,F ; now MyReg = 01100110

Example: We can use this instruction to swap the upper and lower nibbles.
MyReg SET 0x20 ; set aside loc 20H for MyReg
MOVlw 0x36 ; WREG = 36H = 00110110
MOVwf MyReg ; MyReg = 36H = 00110110
RRncf MyReg,F ; now MyReg = 00011011
RRncf MyReg,F ; now MyReg = 10001101
RRncf MyReg,F ; now MyReg = 11000110
RRncf MyReg,F ; now MyReg = 01100011

<table>
<thead>
<tr>
<th>SETF</th>
<th>Set fileReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Set</td>
</tr>
<tr>
<td>Syntax:</td>
<td>SETF f, a</td>
</tr>
</tbody>
</table>

This instruction sets the entire byte in fileReg to HIGH. All bits of the register are set to 1.

Examples:
SETF MyReg ; MyReg = 11111111
SETF TRISB ; TRISB = FFH, (makes PORTB input)
SETF PORTC ; PORTC = 1111 1111

Notice that in this instruction, the result can be placed in fileReg only and there is no option for WREG to be used as the destination for the result.

<table>
<thead>
<tr>
<th>SLEEP</th>
<th>Enter Sleep mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Put the CPU into sleep mode</td>
</tr>
<tr>
<td>Syntax:</td>
<td>SLEEP</td>
</tr>
</tbody>
</table>

This instruction stops the oscillator and puts the CPU into sleep mode. It also resets the Watchdog Timer (WDT). The WDT is used mainly with the SLEEP instruction. Upon execution of the SLEEP instruction, the entire microcontroller goes into sleep mode by shutting down the main oscillator and by stopping the Program Counter from fetching the next instruction after SLEEP. There are two ways to get out of sleep mode: (a) an external event via hardware interrupt, (b) the internal WDT interrupt. Upon wake-up from a WDT interrupt, the microcontroller resumes operation by executing the next instruction after SLEEP.

Check the Microchip Corp. website for application notes on WDT.
SUBFWB  Subtract fileReg from WREG with borrow

Function:  WREG – fileReg – #borrow ; #borrow is inverted carry
Syntax:  SUBFWB f, d, a

This subtracts fileReg and the Carry (borrow) flag from WREG and puts the result in WREG (d = 0) or fileReg (d = 1). The steps for subtraction performed by the internal hardware of the CPU are as follows:

1. Take the 2’s complement of the fileReg byte.
2. Add this to register WREG.
3. Add the inverted Carry (borrow) flag to the result.
4. Ignore the Carry.
5. Examine the N (negative) flag for positive or negative result.

Example:
MyReg SET 0x20 ;set aside loc 0x20 for MyReg
BSF STATUS,C ;make Carry = 1
MOVLW 0x45 ;WREG 45H
MOVWF MyReg ;MYReg = 45H
MOVLW 0x23
SUBWF MyReg ;WREG = 45H - 23H - 0 = 22H

45H 0100 0101
-23H 0010 0011 2’s comp + 1101 1101
Inverted carry +

----------
+22H 0010 0010

Because D7 (the N flag) is 0, the result is positive.

This instruction sets the negative flag according to the following:

\[
\begin{align*}
N &\quad \text{WREG} > (\text{fileReg} + \#C) \quad 0 \quad \text{the result is positive} \\
&\quad \text{WREG} = (\text{fileReg} + \#C) \quad 0 \quad \text{the result is 0} \\
&\quad \text{WREG} < (\text{fileReg} + \#C) \quad 1 \quad \text{the result is negative and in 2’s comp}
\end{align*}
\]

SUBLW  Subtract WREG from Literal value

Function:  Subtract WREG from literal value k (WREG = k – WREG)
Syntax:  SUBLW k

This subtracts the WREG value from the literal value k and puts the result in WREG. The steps for subtraction performed by the internal hardware of the CPU are as follows:

1. Take the 2’s complement of the WREG value.
2. Add it to literal value k.
3. Ignore the Carry.
4. Examine the N (negative) flag for positive or negative result.
MOVLW 0x23 ;WREG 23H  
SUBLW 0x45 ;WREG = 45H - 23H = 22H

45H 0100 0101 0100 0101
-23H 0010 0011 2's comp +1101 1101
-------  ------------------
+22H 0010 0010

Because D7 (the N flag) is 0, the result is positive.

This instruction sets the negative flag according to the following:

<table>
<thead>
<tr>
<th>N</th>
<th>Literal value k &gt; WREG</th>
<th>0</th>
<th>the result is positive</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Literal value k = WREG</td>
<td>0</td>
<td>the result is 0</td>
</tr>
<tr>
<td></td>
<td>Literal value &lt; WREG</td>
<td>1</td>
<td>the result is negative and in 2's comp</td>
</tr>
</tbody>
</table>

Example:

MOVLW 0x98 ;WREG 98H
SUBLW 0x66 ;WREG = 66H - 98H = CEH

66H 0110 0110 0110 0110
-98H 1001 1000 2's comp +0110 1000
-------  ------------------
CEH 1100 1110

Because D7 (the N flag) is 1, the result is negative and in 2's comp.

<table>
<thead>
<tr>
<th>SUBWF</th>
<th>Subtract WREG from fileReg</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Subtract WREG from fileReg (Dest = fileReg - WREG)</td>
</tr>
<tr>
<td>Syntax:</td>
<td>SUBWF f, d, a</td>
</tr>
</tbody>
</table>

This subtracts the WREG value from the fileReg value and puts the result in either WREG (d = 0) or fileReg (d = 1). The steps for subtraction performed by the internal hardware of the CPU are as follows:

1. Take the 2's complement of the WREG byte.
2. Add this to the fileReg register.
3. Ignore the carry.
4. Examine the N (negative) flag for positive or negative result.

Example:

MyReg SET 0x20 ;set aside loc 0x20 for MyReg
MOVLW 0x45 ;WREG 45H
MOVWF MyReg ;MYReg = 45H
MOVLW 0x23 ;WREG = 23H
SUBWF MyReg,F ;MyReg = 45H - 23H = 22H
Because D7 (the N flag) is 0, the result is positive.

This instruction sets the negative flag according to the following:

\[ \begin{align*}
    &\text{N} \\
    &\text{fileReg} > \text{WREG} \quad 0 \quad \text{the result is positive} \\
    &\text{fileReg} = \text{WREG} \quad 0 \quad \text{the result is 0} \\
    &\text{fileReg} < \text{WREG} \quad 1 \quad \text{the result is negative and in 2's comp}
\end{align*} \]

**SUBWFB**  
**Subtract WREG from fileReg with borrow**

Function: \( \text{Dest} = \text{fileReg} - \text{WREG} - \#\text{borrow} ; \#\text{borrow} \text{ is inverted carry} \)
Syntax: SUBWFB f, d, a

This subtracts the WREG value and the inverted borrow (carry) flag from the fileReg value and puts the result in WREG (if \( d = 0 \)), or fileReg (if \( d = 1 \)). The steps for subtraction performed by the internal hardware of the CPU are as follows:

1. Take the 2's complement of WREG.
2. Add this to fileReg.
3. Add the inverted Carry flag to the result.
4. Ignore the carry.
5. Examine the N (negative) flag for positive or negative result.

Example:

```c
MyReg SET 0x20 ;set aside loc 0x20 for MyReg
BSF STATUS, C ;C = 1
MOVLW 0x45 ;WREG 45H
MOVWF MyReg ;MyReg = 45H
MOVLW 0x23 ;WREG = 23H
SUBWFB MyReg, F ;MyReg = 45H - 23H - 0 = 22H
```

\[
\begin{align*}
    &45H \quad 0100 \ 0101 \\
    &-23H \quad 0010 \ 0011 \ 2's \ comp \quad +1101 \ 1101 \\
    &\text{---------} \\
    &+22H \quad 0010 \ 0010 \\
\end{align*}
\]

Because D7 (the N flag) is 0, the result is positive.
This instruction sets the negative flag according to the following:

<table>
<thead>
<tr>
<th>Condition</th>
<th>Flag Set</th>
<th>Result Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>fileReg &gt; (WREG + #C)</td>
<td>0</td>
<td>the result is positive</td>
</tr>
<tr>
<td>fileReg = (WREG + #C)</td>
<td>0</td>
<td>the result is 0</td>
</tr>
<tr>
<td>fileReg &lt; (WREG + #C)</td>
<td>1</td>
<td>the result is negative and in 2’s comp.</td>
</tr>
</tbody>
</table>

**SWAPF**  
**Swap Nibbles in fileReg**  

Function: Swap nibbles within fileReg  
Syntax: SAWPF f, d, a

The SWAPF instruction interchanges the lower nibble (D0–D3) with the upper nibble (D4–D7) inside fileReg. The result is placed in WREG (d = 0) or fileReg (d = 1).

Example:  
MyReg SET 0x20 ;set aside loc 20H for MyReg  
MOVlw 0x59H ;W = 59H (0101 1001 in binary)  
MOVWF MyReg ;MyReg = 59H (0101 1001)  
SWAPF MyReg,F ;MyReg = 95H (1001 0101)

**TBLRD**  
**Table Read**  

Function: Read a byte from ROM to the TABLAT register  
Syntax: TBLRD *  
TBLRD *+  
TBLRD *-  
TBLRD +*

This instruction moves (copies) a byte of data located in program (code) ROM into the TableLatch (TABLAT) register. This allows us to put strings of data, such as look-up table elements, in the code space and read them into the CPU. The address of the desired byte in the program space (on-chip ROM) is held by the TBLPTR register. Table A-6 shows the auto-increment feature of the TBLRD instruction.

**Table A-6: PIC18 Table Read Instructions**

<table>
<thead>
<tr>
<th>Instruction</th>
<th>Function</th>
<th>Note</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBLRD*</td>
<td>Table Read</td>
<td>After read, TBLPTR stays the same</td>
</tr>
<tr>
<td>TBLRD*+</td>
<td>Table Read with post-increment</td>
<td>(Read and increment TBLPTR)</td>
</tr>
<tr>
<td>TBLRD*-</td>
<td>Table Read with post-decrement</td>
<td>(Read and decrement TBLPTR)</td>
</tr>
<tr>
<td>TBLRD+*</td>
<td>Table Read with pre-increment</td>
<td>(increment TBLPTR and read)</td>
</tr>
</tbody>
</table>

Note: A byte of data is read into the TABLAT register from code space pointed to by TBLPTR.

Example: Assume that an ASCII character string is stored in the on-chip ROM program memory starting at address 500H. Write a program to bring each character into the CPU and send it to PORTB.

```
ORG 0000H ;burn into ROM starting at 0
```
MOVLW LOW(MESSAGE) ;WREG = 00 low-byte addr.
MOVF TBLPTRL ;look-up table low-byte addr
MOVLW HIGH(MESSAGE) ;WREG = 05 = high-byte addr
MOVF TBLPTRH ;look-up table high-byte addr
CLRF TBLPTRU ;clear upper 5 bits

B8 TBLRD*+ ;read the table, then increment TBLPTR
MOVF TABLAT,W ;copy to WREG (Z = 1 if null)
BZ EXIT ;exit if end of string
MOVF PORTB ;copy WREG to PORTB
BRA B8
EXIT GOTO EXIT

; --------------- message
ORG 0x500 ;data burned starting at 0x500
ORG 0x500
MESSAGE DB "The earth is but one country and "
DB "mankind its citizens","Baha'u'llah",0
END

In the program above, the TBLPTR holds the address of the desired byte. After the execution of the TBLRD*+ instruction, register TABLAT has the character. Notice that TBLPTR is incremented automatically to point to the next character in the MESSAGE table.

<table>
<thead>
<tr>
<th>TBLWT</th>
<th>Table Write</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Write to Flash a block of data</td>
</tr>
<tr>
<td>Syntax:</td>
<td>TBLWT*</td>
</tr>
<tr>
<td></td>
<td>TBLWT*++</td>
</tr>
<tr>
<td></td>
<td>TBLWT*-</td>
</tr>
<tr>
<td></td>
<td>TBLWT++</td>
</tr>
</tbody>
</table>

This instruction writes a block of data to the program (code) space assuming that the on-chip program ROM is of Flash type. The address of the desired location in Flash ROM is held by the TBLPTR register. The process of writing to Flash ROM using the TBLWT instruction is discussed in Section 14.3 of Chapter 14.

<table>
<thead>
<tr>
<th>TSTFSZ</th>
<th>Test fileReg, Skip if Zero</th>
</tr>
</thead>
<tbody>
<tr>
<td>Function:</td>
<td>Test fileReg for zero value and skip if it is zero</td>
</tr>
<tr>
<td>Syntax:</td>
<td>TSTFSZ f, a</td>
</tr>
</tbody>
</table>

This instruction tests the entire contents of fileReg for value zero and skips the next instruction if fileReg has zero in it.

Example: Test PORTB for zero continuously.
SETF TRISB ;make PORTB an input
CLRF TRISD ;make PORTD an output
BACK TSTFSZ PORTB
BRA BACK
MOVFF PORTB, PORTD

Example: Toggle PORTB 250 times.

COUNTER SET 0x40 ;loc 40H for COUNTER
SETF TRISB ;PORTB as output
MOVLW D'250' ;WREG = 250
MOVWF COUNTER ;COUNTER = 250
MOVLW 0x55 ;WREG = 55H
MOVWF PORTB
BACK COMF PORTB,F ;toggle PORTB
DECF COUNTER,F ;decrement COUNTER
TSTFSZ COUNTER ;test counter for 0
BRA BACK ;keep doing it
......

XORLW Ex-Or Literal with WREG

Function: Logical exclusive-OR Literal k and WREG
Syntax: XORLW k

This performs a logical exclusive-OR on the Literal value and WREG operands, bit by bit, storing the result in WREG.

Example:

<table>
<thead>
<tr>
<th>A</th>
<th>B</th>
<th>A XOR B</th>
</tr>
</thead>
<tbody>
<tr>
<td>0</td>
<td>0</td>
<td>0</td>
</tr>
<tr>
<td>0</td>
<td>1</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>1</td>
<td>0</td>
</tr>
</tbody>
</table>

MOVLW 0x39 ;WREG = 39H
XORLW 0x09 ;WREG = 39H ORed with 09 ;now, WREG = 30H

39H 0011 1001
09H 0000 1001
30  0011 0000

Example:

MOVLW 0x32 ;WREG = 32H
XORLW 0x50 ;(now, WREG = 62H)

32H 0011 0010
50H 0101 0000
62H 0110 0010

XORWF Ex-Or WREG with fileReg

Function: Logical exclusive-OR fileReg and WREG
Syntax: XORWF f, d, a

This performs a logical exclusive-OR on the operands, bit by bit, storing
the result in the destination. The destination can be WREG (d = 0), or fileReg (d = 1).

Example:

```
MyReg SET 0x20 ;set aside loc 20h for MyReg
MOVLW 0x39 ;WREG = 39H
MOVWF MyReg ;MyReg = 39H
MOVLW 0x09 ;WREG = 09H
XORWF MyReg,F ;MyReg = 39H ORed with 09
 ;MyReg = 30H
```

```
39H 0011 1001
09H 0000 1001
30 0011 0000
```

Example:

```
MyReg SET 0x15 ;set aside loc 15 for MyReg
MOVLW 0x32 ;WREG = 32H
MOVWF MyReg ;MyReg = 32H
MOVLW 0x50 ;WREG = 50H
XORWF MyReg,F ;now W = 62H
```

```
32H 0011 0010
50H 0101 0000
62H 0110 0010.
```

We can place the result in WREG.

Example:

```
MyReg SET 0x15 ;set aside loc 15 for MyReg
MOVLW 0x44 ;WREG = 44H
MOVWF MyReg ;MyReg = 44H
MOVLW 0x67 ;WREG = 67H
XORWF MyReg ;now W = 23H, and MyReg = 44H
```

```
44H 0100 0100
67H 0110 0111
23H 0010 0011
```
OVERVIEW

This appendix shows the basics of wire wrapping.
BASICS OF WIRE WRAPPING

*Note:* For this tutorial appendix, you will need the following:
Wire-wrapping tool (Radio Shack part number 276-1570)
30-gauge (30-AWG) wire for wire wrapping
(Thanks to Shannon Looper and Greg Boyle for their assistance on this section.)

The following describes the basics of wire wrapping:

1. There are several different types of wire-wrap tools available. The best one is available from Radio Shack for less than $10. The part number for the Radio Shack model is 276-1570. This tool combines the wrap and unwrap functions in the same end of the tool and includes a separate stripper. We found this to be much easier to use than the tools that combined all these features on one two-ended shaft. There are also wire-wrap guns, which are, of course, more expensive.

2. Wire-wrapping wire is available prestripped in various lengths or in bulk on a spool. The prestripped wire is usually more expensive and you are restricted to the different wire lengths you can afford to buy. Bulk wire can be cut to any length you wish, which allows each wire to be custom fit.

3. Several different types of wire-wrap boards are available. These are usually called perfboards or wire-wrap boards. These types of boards are sold at many electronics stores (such as Radio Shack). The best type of board has plating around the holes on the bottom of the board. These boards are better because the sockets and pins can be soldered to the board, which makes the circuit more mechanically stable.

4. Choose a board that is large enough to accommodate all the parts in your design with room to spare so that the wiring does not become too cluttered. If you wish to expand your project in the future, you should be sure to include enough room on the original board for the complete circuit. Also, if possible, the layout of the IC on the board needs to be such that signals go from left to right just like the schematics.

5. To make the wiring easier and to keep pressure off the pins, install one standoff on each corner of the board. You may also wish to put standoffs on the top of the board to add stability when the board is on its back.

6. For power hook-up, use some type of standard binding post. Solder a few single wire-wrap pins to each power post to make circuit connections (to at least one pin for each IC in the circuit).

7. To further reduce problems with power, each IC must have its own connection to the main power of the board. If your perfboard does not have built-in power buses, run a separate power and ground wire from each IC to the main power. In other words, DO NOT daisy chain (chip-to-chip connection is called daisy chain) power connections, as each connection down the line will have more wire and more resistance to get power through. See Figure B-1. However, daisy chaining is acceptable for other connections such as data, address, and control buses.

8. You must use wire-wrap sockets. These sockets have long square pins whose edges will cut into the wire as it is wrapped around the pin.
9. Wire wrapping will not work on round legs. If you need to wrap to components, such as capacitors, that have round legs, you must also solder these connections. The best way to connect single components is to install individual wire-wrap pins into the board and then solder the components to the pins. An alternate method is to use an empty IC socket to hold small components such as resistors and wrap them to the socket.

10. The wire should be stripped about 1 inch. This will allow 7 to 10 turns for each connection. The first turn or turn-and-a-half should be insulated. This prevents stripped wire from coming in contact with other pins. This can be accomplished by inserting the wire as far as it will go into the tool before making the connection.

11. Try to keep wire lengths to a minimum. This prevents the circuit from looking like a bird nest. Be neat and use color coding as much as possible. Use only red wires for $V_{CC}$ and black wires for ground connections. Also use different colors for data, address, and control signal connections. These suggestions will make troubleshooting much easier.

12. It is standard practice to connect all power lines first and check them for continuity. This will eliminate trouble later on.

13. It's also a good idea to mark the pin orientation on the bottom of the board. Plastic templates are available with pin numbers preprinted on them specifically for this purpose, or you can make your own from paper. Forgetting to reverse pin order when looking at the bottom of the board is a very common mistake when wire wrapping circuits.

14. To prevent damage to your circuit, place a diode (such as IN5338) in reverse bias across the supply. If the power gets hooked up backwards, the diode will be forward biased and will act as a short, keeping the reversed voltage from your circuit.

15. In digital circuits, there can be a problem with current demand on the power supply. To filter the noise on the power supply, a 100 μF electrolytic capacitor and a 0.1 μF monolithic capacitor are connected from $V_{CC}$ to ground, in parallel with each other, at the entry point of the power supply to the board. These two together will filter both the high- and the low-frequency noises. Instead of using two capacitors in parallel, you can use a single 20–100 μF tantalum capacitor. Remember that the long lead is the positive one.

16. To filter the transient current, use a 0.1 μF monolithic capacitor for each IC. Place the 0.1 μF monolithic capacitor between $V_{CC}$ and ground of each IC. Make sure the leads are as short as possible.

![Figure B-1. Daisy Chain Connection (not recommended for power lines)](image-url)
This appendix provides an overview of IC technology and PIC18 interfacing. In addition, we look at the microcontroller-based system as a whole and examine some general issues in system design.

First, in Section C.1, we provide an overview of IC technology. Then, in Section C.2, the internal details of PIC18 I/O ports and interfacing are discussed. Section C.3 examines system design issues.
C.1: OVERVIEW OF IC TECHNOLOGY

In this section we examine IC technology and discuss some major developments in advanced logic families. Because this is an overview, it is assumed that the reader is familiar with logic families on the level presented in basic digital electronics books.

Transistors

The transistor was invented in 1947 by three scientists at Bell Laboratory. In the 1950s, transistors replaced vacuum tubes in many electronics systems, including computers. It was not until 1959 that the first integrated circuit was successfully fabricated and tested by Jack Kilby of Texas Instruments. Prior to the invention of the IC, the use of transistors, along with other discrete components such as capacitors and resistors, was common in computer design. Early transistors were made of germanium, which was later abandoned in favor of silicon. This was because the slightest rise in temperature resulted in massive current flows in germanium-based transistors. In semiconductor terms, it is because the band gap of germanium is much smaller than that of silicon, resulting in a massive flow of electrons from the valence band to the conduction band when the temperature rises even slightly. By the late 1960s and early 1970s, the use of the silicon-based IC was widespread in mainframes and minicomputers. Transistors and ICs at first were based on P-type materials. Later on, because the speed of electrons is much higher (about two-and-a-half times) than the speed of holes, N-type devices replaced P-type devices. By the mid-1970s, NPN and NMOS transistors had replaced the slower PNP and PMOS transistors in every sector of the electronics industry, including in the design of microprocessors and computers. Since the early 1980s, CMOS (complementary MOS) has become the dominant technology of IC design. Next we provide an overview of differences between MOS and bipolar transistors. See Figure C-1.

Figure C-1. Bipolar vs. MOS Transistors
MOS vs. bipolar transistors

There are two types of transistors: bipolar and MOS (metal-oxide semiconductor). Both have three leads. In bipolar transistors, the three leads are referred to as the emitter, base, and collector, while in MOS transistors they are named source, gate, and drain. In bipolar transistors, the carrier flows from the emitter to the collector, and the base is used as a flow controller. In MOS transistors, the carrier flows from the source to the drain, and the gate is used as a flow controller. In NPN-type bipolar transistors, the electron carrier leaving the emitter must overcome two voltage barriers before it reaches the collector (see Figure C-1). One is the N-P junction of the emitter-base and the other is the P-N junction of the base-collector. The voltage barrier of the base-collector is the most difficult one for the electrons to overcome (because it is reverse-biased) and it causes the most power dissipation. This led to the design of the unipolar type transistor called MOS. In N-channel MOS transistors, the electrons leave the source and reach the drain without going through any voltage barrier. The absence of any voltage barrier in the path of the carrier is one reason why MOS dissipates much less power than bipolar transistors. The low power dissipation of MOS allows millions of transistors to fit on a single IC chip. In today's technology, putting 10 million transistors into an IC is common, and it is all because of MOS technology. Without the MOS transistor, the advent of desktop personal computers would not have been possible, at least not so soon. The bipolar transistors in both the mainframes and minicomputers of the 1960s and 1970s were bulky and required expensive cooling systems and large rooms. MOS transistors do have one major drawback: They are slower than bipolar transistors. This is due partly to the gate capacitance of the MOS transistor. For a MOS to be turned on, the input capacitor of the gate takes time to charge up to the turn-on (threshold) voltage, leading to a longer propagation delay.

Overview of logic families

Logic families are judged according to (1) speed, (2) power dissipation, (3) noise immunity, (4) input/output interface compatibility, and (5) cost. Desirable qualities are high speed, low power dissipation, and high noise immunity (because it prevents the occurrence of false logic signals during switching transition). In interfacing logic families, the more inputs that can be driven by a single output, the better. This means that high-driving-capability outputs are desired. This, plus the fact that the input and output voltage levels of MOS and bipolar transistors are not compatible mean that one must be concerned with the ability of one logic family to drive the other one. In terms of the cost of a given logic family, it is high during the early years of its introduction but it declines as production and use rise.

The case of inverters

As an example of logic gates, we look at a simple inverter. In a one-transistor inverter, the transistor plays the role of a switch, and R is the pull-up resistor. See Figure C-2. For this inverter to work most effectively in digital circuits, however, the R value must be high when the transistor is “on” to limit the current flow from Vcc to ground in order to have low power dissipation (P = VI, where V
In other words, the lower the I, the lower the power dissipation. On the other hand, when the transistor is "off", R must be a small value to limit the voltage drop across R, thereby making sure that $V_{OUT}$ is close to $V_{CC}$. This is a contradictory demand on R. This is one reason that logic gate designers use active components (transistors) instead of passive components (resistors) to implement the pull-up resistor R.

Figure C-2. One-Transistor Inverter with Pull-up Resistor

The case of a TTL inverter with totem-pole output is shown in Figure C-3. In Figure C-3, Q3 plays the role of a pull-up resistor.

Figure C-3. TTL Inverter with Totem-Pole Output

**CMOS inverter**

In the case of CMOS-based logic gates, PMOS and NMOS are used to construct a CMOS (complementary MOS) inverter as shown in Figure C-4. In CMOS inverters, when the PMOS transistor is off, it provides a very high impedance path, making leakage current almost zero (about 10 nA); when the PMOS is on, it provides a low resistance on the path of $V_{DD}$ to load. Because the speed of the hole is slower than that of the electron, the PMOS transistor is wider to compensate for this disparity; therefore, PMOS transistors take more space than NMOS transistors in the CMOS gates. At the end of this section we will see an open-collector gate in which the pull-up resistor is provided externally, thereby allowing system designers to choose the value of the pull-up resistor.
In 1968 the first logic family made of bipolar transistors was marketed. It was commonly referred to as the standard TTL (transistor-transistor logic) family. The first MOS-based logic family, the CD4000/74C series, was marketed in 1970. The addition of the Schottky diode to the base-collector of bipolar transistors in the early 1970s gave rise to the S family. The Schottky diode shortens the propagation delay of the TTL family by preventing the collector from going into what is called deep saturation. Table C-1 lists major characteristics of some logic families. In Table C-1, note that as the CMOS circuit's operating frequency rises, the power dissipation also increases. This is not the case for bipolar-based TTL.

![CMOS Inverter Diagram](image)

**Input/output characteristics of some logic families**

<table>
<thead>
<tr>
<th>Characteristic</th>
<th>STD TTL</th>
<th>LSTTL</th>
<th>ALSTTL</th>
<th>HCMOS</th>
</tr>
</thead>
<tbody>
<tr>
<td>V&lt;sub&gt;CC&lt;/sub&gt;</td>
<td>5 V</td>
<td>5 V</td>
<td>5 V</td>
<td>5 V</td>
</tr>
<tr>
<td>V&lt;sub&gt;IH&lt;/sub&gt;</td>
<td>2.0 V</td>
<td>2.0 V</td>
<td>2.0 V</td>
<td>3.15 V</td>
</tr>
<tr>
<td>V&lt;sub&gt;IL&lt;/sub&gt;</td>
<td>0.8 V</td>
<td>0.8 V</td>
<td>0.8 V</td>
<td>1.1 V</td>
</tr>
<tr>
<td>V&lt;sub&gt;OH&lt;/sub&gt;</td>
<td>2.4 V</td>
<td>2.7 V</td>
<td>2.7 V</td>
<td>3.7 V</td>
</tr>
<tr>
<td>V&lt;sub&gt;OL&lt;/sub&gt;</td>
<td>0.4 V</td>
<td>0.5 V</td>
<td>0.4 V</td>
<td>0.4 V</td>
</tr>
<tr>
<td>I&lt;sub&gt;IL&lt;/sub&gt;</td>
<td>-1.6 mA</td>
<td>-0.36 mA</td>
<td>-0.2 mA</td>
<td>-1 μA</td>
</tr>
<tr>
<td>I&lt;sub&gt;IH&lt;/sub&gt;</td>
<td>40 μA</td>
<td>20 μA</td>
<td>20 μA</td>
<td>1 μA</td>
</tr>
<tr>
<td>I&lt;sub&gt;OL&lt;/sub&gt;</td>
<td>16 mA</td>
<td>8 mA</td>
<td>4 mA</td>
<td>4 mA</td>
</tr>
<tr>
<td>I&lt;sub&gt;OH&lt;/sub&gt;</td>
<td>-400 μA</td>
<td>-400 μA</td>
<td>-400 μA</td>
<td>4 mA</td>
</tr>
<tr>
<td>Propagation delay</td>
<td>10 ns</td>
<td>9.5 ns</td>
<td>4 ns</td>
<td>9 ns</td>
</tr>
<tr>
<td>Static power dissipation (f = 0)</td>
<td>10 mW</td>
<td>2 mW</td>
<td>1 mW</td>
<td>0.0025 nW</td>
</tr>
<tr>
<td>Dynamic power dissipation</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>at f = 100 kHz</td>
<td>10 mW</td>
<td>2 mW</td>
<td>1 mW</td>
<td>0.17 mW</td>
</tr>
</tbody>
</table>
History of logic families

Early logic families and microprocessors required both positive and negative power voltages. In the mid-1970s, 5 V $V_{CC}$ became standard. In the late 1970s, advances in IC technology allowed combining the speed and drive of the S family with the lower power of LS to form a new logic family called FAST (Fairchild Advanced Schottky TTL). In 1985, AC/ACT (Advanced CMOS Technology), a much higher speed version of HCMOS, was introduced. With the introduction of FCT (Fast CMOS Technology) in 1986, the speed gap between CMOS and TTL at last was closed. Because FCT is the CMOS version of FAST, it has the low power consumption of CMOS but the speed is comparable with TTL. Table C-2 provides an overview of logic families up to FCT.

Table C-2: Logic Family Overview

<table>
<thead>
<tr>
<th>Product</th>
<th>Year Introduced</th>
<th>Speed (ns)</th>
<th>Static Supply Current (mA)</th>
<th>High/Low Family Drive (mA)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Std TTL</td>
<td>1968</td>
<td>40</td>
<td>30</td>
<td>-2/32</td>
</tr>
<tr>
<td>CD4K/74C</td>
<td>1970</td>
<td>70</td>
<td>0.3</td>
<td>-0.48/6.4</td>
</tr>
<tr>
<td>LS/S</td>
<td>1971</td>
<td>18</td>
<td>54</td>
<td>-15/24</td>
</tr>
<tr>
<td>HC/HCT</td>
<td>1977</td>
<td>25</td>
<td>0.08</td>
<td>-6/-6</td>
</tr>
<tr>
<td>FAST</td>
<td>1978</td>
<td>6.5</td>
<td>90</td>
<td>-15/64</td>
</tr>
<tr>
<td>AS</td>
<td>1980</td>
<td>6.2</td>
<td>90</td>
<td>-15/64</td>
</tr>
<tr>
<td>ALS</td>
<td>1980</td>
<td>10</td>
<td>27</td>
<td>-15/64</td>
</tr>
<tr>
<td>AC/ACT</td>
<td>1985</td>
<td>10</td>
<td>0.08</td>
<td>-24/24</td>
</tr>
<tr>
<td>FCT</td>
<td>1986</td>
<td>6.5</td>
<td>1.5</td>
<td>-15/64</td>
</tr>
</tbody>
</table>


Recent advances in logic families

As the speed of high-performance microprocessors reached 25 MHz, it shortened the CPU's cycle time, leaving less time for the path delay. Designers normally allocate no more than 25% of a CPU's cycle time budget to path delay. Following this rule means that there must be a corresponding decline in the propagation delay of logic families used in the address and data path as the system frequency is increased. In recent years, many semiconductor manufacturers have responded to this need by providing logic families that have high speed, low noise, and high drive I/O. Table C-3 provides the characteristics of high-performance logic families introduced in recent years. ACQ/ACTQ are the second-generation advanced CMOS (ACMOS) with much lower noise. While ACQ has the CMOS input level, ACTQ is equipped with TTL-level input. The FCTx and FCTx-T are second-generation FCT with much higher speed. The “x” in the FCTx and FCTx-T refers to various speed grades, such as A, B, and C, where A means low speed and C means high speed. For designers who are well versed in using the FAST logic family, FASTr is an ideal choice because it is faster than FAST, has higher driving capability ($I_{OL}$, $I_{OH}$), and produces much lower noise than FAST. At the time of this writing, next to ECL and gallium arsenide logic gates, FASTr is the fastest logic family in the market (with the 5 V $V_{CC}$), but the power consumption is high relative to other logic families, as shown in Table C-3. The combining of
high-speed bipolar TTL and the low power consumption of CMOS has given birth to what is called BICMOS. Although BICMOS seems to be the future trend in IC design, at this time it is expensive due to extra steps required in BICMOS IC fabrication, but in some cases there is no other choice. (For example, Intel's Pentium microprocessor, a BICMOS product, had to use high-speed bipolar transistors to speed up some of the internal functions.) Table C-3 provides advanced logic characteristics. The “x” is for different speeds designated as A, B, and C. A is the slowest one while C is the fastest one. The above data is for the 74244 buffer.

Table C-3: Advanced Logic General Characteristics

<table>
<thead>
<tr>
<th>Family</th>
<th>Year</th>
<th>Number of Suppliers</th>
<th>Tech Base</th>
<th>I/O Level</th>
<th>Speed (ns)</th>
<th>Static Current</th>
<th>I\textsubscript{OH}/I\textsubscript{OL}</th>
</tr>
</thead>
<tbody>
<tr>
<td>ACQ</td>
<td>1989</td>
<td>2</td>
<td>CMOS</td>
<td>CMOS/CMOS</td>
<td>6.0</td>
<td>80 µA</td>
<td>-24/24 mA</td>
</tr>
<tr>
<td>ACTQ</td>
<td>1989</td>
<td>2</td>
<td>CMOS</td>
<td>TTL/CMOS</td>
<td>7.5</td>
<td>80 µA</td>
<td>-24/24 mA</td>
</tr>
<tr>
<td>FCTx</td>
<td>1987</td>
<td>3</td>
<td>CMOS</td>
<td>TTL/CMOS</td>
<td>4.1–4.8</td>
<td>1.5 mA</td>
<td>-15/64 mA</td>
</tr>
<tr>
<td>FCTxT</td>
<td>1990</td>
<td>2</td>
<td>CMOS</td>
<td>TTL/TTL</td>
<td>4.1–4.8</td>
<td>1.5 mA</td>
<td>-15/64 mA</td>
</tr>
<tr>
<td>FASTr</td>
<td>1990</td>
<td>1</td>
<td>Bipolar</td>
<td>TTL/TTL</td>
<td>3.9</td>
<td>50 mA</td>
<td>-15/64 mA</td>
</tr>
<tr>
<td>BCT</td>
<td>1987</td>
<td>2</td>
<td>BICMOS</td>
<td>TTL/TTL</td>
<td>5.5</td>
<td>10 mA</td>
<td>-15/64 mA</td>
</tr>
</tbody>
</table>

Since the late 70s, the use of a +5 V power supply has become standard in all microprocessors and microcontrollers. To reduce power consumption, 3.3 V \( V_{CC} \) is being embraced by many designers. The lowering of \( V_{CC} \) to 3.3 V has two major advantages: (1) it lowers the power consumption, prolonging the life of the battery in systems using a battery, and (2) it allows a further reduction of line size (design rule) to submicron dimensions. This reduction results in putting more transistors in a given die size. As fabrication processes improve, the decline in the line size is reaching submicron level and transistor densities are approaching 1 billion transistors.

Open-collector and open-drain gates

To allow multiple outputs to be connected together, we use open-collector logic gates. In such cases, an external resistor will serve as load. This is shown in Figures C-5 and C-6.

Figure C-5. Open Collector

Figure C-6. Open Drain
SECTION C.2: PIC18 I/O PORT STRUCTURE AND INTERFACING

In interfacing the PIC18 microcontroller with other IC chips or devices, fan-out is the most important issue. To understand the PIC18 fan-out we must first understand the port structure of the PIC18. This section provides a detailed discussion of the PIC18 port structure and its fan-out. It is very critical that we understand the I/O port structure of the PIC18 lest we damage it while trying to interface it with an external device.

IC fan-out

When connecting IC chips together, we need to find out how many input pins can be driven by a single output pin. This is a very important issue and involves the discussion of what is called IC fan-out. The IC fan-out must be addressed for both logic "0" and logic "1" outputs. See Example C-1. Fan-out for logic LOW and fan-out for logic HIGH are defined as follows:

\[
\text{fan-out (of LOW)} = \frac{I_{OL}}{I_{IL}} \quad \text{fan-out (of HIGH)} = \frac{I_{OH}}{I_{IH}}
\]

Of the above two values, the lower number is used to ensure the proper noise margin. Figure C-7 shows the sinking and sourcing of current when ICs are connected together.

![Figure C-7. Current Sinking and Sourcing in TTL](image)

Notice that in Figure C-7, as the number of input pins connected to a single output increases, \( I_{OL} \) rises, which causes \( V_{OL} \) to rise. If this continues, the rise of \( V_{OL} \) makes the noise margin smaller, and this results in the occurrence of false logic due to the slightest noise.
Example C-1

Find how many unit loads (UL) can be driven by the output of the LS logic family.

Solution:

The unit load is defined as $I_{UL} = 1.6 \text{ mA}$ and $I_{IH} = 40 \mu\text{A}$. Table C-1 shows $I_{OH} = 400 \mu\text{A}$ and $I_{OL} = 8 \text{ mA}$ for the LS family. Therefore, we have

$$\text{fan-out (LOW)} = \frac{I_{OL}}{I_{UL}} = \frac{8 \text{ mA}}{1.6 \text{ mA}} = 5$$

$$\text{fan-out (HIGH)} = \frac{I_{OH}}{I_{IH}} = \frac{400 \mu\text{A}}{40 \mu\text{A}} = 10$$

This means that the fan-out is 5. In other words, the LS output must not be connected to more than 5 inputs with unit load characteristics.

74LS244 and 74LS245 buffers/drivers

In cases where the receiver current requirements exceed the driver’s capability, we must use buffers/drivers such as the 74LS245 and 74LS244. Figure C-8 shows the internal gates for the 74LS244 and 74LS245. The 74LS245 is used for bidirectional data buses, and the 74LS244 is used for unidirectional address buses.

[Figure C-8 (a). 74LS244 Octal Buffer]

[Figure C-8 (b). 74LS245 Bidirectional Buffer]
Tri-state buffer

Notice that the 74LS244 is simply 8 tri-state buffers in a single chip. As shown in Figure C-9 a tri-state buffer has a single input, a single output, and the enable control input. By activating the enable, data at the input is transferred to the output. The enable can be an active-LOW or an active-HIGH. Notice that the enable input for the 74LS244 is an active-LOW whereas the enable input pin for Figure C-9 is active-HIGH.

74LS245 and 74LS244 fan-out

It must be noted that the output of the 74LS245 and 74LS244 can sink and source a much larger amount of current than that of other LS gates. See Table C-4. That is the reason we use these buffers for driver when a signal is travelling a long distance through a cable or it has to drive many inputs.

Table C-4: Electrical Specifications for Buffers/Drivers

<table>
<thead>
<tr>
<th></th>
<th>I_{OH} (mA)</th>
<th>I_{OL} (mA)</th>
</tr>
</thead>
<tbody>
<tr>
<td>74LS244</td>
<td>3</td>
<td>12</td>
</tr>
<tr>
<td>74LS245</td>
<td>3</td>
<td>12</td>
</tr>
</tbody>
</table>

After this background on the fan-out, next we discuss the structure of PIC18 ports.

PIC18 port structure and operation

Because all the ports of the PIC18 are bidirectional they all have the following four components in their structure:

1. Data latch
2. Output driver
3. Input buffer
4. TRIS latch

Figure C-10 shows the structure of a port and its four components. Notice that in Figure C-10, the PIC18 ports have both the latch and buffer. Now the question is, in reading the port, are we reading the status of the input pin or are we read-
Inputting (Reading) 1 from a Pin in the PIC18

The status of the latch? That is an extremely important question and its answer depends on which instruction we are using. Therefore, when reading the ports there are two possibilities: (1) reading the input pin, or (2) reading the latch. The above distinction is very important and must be understood lest you damage the PIC18 port. Each is described next.

**Reading the pin when TRIS = 1 (Input)**

As we stated in Chapter 4, to make any bits of any port of the PIC18 an input port, we first must write a 1 (logic HIGH) to the TRIS bit. Look at the following sequence of events to see why:

1. As can be seen from Figure C-10, a 1 written to the TRIS latch has “HIGH” on its Q. Therefore, Q = 1 and Q = 0. Because Q = 1, it turns off the P transistor.
2. Because Q = 0 and is connected to the gate of the N transistor, the N transistor is off.
3. When both transistors are off, they block any path to the ground or VCC for any signal connected to the input pin, and the input signal is directed to the buffer.
4. When reading the input port in instructions such as “MOVFW PORTB” we are really reading the data present at the pin. In other words, it is bringing into the CPU the status of the external pin. This instruction activates the read pin of buffer and lets data at the pins flow into the CPU’s internal bus. Figures C-10 and C-11 show HIGH and LOW signals at the input, respectively.
The above discussion showed why we must write a “HIGH” to a port’s TRIS bits in order to make it an input port. What happens if we write a “0” to TRIS that was configured as an input port? From Figure C-12 we see that when TRIS = 0, if we write a 0 to the Data latch, then Q = 0 and \( \overline{Q} = 1 \). As a result of \( \overline{Q} = 1 \), the N transistor is “on” and the P transistor is “off.” If N is “on,” it provides the path to ground for the input pin. Therefore, any attempt to read the input pin will always get the “LOW” ground signal. Figure C-13 shows what happens when we write “HIGH” to output port (Data latch) when TRIS = 0. Writing 1 to the Data latch makes \( \overline{Q} = 0 \). As a result of that, the P transistor is “on” and the N transistor is “off,” which allows a 1 to be provided to the output pin. Therefore, any attempt to read the input pin will always get the “HIGH” signal.

**Avoid damaging the port**

The following methods can be used as precautions to prevent damage to the PIC18 ports:

1. Have a 10k ohms resistor on the V\(_{CC}\) path to limit current flow.
2. Connect any input switch to a 74LS244 tri-state buffer before it is fed to the PIC18 pin.

The above points are extremely important and must be emphasized because many people damage their ports and afterwards wonder how it happened. We must also use the right instruction when we want to read the status of an input pin. Table C-5 shows the list of instructions in which reading the port reads the status of the input pin.

736
Figure C-12. Outputting (Writing) 0 to a Pin in the PIC18

Figure C-13. Outputting (Writing) 1 to a Pin in the PIC18
Table C-5: Some of the Instructions Reading the Status of Input Port

<table>
<thead>
<tr>
<th>Mnemonics</th>
<th>Examples</th>
</tr>
</thead>
<tbody>
<tr>
<td>MOVFW PORTx</td>
<td>MOVFW PORTB</td>
</tr>
<tr>
<td>TSTFSZ f</td>
<td>TSTFSZ PORTC</td>
</tr>
<tr>
<td>BTFSS f, b</td>
<td>BTFSS PORTD, 0</td>
</tr>
<tr>
<td>BTFSC f, b</td>
<td>BTFSC PORTB, 7</td>
</tr>
<tr>
<td>CPFSEQ f</td>
<td>CPFSEQ PORTB</td>
</tr>
</tbody>
</table>

PIC18 port fan-out

Now that we are familiar with the port structure of the PIC18, we need to examine the fan-out for the PIC18 microcontroller. While the early chips were based on NMOS IC technology, today's PIC18 microcontrollers are all based on CMOS technology. Note, however, that while the core of the PIC18 microcontroller is CMOS, the circuitry driving its pins is all TTL compatible. That is, the PIC18 is a CMOS-based product with TTL-compatible pins. All the ports of the PIC18 have the same I/O structure, and therefore the same fan-out. Table C-6 provides the I/O characteristics of PIC18F458 ports.

Table C-6: PIC18 Fan-out for PORTS

<table>
<thead>
<tr>
<th>Pin</th>
<th>Fan-out</th>
</tr>
</thead>
<tbody>
<tr>
<td>IOL</td>
<td>8.5 mA</td>
</tr>
<tr>
<td>IOH</td>
<td>-3 mA</td>
</tr>
<tr>
<td>III</td>
<td>1 μA</td>
</tr>
<tr>
<td>IIH</td>
<td>1 μA</td>
</tr>
</tbody>
</table>

Note: Negative current is defined as current sourced by the pin.

74LS244 driving an output pin

In some cases, when an PIC18 port is driving multiple inputs, or driving a single input via a long wire or cable (e.g., printer cable), we can use the 74LS244 as a driver. When driving an off-board circuit, placing the 74LS244 buffer between your PIC18 and the circuit is essential because the PIC18 lacks sufficient current. See Figure C-14.
SECTION C.3: SYSTEM DESIGN ISSUES

In addition to fan-out, the other issues related to system design are power dissipation, ground bounce, $V_{CC}$ bounce, crosstalk, and transmission lines. In this section we provide an overview of these topics.

Power dissipation considerations

Power dissipation of a system is a major concern of system designers, especially for laptop and hand-held systems in which batteries provide the power. Power dissipation is a function of frequency and voltage as shown below:

$$Q = CV$$

$$\frac{Q}{T} = \frac{CV}{T}$$

since

$$F = \frac{1}{T}$$

and

$$I = \frac{Q}{T}$$

now

$$P = VI = CV^2F$$

In the above equations, the effects of frequency and $V_{CC}$ voltage should be noted. While the power dissipation goes up linearly with frequency, the impact of the power supply voltage is much more pronounced (squared). See Example C-2.

Example C-2

Compare the power consumption of two microcontroller-based systems. One uses 5 V and the other uses 3 V for $V_{CC}$.

Solution:
Because $P = VI$, by substituting $I = V/R$ we have $P = V^2/R$. Assuming that $R = 1$, we have $P = 5^2 = 25$ W and $P = 3^2 = 9$ W. This results in using 16 W less power, which means power saving of 64%. ($16/25 \times 100$) for systems using 3 V for power source.

Dynamic and static currents

Two major types of currents flow through an IC: dynamic and static. A dynamic current is $I_{DC} = CVF$. It is a function of the frequency under which the component is working. This means that as the frequency goes up, the dynamic current and power dissipation go up. The static current, also called DC, is the current consumption of the component when it is inactive (not selected). The dynamic current dissipation is much higher than the static current consumption. To reduce power consumption, many microcontrollers, including the PIC18, have power-saving modes. In the PIC18, the power saving mode is called sleep mode. We describe the sleep mode next.
**Sleep mode**

In sleep mode the on-chip oscillator is frozen, which cuts off frequency to the CPU and peripheral functions, such as serial ports, interrupts, and timers. Notice that while this mode brings power consumption down to an absolute minimum, the contents of RAM and the SFR registers are saved and remain unchanged.

**Ground bounce**

One of the major issues that designers of high-frequency systems must grapple with is ground bounce. Before we define ground bounce, we will discuss lead inductance of IC pins. There is a certain amount of capacitance, resistance, and inductance associated with each pin of the IC. The size of these elements varies depending on many factors such as length, area, and so on.

The inductance of the pins is commonly referred to as *self-inductance* because there is also what is called *mutual inductance*, as we will show below. Of the three components of capacitor, resistor, and inductor, the property of self-inductance is the one that causes the most problems in high-frequency system design because it can result in ground bounce. Ground bounce occurs when a massive amount of current flows through the ground pin caused by many outputs changing from HIGH to LOW all at the same time. See Figure C-15(a). The voltage is related to the inductance of the ground lead as follows:

\[ V = L \frac{di}{dt} \]

As we increase the system frequency, the rate of dynamic current, \( di/dt \), is also increased, resulting in an increase in the inductance voltage \( L \ (di/dt) \) of the ground pin. Because the LOW state (ground) has a small noise margin, any extra voltage due to the inductance can cause a false signal. To reduce the effect of ground bounce, the following steps must be taken where possible:

1. The \( V_{CC} \) and ground pins of the chip must be located in the middle rather than at opposite ends of the IC chip (the 14-pin TTL logic IC uses pins 14 and 7 for ground and \( V_{CC} \)). This is exactly what we see in high-performance logic gates such as Texas Instruments' advanced logic AC11000 and ACT11000 families. For example, the ACT11013 is a 14-pin DIP chip in which pin numbers 4 and 11 are used for the ground and \( V_{CC} \), instead of 7 and 14 as in the traditional TTL family. We can also use the SOIC packages instead of DIP.

2. Another solution is to use as many pins for ground and \( V_{CC} \) as possible to reduce the lead length. This is exactly why all high-performance microprocessors and logic families use many pins for \( V_{CC} \) and ground instead of the traditional single pin for \( V_{CC} \) and single pin for GND. For example, in the case of Intel's Pentium processor there are over 50 pins for ground, and another 50 pins for \( V_{CC} \).
The above discussion of ground bounce is also applicable to $V_{CC}$ when a large number of outputs changes from the LOW to the HIGH state; this is referred to as $V_{CC}$ bounce. However, the effect of $V_{CC}$ bounce is not as severe as ground bounce because the HIGH ("1") state has a wider noise margin than the LOW ("0") state.

**Filtering the transient currents using decoupling capacitors**

In the TTL family, the change of the output from LOW to HIGH can cause what is called *transient current*. In a totem-pole output in which the output is LOW, Q4 is on and saturated, whereas Q3 is off. By changing the output from the LOW to the HIGH state, Q3 turns on and Q4 turns off. This means that there is a time when both transistors are on and drawing current from $V_{CC}$. The amount of current depends on the $R_{ON}$ values of the two transistors, which in turn depend on the internal parameters of the transistors. The net effect of this, however, is a large amount of current in the form of a spike for the output current, as shown in Figure C-15(b). To filter the transient current, a $0.01 \mu F$ or $0.1 \mu F$ ceramic disk capacitor can be placed between the $V_{CC}$ and ground for each TTL IC. The lead for this capacitor, however, should be as small as possible because a long lead results in a large self-inductance, and that results in a spike on the $V_{CC}$ line [$V = L \frac{di}{dt}$]. This spike is called $V_{CC}$ bounce. The ceramic capacitor for each IC is referred to as a *decoupling capacitor*. There is also a bulk decoupling capacitor, as described next.

**Bulk decoupling capacitor**

If many IC chips change state at the same time, the combined currents drawn from the board's $V_{CC}$ power supply can be massive and may cause a fluctuation of $V_{CC}$ on the board where all the ICs are mounted. To eliminate this, a relatively large decoupling tantalum capacitor is placed between the $V_{CC}$ and ground lines. The size and location of this tantalum capacitor varies depending on the number of ICs on the board and the amount of current drawn by each IC, but it is
common to have a single 22 μF to 47 μF capacitor for each of the 16 devices, placed between the \( V_{CC} \) and ground lines.

**Crosstalk**

Crosstalk is due to mutual inductance. See Figure C-16. Previously, we discussed self-inductance, which is inherent in a piece of conductor. *Mutual inductance* is caused by two electric lines running parallel to each other. The mutual inductance is a function of the length of two conductors running in parallel, the distance between them, and the medium material placed between them. The effect of crosstalk can be reduced by increasing the distance between the parallel or adjacent lines (in printed circuit boards, they will be traces). In many cases, such as printer and disk drive cables, there is a dedicated ground for each signal. Placing ground lines (traces) between signal lines reduces the effect of crosstalk. This method is used even in some ACT logic families where a \( V_{CC} \) and a GND pin are next to each other. Crosstalk is also called *EMI* (electromagnetic interference). This is in contrast to *ESI* (electrostatic interference), which is caused by capacitive coupling between two adjacent conductors.

**Transmission line ringing**

The square wave used in digital circuits is in reality made of a single fundamental pulse and many harmonics of various amplitudes. When this signal travels on the line, not all the harmonics respond in the same way to the capacitance, inductance, and resistance of the line. This causes what is called *ringing*, which depends on the thickness and the length of the line driver, among other factors. To reduce the effect of ringing, the line drivers are terminated by putting a resistor at the end of the line. See Figure C-17. There are three major methods of line driver termination: parallel, serial, and Thevenin.

In serial termination, resistors of 30–50 ohms are used to terminate the line. The parallel and Thevenin methods are used in cases where there is a need to match the impedance of the line with the load impedance. This requires a detailed analysis of the signal traces and load impedance, which is beyond the scope of this book. In high-frequency systems, wire traces on the printed circuit board (PCB) behave like transmission lines, causing ringing. The severity of this ringing depends on the speed and the logic family used. Table C-7 provides the length of the traces, beyond which the traces must be looked at as transmission lines.
Table C-7: Line Length Beyond Which Traces Behave Like Transmission Lines

<table>
<thead>
<tr>
<th>Logic Family</th>
<th>Line Length (in.)</th>
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</thead>
<tbody>
<tr>
<td>LS</td>
<td>25</td>
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<tr>
<td>S, AS</td>
<td>11</td>
</tr>
<tr>
<td>F, ACT</td>
<td>8</td>
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<tr>
<td>AS, ECL</td>
<td>6</td>
</tr>
<tr>
<td>FCT, FCTA</td>
<td>5</td>
</tr>
</tbody>
</table>

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This appendix provides an introduction to writing flowcharts and pseudocode.
Flowcharts

If you have taken any previous programming courses, you are probably familiar with flowcharting. Flowcharts use graphic symbols to represent different types of program operations. These symbols are connected together into a flowchart to show the flow of execution of a program. Figure D-1 shows some of the more commonly used symbols. Flowchart templates are available to help you draw the symbols quickly and neatly.

Pseudocode

Flowcharting has been standard practice in industry for decades. However, some find limitations in using flowcharts, such as the fact that you can't write much in the little boxes, and it is hard to get the "big picture" of what the program does without getting bogged down in the details. An alternative to using flowcharts is pseudocode, which involves writing brief descriptions of the flow of the code. Figures D-2 through D-6 show flowcharts and pseudocode for commonly used control structures.
Structured programming uses three basic types of program control structures: sequence, control, and iteration. Sequence is simply executing instructions one after another. Figure D-2 shows how sequence can be represented in pseudocode and flowcharts.

Figures D-3 and D-4 show two control programming structures: IF-THEN-ELSE and IF-THEN in both pseudocode and flowcharts.

Note in Figures D-2 through D-6 that “statement” can indicate one statement or a group of statements.

Figures D-5 and D-6 show two iteration control structures: REPEAT UNTIL and WHILE DO. Both structures execute a statement or group of statements repeatedly. The difference between them is that the REPEAT UNTIL structure always executes the statement(s) at least once, and checks the condition after each iteration, whereas the WHILE DO may not execute the statement(s) at all because the condition is checked at the beginning of each iteration.

**Figure D-3. IF THEN ELSE Pseudocode versus Flowchart**

```
IF (condition) THEN
    Statement 1
ELSE
    Statement 2
```

**Figure D-4. IF THEN Pseudocode versus Flowchart**

```
IF (condition) THEN
    Statement
```

APPENDIX D: FLOWCHARTS AND PSEUDOCODE
Program D-1 finds the sum of a series of bytes. Compare the flowchart versus the pseudocode for Program D-1 (shown in Figure D-7). In this example, more program details are given than one usually finds. For example, this shows steps for initializing and decrementing counters. Another programmer may not include these steps in the flowchart or pseudocode. It is important to remember that the purpose of flowcharts or pseudocode is to show the flow of the program and what the program does, not the specific Assembly language instructions that accomplish the program's objectives. Notice also that the pseudocode gives the same information in a much more compact form than does the flowchart. It is important to note that sometimes pseudocode is written in layers, so that the outer level or layer shows the flow of the program and subsequent levels show more details of how the program accomplishes its assigned tasks.
Count = 5
Address = 40H
Repeat
   Add next byte
   Increment address
   Decrement counter
Until Count = 0
Store Sum

Figure D-7. Pseudocode versus Flowchart for Program D-1

<table>
<thead>
<tr>
<th>COUNTVAL</th>
<th>EQU 5</th>
<th>COUNT = 5</th>
</tr>
</thead>
<tbody>
<tr>
<td>COUNTREG</td>
<td>SET 0x20</td>
<td>set aside location 20H for counter</td>
</tr>
<tr>
<td>SUM</td>
<td>SET 0x30</td>
<td>set aside location 30H for sum</td>
</tr>
<tr>
<td>MOVLW</td>
<td>COUNTVAL</td>
<td>WREG = 5</td>
</tr>
<tr>
<td>MOVWF</td>
<td>COUNTREG</td>
<td>load the counter</td>
</tr>
<tr>
<td>LFSR</td>
<td>0,0x40</td>
<td>load pointer. FSR0 = 40H, RAM address</td>
</tr>
<tr>
<td>CLR</td>
<td>WREG</td>
<td>clear WREG</td>
</tr>
<tr>
<td>B5</td>
<td>ADDWF</td>
<td>POSTINC0, W</td>
</tr>
<tr>
<td>DECF</td>
<td>COUNTREG,F</td>
<td>decrement counter</td>
</tr>
<tr>
<td>BNZ B5</td>
<td>MOVWF</td>
<td>SUM</td>
</tr>
<tr>
<td></td>
<td>SUM</td>
<td>store WREG in SUM</td>
</tr>
</tbody>
</table>

Program D-1
# APPENDIX E.1

## PIC18 PRIMER FOR x86 PROGRAMMERS

<table>
<thead>
<tr>
<th>x86</th>
<th>PIC18</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>8-bit registers:</strong></td>
<td><strong>WREG and up to 256 RAM locations in Access Bank</strong></td>
</tr>
<tr>
<td>AL, AH, BL, BH,</td>
<td></td>
</tr>
<tr>
<td>CL, CH, DL, DH</td>
<td></td>
</tr>
<tr>
<td><strong>16-bit (data pointer):</strong></td>
<td><strong>TBLPTR</strong></td>
</tr>
<tr>
<td>BX, SI, DI</td>
<td></td>
</tr>
<tr>
<td><strong>Program Counter:</strong></td>
<td><strong>PC (21-bit)</strong></td>
</tr>
<tr>
<td>IP (16-bit)</td>
<td></td>
</tr>
</tbody>
</table>

### Input:

- MOV DX, port addr
- IN AL, DX
- MOVFW PORTx ; (x = A, B, ..G)

### Output:

- MOV DX, port addr
- OUT DX, AL
- MOVFW PORTx ; (x = A, B, ..G)

### Loop:

- DEC CL
- JNZ TARGET
- DECF MyReg, F
- BNZ TARGET

### Stack pointer:

- SP (16-bit)
- SP (21-bit)

As we PUSH data onto the stack, it decrements the SP. Push increments the SP. (Used exclusively for saving PC)

As we POP data from the stack, it increments the SP. Pop decrements the SP. (Used exclusively for retrieving PC)

### Data movement:

**From the code segment:**

- MOV AL, CS: [SI]  
  TBLRD

**From the data segment:**

- MOV AL, [SI]  
  MOVFW FSRx

**From RAM:**

- MOV AL, [SI]  
  MOVFW FSRx  
  (Use SI, DI, or BX only.)

**To RAM:**

- MOV [SI], AL  
  MOVWF FSRx
APPENDIX E.2

PIC18 PRIMER FOR 8051 PROGRAMMERS

<table>
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<tr>
<th>8051</th>
<th>PIC18</th>
</tr>
</thead>
<tbody>
<tr>
<td>8-bit registers:</td>
<td>A, B, R0, R1, .....R7</td>
</tr>
<tr>
<td>16-bit (data pointer):</td>
<td>DPTR</td>
</tr>
<tr>
<td>Program Counter:</td>
<td>PC (16-bit)</td>
</tr>
</tbody>
</table>

Input:

MOV A, Pn ; (n=0 - 3) | MOVFW PORTx ; (x = A,B,..G) |

Output:

MOV Pn, A ; (n=0 - 3) | MOVWF PORTx ; (x = A,B,..G) |

Loop:

DJNZ R3, TARGET | DECF MyReg, F |
(Using R0-R7) | BNZ TARGET |

Stack pointer:

SP (8-bit) | SP (21-bit) |

As we PUSH data onto the stack, it increments the SP. | Push increments the SP. (Used exclusively for saving PC)
As we POP data from the stack, it decrements the SP. | Pop decrements the SP. (Used exclusively for retrieving PC)

Data movement:

From the code segment:

MOVCA, @A+PC | TBLRD |

From the data segment:

MOVX A, @DPTR | MOVFW FSRx |

From RAM:

MOV A, @R0 | MOVFW FSRx |
(Use R0 or R1 only) |

To RAM:

MOV @R0, A | MOVWF FSRx |
(Use R0 or R1 only) |
# APPENDIX F

## ASCII CODES

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<th>Dec</th>
<th>Hex</th>
<th>Ch</th>
<th>Code</th>
</tr>
</thead>
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<thead>
<tr>
<th>Ctrl</th>
<th>Dec</th>
<th>Hex</th>
<th>Ch</th>
<th>Code</th>
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<tr>
<td><code>^E</code></td>
<td>96</td>
<td>60</td>
<td>`</td>
<td>776</td>
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<tr>
<td><code>^A</code></td>
<td>97</td>
<td>61</td>
<td>a</td>
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<tr>
<td><code>^B</code></td>
<td>98</td>
<td>62</td>
<td>b</td>
<td>778</td>
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<tr>
<td><code>^C</code></td>
<td>99</td>
<td>63</td>
<td>c</td>
<td>779</td>
</tr>
<tr>
<td><code>^D</code></td>
<td>100</td>
<td>64</td>
<td>d</td>
<td>780</td>
</tr>
<tr>
<td><code>^E</code></td>
<td>101</td>
<td>65</td>
<td>e</td>
<td>781</td>
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<tr>
<td><code>^F</code></td>
<td>102</td>
<td>66</td>
<td>f</td>
<td>782</td>
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<tr>
<td><code>^G</code></td>
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<td>67</td>
<td>g</td>
<td>783</td>
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<td>104</td>
<td>68</td>
<td>h</td>
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<td><code>^I</code></td>
<td>105</td>
<td>69</td>
<td>i</td>
<td>785</td>
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<tr>
<td><code>^J</code></td>
<td>106</td>
<td>6A</td>
<td>j</td>
<td>786</td>
</tr>
<tr>
<td><code>^K</code></td>
<td>107</td>
<td>6B</td>
<td>k</td>
<td>787</td>
</tr>
<tr>
<td><code>^L</code></td>
<td>108</td>
<td>6C</td>
<td>l</td>
<td>788</td>
</tr>
<tr>
<td><code>^M</code></td>
<td>109</td>
<td>6D</td>
<td>m</td>
<td>789</td>
</tr>
<tr>
<td><code>^N</code></td>
<td>110</td>
<td>6E</td>
<td>n</td>
<td>790</td>
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<td><code>^O</code></td>
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<td>6F</td>
<td>o</td>
<td>791</td>
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<tr>
<td><code>^P</code></td>
<td>112</td>
<td>70</td>
<td>p</td>
<td>792</td>
</tr>
<tr>
<td><code>^Q</code></td>
<td>113</td>
<td>71</td>
<td>q</td>
<td>793</td>
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<tr>
<td><code>^R</code></td>
<td>114</td>
<td>72</td>
<td>r</td>
<td>794</td>
</tr>
<tr>
<td><code>^S</code></td>
<td>115</td>
<td>73</td>
<td>s</td>
<td>795</td>
</tr>
<tr>
<td><code>^T</code></td>
<td>116</td>
<td>74</td>
<td>t</td>
<td>796</td>
</tr>
<tr>
<td><code>^U</code></td>
<td>117</td>
<td>75</td>
<td>u</td>
<td>797</td>
</tr>
<tr>
<td><code>^V</code></td>
<td>118</td>
<td>76</td>
<td>v</td>
<td>798</td>
</tr>
<tr>
<td><code>^W</code></td>
<td>119</td>
<td>77</td>
<td>w</td>
<td>799</td>
</tr>
<tr>
<td><code>^X</code></td>
<td>120</td>
<td>78</td>
<td>x</td>
<td>800</td>
</tr>
<tr>
<td><code>^Y</code></td>
<td>121</td>
<td>79</td>
<td>y</td>
<td>801</td>
</tr>
<tr>
<td><code>^Z</code></td>
<td>122</td>
<td>7A</td>
<td>z</td>
<td>802</td>
</tr>
<tr>
<td><code>^I</code></td>
<td>123</td>
<td>7B</td>
<td>(</td>
<td>803</td>
</tr>
<tr>
<td><code>\</code></td>
<td>124</td>
<td>7C</td>
<td>)</td>
<td>804</td>
</tr>
<tr>
<td><code>^J</code></td>
<td>125</td>
<td>7D</td>
<td>`</td>
<td>805</td>
</tr>
<tr>
<td><code>^\</code></td>
<td>126</td>
<td>7E</td>
<td>~</td>
<td>806</td>
</tr>
<tr>
<td><code>-</code></td>
<td>127</td>
<td>7F</td>
<td>o</td>
<td>807</td>
</tr>
</tbody>
</table>
This appendix provides various sources for PIC18 assemblers and trainers. In addition, it lists some suppliers for chips and other hardware needs. While these are all established products from well-known companies, neither the authors nor the publisher assumes responsibility for any problem that may arise with any of them. You are neither encouraged nor discouraged from purchasing any of the products mentioned; you must make your own judgment in evaluating the products. This list is simply provided as a service to the reader. It also must be noted that the list of products is by no means complete or exhaustive.

**PIC18 assemblers**

The PIC18 assembler is provided by Microchip and other companies. Some of the companies provide shareware versions of their products, which you can download from their Web sites. However, the size of code for these shareware versions is limited to a few KB. Figure G-1 lists some suppliers of assemblers.

**PIC18 trainers**

There are many companies that produce and market PIC18 trainers. Figure G-2 provides a list of some of them.

---

**Figure G-1. Suppliers of Assemblers and Compilers**

- Microchip Corp.
  - www.microchip.com
- Custom Computer Services Inc.
  - www.ccsinfo.com

**Figure G-2. Trainer Suppliers**

- Microchip Corp.
  - www.microchip.com
  - www.MicroDigitalEd.com
- Custom Computer Services Inc.
  - www.ccsinfo.com
- RSR Electronics
  - www.elexp.com
## Parts Suppliers

Figure G-3 provides a list of suppliers for many electronics parts.

<table>
<thead>
<tr>
<th>Supplier</th>
<th>Address</th>
<th>Phone Numbers</th>
<th>Website</th>
</tr>
</thead>
<tbody>
<tr>
<td>RSR Electronics</td>
<td>365 Blair Road, Avenel, NJ 07001</td>
<td>(732) 381-1572, 1-800-972-2225</td>
<td><a href="http://www.rsrelectronics.com">www.rsrelectronics.com</a></td>
</tr>
<tr>
<td>Electronix Express</td>
<td>N. Main St., Mansfield, TX 76063</td>
<td>1-800-346-6873</td>
<td><a href="http://www.electronixexpress.com">www.electronixexpress.com</a></td>
</tr>
<tr>
<td>Mouser Electronics</td>
<td>958 N. Main St., Mansfield, TX 76063</td>
<td>1-800-346-6873</td>
<td><a href="http://www.mouser.com">www.mouser.com</a></td>
</tr>
<tr>
<td>Jameco Electronic</td>
<td>1355 Shoreway Road, Belmont, CA 94002-4100</td>
<td>1-800-831-4242</td>
<td><a href="http://www.jameco.com">www.jameco.com</a></td>
</tr>
<tr>
<td>Electronix</td>
<td>11342 IH-35 North, San Antonio, TX 78233</td>
<td>(210) 637-3264, 1-800-531-5369</td>
<td><a href="http://www.elexp.com">www.elexp.com</a></td>
</tr>
<tr>
<td>Altex Electronics</td>
<td><a href="http://www.altex.com">www.altex.com</a></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Digi-Key</td>
<td>1-800-344-4539 (1-800-DIGI-KEY), 1-800-276-2206 (orders only)</td>
<td>(210) 637-3264, (972) 271-5546, (972) 271-2462</td>
<td><a href="http://www.digikey.com">www.digikey.com</a></td>
</tr>
<tr>
<td>Radio Shack</td>
<td><a href="http://www.radioshack.com">www.radioshack.com</a></td>
<td></td>
<td></td>
</tr>
<tr>
<td>B. G. Micro</td>
<td>P. O. Box 280298, Dallas, TX 75228</td>
<td>1-800-276-2206 (orders only), (972) 271-5546, (972) 271-2462</td>
<td><a href="http://www.bgmicro.com">www.bgmicro.com</a></td>
</tr>
<tr>
<td>JDR Microdevices</td>
<td>1-800-538-5000, (408) 494-1400, 1-800-538-5005, (408) 494-1420</td>
<td>(972) 242-8702</td>
<td><a href="http://www.jdr.com">www.jdr.com</a></td>
</tr>
<tr>
<td>Tanner Electronics</td>
<td>1100 Valwood Parkway, Suite #100, Carrollton, TX 75006</td>
<td>(972) 242-8702</td>
<td><a href="http://www.tannerelectronics.com">www.tannerelectronics.com</a></td>
</tr>
</tbody>
</table>

Figure G-3. Electronics Suppliers
APPENDIX H

DATA SHEETS
25.0 INSTRUCTION SET SUMMARY

PIC18F2480/2580/4480/4580 devices incorporate the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

25.1 Standard Instruction Set

The standard PIC18 instruction set adds many enhancements to the previous PICmicro® instruction sets, while maintaining an easy migration from these PICmicro instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal operations
- Control operations

The PIC18 instruction set summary in Table 25-2 lists byte-oriented, bit-oriented, literal and control operations. Table 25-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

1. The file register (specified by 'f')
2. The destination of the result (specified by 'd')
3. The accessed memory (specified by 'a')

The file register designator 'f' specifies which file register is to be used by the instruction. The destination designator 'd' specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All bit-oriented instructions have three operands:

1. The file register (specified by 'f')
2. The bit in the file register (specified by 'b')
3. The accessed memory (specified by 'a')

The bit field designator 'b' selects the number of the bit affected by the operation, while the file register designator 'f' represents the number of the file in which the bit is located.

The literal instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The control instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSBs are '1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the program counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double-word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 µs. If a conditional test is true, or the program counter is changed as a result of an instruction, the instruction execution time is 2 µs. Two-word branch instructions (if true) would take 3 µs.

Figure 25-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 25-2, lists the standard instructions recognized by the Microchip MPASM™ Assembler.

Section 25.1.1 “Standard Instruction Set” provides a description of each instruction.
<table>
<thead>
<tr>
<th>Field</th>
<th>Description</th>
</tr>
</thead>
</table>
| a     | RAM access bit  
|       | a = 0: RAM location in Access RAM (BSR register is ignored)  
|       | a = 1: RAM bank is specified by BSR register  |
| bbb   | Bit address within an 8-bit file register (0 to 7).  |
| BSR   | Bank Select Register. Used to select the current RAM bank.  |
| C, DC, Z, OV, N | ALU status bits: Carry, Digit Carry, Zero, Overflow, Negative. |
| d     | Destination select bit  
|       | d = 0: store result in WREG  
|       | d = 1: store result in file register f  |
| dest  | Destination: either the WREG register or the specified register file location.  |
| f     | 8-bit Register file address (00h to FFh), or 2-bit FSR designator (0h to 3h).  |
| f, s  | 12-bit Register file address (000h to FFFFh). This is the source address.  |
| f, t  | 12-bit Register file address (000h to FFFFh). This is the destination address.  |
| GIE   | Global Interrupt Enable bit.  |
| k     | Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).  |
| label | Label name  |
| mm    | The mode of the TBLPTR register for the table read and table write instructions.  
|       | Only used with table read and table write instructions:  
|       | * No change to register (such as TBLPTR with table reads and writes)  
|       | + Post-Increment register (such as TBLPTR with table reads and writes)  
|       | - Post-Decrement register (such as TBLPTR with table reads and writes)  
|       | + Pre-Increment register (such as TBLPTR with table reads and writes)  |
| n     | The relative address (2's complement number) for relative branch instructions or the direct address for Cal/ Branch and Return instructions  |
| PC    | Program Counter.  |
| PCL   | Program Counter Low Byte.  |
| PCH   | Program Counter High Byte.  |
| PCLATH| Program Counter High Byte Latch.  |
| PCLATU| Program Counter Upper Byte Latch.  |
| PD    | Power-down bit.  |
| PRODH | Product of Multiply High Byte.  |
| PRODL | Product of Multiply Low Byte.  |
| s     | Fast Call/Return mode select bit  
|       | s = 0: do not update into/from shadow registers  
|       | s = 1: certain registers loaded into/from shadow registers (Fast mode)  |
| TBLPTR| 21-bit Table Pointer (points to a Program Memory location).  |
| TABLET| 8-bit Table Latch.  |
| TO    | Time-out bit.  |
| TOS   | Top-of-Stack.  |
| u     | Unused or unchanged.  |
| WDT   | Watchdog Timer.  |
| WREG  | Working register (accumulator).  |
| x     | Don’t care (‘0’ or ‘1’). The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.  |
| z, s  | 7-bit offset value for indirect addressing of register files (source).  |
| z, t  | 7-bit offset value for indirect addressing of register files (destination).  |
| (     | Optional argument.  |
|  )    | Indicates an indexed address.  |
| [text]| The contents of text.  |
| [expr]<n> | Specifies bit n of the register indicated by the pointer expr.  |
| →     | Assigned to.  |
| < >   | Register bit field.  |
| e     | In the set of.  |
| italics| User defined term (font is Courier).  |
**FIGURE 25-1: GENERAL FORMAT FOR INSTRUCTIONS**

<table>
<thead>
<tr>
<th>Byte-oriented file register operations</th>
<th>Example Instruction</th>
</tr>
</thead>
<tbody>
<tr>
<td>15 10 9 8 7 0</td>
<td>ADDWF MYREG, W, B</td>
</tr>
<tr>
<td>OPCODE</td>
<td>d</td>
</tr>
<tr>
<td>d = 0 for result destination to be WREG register</td>
<td></td>
</tr>
<tr>
<td>d = 1 for result destination to be file register (f)</td>
<td></td>
</tr>
<tr>
<td>a = 0 to force Access Bank</td>
<td></td>
</tr>
<tr>
<td>a = 1 for BSR to select bank</td>
<td></td>
</tr>
<tr>
<td>f = 8-bit file register address</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Byte to Byte move operations (2-word)</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>15 12 11 0</td>
<td>MOVFF MYREG1, MYREG2</td>
</tr>
<tr>
<td>OPCODE</td>
<td>f (Source FILE #)</td>
</tr>
<tr>
<td>15 12 11 0</td>
<td>f (Destination FILE #)</td>
</tr>
<tr>
<td>f = 12-bit file register address</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Bit-oriented file register operations</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>15 12 11 9 8 7 0</td>
<td>BSF MYREG, bit, B</td>
</tr>
<tr>
<td>OPCODE</td>
<td>b (BIT #)</td>
</tr>
<tr>
<td>b = 3-bit position of bit in file register (f)</td>
<td></td>
</tr>
<tr>
<td>a = 0 to force Access Bank</td>
<td></td>
</tr>
<tr>
<td>a = 1 for BSR to select bank</td>
<td></td>
</tr>
<tr>
<td>f = 8-bit file register address</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Literal operations</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>15 8 7 0</td>
<td>MOVLW 7FH</td>
</tr>
<tr>
<td>OPCODE</td>
<td>k (literal)</td>
</tr>
<tr>
<td>k = 8-bit immediate value</td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Control operations</th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>CALL, GOTO and Branch operations</td>
<td></td>
</tr>
<tr>
<td>15 8 7 0</td>
<td>GOTO Label</td>
</tr>
<tr>
<td>OPCODE</td>
<td>n&lt;7:0&gt; (literal)</td>
</tr>
<tr>
<td>15 12 11 0</td>
<td>n&lt;19:8&gt; (literal)</td>
</tr>
<tr>
<td>n = 20-bit immediate value</td>
<td></td>
</tr>
</tbody>
</table>

| 15 8 7 0  | CALL MYFUNC  |
| OPCODE | S | n<7:0> (literal) |
| 15 12 11 0  | n<19:8> (literal) |
| S = Fast bit |

| 15 11 10 0  | BRA MYFUNC  |
| OPCODE | n<10:0> (literal) |

| 15 8 7 0  | BC MYFUNC  |
| OPCODE | n<7:0> (literal) |
### TABLE 25-2: PIC18FXXXX INSTRUCTION SET

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>16-Bit Instruction Word</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>MSb</td>
<td>LSb</td>
<td></td>
</tr>
<tr>
<td><strong>BYTE-ORIENTED OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDWF f, d, a</td>
<td>Add WREG and f</td>
<td>1</td>
<td>001001da fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>ADDWF f d a</td>
<td>Add WREG and Carry bit to f</td>
<td>1</td>
<td>001000da fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>ANDWF f, d, a</td>
<td>AND WREG with f</td>
<td>1</td>
<td>000101da fffe fffe</td>
<td>Z, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>CLRF f, a</td>
<td>Clear f</td>
<td>1</td>
<td>0110101a fffe fffe</td>
<td>Z</td>
<td>2</td>
</tr>
<tr>
<td>COMF f, d, a</td>
<td>Complement f</td>
<td>1</td>
<td>000111da fffe fffe</td>
<td>Z, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>CPFSEQ f, a</td>
<td>Compare f with WREG, skip =</td>
<td>(2 or 3)</td>
<td>0110001a fffe fffe</td>
<td>None</td>
<td>4</td>
</tr>
<tr>
<td>CPFSGT f, a</td>
<td>Compare f with WREG, skip &gt;</td>
<td>(2 or 3)</td>
<td>0110010a fffe fffe</td>
<td>None</td>
<td>4</td>
</tr>
<tr>
<td>CPFSLt f, a</td>
<td>Compare f with WREG, skip &lt;</td>
<td>(2 or 3)</td>
<td>0110000a fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>DECF f, d, a</td>
<td>Decrement f</td>
<td>1</td>
<td>000001da fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>DECFSZ f, d, a</td>
<td>Decrement f, Skip if 0</td>
<td>(2 or 3)</td>
<td>001011da fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>DCFSNZ f, d, a</td>
<td>Decrement f, Skip if Not 0</td>
<td>(2 or 3)</td>
<td>010011da fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>INCF f, d, a</td>
<td>Increment f</td>
<td>1</td>
<td>0011010a fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>INCFSZ f, d, a</td>
<td>Increment f, Skip if 0</td>
<td>(2 or 3)</td>
<td>001110da fffe fffe</td>
<td>None</td>
<td>4</td>
</tr>
<tr>
<td>INFSNZ f, d, a</td>
<td>Increment f, Skip if Not 0</td>
<td>(2 or 3)</td>
<td>010010da fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>IORWF f, d, a</td>
<td>Inclusive OR WREG with f</td>
<td>1</td>
<td>000100da fffe fffe</td>
<td>Z, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>MOVf f, d</td>
<td>Move f</td>
<td>1</td>
<td>010100da fffe fffe</td>
<td>Z, N</td>
<td>1</td>
</tr>
<tr>
<td>MOVFF f, d</td>
<td>Move (source) to 1st word f</td>
<td>2</td>
<td>1100 fffe fffe</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td></td>
<td>(destination) 2nd word</td>
<td></td>
<td>1111 fffe fffe</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>MOVWF f, a</td>
<td>Move WREG to f</td>
<td>1</td>
<td>0110111a fffe fffe</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>MULWF f, a</td>
<td>Multiply WREG with f</td>
<td>1</td>
<td>0000011a fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>NEGf a</td>
<td>Negate f</td>
<td>1</td>
<td>0110101a fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td></td>
</tr>
<tr>
<td>RLCF f, d, a</td>
<td>Rotate Left f through Carry</td>
<td>1</td>
<td>001101da fffe fffe</td>
<td>C, Z, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>RLLCF f, d, a</td>
<td>Rotate Left f (No Carry)</td>
<td>1</td>
<td>010010da fffe fffe</td>
<td>Z, N</td>
<td></td>
</tr>
<tr>
<td>RRCF f, d, a</td>
<td>Rotate Right f through Carry</td>
<td>1</td>
<td>001100da fffe fffe</td>
<td>C, Z, N</td>
<td></td>
</tr>
<tr>
<td>RRNCf f, d, a</td>
<td>Rotate Right f (No Carry)</td>
<td>1</td>
<td>0100100a fffe fffe</td>
<td>Z, N</td>
<td></td>
</tr>
<tr>
<td>SETf a</td>
<td>Set f</td>
<td>1</td>
<td>0110100a fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>SUBFWB f, d, a</td>
<td>Subtract f from WREG with borrow</td>
<td>1</td>
<td>010101da fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td></td>
</tr>
<tr>
<td>SUBWF f, d, a</td>
<td>Subtract WREG from f</td>
<td>1</td>
<td>010111da fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>SUBWFB f, d, a</td>
<td>Subtract WREG from f with borrow</td>
<td>1</td>
<td>0101110a fffe fffe</td>
<td>C, DC, Z, OV, N</td>
<td>1, 2</td>
</tr>
<tr>
<td>SWAPF f, d, a</td>
<td>Swap nibbles in f</td>
<td>1</td>
<td>001110da fffe fffe</td>
<td>None</td>
<td>4</td>
</tr>
<tr>
<td>TSTFSZ f, d, a</td>
<td>Test f, skip if 0</td>
<td>(2 or 3)</td>
<td>0110111a fffe fffe</td>
<td>None</td>
<td>1, 2</td>
</tr>
<tr>
<td>XORWF f, d, a</td>
<td>Exclusive OR WREG with f</td>
<td>1</td>
<td>000110da fffe fffe</td>
<td>Z, N</td>
<td></td>
</tr>
</tbody>
</table>

**Note 1:** When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

**Note 2:** If this instruction is executed on the TMRO register (and where applicable, 'd' = 1), the prescaler will be cleared if assigned.

**Note 3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**Note 4:** Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

**Note 5:** If the table write starts the write cycle to internal memory, the write will continue until terminated.
## TABLE 25-2: PIC18FXXXX INSTRUCTION SET (CONTINUED)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>16-Bit Instruction Word</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>MSb</td>
<td>LSb</td>
<td></td>
</tr>
<tr>
<td><strong>BIT-ORIENTED OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BCF f, b, a</td>
<td>Bit Clear f</td>
<td>1</td>
<td>1001 bbbba</td>
<td>ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td>BSF f, b, a</td>
<td>Bit Set f</td>
<td>1</td>
<td>1000 bbbba</td>
<td>ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td>BTSC f, b, a</td>
<td>Bit Test f, Skip if Clear</td>
<td>(2 or 3)</td>
<td>1011 bbbba</td>
<td>ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td>BTSS f, b, a</td>
<td>Bit Test f, Skip if Set</td>
<td>(2 or 3)</td>
<td>1010 bbbba</td>
<td>ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td>STG f, d, a</td>
<td>Bit Toggle f</td>
<td>1</td>
<td>0111 bbbba</td>
<td>ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td><strong>CONTROL OPERATIONS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>BC n</td>
<td>Branch if Carry</td>
<td>1 (2)</td>
<td>1110 0010</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BN n</td>
<td>Branch if Negative</td>
<td>1 (2)</td>
<td>1110 0110</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BNC n</td>
<td>Branch if Not Carry</td>
<td>1 (2)</td>
<td>1110 0111</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BNN n</td>
<td>Branch if Not Negative</td>
<td>1 (2)</td>
<td>1110 0111</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BNOV n</td>
<td>Branch if Not Overflow</td>
<td>1 (2)</td>
<td>1110 0101</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BNZ n</td>
<td>Branch if Not Zero</td>
<td>1 (2)</td>
<td>1110 0011</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BOV n</td>
<td>Branch if Overflow</td>
<td>1 (2)</td>
<td>1110 0100</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BRA n</td>
<td>Branch Unconditionally</td>
<td>2</td>
<td>1110 1000</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>BZ n</td>
<td>Branch if Zero</td>
<td>1 (2)</td>
<td>1110 0000</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>CALL n, s</td>
<td>Call subroutine 1st word</td>
<td>2</td>
<td>1110 110s</td>
<td>kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td></td>
<td>2nd word</td>
<td></td>
<td>1111 kkkk</td>
<td>kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>CLRWD T</td>
<td>Clear Watchdog Timer</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0100</td>
<td>TO, PD</td>
</tr>
<tr>
<td>DAW —</td>
<td>Decimal Adjust WREG</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0111</td>
<td>C</td>
</tr>
<tr>
<td>GOTO n</td>
<td>Go to address 1st word</td>
<td>2</td>
<td>1110 1111</td>
<td>kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td></td>
<td>2nd word</td>
<td></td>
<td>1111 kkkk</td>
<td>kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>NOP</td>
<td>No Operation</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0000</td>
<td>None</td>
</tr>
<tr>
<td>NOP —</td>
<td>No Operation</td>
<td>1</td>
<td>1111 xxxxx</td>
<td>xxxxx xxxxx</td>
<td>None</td>
</tr>
<tr>
<td>POP —</td>
<td>Pop top of return stack (TOS)</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0110</td>
<td>None</td>
</tr>
<tr>
<td>PUSH —</td>
<td>Push top of return stack (TOS)</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0101</td>
<td>None</td>
</tr>
<tr>
<td>RCALL n</td>
<td>Relative Call</td>
<td>2</td>
<td>1101 111n</td>
<td>nnnn nnnn</td>
<td>None</td>
</tr>
<tr>
<td>RESET</td>
<td>Software device Reset</td>
<td>1</td>
<td>0000 0000</td>
<td>1111 1111</td>
<td>All</td>
</tr>
<tr>
<td>RETFIE s</td>
<td>Return from interrupt enable</td>
<td>2</td>
<td>0000 0000</td>
<td>0001 0001</td>
<td>GIE/GIEH, PEIE/GIEL</td>
</tr>
<tr>
<td>RETLW k</td>
<td>Return with literal in WREG</td>
<td>2</td>
<td>0000 1100</td>
<td>kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>RETURN s</td>
<td>Return from Subroutine</td>
<td>2</td>
<td>0000 0000</td>
<td>0000 001s</td>
<td>None</td>
</tr>
<tr>
<td>SLEEP —</td>
<td>Go to Standby mode</td>
<td>1</td>
<td>0000 0000</td>
<td>0000 0011</td>
<td>TO, PD</td>
</tr>
</tbody>
</table>

**Note 1:** When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

**2:** If this instruction is executed on the TMR0 register (and where applicable, 'd' = 1), the prescaler will be cleared if assigned.

**3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**4:** Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

**5:** If the table write starts the write cycle to external memory, the write will continue until terminated.
### TABLE 25-2: PIC18FXXXX INSTRUCTION SET (CONTINUED)

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>16-Bit Instruction Word</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>MSb</td>
<td></td>
<td></td>
</tr>
<tr>
<td>LITERAL OPERATIONS</td>
<td></td>
<td></td>
<td>LSb</td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADDLW k</td>
<td>Add literal and WREG</td>
<td>1</td>
<td>0000 1111 kkkk kkkk</td>
<td>C, DC, Z, OV, N</td>
<td></td>
</tr>
<tr>
<td>ANDLW k</td>
<td>AND literal with WREG</td>
<td>1</td>
<td>0000 1011 kkkk kkkk</td>
<td>Z, N</td>
<td></td>
</tr>
<tr>
<td>IORLW k</td>
<td>Inclusive OR literal with WREG</td>
<td>1</td>
<td>0000 1001 kkkk kkkk</td>
<td>Z, N</td>
<td></td>
</tr>
<tr>
<td>LFSR f, k</td>
<td>Move literal (12-bit) 2nd word</td>
<td>2</td>
<td>1110 1110 00ff kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td></td>
<td>to FSR(f)</td>
<td>1st word</td>
<td>1111 0000 kkkk kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>MOVLB k</td>
<td>Move literal to BSR&lt;3:0&gt;</td>
<td>1</td>
<td>0000 0001 0000 kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>MOVLW k</td>
<td>Move literal to WREG</td>
<td>1</td>
<td>0000 1110 kkkk kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>MULLW k</td>
<td>Multiply literal with WREG</td>
<td>1</td>
<td>0000 1101 kkkk kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>RETLW k</td>
<td>Return with literal in WREG</td>
<td>2</td>
<td>0000 1100 kkkk kkkk</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>SUBLW k</td>
<td>Subtract WREG from literal</td>
<td>1</td>
<td>0000 1000 kkkk kkkk</td>
<td>C, DC, Z, OV, N</td>
<td></td>
</tr>
<tr>
<td>XORLW k</td>
<td>Exclusive OR literal with WREG</td>
<td>1</td>
<td>0000 1010 kkkk kkkk</td>
<td>Z, N</td>
<td></td>
</tr>
</tbody>
</table>

#### DATA MEMORY ↔ PROGRAM MEMORY OPERATIONS

<table>
<thead>
<tr>
<th>Mnemonic</th>
<th>Description</th>
<th>Cycles</th>
<th>16-Bit Instruction Word</th>
<th>Status Affected</th>
<th>Notes</th>
</tr>
</thead>
<tbody>
<tr>
<td>TBLRD*</td>
<td>Table Read</td>
<td>2</td>
<td>0000 0000 0000 1000</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>TBLRD**</td>
<td>Table Read with post-increment</td>
<td>2</td>
<td>0000 0000 0000 1001</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>TBLRD*-</td>
<td>Table Read with post-decrement</td>
<td>2</td>
<td>0000 0000 0000 1010</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>TBLRD**-</td>
<td>Table Read with pre-decrement</td>
<td>2</td>
<td>0000 0000 0000 1101</td>
<td>None</td>
<td></td>
</tr>
<tr>
<td>TBLWT*</td>
<td>Table Write</td>
<td>2</td>
<td>0000 0000 0000 1110</td>
<td>None</td>
<td>5</td>
</tr>
<tr>
<td>TBLWT**</td>
<td>Table Write with post-increment</td>
<td>2</td>
<td>0000 0000 0000 1101</td>
<td>None</td>
<td>5</td>
</tr>
<tr>
<td>TBLWT*-</td>
<td>Table Write with post-decrement</td>
<td>2</td>
<td>0000 0000 0000 1110</td>
<td>None</td>
<td>5</td>
</tr>
<tr>
<td>TBLWT**-</td>
<td>Table Write with pre-increment</td>
<td>2</td>
<td>0000 0000 0000 1111</td>
<td>None</td>
<td>5</td>
</tr>
</tbody>
</table>

**Note 1:** When a Port register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is ‘1’ for a pin configured as input and is driven low by an external device, the data will be written back with a ‘0’.

2: If this instruction is executed on the TMR0 register (and where applicable, ‘d’ = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

5: If the table write starts the write cycle to internal memory, the write will continue until terminated.
### 25.1.1 STANDARD INSTRUCTION SET

#### ADDLW

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>ADDLW k</th>
</tr>
</thead>
<tbody>
<tr>
<td>Operands:</td>
<td>0 ≤ k ≤ 255</td>
</tr>
<tr>
<td>Operation:</td>
<td>(W) + k → W</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>N, OV, C, DC, Z</td>
</tr>
<tr>
<td>Encoding:</td>
<td>0000 1111 kkkk kkkk</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of W are added to the 8-bit literal 'k' and the result is placed in W.</td>
</tr>
<tr>
<td>Words:</td>
<td>1</td>
</tr>
<tr>
<td>Cycles:</td>
<td>1</td>
</tr>
</tbody>
</table>

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write to W</td>
</tr>
</tbody>
</table>

**Example:**

Add L W 15h

Before Instruction

W = 10h

After Instruction

W = 25h

#### ADDWF

<table>
<thead>
<tr>
<th>Syntax:</th>
<th>ADDWF f{,.d,.a}</th>
</tr>
</thead>
</table>
| Operands:     | 0 ≤ f ≤ 255  
d e [0,1]  
a e [0,1] |
| Operation:    | (W) + f → dest |
| Status Affected: | N, OV, C, DC, Z |
| Encoding:     | 0010 0010 ffff ffff |
| Description:  | Add W to register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default). |
| If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). |
| If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \( f < 95 \) (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details. |
| Words:        | 1 |
| Cycles:       | 1 |

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**

Add W to f

Before Instruction

W = 17h  
REG = 0C2h

After Instruction

W = 0D9h  
REG = 0C2h

**Note:** All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: `{label} instruction argument(s):`
ADDWFC

ADD W and Carry bit to f

Syntax:
ADDWFC f{,d{,a}}

Operands:
0 ≤ f ≤ 255
d ∈ [0,1]
a ∈ [0,1]

Operation:
(W) + (f) + (C) → dest

Status Affected:
N, OV, C, DC, Z

Encoding:
0010 00da EEEE EEEE

Description:
Add W, the Carry flag and data memory location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'.

Example:
Before Instruction
REG = 02h
W = 40h
After Instruction
REG = 02h
W = 50h

ANDLW

AND Literal with W

Syntax:
ANDLW k

Operands:
0 ≤ k ≤ 255

Operation:
(W) .AND. k → W

Status Affected:
N, Z

Encoding:
0000 1011 kkkk kkkk

Description:
The contents of W are ANDed with the 8-bit literal 'k'. The result is placed in W.

Words:
1

Cycles:
1

Q Cycle Activity:

Example:
ANDLW 05Fh
Before Instruction
W = A3h
After Instruction
W = 03h
ANDWF

**Syntax:**
\[
\text{ANDWF } f\{,d\{,a\}\}\]

**Operands:**
- \(0 \leq f \leq 255\)
- \(d \in \{0,1\}\)
- \(a \in \{0,1\}\)

**Operation:**
\((W) \text{ AND } (f) \rightarrow \text{ dest}\)

**Status Affected:**
- \(N\)
- \(Z\)

**Encoding:**
\[
\begin{array}{cccc}
0001 & 01da & ffff & ffff \\
\end{array}
\]

**Description:**
The contents of \(W\) are AND'ed with register \(f\). If \(d\) is '0', the result is stored in \(W\). If \(d\) is '1', the result is stored back in register \(f\) (default).

If \(a\) is '0', the Access Bank is selected. If \(a\) is '1', the BSR is used to select the GPR bank (default).

If \(a\) is '0' and the extended instruction set is enabled, this instruction operates in indexed Literal Offset Addressing mode whenever \(f \leq 95\) (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:**
- 1

**Cycles:**
- 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**
Before Instruction
- \(W = 17h\)
- \(REG = C2h\)

After Instruction
- \(W = 02h\)
- \(REG = C2h\)

BC

**Branch if Carry**

**Syntax:**
\[
\text{BC } n\]

**Operands:**
- \(-128 \leq n \leq 127\)

**Operation:**
- If Carry bit is '1'
  \((PC) + 2 + 2n \rightarrow PC\)

**Status Affected:**
- None

**Encoding:**
\[
\begin{array}{c|c|c|c}
1110 & 0010 & nnnn & nnnn \\
\end{array}
\]

**Description:**
If the Carry bit is '1', then the program will branch.

The 2's complement number \(2n\) is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be \(PC + 2 + 2n\). This instruction is then a two-cycle instruction.

**Words:**
- 1

**Cycles:**
- 1(2)

**Q Cycle Activity:**

If Jump:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

If No Jump:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**
Before Instruction
- \(PC = \text{ address (HERE)}\)
- If Carry\(1; PC = \text{ address (HERE + 12)}\)
- If Carry\(0; PC = \text{ address (HERE + 2)}\)
BCF  Bit Clear f

Syntax:  BCF  f, b (a)
Operands:  0 ≤ f ≤ 255
          0 ≤ b ≤ 7
          a ∈ [0, 1]
Operation:  0 → fÆb
Status Affected:  None
Encoding:  1001  bbba  ffff  ffff
Description:  Bit 'b' in register 'f' is cleared.
If 'a' is '0', the Access Bank is selected.  If 'a' is '1', the BSR is used to select the
GPR bank (default).  If 'a' is '0' and the extended instruction set is enabled, this instruction operates
in Indexed Literal Offset addressing mode whenever f ≤ 95 (5Fh).  See Section 25.2.3
"Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

Words:  1
Cycles:  1
Q Cycle Activity:

Example:  BCF  FLAG_REG, 7, 0
Before Instruction
  FLAG_REG = C7h
After Instruction
  FLAG_REG = 47h

BN  Branch if Negative

Syntax:  BN  n
Operands:  -128 ≤ n ≤ 127
Operation:  if Negative bit is '1'  
            (PC) + 2 + 2n → PC
Status Affected:  None
Encoding:  1110  0110  nnnn  nnnn
Description:  If the Negative bit is '1', then the program will branch.
The 2's complement number '2n' is added to the PC.  Since the PC will have
incremented to fetch the next instruction, the new address will be
PC + 2 + 2n.  This instruction is then a
two-cycle instruction.

Words:  1
Cycles:  1
Q Cycle Activity:

Example:  HERE  BN  Jump
Before Instruction
  PC = address (HERE)
After Instruction
  If Negative = 1;
    PC = address (Jump)
  If Negative = 0;
    PC = address (HERE + 2)
### BNC Branch if Not Carry

**Syntax:**
```
BNC n
```

**Operands:**
```
-128 ≤ n ≤ 127
```

**Operation:**
```
if Carry bit is '0'

(PC) + 2 + 2n → PC
```

**Status Affected:**
```
None
```

**Encoding:**
```
1110 0011 mnnn mnnn
```

**Description:**
If the Carry bit is '0', then the program will branch.

The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

**Words:**
1

**Cycles:**
1(2)

<table>
<thead>
<tr>
<th>Q Cycle Activity:</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>If Jump:</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Decode</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Read literal 'n'</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Process Data</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Write to PC</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>If No Jump:</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Read literal 'n'</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Process Data</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Example:**

Before Instruction
```
PC = address (HERE)
```

After Instruction
```
If Carry = 0;
PC = address (Jump)
If Carry = 1;
PC = address (HERE + 2)
```

### BNN Branch if Not Negative

**Syntax:**
```
BNN n
```

**Operands:**
```
-128 ≤ n ≤ 127
```

**Operation:**
```
if Negative bit is '0'

(PC) + 2 + 2n → PC
```

**Status Affected:**
```
None
```

**Encoding:**
```
1110 0111 mnnn mnnn
```

**Description:**
If the Negative bit is '0', then the program will branch.

The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

**Words:**
1

**Cycles:**
1(2)

<table>
<thead>
<tr>
<th>Q Cycle Activity:</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>If Jump:</td>
<td></td>
<td></td>
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<td></td>
</tr>
<tr>
<td>Decode</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Read literal 'n'</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Process Data</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Write to PC</td>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>If No Jump:</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Read literal 'n'</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Process Data</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>No operation</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Example:**

Before Instruction
```
PC = address (HERE)
```

After Instruction
```
If Negative = 0;
PC = address (Jump)
If Negative = 1;
PC = address (HERE + 2)
```
**BNOV**  
Branch if Not Overflow

**Syntax:** BNOV n  
**Operands:** -128 ≤ n ≤ 127  
**Operation:** if Overflow bit is '0'  
(PC) + 2 + 2n → PC  
**Status Affected:** None  
**Encoding:**  

<table>
<thead>
<tr>
<th>1110</th>
<th>0101</th>
<th>nnnn</th>
<th>nnnn</th>
</tr>
</thead>
</table>

**Description:** If the Overflow bit is '0', then the program will branch. The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

**Words:** 1  
**Cycles:** 1(2)  
**Q Cycle Activity:**  
If Jump:  
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

If No Jump:  
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**  
Before Instruction  
PC = address (HERE)  
After Instruction  
If Overflow = 0;  
PC = address (Jump)  
If Overflow = 1;  
PC = address (HERE + 2)

**BNZ**  
Branch if Not Zero

**Syntax:** BNZ n  
**Operands:** -128 ≤ n ≤ 127  
**Operation:** if Zero bit is '0'  
(PC) + 2 + 2n → PC  
**Status Affected:** None  
**Encoding:**  

<table>
<thead>
<tr>
<th>1110</th>
<th>0001</th>
<th>nnnn</th>
<th>nnnn</th>
</tr>
</thead>
</table>

**Description:** If the Zero bit is '0', then the program will branch. The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

**Words:** 1  
**Cycles:** 1(2)  
**Q Cycle Activity:**  
If Jump:  
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

If No Jump:  
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**  
Before Instruction  
PC = address (HERE)  
After Instruction  
If Zero = 0;  
PC = address (Jump)  
If Zero = 1;  
PC = address (HERE + 2)
### BRA: Unconditional Branch

**Syntax:**
```
BRA n
```

**Operands:**
\[-1024 \leq n \leq 1023\]

**Operation:**
```
(PC) + 2 + 2n \rightarrow PC
```

**Status Affected:** None

**Encoding:**
```
1101 0nnn nnnn
```

**Description:**
Add the 2's complement number \(2n\) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be \(PC + 2 + 2n\). This instruction is a two-cycle instruction.

**Words:** 1

**Cycles:** 2

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**
Before Instruction
```
PC = address (HERE)
```

After Instruction
```
PC = address (Jump)
```

### BSF: Bit Set f

**Syntax:**
```
BSF f, b [a]
```

**Operands:**
\[0 \leq f \leq 255\]
\[0 \leq b \leq 7\]
\[a \in \{0,1\}\]

**Operation:**
```
1 \rightarrow f\langle b\rangle
```

**Status Affected:** None

**Encoding:**
```
1000 bbb a ffff ffff
```

**Description:**
Bit 'b' in register 'f' is set.
- If 'a' is '0', the Access Bank is selected.
- If 'a' is '1', the BSR is used to select the GPR bank (default).
- If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \(f \leq 95\) (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write register 'f'</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**
```
BRA
```

Before Instruction
```
FLAG_REG = 0Ah
```

After Instruction
```
FLAG_REG = 8Ah
```
# BTFSC

**Bit Test File, Skip if Clear**

**Syntax:**

```
BTFSC f, b (a)
```

**Operands:**

- \(0 \leq f \leq 255\)
- \(0 \leq b \leq 7\)
- \(a \in [0, 1]\)

**Operation:**

- `skip if \((f < b) = 0\)`

**Status Affected:**

None

**Encoding:**

| 1011 | bbba | ffff | ffff |

**Description:**

If bit 'b' in register 'f' is '0', the next instruction is skipped. If bit 'b' is '0', the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction.

- If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).
- If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \(f \leq 95\) (6Fh).

See Section 26.2.2 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:**

1

**Cycles:**

1(2)

**Note:**

3 cycles if skip and followed by a 2-word instruction.

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

```
HERE BTFSC FLAG, 1, 0
FALSE :
TRUE :
```

Before Instruction

| PC   | address (HERE) |
After Instruction

- `IF FLAG<1> = 0;`
- `PC = address (TRUE)`
- `IF FLAG<1> = 1;`
- `PC = address (FALSE)`

# BTFSS

**Bit Test File, Skip if Set**

**Syntax:**

```
BTFSS f, b (a)
```

**Operands:**

- \(0 \leq f \leq 255\)
- \(0 \leq b < 7\)
- \(a \in [0, 1]\)

**Operation:**

- `skip if \((f < b) = 1\)`

**Status Affected:**

None

**Encoding:**

| 1010 | bbba | ffff | ffff |

**Description:**

If bit 'b' in register 'f' is '1', the next instruction is skipped. If bit 'b' is '1', the next instruction fetched during the current instruction execution is discarded and a NOP is executed instead, making this a two-cycle instruction.

- If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).
- If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \(f \leq 95\) (6Fh).

See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:**

1

**Cycles:**

1(2)

**Note:**

3 cycles if skip and followed by a 2-word instruction.

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

```
HERE BTFSS FLAG, 1, 0
FALSE :
TRUE :
```

Before Instruction

| PC   | address (HERE) |
After Instruction

- `IF FLAG<1> = 0;`
- `PC = address (FALSE)`
- `IF FLAG<1> = 1;`
- `PC = address (TRUE)`

---

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BTG  Bit Toggle f

Syntax:  BTG f, b [,a]  
Operands:  0 ≤ f ≤ 255  
         0 ≤ b < 7  
         a ∈ [0,1]  
Operation:  \((f < b) \rightarrow f < b)\)  
Status Affected:  None  
Encoding:  \[\begin{array}{cccc}
0111 & \text{bbba} & \text{efff} & \text{efff} \\
\end{array}\]  
Description:  Bit 'b' in data memory location 'f' is inverted.  
If 'a' is '0', the Access Bank is selected.  
If 'a' is '1', the BSR is used to select the GPR bank (default).  
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

Words:  1  
Cycles:  1  
Q Cycle Activity:  

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to register 'f'</td>
</tr>
</tbody>
</table>

Example:  BTG PORTC, 4, 0  
Before Instruction:  
PORTC = 0111 0101 [75h]  
After Instruction:  
PORTC = 0110 0101 [65h]

BOV  Branch if Overflow

Syntax:  BOV n  
Operands:  -128 ≤ n ≤ 127  
Operation:  if Overflow bit is '1', \((PC) + 2 + 2n \rightarrow PC\)  
Status Affected:  None  
Encoding:  \[\begin{array}{cccc}
1110 & 0100 & \text{nnnn} & \text{nnnn} \\
\end{array}\]  
Description:  If the Overflow bit is '1', then the program will branch.  
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be \(PC + 2 + 2n\). This instruction is then a two-cycle instruction.

Words:  1  
Cycles:  1(2)  
Q Cycle Activity:  
If Jump:  

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

If No Jump:  

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

Example:  HERE BOV Jump  
Before Instruction:  
PC = address (HERE)  
After Instruction:  
If Overflow = 1;  
PC = address (Jump)  
If Overflow = 0;  
PC = address (HERE + 2)
### BZ Branch if Zero

**Syntax:**
BZ n

**Operands:**
-128 ≤ n ≤ 127

**Operation:**
If Zero bit is '1',
(PC) + 2 + 2n → PC

**Status Affected:**
None

**Encoding:**
```
1110 0000 nnnn nnn
```

**Description:**
If the Zero bit is '1', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

**Words:**
1

**Cycles:**
1(2)

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'n'</td>
<td>Process Data</td>
<td>Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**
Before Instruction
PC = address (HERE)

After Instruction
If Zero = 1;
PC = address (Jump)
If Zero = 0;
PC = address (HERE + 2)

### CALL Subroutine Call

**Syntax:**
CALL k {,s}

**Operands:**
0 ≤ k ≤ 1048575
s ∈ [0,1]

**Operation:**
(PC) + 4 → TOS, k → PC<20:1>,
if s = 1
(W) → WS,
(Status) → STATUSS,
(BSR) → BSRS

**Status Affected:**
None

**Encoding:**
```
1st word (k<7:0>)
1110 110s kkkk kkkk

2nd word (k<19:8>)
1112 kkkk kkkk kkkk
```

**Description:**
Subroutine call of entire 2-Mbyte memory range. First, return address (PC + 4) is pushed onto the return stack. If 's' = 1, the W, Status and BSR registers are also pushed into their respective shadow registers, WS, STATUSS and BSRS. If 's' = 0, no update occurs (default). Then, the 20-bit value 'k' is loaded into PC<20:1>. CALL is a two-cycle instruction.

**Words:**
2

**Cycles:**
2

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'&lt;7:0&gt;,</td>
<td>Push PC to stack</td>
<td>Read literal 'k'&lt;19:8&gt;, Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**
Before Instruction
PC = address (HERE)

After Instruction
PC = address (HERE)
TOS = address (HERE + 4)
WS = W
BSRS = BSR
STATUSS = Status
CLRF

Clear f

Syntax:  CLRF f{,a)

Operands:  0 ≤ f ≤ 255
          a ∈ [0,1]

Operation:  000h → f
          1 → Z

Status Affected:  Z

Encoding:  0110 101a fffe fffe

Description:  Clears the contents of the specified register.
              If 'a' is '0', the Access Bank is selected.
              If 'a' is '1', the BSR is used to select the
              GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates
in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See
Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed
Literal Offset Mode" for details.

Words:  1
Cycles:  1

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register f</td>
<td>Process Data</td>
<td>Write register f</td>
</tr>
</tbody>
</table>

Example:

Before Instruction
FLAG_REG = 5Ah
After Instruction
FLAG_REG = 00h

CLRWDWT
Clear Watchdog Timer

Syntax:  CLRWDWT

Operands:  None

Operation:  000h → WDT,
            000h → WDT postscaler,
            1 → TO,
            1 → PD

Status Affected:  TO, PD

Encoding:  0000 0000 0000 0100

Description:  CLRWDWT instruction resets the
              Watchdog Timer. It also resets the
              postscaler of the WDT. Status bits TO
              and PD are set.

Words:  1
Cycles:  1

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>No operation</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

Example:  CLRWDWT

Before Instruction
WDT Counter = ?
After Instruction
WDT Counter = 00h
WDT Postscaler = 0
TO = 1
PD = 1
COMF  Complement f

Syntax:  COMF f{.d{.a}}
Operands:  0 ≤ f ≤ 255
           d ∈ [0,1]
           a ∈ [0,1]
Operation:  (f) → dest
Status Affected:  N, Z
Encodings:  0001 111a  ffff  ffff
Description:  The contents of register 'f' are complemented. If 'd' is '1', the result is stored in W. If 'd' is '0', the result is stored back in register 'f' (default).
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever |f| ≤ 95 (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.
Words:  1
Cycles:  1
Q Cycle Activity:  

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

Example:  COMP  REG, 0, 0
Before Instruction
REG = 13h
After Instruction
REG = 13h
W = ECh

CPFSEQ  Compare f with W, Skip if f = W

Syntax:  CPFSEQ f{.a}
Operands:  0 ≤ f ≤ 255
           a ∈ [0,1]
Operation:  (f) − (W),
skip if (f) = (W)
(unsigned comparison)
Status Affected:  None
Encoding:  0110 001a  ffff  ffff
Description:  Compares the contents of data memory location 'f' to the contents of W by performing an unsigned subtraction. If 'f' = W, then the fetched instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever |f| ≤ 95 (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.
Words:  1
Cycles:  1(2)
Note:  3 cycles if skip and followed by a 2-word instruction.
Q Cycle Activity:  

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

If skip:
Q1 Q2 Q3 Q4
No operation No operation No operation No operation

If skip and followed by 2-word instruction:
Q1 Q2 Q3 Q4
No operation No operation No operation No operation

Example:  HERE  CPFSEQ REG, 0
NEQUAL :
EQUAL :
Before Instruction
PC Address = HERE
W = ?
REG = ?
After Instruction
If REG ≠ W:
PC = Address (EQUAL)
If REG = W:
PC = Address (NEQUAL)
**PIC18F2480/2580/4480/4580**

<table>
<thead>
<tr>
<th>CPFSGT</th>
<th>Compare f with W, Skip if f &gt; W</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>CPFSGT f,a</td>
</tr>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 255, a ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>f - (W), skip if (f) &gt; (W)</td>
</tr>
<tr>
<td>(unsigned comparison)</td>
<td></td>
</tr>
</tbody>
</table>

**Status Affected:** None

**Encoding:**

<table>
<thead>
<tr>
<th>Code</th>
</tr>
</thead>
<tbody>
<tr>
<td>f110a</td>
</tr>
</tbody>
</table>

**Description:**

Comparing the contents of data memory location 'f' to the contents of the W by performing an unsigned subtraction. If the contents of 'f' are greater than the contents of WREG, then the fetched instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:** 1

**Cycles:** 1(2)

**Note:** 3 cycles if skip and followed by a 2-word instruction.

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
<tr>
<td>If skip:</td>
<td>Q1</td>
<td>Q2</td>
<td>Q3</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
<tr>
<td>If skip and followed by 2-word instruction:</td>
<td>Q1</td>
<td>Q2</td>
<td>Q3</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

<table>
<thead>
<tr>
<th>Before Instruction</th>
<th>After Instruction</th>
</tr>
</thead>
<tbody>
<tr>
<td>PC = Address (HERE)</td>
<td>PC = ?</td>
</tr>
<tr>
<td>W = ?</td>
<td>If REG &lt; W, PC = Address (LESS)</td>
</tr>
<tr>
<td>If REG ≥ W, PC = Address (GREATER)</td>
<td>If REG ≥ W, PC = Address (NLLESS)</td>
</tr>
</tbody>
</table>

---

<table>
<thead>
<tr>
<th>CPFSLT</th>
<th>Compare f with W, Skip if f &lt; W</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>CPFSLT f,a</td>
</tr>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 255, a ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>f - (W), skip if (f) &lt; (W)</td>
</tr>
<tr>
<td>(unsigned comparison)</td>
<td></td>
</tr>
</tbody>
</table>

**Status Affected:** None

**Encoding:**

<table>
<thead>
<tr>
<th>Code</th>
</tr>
</thead>
<tbody>
<tr>
<td>f110a</td>
</tr>
</tbody>
</table>

**Description:**

Comparing the contents of data memory location 'f' to the contents of the W by performing an unsigned subtraction. If the contents of 'f' are less than the contents of W, then the fetched instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

**Words:** 1

**Cycles:** 1(2)

**Note:** 3 cycles if skip and followed by a 2-word instruction.

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
<tr>
<td>If skip:</td>
<td>Q1</td>
<td>Q2</td>
<td>Q3</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
<tr>
<td>If skip and followed by 2-word instruction:</td>
<td>Q1</td>
<td>Q2</td>
<td>Q3</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

<table>
<thead>
<tr>
<th>Before Instruction</th>
<th>After Instruction</th>
</tr>
</thead>
<tbody>
<tr>
<td>PC = Address (HERE)</td>
<td>PC = ?</td>
</tr>
<tr>
<td>W = ?</td>
<td>If REG &lt; W, PC = Address (LESS)</td>
</tr>
<tr>
<td>If REG ≥ W, PC = Address (GREATER)</td>
<td>If REG ≥ W, PC = Address (NLLESS)</td>
</tr>
</tbody>
</table>

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APPENDIX H: DATA SHEETS
**DAW**

**Decimal Adjust W Register**

**Syntax:**

DAW

**Operands:**

None

**Operation:**

If [W<3:0> >9] or [DC = 1] then

(W<3:0>) + 6 → W<3:0>;

else

(W<3:0>) → W<3:0>;

If [W<7:4> >9] or [C = 1] then

(W<7:4>) + 6 → W<7:4>;

else

(W<7:4>) → W<7:4>;

**Status Affected:**

C

**Encoding:**

| 0000 | 0000 | 0000 | 0111 |

**Description:**

DAW adjusts the eight-bit value in W, resulting from the earlier addition of two variables (each in packed BCD format) and produces a correct packed BCD result.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register W</td>
<td>Process Data</td>
<td>Write W</td>
</tr>
</tbody>
</table>

**Example 1:**

Before Instruction

W = A5h
C = 0
DC = 0

After Instruction

W = 05h
C = 1
DC = 0

**Example 2:**

Before Instruction

W = CEh
C = 0
DC = 0

After Instruction

W = 34h
C = 1
DC = 0

**DECF**

**Decrement f**

**Syntax:**

DECF f (d1 [a])

**Operands:**

0 ≤ f ≤ 255
d1 ∈ [0,1]
a ∈ [0,1]

**Operation:**

(f) – 1 → dest

**Status Affected:**

C, DC, N, OV, Z

**Encoding:**

| 0000 | 010a | fffe | fffe |

**Description:**

Decrement register f. If ‘d’ is ‘0’, the result is stored in W. If ‘d’ is ‘1’, the result is stored back in register ‘f’ (default).

If ‘a’ is ‘0’, the Access Bank is selected. If ‘a’ is ‘1’, the BSR is used to select the GPR bank (default).

If ‘a’ is ‘0’ and the extended instruction set is enabled, this instruction operates in indexed literal offset addressing mode whenever f ≤ 95 (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register f’</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**

Before Instruction

CNT = 01h
Z = 0

After Instruction

CNT = 00h
Z = 1
<table>
<thead>
<tr>
<th>DECFSZ</th>
<th>Decrement f, Skip if 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>DECFSZ f,d (a)</td>
</tr>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 255</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td></td>
<td>a ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(f) → dest,</td>
</tr>
<tr>
<td></td>
<td>skip if result = 0</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>None</td>
</tr>
<tr>
<td>Encoding:</td>
<td>0100 1ida fffe fffe</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If the result is '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>DCFSNZ</th>
<th>Decrement f, Skip if not 0</th>
</tr>
</thead>
<tbody>
<tr>
<td>Syntax:</td>
<td>DCFSNZ f,d (a)</td>
</tr>
<tr>
<td>Operands:</td>
<td>0 ≤ f ≤ 255</td>
</tr>
<tr>
<td></td>
<td>d ∈ [0,1]</td>
</tr>
<tr>
<td></td>
<td>a ∈ [0,1]</td>
</tr>
<tr>
<td>Operation:</td>
<td>(f) → dest,</td>
</tr>
<tr>
<td></td>
<td>skip if result = 0</td>
</tr>
<tr>
<td>Status Affected:</td>
<td>None</td>
</tr>
<tr>
<td>Encoding:</td>
<td>0100 1ida fffe fffe</td>
</tr>
<tr>
<td>Description:</td>
<td>The contents of register 'f' are decremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.</td>
</tr>
</tbody>
</table>

Example: HERE DECFSZ CNT, 1, 1 CONTINUE GOTO LOOP

Example: HERE DCFSNZ TEMP, 1, 0 ZERO: ; NZERO: ;

Example: Before instruction PC = Address (HERE)

Before instruction PC = Address (HERE)

Before instruction TEMP = ?

After instruction TEMP = ?

After instruction TEMP = ?

After instruction TEMP = ?

PC = Address (ZERO)

PC = Address (NZERO)
GOTO | Unconditional Branch
--- | ---
Syntax: | GOTO k
Operands: | 0 ≤ k ≤ 1048575
Operation: | k → PC<20:1>
Status Affected: | None

Encoding:
1st word (k<7:0>)

<table>
<thead>
<tr>
<th>1110</th>
<th>k&lt;7:0&gt;</th>
</tr>
</thead>
</table>

2nd word (k<19:8>)

| 1111 | k<19:8> |

Description: GOTO allows an unconditional branch anywhere within entire 2-Mbyte memory range. The 20-bit value 'k' is loaded into PC<20:1>. GOTO is always a two-cycle instruction.

Words: 2
Cycles: 2

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'&lt;7:0&gt;</td>
<td>No operation</td>
<td>Read literal 'k'&lt;19:8&gt;, Write to PC</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td></td>
</tr>
</tbody>
</table>

Example: GOTO THERE

After Instruction
PC = Address (THERE)

INCF | Increment f
--- | ---
Syntax: | INCF f{,d{,a}}
Operands: | 0 ≤ f ≤ 255
d ∈ [0,1]
a ∈ [0,1]
Operation: | (f) + 1 → dest
Status Affected: | C, DC, N, OV, Z

Encoding:

| 0010 | 10da | ffff | ffff |

Description: The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

Words: 1
Cycles: 1

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

Example: INCF CNT, 1, 0

Before Instruction
CNT = FFh
Z = 0
C = ?
DC = ?

After Instruction
CNT = 00h
Z = 1
C = 1
DC = 1
### INCFSZ

**Increment f, Skip if 0**

**Syntax:**

\[
\text{INCFSZ } f \langle d (a) \rangle
\]

**Operands:**

\[
0 \leq f \leq 255 \\
d \in [0, 1] \\
a \in [0, 1]
\]

**Operation:**

\[
(f) + 1 \rightarrow \text{dest, skip if result } = 0
\]

**Status Affected:** None

**Encoding:**

| 0011 | lida | ffff | ffff |

**Description:**

The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).

If the result is '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.

If 'f' is '0', the Access Bank is selected. If 'f' is '1', the BSR is used to select the GPR bank (default).

If 'f' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f s 95 (5Fh). See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**

1

**Cycles:**

(1(2)

**Note:** 3 cycles if skip and followed by a 2-word instruction.

#### Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Example:**

**Before Instruction**

- PC = Address (HERE)

**After Instruction**

- CNT = CNT + 1
- If CNT = 0; PC = Address (ZERO)
- If CNT = 0; PC = Address (NZERO)

### INFSNZ

**Increment f, Skip if not 0**

**Syntax:**

\[
\text{INFSNZ } f \langle d (a) \rangle
\]

**Operands:**

\[
0 \leq f \leq 255 \\
d \in [0, 1] \\
a \in [0, 1]
\]

**Operation:**

\[
(f) + 1 \rightarrow \text{dest, skip if result } \neq 0
\]

**Status Affected:** None

**Encoding:**

| 0100 | lida | ffff | ffff |

**Description:**

The contents of register 'f' are incremented. If 'd' is '0', the result is placed back in register 'f' (default).

If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction.

If 'f' is '0', the Access Bank is selected. If 'f' is '1', the BSR is used to select the GPR bank (default).

If 'f' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f s 95 (5Fh). See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**

1

**Cycles:**

(1(2)

**Note:** 3 cycles if skip and followed by a 2-word instruction.

#### Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Example:**

**Before Instruction**

- PC = Address (HERE)

**After Instruction**

- REG = REG + 1
- If REG = 0; PC = Address (NZERO)
- If REG = 0; PC = Address (ZERO)
IORLW  Inclusive OR Literal with W

Syntax:    IORLW k
Operands:  0 ≤ k ≤ 255
Operation: (W).OR. k → W
Status Affected: N, Z
Encoding:
Description: The contents of W are ORed with the eight-bit literal 'k'. The result is placed in W.
Words: 1
Cycles: 1
Q Cycle Activity:

Example:    IORLW 35h
Before Instruction
W = 9Ah
After Instruction
W = BFh

IORWF  Inclusive OR W with f

Syntax:    IORWF f{,d{,a}}
Operands:  0 ≤ f ≤ 255
d ∈ [0,1]
a ∈ [0,1]
Operation: (W).OR. (f) → dest
Status Affected: N, Z
Encoding:
Description: Inclusive OR W with register 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f < 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.
Words: 1
Cycles: 1
Q Cycle Activity:

Example:    IORWF RESULT, 0, 1
Before Instruction
RESULT = 13h
W = 91h
After Instruction
RESULT = 13h
W = 93h
LFSR

Syntax: \( \text{LFSR } f, k \)

Operands: \( 0 < f < 2 \)
\( 0 < k < 4095 \)

Operation: \( k \rightarrow \text{FSR}_f \)

Status Affected: None

Encoding:

| 1111 | 1110 | 00ff | k_f
|------|------|------|------
| 1111 | 0000 | k_f  | k_f

Description: The 12-bit literal 'k' is loaded into the file select register pointed to by 'f'.

Words: 2
Cycles: 2

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Decode</th>
<th>Read literal 'k' MSB</th>
<th>Process Data</th>
<th>Write literal 'k' MSB to FSRf</th>
</tr>
</thead>
<tbody>
<tr>
<td>Example: LFSR 2, 3Ah</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>After Instruction</td>
<td>FSR2H = 03h</td>
<td>FSR2L = ABh</td>
<td></td>
</tr>
</tbody>
</table>

MOVF

Syntax: \( \text{MOVF } f(d[a]) \)

Operands: \( 0 < f < 255 \)
\( d \in [0,1] \)
\( a \in [0,1] \)

Operation: \( f \rightarrow \text{dest} \)

Status Affected: N, Z

Encoding:

| 0101 | 00da | ffff | ffff |

Description: The contents of register 'f' are moved to a destination dependent upon the status of 'd'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). Location 'f' can be anywhere in the 256-byte bank.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \( f \leq 95 \) (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

Words: 1
Cycles: 1

Q Cycle Activity:

<table>
<thead>
<tr>
<th>Decode</th>
<th>Read register 'f'</th>
<th>Process Data</th>
<th>Write W</th>
</tr>
</thead>
<tbody>
<tr>
<td>Example: MOVF REG, 0, 0</td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Before Instruction</td>
<td>REG = 22h</td>
<td>W = FFFh</td>
<td></td>
</tr>
<tr>
<td>After Instruction</td>
<td>REG = 22h</td>
<td>W = 22h</td>
<td></td>
</tr>
</tbody>
</table>
**MOVFF**

**Move f to f**

- **Syntax:** MOVFF f,g
- **Operands:** 0 ≤ f ≤ 4095, 0 ≤ g ≤ 4095
- **Operation:** (f) → g
- **Status Affected:** None
- **Encoding:**
  - 1st word (source):
    - 1100: f
    - 1101: f
  - 2nd word (destin.):
    - The contents of source register f are moved to destination register g.
- **Description:**
  - The contents of source register f are moved to destination register g.
  - Location of source f can be anywhere in the 4096-byte data space (000h to FFFFh) and location of destination g can be anywhere from 000h to FFFFh.
  - Either source or destination can be W (a useful special situation).
  - MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port).
  - The MOVFF instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.

- **Words:** 2
- **Cycles:** 2 (3)
- **Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Decode</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>No operation</td>
<td>Register 'f' (src)</td>
<td>Process Data</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

- **Before Instruction**
  - REG1 = 33h
  - REG2 = 11h
- **After Instruction**
  - REG1 = 11h
  - REG2 = 33h

---

**MOVLB**

**Move Literal to Low Nibble in BSR**

- **Syntax:** MOVLW k
- **Operands:** 0 ≤ k ≤ 255
- **Operation:** k → BSR
- **Status Affected:** None
- **Encoding:**
  - 0000 0001 kkkk kkkk
- **Description:** The eight-bit literal k is loaded into the Bank Select Register (BSR). The value of BSR<7:4> always remains '0', regardless of the value of k7,k4.
- **Words:** 1
- **Cycles:** 1
- **Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write literal 'k' to BSR</td>
</tr>
</tbody>
</table>

**Example:**

- **Before Instruction**
  - BSR Register = 02h
- **After Instruction**
  - BSR Register = 05h
### MOVWL

**Move Literal to W**

**Syntax:**
```plaintext
MOVLW k
```

**Operands:**

- `0 ≤ k ≤ 255`

**Operation:**

- `k → W`

**Status Affected:**

- None

**Encoding:**

```
0000 1110 kkkk
```

**Description:**

The eight-bit literal `k` is loaded into `W`.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Decode</th>
<th>Read literal 'k'</th>
<th>Process Data</th>
<th>Write to W</th>
</tr>
</thead>
</table>

**Example:**

- **Before Instruction**
  
  - `W` = `5Ah`

- **After Instruction**
  
  - `W` = `5Ah`

### MOVWF

**Move W to f**

**Syntax:**

```plaintext
MOVWF f{,a}
```

**Operands:**

- `0 ≤ f ≤ 255`

- `a ∈ [0,1]`

**Operation:**

- `(W) → f`

**Status Affected:**

- None

**Encoding:**

```
0110 1110 ffff ffff
```

**Description:**

Move data from `W` to register `f`. Location `f` can be anywhere in the 256-byte bank.

- If `a` is '0', the Access Bank is selected.
- If `a` is '1', the BSR is used to select the GPR bank (default).

If `a` is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever `f < 95 (5Fh)`. See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write register 'f'</td>
</tr>
</tbody>
</table>

**Example:**

- **Before Instruction**
  
  - `W` = `4Fh`
  - `REG` = `FFh`

- **After Instruction**
  
  - `W` = `4Fh`
  - `REG` = `4Fh`
### MULLW

**Multiply Literal with W**

| Syntax: | MULLW k |
| Operands: | 0 ≤ k ≤ 255 |
| Operation: | (W) × k → PRODH:PRODL |
| Status Affected: | None |
| Encoding: | 0000 1101 kkkk kkkk |
| Description: | An unsigned multiplication is carried out between the contents of W and the 8-bit literal 'k'. The 16-bit result is placed in the PRODH:PRODL register pair. PRODH contains the high byte. None of the status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected. |

**Words:** 1  
**Cycles:** 1

#### Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write registers PRODH:PRODL</td>
</tr>
</tbody>
</table>

#### Example:

<table>
<thead>
<tr>
<th>W</th>
<th>PRODH</th>
<th>PRODL</th>
</tr>
</thead>
<tbody>
<tr>
<td>E2h</td>
<td>?</td>
<td>?</td>
</tr>
</tbody>
</table>

Before Instruction:

After Instruction:

<table>
<thead>
<tr>
<th>W</th>
<th>PRODH</th>
<th>PRODL</th>
</tr>
</thead>
<tbody>
<tr>
<td>E2h</td>
<td>ADh</td>
<td>00h</td>
</tr>
</tbody>
</table>

### MULWF

**Multiply W with f**

| Syntax: | MULWF f(a) |
| Operands: | 0 ≤ f ≤ 255 |
| a ∈ [0,1] |
| Operation: | (W) × f → PRODH:PRODL |
| Status Affected: | None |
| Encoding: | 0000 001a ffff ffff |
| Description: | An unsigned multiplication is carried out between the contents of W and the register file location 'f'. The 16-bit result is stored in the PRODH:PRODL register pair. PRODH contains the high byte. Both W and 'f' are unchanged. None of the status flags are affected. Note that neither overflow nor carry is possible in this operation. A zero result is possible but not detected. |

**Words:** 1  
**Cycles:** 1

#### Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write registers PRODH:PRODL</td>
</tr>
</tbody>
</table>

#### Example:

<table>
<thead>
<tr>
<th>W</th>
<th>REG</th>
<th>PRODH</th>
<th>PRODL</th>
</tr>
</thead>
<tbody>
<tr>
<td>C4h</td>
<td>B5h</td>
<td>?</td>
<td>?</td>
</tr>
</tbody>
</table>

Before Instruction:

After Instruction:

<table>
<thead>
<tr>
<th>W</th>
<th>REG</th>
<th>PRODH</th>
<th>PRODL</th>
</tr>
</thead>
<tbody>
<tr>
<td>C4h</td>
<td>B5h</td>
<td>8Ah</td>
<td>9Ah</td>
</tr>
</tbody>
</table>
### NEG F

**Syntax:** NEG F {a}

**Operands:** 0 ≤ f ≤ 255

**Operation:** (T) + 1 → f

**Status Affected:** N, OV, C, DC, Z

**Encoding:**

| Code | 0110 | 110a | fffe | fffe |

**Description:** Location T is negated using two's complement. The result is placed in the data memory location T.

- If 'a' is 'o', the Access Bank is selected.
- If 'a' is '1', the BSR is used to select the GPR bank (default).
- If 'a' is 'o' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 65 (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register f</td>
<td>Process Data</td>
<td>Write register T</td>
</tr>
</tbody>
</table>

**Example:**

Before Instruction

REG = 0011 1010 [3Ah]

After Instruction

REG = 1100 0110 [C6h]

---

### NOP

**Syntax:** NOP

**Operands:** None

**Operation:** No operation

**Status Affected:** None

**Encoding:**

<table>
<thead>
<tr>
<th>Code</th>
<th>0000</th>
<th>0000</th>
<th>0000</th>
<th>0000</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>1111</td>
<td>xxxx</td>
<td>xxxx</td>
<td>xxxx</td>
</tr>
</tbody>
</table>

**Description:** No operation.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

**Example:**

None.
POP
Syntax: POP
Operands: None
Operation: (TOS) → bit bucket
Status Affected: None
Encoding: 0000 0000 0000 0110
Description: The TOS value is pulled off the return stack and is discarded. The TOS value then becomes the previous value that was pushed onto the return stack. This instruction is provided to enable the user to properly manage the return stack to incorporate a software stack.

Words: 1
Cycles: 1
Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>No operation</td>
<td>POP TOS value</td>
<td>No operation</td>
</tr>
</tbody>
</table>

Example:

Before Instruction
TOS = 0031A2h
Stack (1 level down) = 014332h

After Instruction
TOS = 014332h
PC = NEW

PUSH
Syntax: PUSH
Operands: None
Operation: (PC + 2) → TOS
Status Affected: None
Encoding: 0000 0000 0000 0101
Description: The PC + 2 is pushed onto the top of the return stack. The previous TOS value is pushed down on the stack. This instruction allows implementing a software stack by modifying TOS and then pushing it onto the return stack.

Words: 1
Cycles: 1
Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>PUSH PC + 2 onto return stack</td>
<td>No operation</td>
<td>No operation</td>
</tr>
</tbody>
</table>

Example:
PUSH

Before Instruction
TOS = 345Ah
PC = 0124h

After Instruction
PC = 0126h
TOS = 0126h
Stack (1 level down) = 345Ah
PIC18F2480/2580/4480/4580

RCALL

Syntax: RCALL n
Operands: -1024 ≤ n ≤ 1023
Operation: (PC) + 2 → TOS,
( PC ) + 2 + 2n → PC
Status Affected: None
Encoding: 1101 lnnn mnnn mnnn
Description: Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.

Words: 1
Cycles: 2
Q Cycle Activity:

Example: HERE RCALL Jump
Before Instruction
PC = Address (HERE)
After Instruction
PC = Address (Jump)
TOS = Address (HERE + 2)

RESET

Syntax: RESET
Operands: None
Operation: Reset all registers and flags that are affected by a MCLR Reset.
Status Affected: All
Encoding: 0000 0000 1111 1111
Description: This instruction provides a way to execute a MCLR Reset in software.

Words: 1
Cycles: 1
Q Cycle Activity:

Example: RESET
After Instruction
Registers = Reset Value
Flags* = Reset Value

APPENDIX H: DATA SHEETS
PIC18F2480/2580/4480/4580

RETFIE

Return from Interrupt

Syntax: RETFIE {s}
Operands: s ∈ [0,1]
Operation: (TOS) → PC, 
1 → GIE/GIEH or PEIE/GIEL,
if s = 1
(WS) → W, 
(STATUS) → Status,
(BSRs) → BSR, 
PCLATU, PCLATH are unchanged.
Status Affected: GIE/GIEH, PEIE/GIEL.
Encoding:

<p>| | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>0000</td>
<td>0000</td>
<td>0001</td>
<td>0000</td>
</tr>
</tbody>
</table>

Description: Return from Interrupt. Stack is popped and Top-of-Stack (TOS) is loaded into the PC. Interrupts are enabled by setting either the high or low priority global interrupt enable bit. If 's' = 1, the contents of the shadow registers, WS, STATUS and BSRs, are loaded into their corresponding registers, W, Status and BSR. If 's' = 0, no update of these registers occurs (default).

Words: 1
Cycles: 2
Q Cycle Activity:

Example: RETFIE 1

After Interrupt
PC = TOS
W = WS
BSR = BSR
Status = STATUS
GIE/GIEH, PEIE/GIEL = 1

RETLW

Return Literal to W

Syntax: RETLW k
Operands: 0 ≤ k ≤ 255
Operation: k → W, (TOS) → PC, PCLATU, PCLATH are unchanged
Status Affected: None
Encoding: 0000 1100 kkkk kkkk
Description: W is loaded with the eight-bit literal 'k'. The program counter is loaded from the top of the stack (the return address). The high address latch (PCLATH) remains unchanged.

Words: 1
Cycles: 2
Q Cycle Activity:

Example:

```
CALL TABLE; W contains table
; offset value
; W now has
; table value

TABLE
ADDWF PCL; W = offset
RETLW k0; Begin table
RETLW k1;
;
RETLW kn; End of table

Before Instruction
W = 07h
After Instruction
W = value of kn
```
**RETURN**

Return from Subroutine

**Syntax:**

```
RETURN {s}
```

**Operands:**

s ∈ [0, 1]

**Operation:**

(TOS) → PC,
if s = 1

(WS) → W,
(STATUSS) → Status,
(BSRS) → BSR,
PCLATU, PCLATH are unchanged

**Status Affected:** None

**Encoding:**

<table>
<thead>
<tr>
<th>0000</th>
<th>0000</th>
<th>0001</th>
<th>001a</th>
</tr>
</thead>
</table>

**Description:**

Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the program counter. If 's' = 1, the contents of the shadow registers, WS, STATUSS and BSRS, are loaded into their corresponding registers, W, Status and BSR. If 's' = 0, no update of these registers occurs (default).

**Words:** 1

**Cycles:** 2

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>Process Data</td>
<td>POP PC from stack</td>
</tr>
</tbody>
</table>

**Example:**

After Interrupt

```
RETURN
```

PC = TOS

---

**RLCF**

Rotate Left f through Carry

**Syntax:**

```
RLCF {d | [a]}
```

**Operands:**

0 ≤ k ≤ 255

d ∈ [0, 1]

**Operation:**

(k<n> → dest<n + 1>,
(k<7>) → C,
(C) → dest<0>

**Status Affected:** C, N, Z

**Encoding:**

<table>
<thead>
<tr>
<th>0011</th>
<th>00da</th>
<th>ffff</th>
<th>ffff</th>
</tr>
</thead>
</table>

**Description:**

The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f' (default).

If 'd' is '0', the Access Bank is selected. If 'd' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever k ≤ 95 (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**

Before Instruction

```
REG  = 1110 0110
C    = 0
```

After Instruction

```
REG  = 1110 0110
W    = 1100 1100
C    = 1
```
PIC18F2480/2580/4480/4580

**RLNCF**

**Syntax:**
```
RLNCF f,d{,a}
```

**Operands:**  
```
0 ≤ f ≤ 255  
d ∈ [0,1]  
a ∈ [0,1]
```

**Operation:**  
```
(f<n>) → dest<n + 1>,  
(f<7>) → dest<0>
```

**Status Affected:**  
```
N, Z
```

**Encoding:**  
```
| 0100 | 01da | ffff | ffff |
```

**Description:**  
The contents of register 'f' are rotated one bit to the left. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f' (default). If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:** 1  
**Cycles:** 1

**Example:**  
```
Before Instruction  
REG = 1010 1011
After Instruction  
REG = 1010 0111
```

---

**RRCF**

**Syntax:**
```
RRCF f,d{,a}
```

**Operands:**  
```
0 ≤ f ≤ 255  
d ∈ [0,1]  
a ∈ [0,1]
```

**Operation:**  
```
(f<n>) → dest<n + 1>,  
(f<7>) → C,  
(C) → dest<7>
```

**Status Affected:**  
```
C, N, Z
```

**Encoding:**  
```
| 0011 | 01da | ffff | ffff |
```

**Description:**  
The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default). If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:** 1  
**Cycles:** 1

**Example:**  
```
Before Instruction  
REG = 1110 0110  
C = 1  
After Instruction  
REG = 1110 0110  
W = 0111 0011  
C = 0
```
### RRNCF

**Syntax:**
RRNCF f{,d{a}}

**Operands:**
- \(0 \leq f \leq 255\)
- \(d \in [0,1]\)
- \(a \in [0,1]\)

**Operation:**
- \((f<n>) \rightarrow \text{dest}<n-1>\)
- \((f<0>) \rightarrow \text{dest}>7\)

**Status Affected:**
- \(N, Z\)

**Encoding:**

<table>
<thead>
<tr>
<th></th>
<th>0100</th>
<th>00da</th>
<th>ffff</th>
<th>ffff</th>
</tr>
</thead>
</table>

**Description:**
The contents of register 'f' are rotated one bit to the right. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).

- If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' is '1', then the bank will be selected as per the BSR value (default).
- If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \(f < 95\) (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Example 1:**
Before Instruction
- REG = 1101 0111

After Instruction
- REG = 1110 1011

**Example 2:**
Before Instruction
- W = ?
- REG = 1101 0111

After Instruction
- W = 1110 1011
- REG = 1101 0111

### SETF

**Syntax:**
SETF f{,a}

**Operands:**
- \(0 \leq f \leq 255\)
- \(a \in [0,1]\)

**Operation:**
- FFh \(\rightarrow f\)

**Status Affected:**
- None

**Encoding:**

|                      | 0110 | 100a | ffff | ffff |

**Description:**
The contents of the specified register are set to FFh.

- If 'a' is '0', the Access Bank is selected.
- If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \(f \leq 95\) (5Fh). See Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Example:**
Before Instruction
- REG = 5Ah

After Instruction
- REG = FFh
SLEEP

Enter Sleep mode

Syntax: SLEEP
Operands: None
Operation: 00h → WDT,
0 → WDT postscaler,
1 → TO,
0 → PD
Status Affected: TO, PD
Encoding: 

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>No operation</td>
<td>Process Data</td>
<td>Go to Sleep</td>
</tr>
</tbody>
</table>

Example: SLEEP
Before Instruction
TO = ?
PD = ?

After Instruction
TO = 1↑
PD = 0

↑ If WDT causes wake-up, this bit is cleared.

SUBFWB

Subtract f from W with Borrow

Syntax: SUBFWB f, [d, a]
Operands: 0 ≤ f ≤ 255
d ∈ [0, 1]
a ∈ [0, 1]
Operation: (W) → (f) → (C) → dest
Status Affected: N, OV, C, DC, Z
Encoding: 

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

Example 1: SUBFWB REG, 1, 0
Before Instruction
REG = 3
W = 2
C = 1
After Instruction
REG = FF
W = 2
Z = 0
N = 1; result is negative

Example 2: SUBFWB REG, 0, 0
Before Instruction
REG = 2
W = 5
C = 1
After Instruction
REG = 2
W = 3
C = 0
N = 0; result is positive

Example 3: SUBFWB REG, 1, 0
Before Instruction
REG = 1
W = 2
C = 0
After Instruction
REG = 0
W = 3
C = 1
Z = 1; result is zero
N = 0
### SUBLW

**Subtract W from Literal**

**Syntax:**

SUBLW k

**Operands:**

0 ≤ k ≤ 255

**Operation:**

k − (W) → W

**Status Affected:**

N, OV, C, DC, Z

**Encoding:**

| 0000 | 1000 | kkkk | kkkk |

**Description:**

W is subtracted from the eight-bit literal 'k'. The result is placed in W.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write to W</td>
</tr>
</tbody>
</table>

**Example 1:**

**Before Instruction**

W = 01h
C = ?

**After Instruction**

W = 01h ; result is positive
C = 1
Z = 0
N = 0

**Example 2:**

**Before Instruction**

W = 02h
C = ?

**After Instruction**

W = 00h ; result is zero
C = 1
Z = 1
N = 0

**Example 3:**

**Before Instruction**

W = 03h
C = ?

**After Instruction**

W = FFh ; (2's complement)
C = 0 ; result is negative
Z = 0
N = 1

### SUBWF

**Subtract W from f**

**Syntax:**

SUBWF f, {d, a} → dest

**Operands:**

0 ≤ f ≤ 255
d ∈ [0, 1]
a ∈ [0, 1]

**Operation:**

(f) − (W) → dest

**Status Affected:**

N, OV, C, DC, Z

**Encoding:**

| 0101 | 110a | ffff | ffff |

**Description:**

Subtract W from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default).

If 'a' is '0', the Access Bank is selected.

If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '1' and the extended instruction set is enabled, this instruction operates in indexed literal offset addressing mode whenever f ≤ 95 (5Fh). See Section 26.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example 1:**

**Before Instruction**

REG = 3
W = 2
C = ?

**After Instruction**

REG = 1
W = 2
C = 1 ; result is positive
Z = 0
N = 0

**Example 2:**

**Before Instruction**

REG = 2
W = 2
C = ?

**After Instruction**

REG = 2
W = 0
C = 1 ; result is zero
Z = 0
N = 0

**Example 3:**

**Before Instruction**

REG = 1
W = 2
C = ?

**After Instruction**

REG = FFh ; (2's complement)
W = 0
C = 0 ; result is negative
Z = 0
N = 1
### SWAPF

**Subtract W from f with Borrow**

**Syntax:**
\[ \text{SWAPF} \quad f\{d \{a\}\} \]

**Operands:**
- \( 0 \leq f \leq 255 \)
- \( d \in [0,1] \)
- \( a \in [0,1] \)

**Operation:**
\( (f) - (W) - (C) \rightarrow \text{dest} \)

**Status Affected:**
- \( N, OV, C, Z, V \)

**Encoding:**
- \( \text{FFFF} \\text{FFFF} \)

**Description:**
Subtract \( W \) and the Carry flag (borrow) from register \( f \) (2's complement method). If 'd' is '0', the result is stored in \( W \). If 'd' is '1', the result is stored back in register \( f \) (default).

- If 'a' is '0', the Access Bank is selected.
- If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \( f \leq 95 \) (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**
1

**Cycles:**
1

<table>
<thead>
<tr>
<th>Q Cycle Activity:</th>
</tr>
</thead>
<tbody>
<tr>
<td>Q1</td>
</tr>
<tr>
<td>\text{Decode}</td>
</tr>
<tr>
<td>Example 1:</td>
</tr>
<tr>
<td>Before Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 00h \quad (0000 \ 0000)</td>
</tr>
<tr>
<td>&amp; \text{C} = 1</td>
</tr>
<tr>
<td>After Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 00h \quad (0000 \ 1011)</td>
</tr>
<tr>
<td>&amp; \text{C} = 1</td>
</tr>
<tr>
<td>&amp; \text{Z} = 0</td>
</tr>
<tr>
<td>&amp; \text{N} = 0 \quad \text{result is positive}</td>
</tr>
<tr>
<td>Example 2:</td>
</tr>
<tr>
<td>Before Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 1Ah \quad (0001 \ 1010)</td>
</tr>
<tr>
<td>&amp; \text{C} = 0</td>
</tr>
<tr>
<td>After Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 00h \quad (0000 \ 0000)</td>
</tr>
<tr>
<td>&amp; \text{C} = 1</td>
</tr>
<tr>
<td>&amp; \text{Z} = 1 \quad \text{result is zero}</td>
</tr>
<tr>
<td>&amp; \text{N} = 0</td>
</tr>
<tr>
<td>Example 3:</td>
</tr>
<tr>
<td>Before Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 0Eh \quad (0000 \ 1110)</td>
</tr>
<tr>
<td>&amp; \text{C} = 1</td>
</tr>
<tr>
<td>After Instruction</td>
</tr>
<tr>
<td>&amp; \text{W} = 0Eh \quad (0000 \ 1110)</td>
</tr>
<tr>
<td>&amp; \text{C} = 0</td>
</tr>
<tr>
<td>&amp; \text{Z} = 0</td>
</tr>
<tr>
<td>&amp; \text{N} = 1 \quad \text{result is negative}</td>
</tr>
</tbody>
</table>

### SWAPF

**Swap f**

**Syntax:**
\[ \text{SWAPF} \quad f\{d \{a\}\} \]

**Operands:**
- \( 0 \leq f \leq 255 \)
- \( d \in [0,1] \)
- \( a \in [0,1] \)

**Operation:**
- \((<3:0>) \rightarrow \text{dest}<7:4>\)
- \((<7:4>) \rightarrow \text{dest}<3:0>\)

**Status Affected:**
None

**Encoding:**
- \( \text{FFFF} \\text{FFFF} \)

**Description:**
Swap upper and lower nibbles of register 'f'. The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in \( W \). If 'd' is '1', the result is placed in register \( f \) (default).

If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever \( f \leq 95 \) (5Fh). See Section 25.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

**Words:**
1

**Cycles:**
1

<table>
<thead>
<tr>
<th>Q Cycle Activity:</th>
</tr>
</thead>
<tbody>
<tr>
<td>Q1</td>
</tr>
<tr>
<td>\text{Decode}</td>
</tr>
<tr>
<td>Example:</td>
</tr>
<tr>
<td>Before Instruction</td>
</tr>
<tr>
<td>After Instruction</td>
</tr>
</tbody>
</table>
TBLRD Table Read

Syntax: TBLRD (*; *+; *-; +*)

Operands: None

Operation:
if TBLRD *,
(Prog Mem (TBLPTR)) → TABLAT;
TBLPTR = No Change;
if TBLRD *+,
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) + 1 → TBLPTR;
if TBLRD *
(TBLPTR) - No Change;
if TBLRD *+
(Prog Mem (TBLPTR)) → TABLAT;
(TBLPTR) - 1 → TBLPTR;

Status Affected: None

Encoding: 0000

Description:
This instruction is used to read the contents of Program Memory (P.M.). To address the program memory, a pointer, called Table Pointer (TBLPTR), is used.

The TBLPTR (a 21-bit pointer) points to each byte in the program memory. TBLPTR has a 2-Mbyte address range.

TBLPTR[0] = 0: Least Significant Byte of Program Memory Word
TBLPTR[0] = 1: Most Significant Byte of Program Memory Word

The TBLRD instruction can modify the value of TBLPTR as follows:
• no change
• post-increment
• post-decrement
• pre-increment

Example 1: TBLRD *+

Before Instruction
TABLAT = 55h
TBLPTR = 00A356h
MEMORY(00A356h) = 34h

After Instruction
TABLAT = 34h
TBLPTR = 00A357h

Example 2: TBLRD *-

Before Instruction
TABLAT = 0Ah
TBLPTR = 01A357h
MEMORY(01A357h) = 12h
MEMORY(01A358h) = 34h

After Instruction
TABLAT = 34h
TBLPTR = 01A358h

APPENDIX H: DATA SHEETS

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**TBLWT Table Write**

**Syntax:**

TBLWT ("*;*+;*-; +")

**Operands:**

None

**Operation:**

if TBLWT*, (TABLAT) → Holding Register;
(TBLPTR) + 1 → TBLPTR;
if TBLWT+*, (TABLAT) → Holding Register;
(TBLPTR) + 1 → TBLPTR;
if TBLWT*-,* (TABLAT) → Holding Register;
(TBLPTR) + 1 → TBLPTR;
if TBLWT+,* (TBLPTR) + 1 → TBLPTR;
(TBLAT) → Holding Register;

**Status Affected:**

None

**Encoding:**

<table>
<thead>
<tr>
<th>0000</th>
<th>0000</th>
<th>0000</th>
<th>linn</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td></td>
<td>=0</td>
</tr>
<tr>
<td>=1</td>
<td>*</td>
<td>=2</td>
<td>*</td>
</tr>
<tr>
<td>=3</td>
<td>+</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Description:**

This instruction uses the 3 LSBs of the TBLPTR to determine which of the 8 holding registers the TABLAT is written to. The holding registers are used to program the contents of Program Memory (P.M.). (Refer to Section 6.0 "Flash Program Memory" for additional details on programming Flash memory.)

The TBLPTR (a 21-bit pointer) points to each byte in the program memory. TBLPTR has a 2-MByte address range. The LSB of the TBLPTR selects which byte of the program memory location to access.

TBLPTR[0] = 0: Least Significant Byte of Program Memory Word
TBLPTR[0] = 1: Most Significant Byte of Program Memory Word

The TBLWT instruction can modify the value of TBLPTR as follows:

- no change
- post-increment
- post-decrement
- pre-increment

**Words:** 1

**Cycles:** 2

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
<td>No operation</td>
</tr>
<tr>
<td>No operation</td>
<td>No operation (Read TABLAT)</td>
<td>No operation</td>
<td>No operation (Write to Holding Register)</td>
</tr>
</tbody>
</table>

Example 1:

**Before Instruction**

TABLAT = 55h
TBLPTR = 00A356h
HOLDING REGISTER (00A356h) = FFh

**After Instructions (table write completion)**

TABLAT = 55h
TBLPTR = 00A357h
HOLDING REGISTER (00A356h) = FFh

Example 2:

**Before Instruction**

TABLAT = 34h
TBLPTR = 01389Ah
HOLDING REGISTER (01389Ah) = FFh

**After Instruction (table write completion)**

TABLAT = 34h
TBLPTR = 01389Bh
HOLDING REGISTER (01389Ah) = FFh
HOLDING REGISTER (01389Bh) = 34h
TSTFSZ  | Test f, Skip if 0
--- | ---
Syntax: | TSTFSZ f (a)
Operands: | 0 ≤ f ≤ 255
Operation: | skip if f = 0
Status Affected: | None
Encoding: | 0110 011a fffe fffe
Description: | If 'f' = 0, the next instruction fetched during the current instruction execution is discarded and a NOP is executed, making this a two-cycle instruction.
Operands: | 0 ≤ f ≤ 255
Operation: | skip if f = 0
Status Affected: | None
Encoding: | 0000 1010 kkkk kkkk
Description: | The contents of W are XORed with the 8-bit literal 'k'. The result is placed in W.

XORLW  | Exclusive OR Literal with W
--- | ---
Syntax: | XORLW k
Operands: | 0 ≤ k ≤ 255
Operation: | (W) XOR k → W
Status Affected: | N, Z
Encoding: | 0000 1010 kkkk kkkk

Before Instruction
W = B5h
After Instruction
W = 1Ah

Example:
XORLW 0AFh
## XORWF

**Exclusive OR W with f**

### Syntax:
```plaintext
XORWF f{,d [,a]}
```

### Operands:
- `0 ≤ f ≤ 255`
- `d ∈ [0,1]`
- `a ∈ [0,1]`

### Operation:
```plaintext
(W) XOR. (f) → dest
```

### Status Affected:
- N, Z

### Encoding:

<p>| | | | |</p>
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>0001</td>
<td>10da</td>
<td>ffff</td>
<td>ffff</td>
</tr>
</tbody>
</table>

### Description:
Exclusive OR the contents of W with register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in the register 'f' (default).

- If 'a' is '0', the Access Bank is selected.
- If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever `f < 95 (5Fh)`. See Section 26.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.

### Words:
1

### Cycles:
1

### Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read</td>
<td>Process</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

### Example:
```plaintext
XORWF REG, 1, 0
```

**Before Instruction**
- `REG = AFh`
- `W = B5h`

**After Instruction**
- `REG = 1Ah`
- `W = B5h`
25.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, PIC18F2480/2580/4480/4580 devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment indirect and indexed addressing operations and the implementation of Indexed Literal Offset Addressing mode for many of the standard PIC18 instructions.

The additional features are disabled by default. To enable them, users must set the XINST configuration bit.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers or use them for indexed addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULINK and SUBULINK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- dynamic allocation and de-allocation of software stack space when entering and leaving subroutines
- function pointer invocation
- software Stack Pointer manipulation
- manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 25-3. Detailed descriptions are provided in Section 25.2.2 “Extended Instruction Set”. The opcode field descriptions in Table 25-1 apply to both the standard and extended PIC18 instruction sets.

![Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.]

25.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of indexed addressing, it is enclosed in square brackets ([ ]). This is done to indicate that the argument is used as an index or offset. MPASM™ Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 25.2.3.1 “Extended Instruction Syntax with Standard PIC18 Commands”.

![Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ({}).]

### TABLE 25-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

<table>
<thead>
<tr>
<th>Mnemonic, Operands</th>
<th>Description</th>
<th>Cycles</th>
<th>16-Bit Instruction Word</th>
<th>Status Affected</th>
</tr>
</thead>
<tbody>
<tr>
<td>ADDFSR f, k</td>
<td>Add literal to FSR</td>
<td>1</td>
<td>1110 1000 ffkk kkkkk</td>
<td>None</td>
</tr>
<tr>
<td>ADDULNK k</td>
<td>Add literal to FSR2 and return</td>
<td>2</td>
<td>1110 1000 11kk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>CALLW</td>
<td>Call subroutine using WREG</td>
<td>2</td>
<td>0000 0000 0001 0100</td>
<td>None</td>
</tr>
<tr>
<td>MOVFSF z&lt;sub&gt;f&lt;/sub&gt;, f&lt;sub&gt;d&lt;/sub&gt;</td>
<td>Move Z&lt;sub&gt;f&lt;/sub&gt; (source) to 1st word f&lt;sub&gt;d&lt;/sub&gt; (destination) 2nd word</td>
<td>1</td>
<td>1110 1011 0zzz kkkkk</td>
<td>None</td>
</tr>
<tr>
<td>MOVSS z&lt;sub&gt;f&lt;/sub&gt;, z&lt;sub&gt;d&lt;/sub&gt;</td>
<td>Move Z&lt;sub&gt;f&lt;/sub&gt; (source) to 1st word Z&lt;sub&gt;d&lt;/sub&gt; (destination) 2nd word</td>
<td>2</td>
<td>1111 ffff ffff ffff</td>
<td>None</td>
</tr>
<tr>
<td>PUSHL k</td>
<td>Store literal at FSR2, decrement FSR2</td>
<td>1</td>
<td>1110 1010 kkkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>SUBFSR f, k</td>
<td>Subtract literal from FSR</td>
<td>1</td>
<td>1110 1001 ffkk kkkk</td>
<td>None</td>
</tr>
<tr>
<td>SUBULNK k</td>
<td>Subtract literal from FSR2 and return</td>
<td>2</td>
<td>1110 1001 11kk kkkk</td>
<td>None</td>
</tr>
</tbody>
</table>
## 25.2.2 EXTENDED INSTRUCTION SET

### ADDFSR  Add Literal to FSR

**Syntax:**
```
ADDFSR f, k
```

**Operands:**
```
0 ≤ k ≤ 63
f ∈ {0, 1, 2}
```

**Operation:**
```
FSR(f) + k → FSR(f)
```

**Status Affected:**
None

**Encoding:**
```
1110 1000 ffkk kkkk
```

**Description:**
The 6-bit literal 'k' is added to the contents of the FSR specified by 'f'.

**Words:**
1

**Cycles:**
1

**Q Cycle Activity:**
```
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write to FSR</td>
</tr>
</tbody>
</table>
```

**Example:**
```
ADDFSR 2, 23h
```

Before Instruction
```
FSR2 = 03FFh
```

After Instruction
```
FSR2 = 0422h
```

### ADDULNK  Add Literal to FSR2 and Return

**Syntax:**
```
ADDULNK k
```

**Operands:**
```
0 ≤ k ≤ 63
```

**Operation:**
```
FSR2 + k → FSR2,
PC = (TOS)
```

**Status Affected:**
None

**Encoding:**
```
1110 1000 llkk kkkk
```

**Description:**
The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.

The instruction takes two cycles to execute; a NOP is performed during the second cycle.

This may be thought of as a special case of the ADDFSR instruction, where f = 3 (binary '111'); it operates only on FSR2.

**Words:**
1

**Cycles:**
2

**Q Cycle Activity:**
```
<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read literal 'k'</td>
<td>Process Data</td>
<td>Write to FSR</td>
</tr>
<tr>
<td>No Operation</td>
<td>No Operation</td>
<td>No Operation</td>
<td>No Operation</td>
</tr>
</tbody>
</table>
```

**Example:**
```
ADDULNK 23h
```

Before Instruction
```
FSR2 = 03FFh
PC = 0100h
TOS = 02AFh
```

After Instruction
```
FSR2 = 0422h
PC = 02AFh
TOS = TOS - 1
```

**Note:** All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction syntax then becomes: (label) instruction argument(s).
CALLW Subroutine Call Using WREG

Syntax: CALLW
Operands: None
Operation: (PC + 2) \rightarrow TOS,
(W) \rightarrow PCL,
(PCLATH) \rightarrow PCH,
(PCLATU) \rightarrow PCU
Status Affected: None
Encoding: 0000 0000 0001 0100
Description: First, the return address (PC + 2) is pushed onto the return stack. Next, the contents of W are written to PCL; the existing value is discarded. Then, the contents of PCLATH and PCLATU are latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while the new next instruction is fetched. Unlike CALL, there is no option to update W, Status or BSR.

Words: 1
Cycles: 2
Q Cycle Activity:

Example: HERE CALLW

Before Instruction
PC = address (HERE)
PCLATH = 10h
PCLATU = 00h
W = 06h
After Instruction
PC = 001006h
PCLATH = 10h
PCLATU = 00h
W = 06h

MOVSF Move Indexed to f

Syntax: MOVSF [z], f
Operands: 0 ≤ z ≤ 127
0 ≤ f ≤ 4095
Operation: ((FSR2) + z) \rightarrow f
Status Affected: None
Encoding: 1st word (source)
2nd word (destin.)

Description: The contents of the source register are moved to destination register 'f'. The actual address of the source register is determined by adding the 7-bit literal offset 'z' in the first word to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f' in the second word. Both addresses can be anywhere in the 4096-byte data space (000h to FFFh).

The MOVSF instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register. If the resultant source address points to an indirect addressing register, the value returned will be 00h.

Words: 2
Cycles: 2
Q Cycle Activity:

Example: MOVSF [03h], REG2

Before Instruction
FSR2 = 80h
Contents of 85h = 33h
REG2 = 11h
After Instruction
FSR2 = 80h
Contents of 85h = 33h
REG2 = 33h
MOVSS  Move Indexed to Indexed

Syntax:  MOVSS [zi], [zj]
Operands:  0 ≤ zi ≤ 127
0 ≤ zj ≤ 127
Operation:  (FSR2) + zj → (FSR2) + zi
Status Affected:  None
Encoding:
1st word (source)  1110 1011 lzzz zzzz
2nd word (dest)  1111 xxxx xzzz zzzz

Description:  The contents of the source register are moved to the destination register. The addresses of the source and destination registers are determined by adding the 7-bit literal offsets 'zi' or 'zj', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space (000h to FFFh).

Words:  2
Cycles:  2
Q Cycle Activity:

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Determine source addr</td>
<td>Determine source addr</td>
<td>Read source reg</td>
</tr>
<tr>
<td>Decode</td>
<td>Determine dest addr</td>
<td>Determine dest addr</td>
<td>Write to dest reg</td>
</tr>
</tbody>
</table>

Example:  MOVSS [05h], [06h]
Before Instruction
FSR2 = 80h
Contents of 05h = 33h
Contents of 06h = 11h
After Instruction
FSR2 = 80h
Contents of 05h = 33h
Contents of 06h = 33h

PUSHL  Store Literal at FSR2, Decrement FSR2

Syntax:  PUSHL k
Operands:  0 ≤ k ≤ 255
Operation:  k → (FSR2), FSR2 − 1 → FSR2
Status Affected:  None
Encoding:

<table>
<thead>
<tr>
<th>Dec</th>
<th>Read 'k'</th>
<th>Process data</th>
<th>Write to destination</th>
</tr>
</thead>
<tbody>
<tr>
<td>1111</td>
<td>1010</td>
<td>kkkk</td>
<td>kkkk</td>
</tr>
</tbody>
</table>

Description:  The 8-bit literal 'k' is written to the data memory address specified by FSR2. FSR2 is decremented by 1 after the operation. This instruction allows users to push values onto a software stack.

Words:  1
Cycles:  1
Q Cycle Activity:

Example:  PUSHL 08h
Before Instruction
FSR2H:FSR2L = 01ECh
Memory (01ECh) = 00h
After Instruction
FSR2H:FSR2L = 01E8h
Memory (01ECh) = 08h
**SUBFSR**  Subtract Literal from FSR

**Syntax:**

\[ \text{SUBFSR } f, k \]

**Operands:**

\[ 0 \leq k \leq 63 \]

\[ f \in \{0, 1, 2\} \]

**Operation:**

\[ \text{FSR}_f - k \rightarrow \text{FSR}_f \]

**Status Affected:**

None

**Encoding:**

1110 1001 fffk kkkk

**Description:**

The 6-bit literal 'k' is subtracted from the contents of the FSR specified by 'f'.

**Words:**

1

**Cycles:**

1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**

**Before Instruction**

| FSR2 = 03FFh |

**After Instruction**

| FSR2 = 03DCh |

---

**SUBULNK**  Subtract Literal from FSR2 and Return

**Syntax:**

\[ \text{SUBULNK } k \]

**Operands:**

\[ 0 \leq k \leq 63 \]

**Operation:**

\[ \text{FSR2} - k \rightarrow \text{FSR2} \]

\[ (\text{TOS}) \rightarrow \text{PC} \]

**Status Affected:**

None

**Encoding:**

1110 1001 11kk kkkk

**Description:**

The 6-bit literal 'k' is subtracted from the contents of the FSR2. A return is then executed by loading the PC with the TOS.

The instruction takes two cycles to execute; a NOP is performed during the second cycle.

This may be thought of as a special case of the SUBFSR instruction, where \( f = 3 \) (binary '11'); it operates only on FSR2.

**Words:**

1

**Cycles:**

2

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register 'f'</td>
<td>Process</td>
<td>Write to destination</td>
</tr>
<tr>
<td>No</td>
<td>No</td>
<td>Operation</td>
<td>No</td>
</tr>
</tbody>
</table>

**Example:**

**Before Instruction**

| FSR2 = 03FFh |

| PC = 0100h |

**After Instruction**

| FSR2 = 03DCh |

| PC = (TOS) |

---

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**APPENDIX H: DATA SHEETS**

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25.2.3 BYTE-ORIENTED AND ASYMMETRIC INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note: Enabling the PIC18 instruction set extension may cause legacy applications to behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing mode (Section 5.8.1 “Indexed Addressing with Literal Offset”). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank (a = 0), or in a GPR bank designated by the BSR (a = 1). When the extended instruction set is enabled and a = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between ‘C’ and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 25.2.3.1 “Extended Instruction Syntax with Standard PIC18 Commands”).

Although the Indexed Literal Offset Addressing mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset Addressing mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

25.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, ‘f’, in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, ‘k’. As already noted, this occurs only when ‘f’ is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM™ Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be ‘0’. This is in contrast to standard operation (extended instruction set disabled) when ‘a’ is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument, ‘d’, functions as before.

In the latest versions of the MPASM assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, /y, or the PE directive in the source listing.

25.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F2480/2580/4480/4580, it is very important to consider the type of code. A large, re-entrant application that is written in ‘C’ and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.
### ADDWF

**ADD W to Indexed (Indexed Literal Offset mode)**

**Syntax:**
ADDWF \([k\] , \d)

**Operands:**
0 ≤ k ≤ 95
\(d \in [0,1]\)
a = 0

**Operation:**
(W) + ((FSR2) + k) → dest

**Status Affected:**
N, OV, C, DC, Z

**Encoding:**
0010 0130 kkkk kkkk

**Description:**
The contents of W are added to the contents of the register indicated by FSR2, offset by the value 'k'.
If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register T (default).

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read ('k)</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**
ADDWF \([OPST], 0\)

Before Instruction
W = 17h
OFST = 2Ch
FSR2 = 0A00h
Contents of 0A2Ch = 20h

After Instruction
W = 37h
Contents of 0A2Ch = 20h

### BSF

**Bit Set Indexed (Indexed Literal Offset mode)**

**Syntax:**
BSF \([k\] , b)

**Operands:**
0 ≤ k ≤ 95
0 ≤ b ≤ 7
a = 0

**Operation:**
1 → ((FSR2 + k)\<b>)

**Status Affected:**
None

**Encoding:**
1000 bbbc kkkk kkkk

**Description:**
Bit 'b' of the register indicated by FSR2, offset by the value 'k', is set.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read register T</td>
<td>Process Data</td>
<td>Write to destination</td>
</tr>
</tbody>
</table>

**Example:**
BSF \([FLAG\_OPST], 7\)

Before Instruction
FLAG\_OPST = 0Ah
FSR2 = 0A00h
Contents of 0A0Ah = 55h

After Instruction
Contents of 0A0Ah = D5h

### SETF

**Set Indexed (Indexed Literal Offset mode)**

**Syntax:**
SETF \([k\]

**Operands:**
0 ≤ k ≤ 95

**Operation:**
FFh → ((FSR2) + k)

**Status Affected:**
None

**Encoding:**
0110 1000 kkkk kkkk

**Description:**
The contents of the register indicated by FSR2, offset by 'k', are set to FFh.

**Words:** 1

**Cycles:** 1

**Q Cycle Activity:**

<table>
<thead>
<tr>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
</tr>
</thead>
<tbody>
<tr>
<td>Decode</td>
<td>Read ('k)</td>
<td>Process Data</td>
<td>Write register</td>
</tr>
</tbody>
</table>

**Example:**
SETF \([OPST]\)

Before Instruction
OFST = 2Ch
FSR2 = 0A00h
Contents of 0A2Ch = 00h

After Instruction
Contents of 0A2Ch = FFh

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25.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB® IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set of the PIC18F2480/2580/4480/4580 family of devices. This includes the MPLAB C18 C compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).

When selecting a target device for software development, MPLAB IDE will automatically set default configuration bits for that device. The default setting for the XINST configuration bit is '0', disabling the extended instruction set and Indexed Literal Offset Addressing mode. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option, or dialog box within the environment, that allows the user to configure the language tool and its settings for the project
- A command line option
- A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.
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